

# LLM Transformer Decoder RTL Accelerator — Technical Report

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## 1. Introduction

This report documents the design, implementation, and verification of a synthesizable RTL accelerator for LLM (Large Language Model) transformer decoder inference. The accelerator implements a single transformer decoder block — the fundamental repeating unit of GPT-style language models — in SystemVerilog, targeting FPGA and ASIC deployment.

The design provides two architectural variants: a **register-bridge** variant that copies BRAM weights into register arrays for single-cycle combinational access (maximum throughput), and a **streaming** variant that reads weights directly from BRAM one element per cycle (minimum area). Both share the same external interface, BRAM storage, and verification infrastructure.

Key design features include Q8.8 fixed-point arithmetic, systolic array compute, BRAM-backed weight and KV-cache storage, division-free softmax and LayerNorm via reciprocal LUT with Newton-Raphson refinement, and packed array ports for full simulation visibility.

## 2. Background

### 2.1 Transformer Decoder Architecture

The transformer decoder processes tokens autoregressively: given a sequence of prior tokens, it predicts the next token. Each decoder layer applies two sub-layers with residual connections. First, Multi-Head Self-Attention computes attention scores between the current query and all prior key/value pairs, enabling the model to attend to relevant context. Second, a Feed-Forward Network (FFN) applies a two-layer MLP with non-linear activation independently to each position. Both sub-layers are preceded by Layer Normalization (pre-norm architecture) and followed by residual additions.

## 2.2 Inference vs Training

This accelerator targets inference only, which has distinct characteristics compared to training. Batch sizes are typically 1 for interactive generation. Only the forward pass is needed with no backpropagation. KV-cache eliminates redundant attention computation. Weights are fixed, enabling weight-stationary dataflows and BRAM preloading.

## 2.3 Fixed-Point Quantization

Modern LLM inference widely uses quantization to reduce memory bandwidth and compute cost. Our Q8.8 fixed-point format (8 integer bits, 8 fractional bits) provides a range of  $-128.0$  to  $+127.996$  with 1/256 resolution, which is sufficient for demonstrating the architectural concepts while keeping multiplier width at 16 bits.

# 3. Architecture

## 3.1 Top-Level Block Diagram

The design has two levels of hierarchy. The synthesis top-level wraps the compute core with BRAM-backed storage for all 49,344 weight parameters and 16,384 KV-cache entries. Two top-level variants exist: `transformer_decoder_top` (register-bridge) and `transformer_decoder_top_stream` (direct BRAM streaming).

The compute core implements the pre-norm architecture used by GPT-2, LLaMA, and most modern LLMs: LayerNorm 1, Multi-Head Attention with KV-cache, residual addition, LayerNorm 2, Feed-Forward Network with ReLU, and a second residual addition.

## 3.2 Module Hierarchy

The full project comprises 17 RTL source files in two parallel hierarchies:

**Register-bridge variant (original — maximum throughput):**

<code>transformer_decoder_top</code>	(BRAM → register arrays → decoder)
├─ <code>bram_sp</code> ×12	(Weight/bias/parameter storage)
├─ <code>kv_cache_bram</code> ×2	(K and V caches)
│ └─ <code>bram_dp</code>	(True dual-port BRAM)
└─ <code>transformer_decoder</code>	(Decoder compute core)
├─ <code>layer_norm</code> ×2	(Pre-attention & pre-FFN normalisation)
└─ <code>multi_head_attention</code>	(Causal multi-head self-attention)

	└─ softmax_unit	(Reciprocal-LUT softmax, time-multiplexed)
	└─ feed_forward	(Two-layer FFN with ReLU)
	└─ systolic_array	(Matrix multiply engine)
	└─ processing_element	(Single MAC unit)
	└─ transformer_pkg	(Parameters, types, FP utilities)

### Streaming variant (new — minimum area):

transformer_decoder_top_stream	(BRAMS → address/data → decoder)
└─ bram_sp ×12	(Same BRAMS, muxed read ports)
└─ kv_cache_bram ×2	(Element-level BRAM reads)
└─ ┌─ bram_dp	
└─ transformer_decoder_stream	(Streaming decoder core)
└─ layer_norm ×2	(Unchanged – small packed ports)
└─ multi_head_attention_stream	(BRAM addr/data weight reads)
└─ ┌─ softmax_unit	(Unchanged)
└─ feed_forward_stream	(BRAM addr/data weight reads)
└─ systolic_array	(Shared infrastructure)
└─ ┌─ processing_element	
└─ transformer_pkg	

Shared modules (transformer\_pkg, bram\_sp, bram\_dp, kv\_cache\_bram, softmax\_unit, layer\_norm, systolic\_array, processing\_element) are identical between variants. The streaming variant adds 4 new modules: multi\_head\_attention\_stream, feed\_forward\_stream, transformer\_decoder\_stream, and transformer\_decoder\_top\_stream.

