Implicit handling of multilayered material substrates in full-wave SCUFF-EM calculations

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$$\epsilon_1, \mu_1$$
 $z=z_1$
 ϵ_2, μ_2
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 $z=z_{N-1}$
 ϵ_N, μ_N
 ϵ_N, μ_N
 ϵ_N, μ_N
 ϵ_N, μ_N

Figure 1: Geometry of the layered substrate. The *n*th layer has relative permittivity and permeability ϵ_n, μ_n , and its lower surface lies at $z = z_n$. The ground plane, if present, lies at $z = z_{\text{GP}}$.

1 Overview

In a previous memo¹ I considered SCUFF-STATIC electrostatics calculations in the presence of a multilayered dielectric substrate. In this memo I extend that discussion to the case of *full-wave* (i.e. nonzero frequencies beyond the quasistatic regime) scattering calculations in the SCUFF-EM core library.

Substrate geometry

As shown in Figure 1, I consider a multilayered substrate consisting of N material layers possibly terminated by a perfectly-conducting ground plane. The uppermost layer (layer 1) is the infinite half-space above the substrate. The nth layer has relative permittivity and permeability ϵ_n, μ_n , and its lower surface lies at $z=z_n$. The ground plane, if present, lies at $z\equiv z_{\rm N}\equiv z_{\rm GP}$. If the ground plane is absent, layer N is an infinite half-space.²

Definition of the substrate DGF

I will use the symbol $\Gamma(\omega; \mathbf{x}_{\text{D}}, \mathbf{x}_{\text{S}})$ for the *total* 6×6 dyadic Green's function relating time-harmonic fields at \mathbf{x}_{D} to sources at \mathbf{x}_{S} : thus, if $\mathcal{S} \equiv \begin{pmatrix} \mathbf{J} \\ \mathbf{M} \end{pmatrix}$ is the 6-vector distribution of free electric and magnetic currents in the presence of the substrate, then the 6-vector of electric and magnetic fields $\mathcal{F} \equiv \begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix}$ is given by

$$m{\mathcal{F}}(\mathbf{x}_{ exttt{ iny D}}) = \int m{\Gamma}(\mathbf{x}_{ exttt{ iny D}}, \mathbf{x}_{ exttt{ iny S}}) \cdot m{\mathcal{S}}(\mathbf{x}_{ exttt{ iny S}}) d\mathbf{x}_{ exttt{ iny S}}.$$

 $^{^1\,\}mathrm{``Implicit}$ handling of multilayered dielectric substrates in SCUFF-STATIC''

²As in the electrostatic case, this means that a finite-thickness substrate consisting of N material layers is described as a stack of N+1 layers in which the bottommost layer is an infinite half-space $(z_{N+1} = -\infty)$ with the material properties of vacuum $(\epsilon_{N+1} = \mu_{N+1} = 1)$.

The 6×6 tensor Γ has a 2×2 block structure:

$$\Gamma = \begin{pmatrix} \Gamma^{\text{EE}} & \Gamma^{\text{EM}} \\ \Gamma^{\text{ME}} & \Gamma^{\text{MM}} \end{pmatrix}$$
 (1a)

with the 3×3 subblocks defined by

$$\Gamma_{ij}^{PQ}(\omega, \mathbf{x}_{D}, \mathbf{x}_{S}) = \begin{pmatrix} i\text{-component of P-type field at } \mathbf{x}_{D} \text{ due to } j\text{-directed} \\ Q\text{-type point current source at } \mathbf{x}_{S}, \text{ all fields and} \\ \text{sources having time dependence } \sim e^{-i\omega t} \end{pmatrix} (1b)$$

Homogeneous DGF In an infinite *homogeneous* medium with relative permittivity and permeability $\{\epsilon^r, \mu^r\}$, Γ reduces to its homogeneous form, for which I will use the symbol Γ^{0r} (where the r index labels the medium, which in this case will be one of the layers in Figure 1, i.e. $r \in \{1, 2, \dots, N\}$):

 $\mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}} \in \text{infinite homogeneous medium } r \implies \Gamma(\omega; \mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}}) = \Gamma^{0r}(\omega; \mathbf{x}_{\mathrm{D}} - \mathbf{x}_{\mathrm{S}})$

where 3

$$\mathbf{\Gamma}^{0r}(\omega, \mathbf{r}) \equiv \begin{pmatrix} ik_r Z_0 Z^r \mathbf{G}(k_r, \mathbf{r}) & ik_r \mathbf{C}(k_r, \mathbf{r}) \\ -ik_r \mathbf{C}(k_r, \mathbf{r}) & \frac{ik_r}{Z_0 Z^r} \mathbf{G}(k_r, \mathbf{r}) \end{pmatrix}$$

$$k_r \equiv \sqrt{\epsilon_0 \epsilon^r \mu_0 \mu^r} \cdot \omega, \quad Z_0 Z^r \equiv \sqrt{\frac{\mu_0 \mu^r}{\epsilon_0 \epsilon^r}},$$

$$G_{ij} = \left(\delta_{ij} - \frac{1}{k^2} \partial_i \partial_j\right) \frac{e^{ik|\mathbf{r}|}}{4\pi |\mathbf{r}|}, \quad C_{ij} = \frac{\varepsilon_{i\ell m}}{ik} \partial_\ell G_{mj}$$
(2)

Inhomogeneous DGF On the other hand, in the presence of the multilayered substrate the full DGF Γ receives corrections, which may be thought of as the fields radiated by surface currents induced on the interfacial surfaces of the substrate, and which I will denote by the symbol \mathcal{G} :

$$\Gamma(\mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}}) = \mathcal{G}(\mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}}) + \begin{cases} \Gamma^{0r}(\mathbf{x}_{\mathrm{D}} - \mathbf{x}_{\mathrm{S}}), & \mathbf{x}_{\mathrm{S}} \in \text{layer r} \\ 0, & \text{otherwise} \end{cases}$$
(3)

