Lecture 9: Stochastic / Predictive Self-Tuning Regulators.

- Minimum Variance Control.
- Moving Average Controller.

The stochastic model

Assume the plant is represented by the special ARMAX model

$$A(q) y(t) = B(q) u(t) + C(q) e(t)$$

where $\{e(t)\}$ is white noise,

$$A(q) = q^n + a_1 q^{n-1} + \dots + a_n, \quad \deg\{A\} = n$$

$$B(q) = b_1 q^{n-d_0} + \dots + b_n,$$
 $\deg\{B\} = n - d_0$

$$C(q) = q^n + c_1 q^{n-1} + \dots + c_n, \qquad \deg\{C\} = n$$

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 $\deg\{A\} = n$ $B(q) = b_1 q^{n-d_0} + \dots + b_n,$ $\deg\{B\} = n - d_0$ $C(q) = q^n + c_1 q^{n-1} + \dots + c_n,$ $\deg\{C\} = n$

Equivalently,

$$y(t) = -a_1 y(t-1) - \cdots - a_n y(t-n)$$
 $+ b_{d_0} u(t-d_0) + \cdots + b_n u(t-n)$ $+ e(t) + c_1 e(t-1) + \cdots + c_n e(t-n)$

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Assuming that *C* is stable is not restrictive:

Typically, such a model is obtained experimentally from spectrum characteristics. It does not change if the unstable roots are substituted by stable, which are symmetrical to them with respect to the unite circle.

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$$(q + a) y(t) = b u(t) + c q e(t - 1)$$

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$$y(t+1) + a y(t) = b u(t) + c e(t)$$

It can be rewritten as

$$(q+a) y(t) = b u(t) + q e_{new}(t)$$

introducing $e_{new(t)} = c e(t-1)$.

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It can be rewritten as

$$y(t+1) + a y(t) = b u(t) + e_{new}(t+1)$$

with $e_{new(t)} = c \, e(t-1)$ being white noise with $var\{e_{new}(t)\} = c^2 \, var\{e(t)\}.$

Consider the model

$$y(t) = -a y(t-1) + b u(t-1) + c e(t-1) + e(t)$$

where |c|<1 and $\{e(t)\}$ is a sequence of random variables with $E\{e(t)\}=0$ and $\mathrm{var}\{e(t)\}=\sigma^2.$

Our goal is to regulate y(t) to 0 as close as possible.

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Our goal is to regulate y(t) to 0 with minimal variance.

In the absence of noise:

$$y(t) = -a y(t-1) + b u(t-1)$$

the best strategy (dead beat design) is

$$u(t) = \frac{a}{b} y(t) \implies y(t+1) = 0$$

What should we do when noise is present?

Consider the model

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In the absence of noise or when c=0:

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the best strategy (dead beat design) is

$$u(t) = \frac{a}{b}y(t) \implies y(t+1) = e(t+1)$$

What should we do when noise dynamics are present?

We have the process

$$y(t+1) + a y(t) = b u(t) + e(t+1) + c e(t)$$

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Let us try to compute e(t) symbolically

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$$e(t) = \frac{q+a}{q+c}y(t) - \frac{b}{q+c}u(t)$$

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Substituting back into the model

$$y(t+1) = -a y(t) + b u(t) + e(t+1) + c \left(\frac{q+a}{q+c} y(t) - \frac{b}{q+c} u(t) \right)$$

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Substituting back into the model

$$y(t+1) = \underbrace{\frac{(c-a) q}{q+c} y(t) + \frac{b q}{q+c} u(t) + e(t+1)}_{q+c}$$

known and independent from e(t+1)!

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Clearly,
$$\operatorname{var}\{y(t+1)\} \geq \operatorname{var}\{e(t+1)\} = \sigma^2$$
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The minimal variance is achieved with:

$$u(t) = -\frac{c-a}{b} y(t).$$

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Additional assumptions:

• The polynomials A and B are monic (right rescaling).

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- $deg\{A\} = deg\{C\} = n \ge 1$ (always can be done).

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- $deg\{A\} = deg\{C\} = n \ge 1$ (always can be done).
- $deg\{A\} deg\{B\} = d_0 \ge 1$ known relative degree.

