


Flat Fading $y = hx + n$

$x \rightarrow \boxed{} \rightarrow y = hx + n$

$$E|x_i|^2 = E x_i x_i^*$$

$$P_{\Sigma} = \sum_{i=1}^{M+1} E|x_i|^2 = t \cdot E[\underbrace{\bar{x} \bar{x}^H}_{R_x}]$$

信号的功率为常数


 Δ Narrowband MIMO
 $y_1 = h_{11}x_1 + h_{21}x_2 + n_1$

$$\begin{aligned} y_1 &= h_{11}x_1 + h_{12}x_2 + n_1 \\ &= (h_{11}, h_{12}) \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + n_1 \Rightarrow \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} n_1 \\ n_2 \end{pmatrix} \\ y_2 &= (h_{21}, h_{22}) \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + n_2 \end{aligned}$$

$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_{m_r} \end{bmatrix} = H_{m_r \times m_b} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{m_b} \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_{m_r} \end{bmatrix}$$

$$R_n = E[\vec{n} \vec{n}^H] = E \begin{bmatrix} n_1 \\ \vdots \\ n_m \end{bmatrix} (n_1^* \dots n_m^*) = \begin{bmatrix} n_1 n_1^* & n_1 n_2^* & \dots & n_1 n_m^* \\ \vdots & n_2 n_2^* & \dots & \vdots \\ \vdots & \vdots & \ddots & \vdots \\ n_m n_1^* & \dots & \dots & n_m n_m^* \end{bmatrix} = \sigma^2 \mathbf{I}$$

① 非对角线上的值都为0

② 对角线上为 $n_i \cdot n_i^* = 6^2 = E |n_i|^2$

$$\sum_{i=1}^M E |x_i|^2 = P \quad R_X = E[\vec{x} \vec{x}^H]$$

$M_r \geq M_t$
 ▲ Parallel Decomposition
 $H \cong M_r \oplus M_t$ Rank(H) = R_H

$$H = U \Lambda V^H$$

$$\begin{aligned} U^H M U &= M & U \cdot U^H &= U^H U = I \\ V^H M V &= M & V \cdot V^H &= V^H V = I \\ \Lambda^H M \Lambda &= M \end{aligned}$$

let $[\wedge]_{i, i} = \delta_i$ $\delta_1 \geq \delta_2 \geq \delta_3 \geq \dots \geq \delta_{2M} > 0$

$$\begin{aligned} \vec{y} &= U \vec{x} + \vec{n} \\ &= U \Lambda V^H \vec{x} + \vec{n} \Rightarrow U^H \vec{y} = \Lambda V^H \vec{x} + U^H \vec{n} \\ \text{let } \begin{cases} \tilde{y} = U^H \vec{y} \\ \tilde{x} = V^H \vec{x} \\ \tilde{n} = U^H \vec{n} \end{cases} &\Rightarrow \tilde{y} = \Lambda \tilde{x} + \tilde{n} \end{aligned}$$

$$\begin{bmatrix} y_1 \\ \vdots \\ y_m \end{bmatrix} = \begin{bmatrix} \sigma_1 & & 0 \\ & \ddots & \\ 0 & & \sigma_m \end{bmatrix} \begin{bmatrix} x_1 \\ \vdots \\ x_m \end{bmatrix} + \begin{bmatrix} \eta_1 \\ \vdots \\ \eta_m \end{bmatrix}$$

$$\hat{y}_i = \sigma_i \hat{x}_i + \hat{n}_i \quad i=1, 2, \dots, R_H \quad (\text{同时可传信号数 } \max = R_H)$$

$\vec{x} \Rightarrow \boxed{H} \Rightarrow \vec{y} = H\vec{x} + \vec{n}$
precoding

把互相干扰的M信道
拆为若干相互不干扰的！

$x \Rightarrow \underline{u} \Rightarrow \underline{\tilde{x}} \Rightarrow \underline{H} \Rightarrow \underline{\tilde{y}} = H\underline{\tilde{x}} + \underline{\tilde{v}} \Rightarrow \underline{U^H} \Rightarrow \underline{\tilde{y}} = U^H \underline{\tilde{y}}$

↓ ↓ ↓ ↓
 想发的信号 真实发的 真接收的 比从 $\underline{y} \rightarrow \underline{\tilde{x}}$ 更易解调

$\Delta MZMO$ 带 CSR/CSIT 的估值容量

$$\begin{aligned} y_i &= \sigma_i x_i + \eta_i \quad i=1, \dots, R_M \\ \eta &= U^H \tilde{\eta} \quad E[\tilde{\eta}] = E[U^H \tilde{\eta}] = U^H E[\tilde{\eta}] = 0 \\ R_{\eta} &= E[\tilde{\eta} \tilde{\eta}^H] = [U^H \tilde{\eta} \tilde{\eta}^H U] = U^H R_{\tilde{\eta}} U \\ &= \sigma^2 U^H U = \sigma^2 I \end{aligned}$$