Like Γ , \mathcal{G} is a 6×6 matrix with a 2×2 block structure:

$$\mathcal{G}(\omega; \mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}}) = \begin{pmatrix} \mathcal{G}^{\mathrm{EE}} & \mathcal{G}^{\mathrm{EM}} \\ \mathcal{G}^{\mathrm{ME}} & \mathcal{G}^{\mathrm{MM}} \end{pmatrix}$$
(4)

with the 3×3 subblocks defined by

$$\mathcal{G}_{ij}^{\text{\tiny PQ}} = \begin{pmatrix} i\text{-component of P-type field at } \mathbf{x}_{\text{\tiny D}} \text{ due to surface currents on substrate interface layers induced by } j\text{-directed Q-type source at } \mathbf{x}_{\text{\tiny S}}. \end{pmatrix}$$

LIBSUBSTRATE is a code for numerical computation of \mathcal{G} .

 $^{^3}$ Cf. Section 3 of the companion memo "LIBSCUFF implementation and Technical Details," http://homerreid.github.io/scuff-em-documentation/tex/lsInnards.pdf

Organization of SCUFF-EM implementation and this memo

The full-wave substrate implementation in SCUFF-EM consists of multiple working parts that fit together in a somewhat modular fashion.

Roughly speaking, the computational problem may be divided into two parts:

- (a) For given source and evaluation (or "destination") points $\{\mathbf{x}_{\mathrm{S}}, \mathbf{x}_{\mathrm{D}}\}$ at a given angular frequency ω in the presence of a multilayer substrate, numerically compute the substrate DGF correction $\mathcal{G}(\omega, \mathbf{x}_{\mathrm{D}}, \mathbf{x}_{\mathrm{S}})$. This task is independent of SCUFF-EM and is implemented by a standalone library called LIBSUBSTRATE, described in Section 2 of this memo.
- (b) For a SCUFF-EM geometry in the presence of a substrate, compute the substrate corrections to the BEM system matrix M and RHS vector v, as well as the substrate corrections to post-processing quantities such as scattered fields. This is done by the file Substrate.cc in LIBSCUFF and is described in Section 3 of this memo.

2 LIBSUBSTRATE: Numerical computation of substrate Green's functions

Numerical evaluation of substrate contributions to dyadic Green's functions is handled by a C++ library called LIBSUBSTRATE. Although this library is packaged and distributed with SCUFF-EM and depends on other support libraries in the SCUFF-EM distribution, it is independent of the particular integral-equation formulation implemented by LIBSCUFF, and thus should be of general utility beyond SCUFF-EM.

2.1 Overview of computational strategy

LIBSUBSTRATE decomposes the problem of computing ${\cal G}$ into several logical steps, as follows:

- 1. Solve a linear system to obtain the Fourier-space representation $\widetilde{\mathcal{G}}(\mathbf{q})$. Here $\mathbf{q} = (q_x, q_y)$ is a 2D Fourier variable. (Section 2.2.)
- **2.** Reduce the two-dimensional integral over \mathbf{q} to a one-dimensional integral over $|\mathbf{q}| \equiv q$. (Section 2.3.)
- **3.** Evaluate the q integral using established methods for evaluating Sommerfeld integrals. (Section $\ref{eq:section}$.)

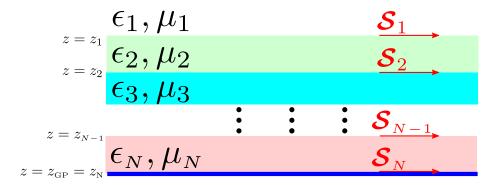


Figure 2: Effective surface-current approach to treatment of multilayer substrate. External field sources induce a distribution of electric and magnetic surface currents $\mathcal{S}_n = \binom{\mathbf{K}_n}{\mathbf{N}_n}$ on the *n*th material interface, and the fields radiated by these effective currents account for the disturbance presented by the substrate.

2.2 Computation of Fourier-space DGF $\widetilde{\mathcal{G}}(\mathbf{q})$

To compute the substrate correction to the fields of external sources, I consider the effective tangential electric and magnetic surface currents \mathbf{K} and \mathbf{N} induced on the interfacial layers by the external field sources (Figure 2). This is the direct extension to full-wave problems of the formalism I used in the electrostatic case, and it comports well with the spirit of surface-integral-equation methods.

More specifically, on the material interface layer at $z=z_n$ I have a four-vector surface-current density $\mathcal{S}_n(\rho)$, where $\rho=(x,y)$ and the components of \mathcal{S} are

$$\boldsymbol{\mathcal{S}}_{n}(\boldsymbol{\rho}) = \begin{pmatrix} K_{x}(\boldsymbol{\rho}) \\ K_{y}(\boldsymbol{\rho}) \\ N_{x}(\boldsymbol{\rho}) \\ N_{y}(\boldsymbol{\rho}) \end{pmatrix}. \tag{5}$$