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- $deg\{A\} = deg\{C\} = n \ge 1$ (always can be done).
- $deg\{A\} deg\{B\} = d_0 \ge 1$ known relative degree.
- The polynomial C is stable (always can be done).
- The polynomial B is stable a serious restriction.

We have the process

$$A(q) y(t) = B(q) u(t) + C(q) e(t)$$

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$$A(q) y(t) = B(q) u(t) + C(q) e(t)$$

Let us solve it for y(t)

$$y(t) = rac{B(q)}{A(q)} u(t) + rac{C(q)}{A(q)} e(t)$$

We have the process

$$A(q) y(t) = B(q) u(t) + C(q) e(t)$$

Shifting by the relative degree

$$y(t+d_0) = rac{B(q)}{A(q)} u(t+d_0) + rac{C(q)}{A(q)} e(t+d_0)$$

We have the process

$$A(q) y(t) = B(q) u(t) + C(q) e(t)$$

Shifting by the relative degree and rewriting

$$y(t+d_0) = rac{B(q)}{A(q)} u(t+d_0) + rac{C(q) q^{d_0-1}}{A(q)} e(t+1)$$

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Shifting by the relative degree and rewriting

$$y(t+d_0) = rac{B(q)}{A(q)} u(t+d_0) + rac{C(q) q^{d_0-1}}{A(q)} e(t+1)$$

Using polynomial long division

$$\left|rac{C(q)\,q^{d_0-1}}{A(q)}=F(q)+rac{G(q)}{A(q)}
ight|$$

where $rac{G(q)}{A(q)}$ is strictly proper and $\deg\{F\}=\deg\{C(q)\,q^{d_0-1}\}-\deg\{A\}=d_0-1.$

We have rewritten the process as

$$y(t+d_0) = rac{B(q)}{A(q)} \, u(t+d_0) + \left(F(q) + rac{G(q)}{A(q)}
ight) \, e(t+1)$$

Finally the process can be represented by

$$y(t+d_0) = \underbrace{\frac{q^{d_0}\,B(q)}{A(q)}}_{ ext{proper fraction}} u(t) + \underbrace{\frac{q\,G(q)}{A(q)}}_{ ext{proper fraction}} e(t) + F(q)\,e(t+1)$$

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 proper fraction

Solving the model for e(t), we have

$$e(t) = \frac{A(q)}{C(q)} y(t) - \frac{B(q)}{C(q)} u(t)$$

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Solving the model for e(t), we have

$$e(t) = \frac{A(q)}{C(q)} y(t) - \frac{B(q)}{C(q)} u(t)$$

After substituting it back

$$y(t+d_0) = rac{q^{d_0}\,B}{A}\,u(t) + rac{q\,G}{A}\,\left(rac{A}{C}\,y(t) - rac{B}{C}\,u(t)
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$$y(t+d_0) = \underbrace{\frac{q^{d_0}\,B(q)}{A(q)}}_{ ext{proper fraction}} u(t) + \underbrace{\frac{q\,G(q)}{A(q)}}_{ ext{proper fraction}} e(t) + F(q)\,e(t+1)$$

Solving the model for e(t), we have

$$e(t) = \frac{A(q)}{C(q)} y(t) - \frac{B(q)}{C(q)} u(t)$$

After substituting it back and collecting terms

$$y(t+d_0) = rac{q\,G}{C}\,y(t) + rac{q\,B}{A\,C}\left(q^{d_0}\,C - G
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$$e(t) = \frac{A(q)}{C(q)} y(t) - \frac{B(q)}{C(q)} u(t)$$

After substituting it back and collecting terms

$$y(t+d_0) = rac{q\,G}{C}\,y(t) + rac{q\,B}{C}\,rac{(q^{d_0}\,C-G)}{A}\,u(t) + F\,e(t+1)$$

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 proper fraction

Solving the model for e(t), we have

$$e(t) = \frac{A(q)}{C(q)} y(t) - \frac{B(q)}{C(q)} u(t)$$

Substituting F(q) from $\frac{C(q)\,q^{d_0-1}}{A(q)}=F(q)+rac{G(q)}{A(q)}$

$$y(t+d_0) = rac{q \, G(q)}{C(q)} \, y(t) + rac{q \, B(q) \, F(q)}{C(q)} \, u(t) + F(q) \, e(t+1)$$

