

## Lecture 13: Passivity-based design

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- Adaptation of Feedforward Gain
- Output feedback
- Strictly positive real functions (*SPR*), Kalman-Yakubovich (*KY*) Lemma
- The Augmented Error

## General Case for State Feedback

Given a continuous time system and the target dynamics

$$\dot{x} = A x + B u, \quad \dot{x}_m = A_m x_m + B_m u_c$$

With the controller and the error signals

$$u(t) = M(\theta) u_c(t) - L(\theta) x(t), \quad e(t) = x(t) - x_m(t)$$

Assuming solvability of the model-matching problem in the nominal case  $\theta = \theta^0$ , the error dynamics are

$$\dot{e} = A_m e + \underbrace{\left( A - A_m - B L \right) (e + x_m(t)) + \left( B M - B_m \right) u_c(t)}_{\Psi(t, e) (\theta - \theta^0)}$$

The adaptation law can be designed using ( $\Gamma = \Gamma^T > 0$ )

$$V(e, \theta - \theta^0) = \frac{1}{2} \left[ e^T P e + (\theta - \theta^0)^T \Gamma^{-1} (\theta - \theta^0) \right]$$

## General Case for State Feedback (cont'd)

The time-derivative of  $V = \frac{1}{2} [e^T P e + (\theta - \theta^0)^T \Gamma^{-1} (\theta - \theta^0)]$  is

$$\dot{V} = \frac{1}{2} e^T [P \mathbf{A}_m + \mathbf{A}_m^T P] e + (\theta - \theta^0)^T \Psi^T P e + (\theta - \theta^0)^T \Gamma^{-1} \dot{\theta}$$

Assuming that  $P = P^T > 0$  is the solution of

$$P \mathbf{A}_m + \mathbf{A}_m^T P = -Q, \quad Q = Q^T > 0$$

and taking the adaptation law as

$$\dot{\theta} = -\Gamma \Psi^T(t, e) P e = -\Gamma \Psi^T(t, e) P (x - x_m)$$

one obtains

$$\frac{d}{dt} V(e(t), \theta(t) - \theta^0) = -\frac{1}{2} e^T(t) Q e(t)$$

and concludes that the signals are bounded and  $\lim_{t \rightarrow \infty} e(t) = 0$ .

## Adaptation of Feedforward Gain

Changing the gain of a stable process

$$y(t) = k \cdot G_0(p) \left( u(t) \right) \longrightarrow y_m(t) = k_0 \cdot G_0(p) \left( u_c(t) \right)$$

can be done using the simple controller  $u(t) = \theta(t) u_c(t)$

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To design an adaptation law, we compute the error signal

$$e(t) = y - y_m = k \cdot G_0(p) (\theta(t) - \theta^0) u_c(t), \quad \theta^0 = k_0/k$$

and proceed as follows.

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and proceed as follows.

First, find a minimal realization  $\{A, B, C\}$

$$G(s) = k \cdot G_0(s) = C^T (sI - A)^{-1} B$$

with known (independent on  $k$ )  $A$  and  $B$ .

## Adaptation of Feedforward Gain (cont'd)

For the state-space model

$$\dot{x} = A x + B (\theta - \theta^0) u_c(t), \quad e(t) = C^T x(t)$$

with a Hurwitz matrix  $A$  (since  $G(s)$  is stable), one can find the solution  $P = P^T > 0$  of the Lyapunov equation

$$P A + A^T P = -Q, \quad Q = Q^T > 0$$

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and consider the Lyapunov function

$$V(x, \theta - \theta^0) = \frac{1}{2} \left[ \gamma x^T P x + (\theta - \theta^0)^2 \right]$$



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Taking

$$\dot{\theta} = -\gamma x^T P B u_c(t)$$

we would have

$$\dot{V} = -\frac{\gamma}{2} x^T Q x \Rightarrow x(t) \rightarrow 0 \Rightarrow e(t) \rightarrow 0$$

## Adaptation of Feedforward Gain (cont'd)

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However, the adaptation law

$$\dot{\theta} = -\gamma x(t)^T P B u_c(t)$$

cannot be used since the signal  $x(t)$  is not available!

We only have  $u_c(t)$  and  $e(t) = y(t) - y_m(t)$ .