Fields in layer interiors. I will adopt the convention that the lower (upper) bounding surface for each region is the positive (negative) bounding surface for that region in the usual sense of SCUFF-EM regions and surfaces (in which the sign of a {surface,region} pair $\{S, \mathcal{R}\}$ is the sign with which surface currents on S contribute to fields in R). Thus, at a point $\mathbf{x} = (\boldsymbol{\rho}, z)$ in the interior of layer $n \ (z_{n-1} > z > z_n)$, the six-vector of total fields $\mathcal{F} = \begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix}$ reads

$$\mathcal{F}_n(\boldsymbol{\rho}, z) = -\Gamma^{0n}(z_{n-1}) \star \mathcal{S}_{n-1} + \Gamma^{0n}(z_n) \star \mathcal{S}_n + \mathcal{F}_n^{\text{ext}}(\boldsymbol{\rho}, z)$$
 (6)

where $\mathcal{F}_n^{\mathrm{ext}}$ are the externally-sourced (incident) fields due to sources in layer n, Γ^{0n} is the 6×6 homogeneous dyadic Green's function for material layer n,

and \star is shorthand for the convolution operation

"
$$\mathcal{F}(\boldsymbol{\rho}, z) \equiv \Gamma(z') \star \mathcal{S}'' \implies \mathcal{F}(\boldsymbol{\rho}, z) = \int \Gamma(\boldsymbol{\rho} - \boldsymbol{\rho}', z - z') \cdot \mathcal{S}(\boldsymbol{\rho}') d\boldsymbol{\rho}'$$
 (7)

where the integral extends over the entire interfacial plane. I will evaluate convolutions of this form using the 2D Fourier representation of Γ^{0n} :

$$\mathbf{\Gamma}^{0n}(\boldsymbol{\rho}, z) = \int \frac{d^2 \mathbf{q}}{(2\pi)^2} \widetilde{\mathbf{\Gamma}^{0n}}(\mathbf{q}, z) e^{i\mathbf{q}\cdot\boldsymbol{\rho}}$$
(8a)

$$\widetilde{\mathbf{\Gamma}^{0n}}(\mathbf{q}, z) = \frac{1}{2} \begin{pmatrix} -\frac{\omega\mu_0\mu_n}{q_{zn}}\widetilde{\mathbf{G}}^{\pm} & +\widetilde{\mathbf{C}}^{\pm} \\ -\widetilde{\mathbf{C}}^{\pm} & -\frac{\omega\epsilon_0\epsilon_n}{q_{zn}}\widetilde{\mathbf{G}}^{\pm} \end{pmatrix} e^{iq_z|z|}$$
(8b)

$$\widetilde{\mathbf{G}}^{\pm}(\mathbf{q},k) = \begin{pmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{pmatrix} - \frac{1}{k^2} \begin{pmatrix} q_x^2 & q_x q_y & \pm q_x q_z \\ q_y q_x & q_y^2 & \pm q_y q_z \\ \pm q_z q_x & \pm q_z q_y & q_z^2 \end{pmatrix}$$
(8c)

$$\widetilde{\mathbf{C}}^{\pm}(\mathbf{q},k) = \begin{pmatrix} 0 & \mp 1 & +q_y/q_z \\ \pm 1 & 0 & -q_x/q_z \\ -q_y/q_z & +q_x/q_z & 0 \end{pmatrix}$$
(8d)

$$k_n \equiv \sqrt{\epsilon_0 \epsilon_n \mu_0 \mu_n} \cdot \omega, \qquad q_z \equiv \sqrt{k^2 - |\mathbf{q}|^2}, \qquad \pm = \text{sign } z.$$
 (8e)

With this representation, convolutions like (7) become products in Fourier space:

$$\Gamma(z') \star \mathcal{S} = \mathcal{F}(\boldsymbol{\rho}, z) = \int \frac{d^2\mathbf{q}}{(2\pi)^2} \widetilde{\mathcal{F}}(\mathbf{q}, z) e^{i\mathbf{q}\cdot\boldsymbol{\rho}}, \quad \text{with} \quad \widetilde{\mathcal{F}}(\mathbf{q}, z) = \widetilde{\Gamma}(\mathbf{q}, z - z') \widetilde{\mathcal{S}}(\mathbf{q})$$

Surface currents from incident fields. To determine the surface currents induced by given incident-field sources, I apply boundary conditions. The boundary condition at $z = z_n$ is that the tangential **E**, **H** fields be continuous: in Fourier space, we have

$$\widetilde{\mathcal{F}}_{\parallel}(\mathbf{q}, z = z_n^+) = \widetilde{\mathcal{F}}_{\parallel}(\mathbf{q}, z = z_n^-)$$
 (9)

The fields just **above** the interface $(z \to z_n^+)$ receive contributions from three sources:

- Surface currents at $z = z_{n-1}$, which contribute with a minus sign and via the Green's function for region n;
- Surface currents at $z = z_n$, which contribute with a plus sign and via the Green's function for region n; and
- external field sources in region n.

The fields just **below** the interface $(z=z_n^-)$ receive contributions from three sources:

- Surface currents at $z=z_n$, which contribute with a minus sign and via the Green's function for region n + 1;
- Surface currents at $z=z_{n+1}$, which contribute with a plus sign and via the Green's function for region n + 1; and
- external field sources in region n+1.