We have rewritten the process as

$$y(t+d_0) = \underbrace{\frac{q\,G(q)}{C(q)}\,y(t) + \frac{q\,B(q)\,F(q)}{C(q)}\,u(t)}_{ ext{known and independent from }\,F(q)\,e(t+1)$$

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$$ext{var}\{y(t+d_0)\} \ge ext{var}\{F(q)\,e(t+1)\} = \left(1+f_1^2+\cdots+f_{d_0-1}^2\right)\,\sigma^2$$
 where $F(q)=q^{d_0-1}+f_1\,q^{d_0-2}+\cdots+f_{d_0-1}.$

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The minimal variance is achieved with:

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The closed-loop system is

$$B(q) (A(q) F(q) + G(q)) y(t) = B(q) F(q) C(q) e(t)$$

We have rewritten the process as

$$y(t+d_0) = \underbrace{\frac{q\,G(q)}{C(q)}\,y(t) + \frac{q\,B(q)\,F(q)}{C(q)}\,u(t)}_{ ext{known and independent from }\,F(q)\,e(t+1)$$

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The minimal variance is achieved with:

$$oxed{u(t) = -rac{G(q)}{B(q)\,F(q)}\,y(t)}.$$

The closed-loop system (using: $q^{d_0-1} C = A F + G$) is

$$q^{d_0-1} B(q) C(q) y(t) = B(q) F(q) C(q) e(t)$$

(1) The closed-loop system

$$q^{d_0-1} B(q) C(q) y(t) = B(q) F(q) C(q) e(t)$$

defines the noise-to-output relation

$$y(t) = rac{B(q) \, F(q) \, C(q)}{q^{d_0 - 1} \, B(q) \, C(q)} \, e(t)$$

(1) The closed-loop system

$$q^{d_0-1} B(q) C(q) y(t) = B(q) F(q) C(q) e(t)$$

defines the noise-to-output relation

$$y(t) = \frac{F(q)}{q^{d_0 - 1}} e(t) = \frac{q^{d_0 - 1} + f_1 q^{d_0 - 2} + \dots + f_{d_0 - 1}}{q^{d_0 - 1}} e(t)$$

(1) The closed-loop system

$$q^{d_0-1} B(q) C(q) y(t) = B(q) F(q) C(q) e(t)$$

defines the noise-to-output relation

$$y(t) = (1 + f_1 q^{-1} + \dots + f_{-d_0+1}) e(t)$$

(1) The closed-loop system

$$q^{d_0-1} B(q) C(q) y(t) = B(q) F(q) C(q) e(t)$$

defines the noise-to-output relation

$$y(t) = e(t) + f_1 e(t-1) + \cdots + f_{-d_0+1} e(t-d_0+1)$$

which is a moving average of order $d_0 - 1$.

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which is a moving average of order $d_0 - 1$.

(2) The closed-loop poles consist of (a) zeros of B(q), (b) zeros of C(q), and (c) $d_0 - 1$ zeros.

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- (2) The closed-loop poles consist of (a) zeros of B(q), (b) zeros of C(q), and (c) $d_0 1$ zeros.
- (3) The minimum variance controller

$$u(t) = -rac{G(q)}{B(q) F(q)} y(t)$$

can be interpreted as a pole-placement controller.

Minimum-variance control as pole-placement

Since

$$q^{d_0-1} C(q) = A(q) F(q) + G(q)$$

the closed-loop characteristic polynomial

$$A_c(q) = q^{d_0-1} B(q) C(q) = A(q) \underbrace{B(q) F(q)}_{R(q)} + B(q) \underbrace{G(q)}_{S(q)}$$

Minimum-variance control as pole-placement

Since

$$q^{d_0-1} C(q) = A(q) F(q) + G(q)$$

the closed-loop characteristic polynomial

$$A_c(q) = q^{d_0 - 1} B(q) C(q) = A(q) \underbrace{B(q) F(q)}_{R(q)} + B(q) \underbrace{G(q)}_{S(q)}$$

Note that for the Diophantine equation $A_c = AR + BS$

$$\deg\{S(q)\} = \deg\{G(q)\} = n-1$$

and S/R is proper since

$$\deg\{R\} = \deg\{B\} + \deg\{F\} = (n - d_0) + (d_0 - 1) = n - 1.$$

Minimum-variance control as pole-placement

Since

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Can we use an analogous pole-placement technique for the case when B(q) has unstable zeros?