一切事務矢野

$$\begin{aligned} \eta &= U^H \tilde{\eta} \quad E \tilde{\eta} = E(U^H \tilde{\eta}) = 0 \\ E[\tilde{\eta} \tilde{\eta}^H] &= E[U^H \tilde{\eta} \tilde{\eta}^H U] = U^H (\sigma^2 I) U = \sigma^2 I \\ \Rightarrow E[\eta_i \eta_j] &= \begin{cases} 0, & i \neq j \\ \sigma^2, & i = j \end{cases} \Rightarrow \tilde{\eta} \sim CN(0, \sigma^2) \end{aligned}$$

$$\begin{aligned} &= U^H U \tilde{x} + U^H \tilde{n} \\ &= U^H H V \tilde{x} + U^H \tilde{n} \\ &= \Lambda \tilde{x} + U^H \tilde{n} \\ \Rightarrow y_i &= \sigma_i \tilde{x}_i + \tilde{n}_i \quad (\text{噪声信号与干扰项}) \\ &\quad \rightarrow \text{上限看 RN} \end{aligned}$$

$$\begin{aligned}
 P &= \text{tr} E[\tilde{x} \tilde{x}^H] \\
 &= \text{tr} E[V \tilde{x} \tilde{x}^H V^H] \\
 &= \text{tr} [V E(\tilde{x} \tilde{x}^H) V^H] \quad (\text{tr}(ABC) = \text{tr}(CAB) = \text{tr}(BCA)) \\
 &= \text{tr} [V^H V E(\tilde{x} \tilde{x}^H)] \\
 &= \text{tr} E(\tilde{x} \tilde{x}^H) \\
 &= \sum_{i=1}^M E|\tilde{x}_i|^2 \Rightarrow M \text{ 个信号, 只有 } R_H \text{ 个能被接收} \\
 &= \sum_{i=1}^{R_H} E|\tilde{x}_i|^2 \Rightarrow \text{把不能接收的信号置0}
 \end{aligned}$$

$$C = \sum_{i=1}^{R_H} B \cdot \log_2 \left(1 + \frac{\sigma_i^2 E |\hat{h}_i|^2}{E |\hat{h}_i|^2} \right) \quad (P_i = E |\hat{h}_i|^2, \sum_{i=1}^{R_H} P_i = P)$$

$$= \sum_{i=1}^{R_H} B \cdot \log_2 \left(1 + \frac{\sigma_i^2 P_i}{\sigma^2} \right) \quad P_i \geq \frac{P_i}{\sigma^2}$$

Lagrangian:

$$J(\lambda) = \sum_{i=1}^{R_H} B \log_2 \left(1 + \sigma_i^2 \frac{P_i}{\sigma^2} \right) - \lambda \left(\sum_{i=1}^{R_H} P_i - P \right)$$

$$C = \max_{\lambda} \max_{\{P_i\}} J(\lambda) \quad (R^+) = \max\{0, A\}$$

fix λ

$$\frac{\partial J(\lambda)}{\partial P_i} = 0 \Rightarrow P_i = \left(\frac{B}{\lambda \ln 2} - \frac{\sigma^2}{\sigma_i^2} \right)^+$$

$$\frac{P_i}{P} = \left(\frac{B}{\lambda P \ln 2} - \frac{\sigma^2}{P \sigma_i^2} \right)^+ = \left(\frac{1}{\gamma_0} - \frac{1}{\gamma_i} \right)^+$$

$$C = \sum_{i=1}^{R_H} B \log_2 \left(1 + \frac{\sigma_i^2 P}{\sigma^2} \cdot \frac{P_i}{P} \right)$$

$$\Rightarrow = \sum_{i=1}^{R_H} B \log_2 \left[1 + \gamma_i \left(\frac{1}{\gamma_0} - \frac{1}{\gamma_i} \right)^+ \right]$$

$$= \sum_{i: \gamma_i \geq \gamma_0} B \log_2 \frac{\gamma_i}{\gamma_0}$$

$$\sum_{i=1}^{R_H} \frac{P_i}{P} = 1 \quad \sum_{i=1}^{R_H} \left(\frac{1}{\gamma_0} - \frac{1}{\gamma_i} \right)^+ = 1 \quad \gamma_i = \frac{P \cdot \sigma_i^2}{\sigma^2} \quad \begin{cases} \sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_{R_H} \\ \gamma_1 \geq \gamma_2 \geq \dots \geq \gamma_{R_H} \\ \frac{1}{\gamma_1} \leq \frac{1}{\gamma_2} \leq \dots \leq \frac{1}{\gamma_{R_H}} \end{cases}$$

① Suppose $\gamma_{R_H} \geq \gamma_0$

$$\sum_{i=1}^{R_H} \left(\frac{1}{\gamma_0} - \frac{1}{\gamma_i} \right) = 1 \Rightarrow \gamma_0 = \frac{R_H}{1 + \sum_{i=1}^{R_H} \frac{1}{\gamma_i}}$$

if $\gamma_0 \leq \gamma_{R_H}$

↓ No

② Suppose $\gamma_{R_H-1} \geq \gamma_0 \geq \gamma_{R_H}$

$$\sum_{i=1}^{R_H-1} \left(\frac{1}{\gamma_0} - \frac{1}{\gamma_i} \right) = 1 \Rightarrow \gamma_0 = \frac{R_H-1}{1 + \sum_{i=1}^{R_H-1} \frac{1}{\gamma_i}}$$

↓

...