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Suppose the matrix  $Q$  is chosen such that

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Suppose the matrix  $Q$  is chosen such that

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Then, we have

$$e(t) = C^T x(t) = x(t)^T C = x(t)^T P B$$

and the adaptation law can be rewritten as

$$\dot{\theta} = -\gamma e(t) u_c(t)$$

Remark: the MIT rule gives:

$$\dot{\theta} = -\gamma e(t) y_m(t)$$

## SPR and KY lemma

Definition: A rational transfer function  $G(s)$  is *positive real* (**PR**) if

$$\operatorname{Re}\{G(s)\} \geq 0 \quad \text{for} \quad \operatorname{Re}\{s\} \geq 0$$

and *strictly positive real* (**SPR**) if

$$\operatorname{Re}\{G(s - \varepsilon)\} \geq 0 \quad \text{for} \quad \operatorname{Re}\{s\} \geq 0$$

for some  $\varepsilon > 0$ .

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for some  $\varepsilon > 0$ .

Lemma (Kalman-Yakubovich): The transfer function

$$G(s) = C^T (s I - A)^{-1} B$$

where  $(A, B)$  is controllable and  $(A, C^T)$  is observable, is SPR if and only if there exist  $P = P^T > 0$  and  $Q = Q^T > 0$  such that

$$P A + A^T P = -Q, \quad P B = C$$

## Strictly Positive Real functions

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Lemma: The rational transfer function  $G(s)$  with real coefficients is **SPR** if and only if

- (i) There are no poles in the closed right-half plane (stable).
- (ii)  $G(\infty) > 0$  OR  $G(\infty) = 0$  &  $\lim_{\omega \rightarrow \infty} \omega^2 \operatorname{Re}\{G(j\omega)\} > 0$ .
- (iii)  $\operatorname{Re}\{G(j\omega)\} \geq 0$  for  $\omega \in \mathbb{R}$ .



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- (iii)  $\operatorname{Re}\{G(j\omega)\} \geq 0$  for  $\omega \in \mathbb{R}$ .

Remark: If the (scalar) rational transfer function  $G(s)$  with real coefficients is **SPR**, then

1. The Nyquist plot lies in the closed right-half plane.
2. The relative degree is zero or one.

## Passivity

Definition: A SISO time-varying nonlinear system

$$\dot{x} = f(t, x, u), \quad y = h(t, x, u)$$

is called **passive** if there is a positive semidefinite **storage function**  $V(t, x) \geq 0$  such that

$$u h(t, x, u) = \boxed{u(t) y(t) \geq \dot{V}} = \frac{\partial}{\partial t} V(t, x) + \frac{\partial}{\partial x} V(t, x) \cdot f(t, x, u)$$

or

$$\int_0^\tau u(s) y(s) ds \geq V(\tau, x(\tau)) - V(0, x(0))$$

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Lemma: If the transfer function  $G(s) = C^T (sI - A)^{-1} B$  with real coefficients is **SPR**, then its minimal state-space realization defines a **passive** system.

Remark: A mechanical system with  $u$  as the generalized force and  $y$  as the generalized velocity defines a **passive** system with total energy as the **storage function**.

## Passivity interpretation

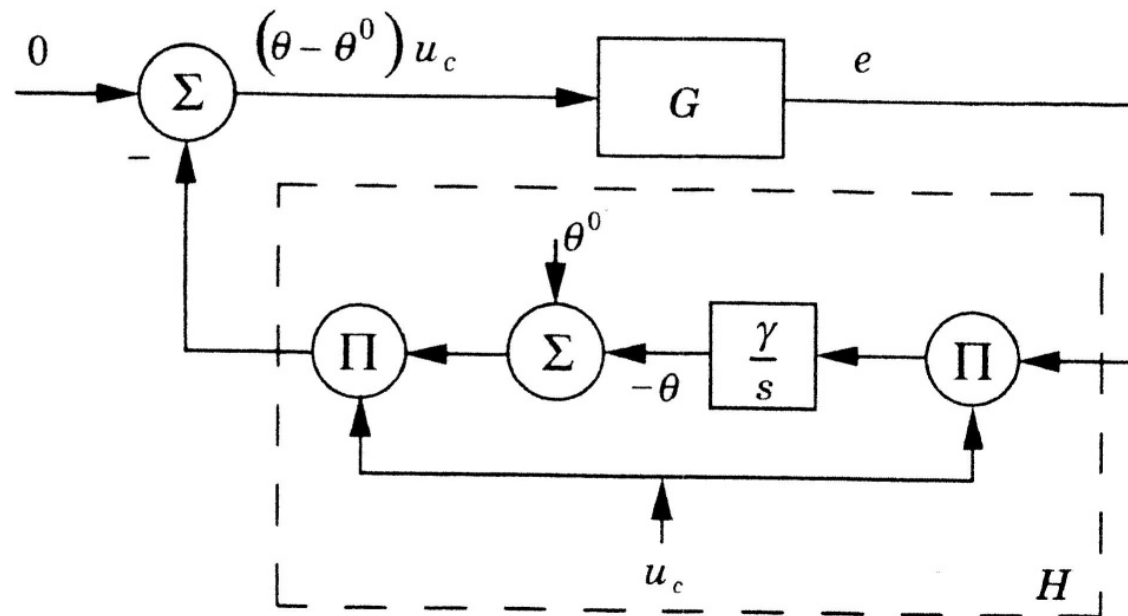
The closed-loop system with adaptation of the feedforward gain