Design the controller

$$u(t) = -rac{S(q)}{R(q)}y(t) \qquad ext{for} \quad A(q)\,y(t) = B(q)\,u(t) + C(q)\,e(t)$$

as follows.

Design the controller

$$u(t) = -rac{S(q)}{R(q)}y(t) \qquad ext{for} \quad A(q)\,y(t) = B(q)\,u(t) + C(q)\,e(t)$$

as follows.

• Factor B as $B = B^+ B^-$, where B^+ is monic and stable.

Design the controller

$$u(t) = -rac{S(q)}{R(q)}y(t) \qquad ext{for} \quad A(q)\,y(t) = B(q)\,u(t) + C(q)\,e(t)$$

as follows.

- Factor B as $B = B^+ B^-$, where B^+ is monic and stable.
- Set $d = \deg\{A\} \deg\{B^+\}$.

Design the controller

$$u(t) = -rac{S(q)}{R(q)}y(t) \qquad ext{for} \quad A(q)\,y(t) = B(q)\,u(t) + C(q)\,e(t)$$

as follows.

- Factor B as $B = B^+ B^-$, where B^+ is monic and stable.
- Set $d = \deg\{A\} \deg\{B^+\}$.
- Find R_p and S solving the Diophantine equation

$$q^{d-1} C = A R_p + B^- S, \qquad d \ge d_0$$

Design the controller

$$u(t) = -rac{S(q)}{R(q)}y(t) \qquad ext{for} \quad A(q)\,y(t) = B(q)\,u(t) + C(q)\,e(t)$$

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- Find R_p and S solving the Diophantine equation

$$q^{d-1} C = A R_p + B^- S, \qquad d \ge d_0$$

• Let $R=R_p\,B^+$, where $\deg\{R_p\}=d-1$.

The obtained transfer function for the closed-loop system is

$$y(t) = q^{1-d} R_p(q) e(t) = e(t) + r_1 e(t-1) + \dots + r_{d-1} e(t-d+1)$$

Example 4.3

Consider the plant described by

$$(q^2 + a_1 q + a_2) y(t) = (b_0 q + b_1) u(t) + (q^2 + c_1 q + c_2) e(t)$$

Example 4.3

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$$(q^2 + a_1 q + a_2) y(t) = (b_0 q + b_1) u(t) + (q^2 + c_1 q + c_2) e(t)$$

1. If $|b_1/b_0| < 1$ minimum variance (MV) controller can be designed noticing that

$$\frac{q^{1-1} (q^2 + c_1 q + c_2)}{q^2 + a_1 q + a_2} = 1 + \frac{(c_1 - a_1) q + (c_2 - a_2)}{q^2 + a_1 q + a_2}$$

as follows

$$u(t) = -\frac{G(q)}{B(q) F(q)} y(t) = -\frac{(c_1 - a_1) q + (c_2 - a_2)}{(b_0 q + b_1) \cdot 1} y(t)$$

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2. If $|b_1/b_0| > 1$ minimum variance (MV) controller cannot be designed but we can apply moving average controller (MA) with d = 2 using the solution of the Diophantine equation

$$q(q^2+c_1 q+c_2) = (q^2+a_1 q+a_2)(q+r_1)+(b_0 q+b_1)(s_0 q+s_1)$$

Next Lecture / Assignments:

Next meeting (May 10, 13:00-15:00, in A208Tekn): Recitations.

Homework problems: Consider the process in Example 4.3 with $a_1 = -1.5$, $a_2 = 0.7$, $b_0 = 1$, $c_1 = -1$, and $c_2 = 0.2$.

Determine the variance of the output in the closed-loop system as a function of b_1 when the Moving Average controller is used. Compare with the lowest achievable variance.