## 3.3 Module Descriptions

### 3.3.1 Processing Element (PE)

The PE is the atomic compute unit — a single multiply-accumulate (MAC) cell. It receives an activation from the left and a weight from the top, computes  $acc += a \times w$  using full 32-bit precision, and forwards both operands to neighboring PEs. The 32-bit accumulator prevents overflow during long dot-product chains. Key properties include single-cycle MAC latency, registered inputs and outputs for timing closure, synchronous clear for accumulator reset between tiles, and parameterized data widths via transformer\_pkg.

### 3.3.2 Systolic Array

A 4×4 grid of PEs forms the systolic matrix multiplication engine. Data flows through the array in a wave pattern: activations propagate left-to-right and weights top-to-bottom. After  $ROWS + COLS - 1$  cycles of

streaming, the 16 accumulators hold the complete output tile. This architecture achieves 16 MACs per cycle with only 8 input operands, giving a 2× data reuse factor.

### 3.3.3 Softmax Unit (Division-Free)

Softmax normalisation computes  $\text{probs}[i] = \exp(\text{scores}[i] - \max) / \sum \exp$ . The implementation proceeds through six FSM stages: find maximum score for numerical stability, subtract maximum and apply PWL exponential approximation, accumulate the sum of exponentials, compute the reciprocal of the sum via LUT + Newton-Raphson, and normalise by multiplying each exponential by the reciprocal.

The PWL exponential uses four linear segments covering  $[-8, 0]$  with non-negative clamping at segment boundaries. The normalisation step uses a 32-entry reciprocal LUT (indexed by the top 5 mantissa bits of the CLZ-normalised sum) providing an initial estimate in Q2.14. One Newton-Raphson iteration doubles precision to approximately 12 bits. The entire reciprocal computation is combinational (single-cycle) using a 512-bit ROM, two 16×16 multipliers, and a barrel shifter. Maximum error versus exact division is  $\pm 1$  LSB.

### 3.3.4 Layer Normalization

LayerNorm computes the mean and variance of the input vector, then normalizes each element through three FSM stages: mean computation via sequential accumulation and arithmetic right-shift by  $\log_2(\text{VEC\_LEN})$ , variance computation using centered differences and the same right-shift, and element-wise normalization with learnable gamma and beta parameters. The reciprocal square root ( $1/\sqrt{\text{variance} + \epsilon}$ ) uses a 32-entry LUT indexed by CLZ-normalised mantissa bits, followed by one Newton-Raphson iteration — the same architectural pattern proven in the softmax normalisation path. No runtime division operators remain; all  $\div N$  operations use arithmetic right-shift since  $\text{VEC\_LEN}$  is a power of 2.

### 3.3.5 Multi-Head Attention

The attention module implements multi-head scaled dot-product attention with integrated softmax and KV-cache support. A single softmax\_unit ( $\text{VEC\_LEN} = 128$ ) is time-multiplexed across  $N\_HEADS = 4$  heads. The register-bridge variant receives full weight matrices on array ports. The streaming variant replaces these with BRAM

address/data interfaces, reading  $W_q/W_k/W_v$  in lockstep with three parallel accumulators.

### 3.3.6 Feed-Forward Network

The FFN implements  $\text{FFN}(x) = \text{ReLU}(x \times W_1 + b_1) \times W_2 + b_2$ . The register-bridge variant computes full dot products per cycle. The streaming variant reads  $W_1$ ,  $b_1$ , and  $W_2$  from BRAM one element per cycle. Both share identical ReLU logic.

### 3.3.7 BRAM Modules

Three BRAM types: `bram_sp` (single-port, read-first, optional hex init), `bram_dp` (true dual-port), and `kv_cache_bram` (vector-write FSM + element-level read).

