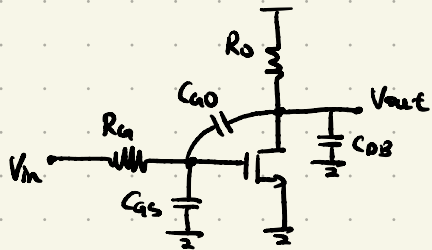


Lec 22

- Dominant - Pole Approximation
- Insights into CE/CS Freq. Response
- Examples
- Freq. Response of Common-Base / Common-gate Stage

Review of Lec 21



From Miller's Theorem:

$$\omega_{p1} = \frac{1}{[(1+g_m R_O) C_{C0} + C_{B0}] R_B}$$

$$\frac{V_{out}}{V_{in}} = \frac{(C_{C0} s - g_m) R_O}{(C_{B0} C_{C0} + C_{C0} C_{E0} + C_{B0} C_{E0}) R_B R_O s^2 + [(1+g_m R_O) C_{C0} + C_{B0}] R_B + (C_{C0} + C_{E0}) R_O] s + 1}$$

- Dominant - Pole Approximation

$$D(s) = \left(\frac{s}{\omega_{p1}} + 1\right) \left(\frac{s}{\omega_{p2}} + 1\right) = \frac{s^2}{\omega_{p1} \omega_{p2}} + \left(\frac{1}{\omega_{p1}} + \frac{1}{\omega_{p2}}\right) s + 1$$

$$\omega_{p1} \ll \omega_{p2} \Rightarrow \frac{1}{\omega_{p1}} \gg \frac{1}{\omega_{p2}}$$

$$\text{coefficient of } s \approx \frac{1}{\omega_{p1}}$$

$$\Rightarrow \omega_{p1} = \frac{1}{[(1+g_m R_O) C_{C0} + C_{B0}] R_B + (C_{C0} + C_{E0}) R_O}$$

$$\omega_{p1} \omega_{p2} = \frac{1}{(C_{C0} C_{E0} + C_{C0} C_{B0} + C_{B0} C_{E0}) R_B R_O}$$

then we can find ω_{p2}

A few points:

$$\textcircled{1} \text{ If } s=0 \Rightarrow \frac{V_{out}}{V_{in}} = -g_m R_O$$

$$\textcircled{2} \text{ If } s \rightarrow \infty \Rightarrow \frac{V_{out}}{V_{in}} \rightarrow 0$$

$\textcircled{3}$ We have 3 caps, but a second-order response. Why?

If we have N caps (no series or parallel combinations), $\Rightarrow N$ -th order diff. equation

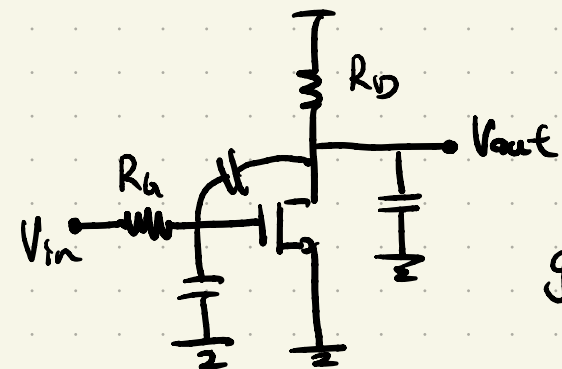
$$d_1 \frac{d^N V_{out}}{dt^N} + d_2 \frac{d^{N-1} V_{out}}{dt^{N-1}} + \dots + d_{N+1} V_{out} = 0$$

Needs N indep. initial conditions.

(Because we can only find 2 initial conditions)

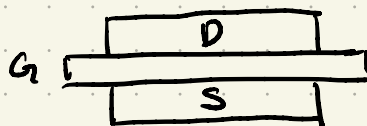
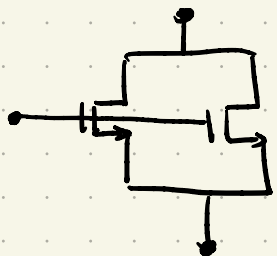
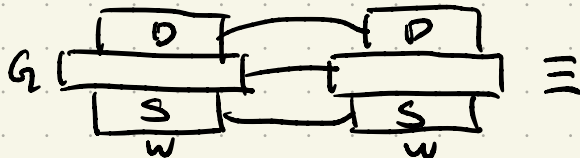
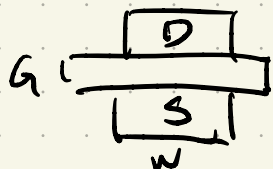
Example

What happens if $W \rightarrow 2W$?
(but I_D is constant)



\Rightarrow All transistor caps are doubled

$$g_m = \sqrt{2\mu_n C_{ox} \frac{W}{L} I_D} \rightarrow \sqrt{2} g_m$$

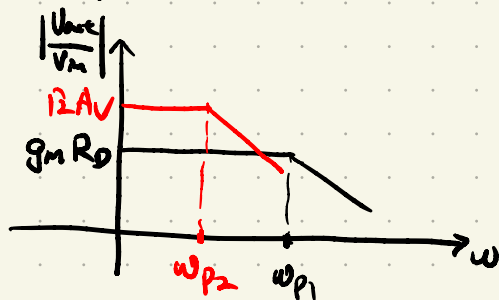


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$$\omega_{p1} = \frac{1}{[C_{gs} + (g_m R_D) C_{gs} + C_{gs}] R_D + (C_{gs} + C_{gs}) R_D}$$

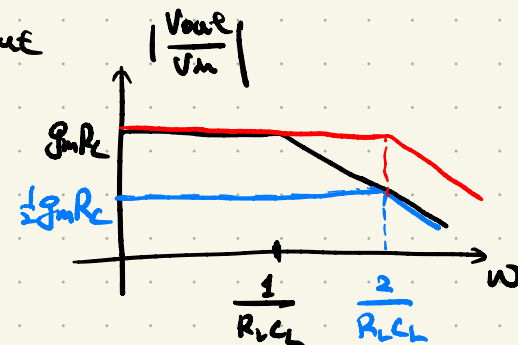
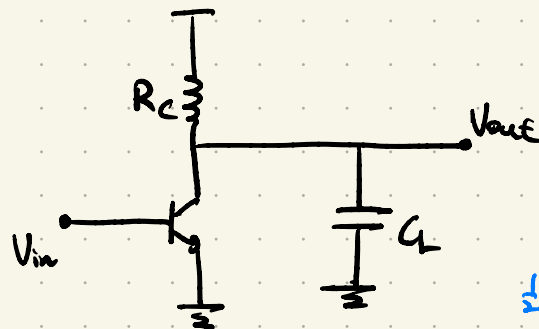
ω_{p1} ↓ by at least a factor of 2

$$A_v \rightarrow \sqrt{2} A_v$$



Example

① If $R_C \rightarrow \frac{R_C}{2}$, what happens?
(I_C const)



Gain \times BW

$$= g_m R_C \frac{1}{R_C C_L} = \frac{g_m}{C_L}$$

② $R_C \rightarrow \frac{R_C}{2}$, $I_C \rightarrow 2I_C$

$$\frac{I_C}{V_T} = g_m \rightarrow 2g_m$$

$$\text{Gain} \times \text{BW} = g_m R_C \cdot \frac{2}{R_C C_L} = 2 \frac{g_m}{C_L}$$

Power cons. = $V_{CC} \cdot 2I_C$ doubled