$$e(t) = y - y_m = G(p) (\theta - \theta^0) u_c(t), \quad \dot{\theta} = -\frac{\gamma}{p} e(t) u_c(t)$$

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**Figure 5.18** Representation of the system with adjustable feedforward gain when using the control law of Eq. 5.40. Compare with Fig. 5.14(b).

## Passivity interpretation (cont'd)

Let us show that this is a feedback connection of two passive systems.

We just need to verify that  $H$ -subsystem

$$\dot{z} = \gamma u_c(t) e, \quad \xi = u_c(t) z$$

is **passive** with the output  $\xi$  and the input  $e$ :

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$$V_H(z) = \frac{1}{2\gamma} z^2$$



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The fact that  $e(t) \rightarrow 0$  can be shown using the Lyapunov function

$$V(x, z) = \frac{\gamma}{2} x^T P x + V_H(z) \quad \Rightarrow \quad \dot{V} \leq -x^T Q x$$

with any other **passive** system defining an adaptation law!

## Relaxing the SPR requirement

Suppose  $G(s) = \frac{B(s)}{A(s)}$  is

- stable, i.e.  $A(s)$  is Hurwitz,
- minimum phase, i.e.  $B(s)$  is Hurwitz as well,
- has relative degree one, i.e.  $\deg\{B\} = \deg\{A\} - 1$ ,

but it is NOT **SPR**.

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- has relative degree one, i.e.  $\deg\{B\} = \deg\{A\} - 1$ ,

but it is NOT **SPR**.

Let us use passivity arguments to design an adaptation law for this case.

First, we construct a polynomial  $C(s)$  such that

- $\deg\{C\} = \deg\{A\} - 1$ ,
- the transfer function  $\frac{C(s)}{A(s)}$  is **SPR**.

## Relaxing the SPR requirement (cont'd)

Introduce the canonical realization of

$$1/A(s) = 1/(s^n + a_1 s^{n-1} + \dots + a_n)$$

$$\dot{z} = A z + B v = \begin{bmatrix} -a_1 & -a_2 & \dots & -a_{n-1} & -a_n \\ 1 & 0 & & 0 & 0 \\ \vdots & & & & \\ 0 & 0 & & 1 & 0 \end{bmatrix} z + \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \end{bmatrix} v$$

## Relaxing the SPR requirement (cont'd)

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Choose any  $Q = Q^T > 0$  and solve

$$P A + A^T P = -Q, \quad P = P^T > 0$$

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Choose any  $Q = Q^T > 0$  and solve

$$P A + A^T P = -Q, \quad P = P^T > 0$$

Choose the matrix  $C = P B$ , i.e.

$$C^T = B^T P = [p_{11}, \dots, p_{1n}] = \text{the first row of } P$$

and take  $C(s) = p_{11} s^{n-1} + \dots + p_{1n}$ .

## Relaxing the SPR requirement (cont'd)

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With the constructed polynomial  $C(s)$  let us define the filter

$$G_c(s) = \frac{C(s)}{B(s)}$$



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which is

- proper:  $\deg\{B\} = \deg\{A\} - 1$  (relative degree one) and  $\deg\{C\} = \deg\{A\} - 1$ ,
- stable:  $B(s)$  is stable (minimum phase property).