Then equation (9) reads (temporarily omitting **q** arguments)

$$-\widetilde{\mathbf{\Gamma}^{0n}}_{\parallel}(z_{n}-z_{n-1})\cdot\widetilde{\boldsymbol{\mathcal{S}}}_{n-1}+\widetilde{\mathbf{\Gamma}^{0n}}_{\parallel}(0^{+})\cdot\widetilde{\boldsymbol{\mathcal{S}}}_{n}+\widetilde{\boldsymbol{\mathcal{F}}}_{n\parallel}^{\mathrm{ext}}(z_{n})$$

$$=-\widetilde{\mathbf{\Gamma}^{0,n+1}}_{\parallel}(0^{-})\cdot\widetilde{\boldsymbol{\mathcal{S}}}_{n}+\widetilde{\mathbf{\Gamma}^{0,n+1}}_{\parallel}(z_{n}-z_{n+1})\cdot\widetilde{\boldsymbol{\mathcal{S}}}_{n+1}+\widetilde{\boldsymbol{\mathcal{F}}}_{n+1\parallel}^{\mathrm{ext}}(z_{n})$$

$$\mathbf{M}_{n,n-1} \cdot \widetilde{\boldsymbol{\mathcal{S}}}_{n-1} + \mathbf{M}_{n,n} \cdot \widetilde{\boldsymbol{\mathcal{S}}}_n + \mathbf{M}_{n,n+1} \cdot \widetilde{\boldsymbol{\mathcal{S}}}_{n+1} = \widetilde{\boldsymbol{\mathcal{F}}}_{n+1\parallel}^{\mathrm{ext}}(z_n) - \widetilde{\boldsymbol{\mathcal{F}}}_{n\parallel}^{\mathrm{ext}}(z_n) \quad (10)$$

with the 4×4 matrix blocks⁴

$$\mathbf{M}_{n,n-1} = -\widetilde{\mathbf{\Gamma}^{0n}}_{\parallel}(z_n - z_{n-1}) \tag{13a}$$

$$\mathbf{M}_{n,n} = +\widetilde{\mathbf{\Gamma}^{0n}}_{\parallel}(0^{+}) + \widetilde{\mathbf{\Gamma}^{0,n+1}}_{\parallel}(0^{-})$$
(13b)

$$\mathbf{M}_{n,n+1} = -\widetilde{\mathbf{\Gamma}^{0,n+1}}_{\parallel}(z_n - z_{n+1}) \tag{13c}$$

Writing down equation (10) equation for all N dielectric interfaces yields a $4N \times 4N$ system of linear equations, with triadiagonal 4×4 block form, relating the surface currents on all layers to the external fields due to sources in all regions:

$$\mathbf{M} \cdot \mathbf{s} = \mathbf{f} \tag{14}$$

$$\mathbf{M}_{n,n} = \sum_{r \in \{n, n+1\}} \frac{1}{2} \begin{pmatrix} -\frac{\omega \epsilon_r}{Z_0 q_{zr}} \mathbf{g}(k_r, \mathbf{q}) & 0\\ 0 & -\frac{\omega \mu_r Z_0}{q_{zr}} \mathbf{g}(k_r, \mathbf{q}) \end{pmatrix}$$
(11)

$$\mathbf{M}_{n,n} = \sum_{r \in \{n,n+1\}} \frac{1}{2} \begin{pmatrix} -\frac{\omega \epsilon_r}{Z_0 q_{zr}} \mathbf{g}(k_r, \mathbf{q}) & 0 \\ 0 & -\frac{\omega \mu_r Z_0}{q_{zr}} \mathbf{g}(k_r, \mathbf{q}) \end{pmatrix}$$

$$\mathbf{M}_{n,n\pm 1} = \frac{1}{2} \begin{pmatrix} -\frac{\omega \epsilon_r}{Z_0 q_{zr}} \mathbf{g}(k_r, \mathbf{q}) & \mathbf{c}^{\pm} \\ -\mathbf{c}^{\pm} & -\frac{\omega \mu_r Z_0}{q_{zn}^*} \mathbf{g}(k_r, \mathbf{q}) \end{pmatrix} e^{iq_{zr}|z_n - z_{n\pm 1}|}$$

$$(12)$$

where I put $r \equiv \begin{cases} n, & \text{for } \mathbf{M}_{n,n-1} \\ n+1, & \text{for } \mathbf{M}_{n,n+1} \end{cases}$ and

$$\mathbf{g}(k; \mathbf{q}) = \mathbf{1} - \frac{\mathbf{q}\mathbf{q}^{\top}}{k^2}, \qquad \mathbf{c}^{\pm} = \begin{pmatrix} 0 & \mp 1 \\ \pm 1 & 0 \end{pmatrix}$$

⁴The 4×4 M blocks here have 2×2 block structure:

where **M** is the $4N \times 4N$ block-tridiagonal matrix (13) and where the 4N-vectors **s**, **f** read

$$\mathbf{s} = \left(egin{array}{c} \widetilde{oldsymbol{\mathcal{S}}}_1 \ \widetilde{oldsymbol{\mathcal{S}}}_2 \ \widetilde{oldsymbol{\mathcal{S}}}_3 \ dots \ \widetilde{oldsymbol{\mathcal{S}}}_N \end{array}
ight), \qquad \mathbf{f} = \left(egin{array}{c} -\widetilde{oldsymbol{\mathcal{F}}}_{1\parallel}(z_1) + \widetilde{oldsymbol{\mathcal{F}}}_{2\parallel}(z_1) \ -\widetilde{oldsymbol{\mathcal{F}}}_{2\parallel}(z_2) + \widetilde{oldsymbol{\mathcal{F}}}_{3\parallel}(z_2) \ -\widetilde{oldsymbol{\mathcal{F}}}_{3\parallel}(z_3) + \widetilde{oldsymbol{\mathcal{F}}}_{3\parallel}(z_4) \ dots \ -\widetilde{oldsymbol{\mathcal{F}}}_{N-1,\parallel}(z_{N-1}) + \widetilde{oldsymbol{\mathcal{F}}}_{N\parallel}(z_{N-1}) \end{array}
ight).$$