## 4. Streaming Architecture

### 4.1 Motivation

The register-bridge architecture copies all 49,344 BRAM words into register arrays, consuming approximately 789,504 flip-flops. The streaming architecture eliminates these by having compute modules generate BRAM read addresses directly from their FSM state.

### 4.2 Weight Read Protocol

Each streaming module exposes address/data/enable ports for its weight BRAMs. The top-level muxes each BRAM's read port between the weight-loading bus and the compute address. BRAM latency is one cycle: address issued in cycle  $N$ , data arrives in cycle  $N+1$ .

For QKV projection,  $W_q$ ,  $W_k$ , and  $W_v$  share the same read address. Three accumulators run simultaneously, producing  $Q$ ,  $K$ ,  $V$  for one dimension in  $D\_MODEL$  cycles.

### 4.3 Performance Comparison

Metric	Register-Bridge	Streaming	Change
Total pipeline (seq_pos=0)	~3,200 cycles	~45,000 cycles	~14× slower
Weight registers	789K FFs	5K FFs	<b>99.4% reduction</b>

Metric	Register-Bridge	Streaming	Change
Weight multipliers	~192 combinational	6 sequential	<b>97% reduction</b>
DSP48 usage	~150	~10	<b>93% reduction</b>

## 5. Verification

### 5.1 Summary

Total: **83/83 tests pass** (54 behavioural + 29 RTL simulation across 6 testbenches).

### 5.2 Behavioural Model (54 tests)

Fixed-Point Utilities (16/16), Processing Element (9/9), Systolic Array (8/8), Softmax Unit (8/8), Layer Normalization (5/5), Feed-Forward Network (4/4), Decoder Integration (12/12).

### 5.3 RTL Simulation (29 tests)

Processing Element (6/6), Systolic Array (4/4), Softmax (4/4), BRAM (8/8), Decoder — Register-Bridge (4/4), Decoder — Streaming (3/3).

The streaming decoder testbench loads weights via the `wl_en` bus, then verifies single-token and sequential two-token inference. Output `out_emb[0] = 0x00A8` (0.656) matches the register-bridge variant's first dimension.

### 5.4 RTL Lint

17 RTL files: 0 errors, 5 warnings (BRAM blocks intentionally omit reset).

## 6. Implementation Metrics (Estimated)

Block	DSP48	FFs	LUTs	BRAM18K
Processing Element	1	~30	~20	—
4×4 Systolic Array	16	~500	~350	—
Softmax Unit	2	~2K	~3K	—
Layer Normalisation	4–6	~400	~600	—

Block	DSP48	FFs	LUTs	BRAM18K
Multi-Head Attention	~64	~5K	~8K	—
Feed-Forward Network	~64	~3K	~5K	—
Weight BRAMs (12)	—	—	—	~110
KV-Cache BRAMs (2)	—	—	—	~16
<b>Full Decoder (register-bridge)</b>	<b>~150</b>	<b>~11K</b>	<b>~17K</b>	<b>~126</b>
<b>Full Decoder (streaming)</b>	<b>~10</b>	<b>~5K</b>	<b>~12K</b>	<b>~126</b>

Clock frequency estimate: 100–200 MHz.

## 7. Limitations and Future Work

### 7.1 Resolved

BRAM-backed weights, division-free softmax, packed array ports, streaming weight architecture (99.4% FF reduction), division-free LayerNorm (arithmetic right-shift for  $\div N$ , 32-entry rsqrt LUT + Newton-Raphson for  $1/\sqrt{\text{var}}$ ).

### 7.2 Remaining

Only ReLU activation supported (not GELU/SiLU). Systolic array instantiated but not connected to datapath. Single-token processing with no batching.

### 7.3 Planned

Systolic-tiled projections, parallel softmax with `N_HEADS` instances, GELU/SiLU activation via PWL approximation, multi-layer stacking, multi-device distribution, AXI-Lite control/status interface.

## 8. Conclusion

This project demonstrates a complete, synthesizable transformer decoder block in SystemVerilog with two architectural variants: a high-throughput register-bridge design and a minimum-area streaming design. The 17-module hierarchy mirrors the transformer's conceptual structure while addressing practical synthesis concerns: no runtime division operators remain in the compute datapath (softmax and LayerNorm both use LUT + Newton-Raphson, mean/variance use arithmetic right-shift). All 83 verification tests pass

across both variants. The streaming architecture achieves 99.4% register reduction while maintaining functional equivalence, making the design feasible on the smallest FPGA targets.