## Relaxing the SPR requirement (cont'd)

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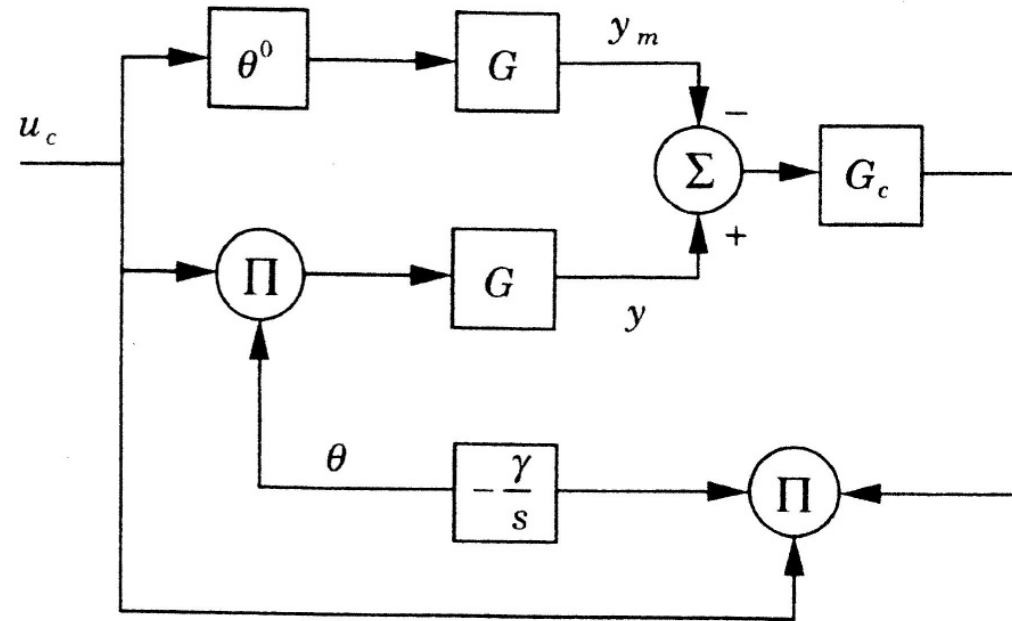
$$G_c(s) G(s) = \frac{C(s)}{B(s)} \frac{B(s)}{A(s)} = \frac{C(s)}{A(s)} \quad \text{is SPR}$$

let

$$e_c(t) = G_c(p) e(t) = G_c(p) G(p) (\theta - \theta^0) u_c(t)$$

and take

$$\dot{\theta} = -\gamma e_c(t) u_c(t).$$



**Figure 5.19** A stable parameter adjustment law is obtained if  $GG_c$  is SPR.

As before, all the signals will be bounded and  $\lim_{t \rightarrow \infty} e_c(t) = 0$ .

Since  $G_c(s)$  is minimum phase, we conclude that  $\lim_{t \rightarrow \infty} e(t) = 0$  as well.

## Relaxing the relative degree restriction

---

In the general case of stable  $G(s)$ , let

$$G(s) = G_1(s) G_2(s)$$

where  $G_1(s)$  is **SPR** and can be taken as  $G_1(s) = 1$ .

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$$e(t) = G(\theta - \theta^0) \mathbf{u}_c(t) = G_1 G_2(\theta - \theta^0) \mathbf{u}_c(t)$$

can be rewritten as

$$e(t) = G_1 [(\theta - \theta^0) G_2 \mathbf{u}_c] - G_1 [(\theta - \theta^0) G_2 \mathbf{u}_c - G_2(\theta - \theta^0) \mathbf{u}_c]$$

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Let us introduce **augmented error**

$$\varepsilon(t) = (\theta(t) - \theta^0) G_2(p) \mathbf{u}_c(t)$$

and **error augmentation**

$$\eta(t) = G_1(p) [(\theta(t) - \theta^0) G_2(p) \mathbf{u}_c(t)] - G(p)(\theta(t) - \theta^0) \mathbf{u}_c(t)$$

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## Adaptation with augmented error

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The adaptation law can be taken as