Solving (14) yields the induced surface currents on all layers in terms of the incident fields:

$$\mathbf{s} = \mathbf{W} \cdot \mathbf{f}$$
 where $\mathbf{W} \equiv \mathbf{M}^{-1}$

or, more explicitly,

$$\widetilde{\boldsymbol{\mathcal{S}}}_n = \sum_m W_{nm} \mathbf{f}_m \tag{15}$$

Surface currents induced by point sources

For DGF computations the incident fields arise from a single point source—say, a *j*-directed source in region s. Then the only nonzero length-4 blocks of the RHS vector in (14) are \mathbf{f}_{s-1} , \mathbf{f}_{s} with components ($\ell = \{1, 2, 4, 5\}$)

$$\left(\mathbf{f}_{s-1}\right)_{\ell} = -\widetilde{\Gamma}_{\ell j}^{0s}(z_{s-1} - z_{s}), \qquad \left(\mathbf{f}_{s}\right)_{\ell} = +\widetilde{\Gamma}_{\ell j}^{0s}(z_{s} - z_{s}) \tag{16}$$

and the surface currents on interface layer n are obtained by solving (15):

$$\widetilde{\boldsymbol{\mathcal{S}}}_{n} = \mathbf{W}_{n,s-1} \, \mathbf{f}_{s-1} + \mathbf{W}_{n,s} \, \mathbf{f}_{s}$$

$$= \sum_{p=0}^{1} (-1)^{p+1} \mathbf{W}_{n,s-1+p} \cdot \widetilde{\boldsymbol{\Gamma}^{0s}}_{\parallel,j} (z_{s} - z_{s-1+p})$$
(17)

Fields due to surface currents

Given the surface currents induced by a j-directed point source at \mathbf{x}_s , I evaluate the fields due to these currents to get the substrate DGF contribution \mathcal{G} . If the evaluation point \mathbf{x}_D lies in region d, then the fields receive contributions from the surface currents at z_{d-1} and z_D , propagated by the homogeneous DGF for region d:

$$\widetilde{\mathcal{F}}(z_{\mathrm{D}}) = -\widetilde{\Gamma^{0d}}(z_{\mathrm{D}} - z_{d-1}) \cdot \widetilde{\mathcal{S}}_{d-1} + \widetilde{\Gamma^{0d}}(z_{\mathrm{D}} - z_{\mathrm{D}}) \cdot \widetilde{\mathcal{S}}_{d}$$

$$= \sum_{q=0}^{1} (-1)^{q+1} \widetilde{\Gamma^{0d}}(z_{\mathrm{D}} - z_{d+q-1}) \cdot \widetilde{\mathcal{S}}_{d+q-1}$$
(18)

(The minus sign in the first term arises because, in my convention, surface currents on the upper surface of a region contribute to the fields in that region with a minus sign). Inserting (17), the i component here—which is the ij component of the substrate DGF—is

$$\widetilde{\mathcal{G}}_{ij}(z_{\text{D}}, z_{\text{S}}) = \sum_{p,q=0}^{1} (-1)^{p+q} \widetilde{\mathbf{\Gamma}^{0d}}_{i,\parallel}(z_{\text{D}} - z_{d-1+q}) \mathbf{W}_{d-1+q,s-1+p} \widetilde{\mathbf{\Gamma}^{0s}}_{\parallel,j}(z_{s-1+p} - z_{\text{S}}).$$
(19)

The calculation of equation (19) is carried out by the routine GetGTwiddle in LIBSUBSTRATE.

Green's functions for potentials

In equation (18) I am computing the 6 components of the **E** and **H** fields produced by the induced surface currents. If instead I compute the *potentials* produced by those currents I obtain a slightly different Green's function. Thus, let $\mathbf{A}^{\mathrm{E}}, \Phi^{\mathrm{E}}$ be the usual vector and scalar potential of an electric-current source in a homogeneous region, and let $\mathbf{A}^{\mathrm{M}}, \Phi^{\mathrm{M}}$ be their counterparts for magnetic-current sources, i.e. if the electric and magnetic volume currents are **J** and **M** then

$$\mathbf{A}^{\mathrm{E}}(\mathbf{x}_{\mathrm{D}}) = \mu \int \mathbf{J}(\mathbf{x}_{\mathrm{S}}) G_{0}(\mathbf{x}_{\mathrm{DS}}) d\mathbf{x}_{\mathrm{S}}, \quad \Phi^{\mathrm{E}}(\mathbf{x}_{\mathrm{D}}) = \frac{1}{i\omega\epsilon} \int (\nabla \cdot \mathbf{J}) G_{0}(\mathbf{x}_{\mathrm{DS}}) d\mathbf{x}_{\mathrm{S}}$$

$$(20a)$$

$$\mathbf{A}^{\mathrm{M}}(\mathbf{x}_{\mathrm{D}}) = \epsilon \int \mathbf{M}(\mathbf{x}_{\mathrm{S}}) G_{0}(\mathbf{x}_{\mathrm{DS}}) d\mathbf{x}_{\mathrm{S}}, \quad \Phi^{\mathrm{M}}(\mathbf{x}_{\mathrm{D}}) = \frac{1}{i\omega\mu} \int (\nabla \cdot \mathbf{M}) G_{0}(\mathbf{x}_{\mathrm{DS}}) d\mathbf{x}_{\mathrm{S}}$$

$$(20b)$$

with $\mathbf{x}_{DS} \equiv \mathbf{x}_{D} - \mathbf{x}_{S}$ and

$$G_0(k; \mathbf{r}) = \frac{e^{ik|\mathbf{r}|}}{4\pi|\mathbf{r}|} = \int \frac{d^2\mathbf{q}}{(2\pi)^2} \widetilde{G}_0(\mathbf{q}, z) e^{i\mathbf{q}\cdot\boldsymbol{\rho}}, \qquad \widetilde{G}_0 = \frac{i}{2q_z} e^{iq_z|z|}.$$

I write

2.3 Reduction of 2D Fourier integrals to 1D (Sommerfeld) integrals

The real-space DGF correction is the inverse Fourier transform of (19):

$$oldsymbol{\mathcal{G}}(oldsymbol{
ho},z_{ ext{ iny D}},z_{ ext{ iny S}}) = \int rac{d^2 \mathbf{q}}{(2\pi)^2} \widetilde{oldsymbol{\mathcal{G}}}(\mathbf{q};z_{ ext{ iny D}};z_{ ext{ iny S}}) e^{i \mathbf{q} \cdot oldsymbol{
ho}}$$

or, in polar coordinates with $(q_x, q_y) = (q \cos \theta_q, q \sin \theta_q), (\rho_x, \rho_y) = (\rho \cos \theta_\rho, \rho \sin \theta_\rho),$

$$\mathcal{G}(\boldsymbol{\rho}) = \int_0^\infty \frac{q \, dq}{2\pi} \int_0^{2\pi} \frac{d\theta_q}{2\pi} \widetilde{\mathcal{G}}(\mathbf{q}) e^{iq\rho \cos(\theta_q - \theta_\rho)}. \tag{21}$$

(Here and for much of this section I suppress $z_{\text{D,S}}$ arguments, but one must remember that they are always there.⁵) The goal of this section is to integrate out the angular variable θ_q to reduce the 2D integral over \mathbf{q} to a 1D integral over $q = |\mathbf{q}|$. In abbreviated form this proceeds as follows:

1. Separate variables by writing $\widetilde{\mathcal{G}}(\mathbf{q})$ as a sum of products of θ_q -independent scalar functions $\widetilde{g}(q)$ times q-independent matrix-valued functions $\mathbf{\Lambda}(\theta_q)$ (Section 2.3.1):

$$\widetilde{\mathcal{G}}(\mathbf{q}) = \sum_{n=1}^{18} \widetilde{g}^{(n)}(q) \mathbf{\Lambda}^{(n)}(\theta_q)$$

2. Evaluate integrals over θ_q analytically to yield Bessel functions $J_{\nu}(q\rho)$ multiplying q-independent matrix-valued functions $\Lambda(\theta_{\rho})$ (Section 2.3.2). After this step (21) reads

$$\mathcal{G}(\rho) = \sum_{m=1}^{22} \underbrace{\left[\int_0^\infty \widetilde{\mathfrak{g}}^{(m)}(q,\rho) \, dq \right]}_{\mathfrak{g}^{(m)}(\rho)} \mathbf{\Lambda}^{(m)}(\theta_\rho)$$
 (22)

where the $\widetilde{\mathfrak{g}}(q,\rho)$ functions are linear combinations of the $\widetilde{g}(q)$ functions times Bessel functions in $q\rho$ and other factors.

3. Evaluate the remaining integrals over q numerically using sophisticated tricks for evaluating Sommereld integrals (Section 2.3.3).

2.3.1 Factor $\hat{\mathcal{G}}$ into q-independent and θ_q -independent terms

I begin by noting that $\hat{\mathcal{G}}(\mathbf{q})$ may be decomposed as a sum of scalar functions of $q = |\mathbf{q}|$ times q-independent matrix-valued functions of $\theta_{\mathbf{q}}$:

$$\widetilde{\mathcal{G}}(\mathbf{q}) = \sum_{n=1}^{18} \widetilde{g}^{(n)}(q) \mathbf{\Lambda}^{(n)}(\theta_{\mathbf{q}})$$
(23)

⁵More specifically, the "g-like" quantities $\mathcal{G}(\rho)$, $\widetilde{\mathcal{G}}(\mathbf{q})$, $\widetilde{\mathcal{G}}(\mathbf{q})$, $\widetilde{\mathfrak{g}}(q)$, and $\mathfrak{g}(\rho)$ all depend on $z_{\mathrm{S,D}}$, but the matrix-valued functions $\Lambda_n(\theta)$ do not.

For example, the upper two quadrants read

$$\begin{split} \widetilde{\boldsymbol{\mathcal{G}}^{\text{EE}}}(\mathbf{q}) = & \widetilde{\boldsymbol{\mathcal{G}}^{\text{EEO}\parallel}}(q) \underbrace{\begin{pmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{pmatrix}}_{\boldsymbol{\Lambda}^{0\parallel}} + \widetilde{\boldsymbol{\mathcal{G}}^{\text{EEO}z}}(q) \underbrace{\begin{pmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1 \end{pmatrix}}_{\boldsymbol{\Lambda}^{0z}} \\ + & \widetilde{\boldsymbol{\mathcal{G}}^{\text{EE1}}}(q) \underbrace{\begin{pmatrix} 0 & 0 & \cos\theta_{\mathbf{q}} \\ 0 & 0 & \sin\theta_{\mathbf{q}} \\ 0 & 0 & 0 \end{pmatrix}}_{\boldsymbol{\Lambda}^{1}(\theta_{\mathbf{q}})} + & \widetilde{\boldsymbol{\mathcal{G}}^{\text{EE1}}}(q) \underbrace{\begin{pmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ \cos\theta_{\mathbf{q}} & \sin\theta_{\mathbf{q}} & 0 \end{pmatrix}}_{\boldsymbol{\Lambda}^{1\top}(\theta_{\mathbf{q}})} \\ + & \widetilde{\boldsymbol{\mathcal{G}}^{\text{EE2}}}(q) \underbrace{\begin{pmatrix} \cos^{2}\theta_{\mathbf{q}} & \cos\theta_{\mathbf{q}}\sin\theta_{\mathbf{q}} & 0 \\ \cos\theta_{\mathbf{q}}\sin\theta_{\mathbf{q}} & \sin^{2}\theta_{\mathbf{q}} & 0 \\ 0 & 0 & 0 \end{pmatrix}}_{\boldsymbol{\Lambda}^{2}(\theta_{\mathbf{q}})} \end{split}$$