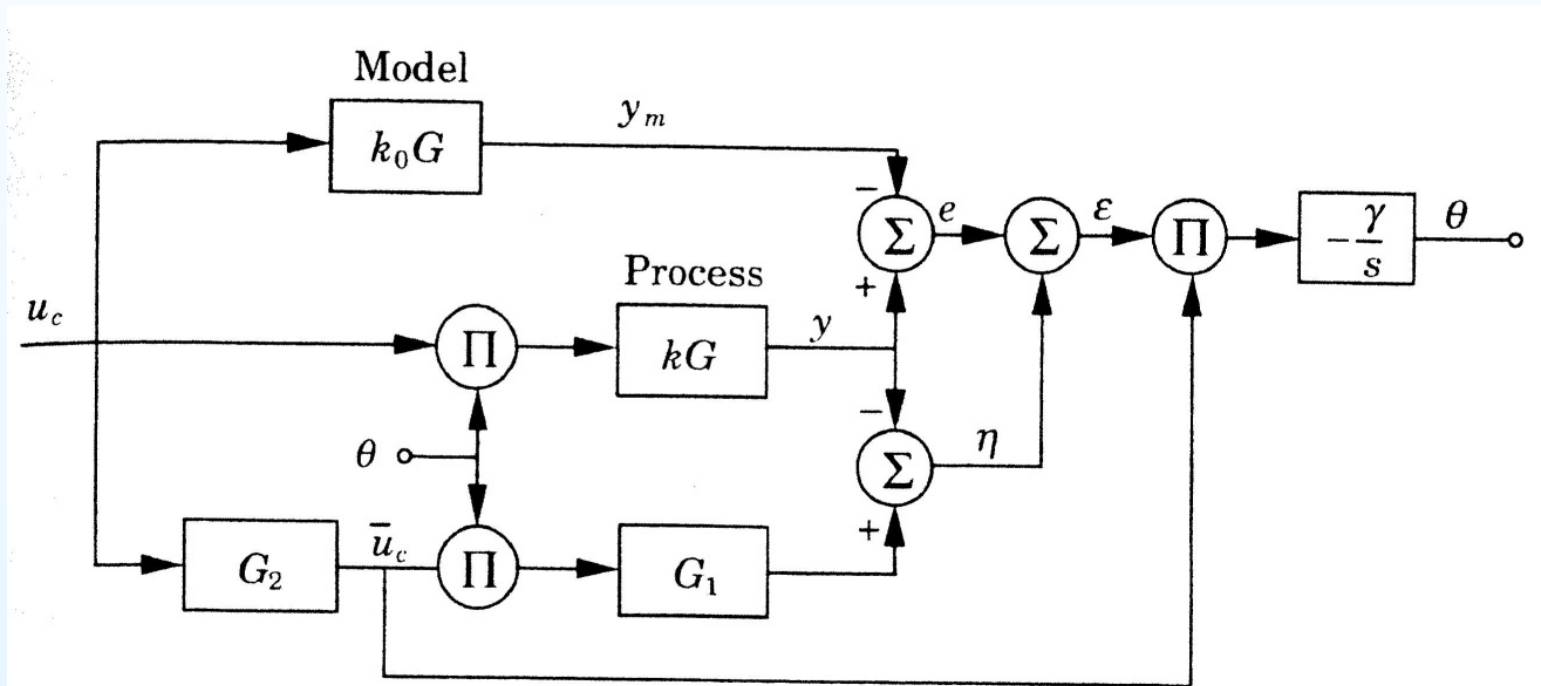
$$\dot{\theta} = -\gamma \varepsilon(t) (G_2(p) u_c(t))$$



## Adaptation with augmented error

The adaptation law can be taken as

$$\dot{\theta} = -\gamma \varepsilon(t) (G_2(p) u_c(t))$$



**Figure 5.20** Block diagram of a model-reference adaptive system based on the augmented error.

It can be shown that with “nice”  $u_c(t)$ :  $e(t) \rightarrow 0$ .

## Next Lecture / Assignments:

Next meeting (May 27, 15:00-17:00, in A206Tekn): Recitations.

### Homework problems:

- Which of the transfer functions:  $G_1(s) = 1/s$ ,  $G_2(s) = 1/(s + a)$  with  $a > 0$ ,  $G_2(s) = 1/(s^2 + s + 1)$  are **PR** / **SPR**?
- Proof that the adaptation law for the problem of adjusting a gain of an SPR transfer function can be taken as a PI law:

$$\theta(t) = -\gamma_1 \mathbf{u}_c(t) e(t) - \gamma_2 \int_0^t \mathbf{u}_c(s) e(s) ds$$

with  $\gamma_1, \gamma_2 > 0$ .