$$\begin{split} \widetilde{\boldsymbol{\mathcal{G}}^{\text{EM}}}(\mathbf{q}) = & \widetilde{\boldsymbol{\mathcal{G}}}^{\text{EM0}\parallel}(q) \underbrace{\left(\begin{array}{ccc} 0 & 1 & 0 \\ -1 & 0 & 0 \\ 0 & 0 & 0 \end{array} \right)}_{\mathbf{\Lambda}^{0\times}} + & \widetilde{\boldsymbol{\mathcal{G}}}^{\text{EM2}}(q) \underbrace{\left(\begin{array}{cccc} \cos\theta_{\mathbf{q}} \sin\theta_{\mathbf{q}} & \sin^2\theta_{\mathbf{q}} & 0 \\ -\cos^2\theta_{\mathbf{q}} & -\cos\theta_{\mathbf{q}} \sin\theta_{\mathbf{q}} & 0 \\ 0 & 0 & 0 \end{array} \right)}_{\mathbf{\Lambda}^{2\times}} \\ + & \widetilde{\boldsymbol{\mathcal{G}}}^{\text{EM1}}(q) \underbrace{\left(\begin{array}{cccc} 0 & 0 & -\sin\theta_{\mathbf{q}} \\ 0 & 0 & +\cos\theta_{\mathbf{q}} \\ 0 & 0 & 1 \end{array} \right)}_{\mathbf{Q}} + & \underbrace{\boldsymbol{\mathcal{G}}^{\text{EM1}}}(q) \underbrace{\left(\begin{array}{cccc} 0 & 0 & 0 \\ 0 & 0 & 0 \\ -\sin\theta_{\mathbf{q}} & \cos\theta_{\mathbf{q}} & 1 \end{array} \right)}_{\mathbf{Q}} \\ \end{split}$$

where the \top superscript indicates matrix transpose. The expressions for $\widetilde{\boldsymbol{\mathcal{G}}}^{^{\mathrm{ME}}}$ and $\widetilde{\boldsymbol{\mathcal{G}}}^{^{\mathrm{MM}}}$ are similar, involving the same Λ matrices with different \widetilde{g} prefactors.

2.3.2 Evaluate θ_{q} integrals

Using Table 3, the $\theta_{\mathbf{q}}$ integral in (21) may be evaluated analytically to yield Bessel-function factors $J_{\nu}(q\rho)$ ($\nu \in \{0,1,2\}$) times Λ matrices, now evaluated

$$\frac{1}{2\pi} \int_0^{2\pi} e^{iq\rho\cos(\theta_q - \theta_\rho)} \left\{ \begin{array}{c} 1 \\ \cos\theta_q \\ \sin\theta_q \\ \cos^2\theta_q \\ \cos\theta_q\sin\theta_q \\ \sin^2\theta_q \end{array} \right\} d\theta_q = \left\{ \begin{array}{c} J_0(q\rho) \\ iJ_1(q\rho)\cos\theta_\rho \\ iJ_1(q\rho)\sin\theta_\rho \\ -J_2(q\rho)\cos^2\theta_\rho + \frac{J_1(q\rho)}{q\rho} \\ -J_2(q\rho)\sin^2\theta_\rho + \frac{J_1(q\rho)}{q\rho} \end{array} \right\},$$

Figure 3: Table of integrals used to reduce 2D integrals over \mathbf{q} to 1D integrals over |q|.

at θ_{ρ} . For example, one term in the expansion of $\mathcal{G}(\rho)$ is

$$\int_{0}^{\infty} \frac{q dq}{2\pi} \widetilde{g}^{\text{EE1}}(q) \underbrace{\int_{0}^{2\pi} \frac{d\theta_{q}}{2\pi} \mathbf{\Lambda}^{1}(\theta_{q}) e^{iq\rho \cos(\theta_{q} - \theta_{\rho})}}_{iJ_{1}(q\rho)\mathbf{\Lambda}^{1}(\theta_{\rho})}$$

$$= \underbrace{\left\{ \int_{0}^{\infty} dq \underbrace{\left[\frac{q}{2\pi} \widetilde{g}^{\text{EE1}}(q) \cdot iJ_{1}(q\rho) \right]}_{\widetilde{\mathfrak{g}}^{\text{EE1}}(\rho)} \right\} \mathbf{\Lambda}^{1}(\theta_{\rho})}_{\mathbf{g}^{\text{EE1}}(\rho)}$$

The second line here defines some new symbols: $\tilde{\mathfrak{g}}$ are functions of q and ρ defined as products of $\tilde{g}(q)$ factors times $J_{\nu}(q\rho)$ factors and other factors, while \mathfrak{g} are functions of ρ obtained by integrating out the q dependence of $\mathfrak{g}(q,\rho)$. The

full set of rules defining the $\widetilde{\mathfrak{g}}$ is

$$\widetilde{\mathfrak{g}}^{\text{EEO}\parallel}(q,\rho) \equiv \frac{q}{2\pi} \left[\widetilde{g}^{\text{EEO}\parallel}(q) J_0(q\rho) + \widetilde{g}^{\text{EE2}}(q) \frac{J_1(q\rho)}{q\rho} \right]$$
(24a)

$$\widetilde{\mathfrak{g}}^{\text{EEOz}}(q,\rho) \equiv \frac{q}{2\pi} \widetilde{g}^{\text{EEOz}}(q) J_0(q\rho)$$
 (24b)

$$\widetilde{\mathfrak{g}}^{\text{EE1}}(q,\rho) \equiv i \frac{q}{2\pi} \widetilde{g}^{\text{EE1}}(q) J_1(q\rho)$$
 (24c)

$$\widetilde{\mathfrak{g}}^{\text{EE1T}}(q,\rho) \equiv i \frac{q}{2\pi} \widetilde{g}^{\text{EE1T}}(q) J_1(q\rho)$$
 (24d)

$$\widetilde{\mathfrak{g}}^{\text{EE2}}(q,\rho) \equiv -\frac{q}{2\pi} \widetilde{g}^{\text{EE2}}(q) J_2(q\rho) \tag{24e}$$

$$\widetilde{\mathfrak{g}}^{\text{EMO}\parallel\times}(q,\rho) \equiv \frac{q}{2\pi} \left[\widetilde{g}^{\text{EMO}\parallel}(q) J_0(q\rho) + \widetilde{g}^{\text{EM2}}(q) \frac{J_1(q\rho)}{q\rho} \right]$$
(24f)

$$\widetilde{\mathfrak{g}}^{\text{EM1}\times}(q,\rho) \equiv i \frac{q}{2\pi} \widetilde{g}^{\text{EM1A}}(q) J_1(q\rho)$$
 (24g)

$$\widetilde{\mathfrak{g}}^{\text{EM1}\times\top}(q,\rho) \equiv i\frac{q}{2\pi}\widetilde{g}^{\text{EM1B}}(q)J_1(q\rho)$$
 (24h)

$$\widetilde{\mathfrak{g}}^{\text{EM2}\times}(q,\rho) \equiv -\frac{q}{2\pi}\widetilde{g}^{\text{EM2}}J_2(q\rho)$$
 (24i)

2.3.3 Evaluate Sommerfeld integrals over q

Assembling the above pieces, the substrate DGF correction ${\cal G}$ is a sum of 22 terms:

$$\mathcal{G}(\rho) = \sum_{m=1}^{22} \mathfrak{g}^{(m)}(\rho) \mathbf{\Lambda}^{(m)}(\theta_{\rho}),$$

where the $\mathfrak{g}^{(m)}(\rho)$ functions are defined by Sommerfeld integrals:

$$\mathfrak{g}^{(m)}(\rho) \equiv \int_0^\infty \widetilde{\mathfrak{g}}^{(m)}(q,\rho) \, dq. \tag{25}$$

 $^{^6}$ This tally treats the integrals of the two integrand terms on the RHS of (24a) as two separate integrals [and similarly for (24f) and the corresponding equations for the ME and MM quadrants]. If the terms are lumped together then the number of distinct $\mathfrak g$ functions is 18.

3 SCUFF-EM integration: Substrate contributions to BEM matrix and RHS vector

3.1 Fields of individual basis functions

$$\mathcal{G}^{ ext{EE}} =$$

3.2 SIE matrix elements: Panel-panel integrals

If S_{α}, S_{β} are two RWGSurfaces exposed to the outermost (ambient) region in a SCUFF-EM geometry, then the elements of the SIE matrix elements corresponding to any pair of basis functions $\{\mathbf{b}_a \in \mathcal{S}_{\alpha}, \mathbf{b}_b \in \mathcal{S}_{\beta}\}$ receive corrections of the form

$$\Delta M_{ab}^{PQ} = \left\langle \mathbf{b}_{a} \middle| \mathcal{G}^{PQ} \middle| \mathbf{b}_{b} \right\rangle$$

$$\equiv \iint \mathbf{b}_{a}(\mathbf{x}_{a}) \cdot \mathcal{G}^{PQ}(\mathbf{x}_{a}, \mathbf{x}_{b}) \cdot \mathbf{b}_{b}(\mathbf{x}_{b}) d\mathbf{x}_{b} d\mathbf{x}_{a}$$
(26)

I will consider two different approaches for evaluating the panel-panel integrals here:

1. The spectral inner approach: In this case I simply evaluate the panel-panel cubature in (26), with values of \mathcal{G} at each cubature point computed via the methods of LIBSUBSTRATE as described in the previous section (possibly accelerated via interpolation tables). I call this the "spectral inner" method because in this case the q integral in the definition of \mathcal{G} is the innermost of 3 integrals. Indeed, inserting equation (22) we have

$$\Delta M_{ab}^{PQ} \equiv \iint \mathbf{b}_a(\mathbf{x}_a) \left\{ \sum \mathfrak{g}^{(m)}(\rho) \mathbf{\Lambda}^{(m)}(\theta_\rho) \right\} \mathbf{b}_b(\mathbf{x}_b) d\mathbf{x}_b d\mathbf{x}_a$$

[where $\rho = (\mathbf{x}_a - \mathbf{x}_b)_{\parallel} = (\rho \cos \theta_{\rho}, \rho \sin \theta_{\rho})$]. Recalling the definition (25), this is a sum of triple integrals:

$$\equiv \iint \mathbf{b}_{a}(\mathbf{x}_{a}) \left\{ \sum \left[\int_{0}^{\infty} \widetilde{\mathfrak{g}}^{(m)}(q,\rho) dq \right] \mathbf{\Lambda}^{(m)}(\theta_{\rho}) \right\} \cdot \mathbf{b}_{b}(\mathbf{x}_{b}) d\mathbf{x}_{b} d\mathbf{x}_{a}.$$
(27)

2. The spectral outer approach: In this case I rearrange the order of integration in (28) so that the q integral is the outermost integral, with an integrand defined for each q by a panel-panel integral involving the spectral-domain GF:

$$\Delta M_{ab}^{PQ} = \int_0^\infty \left\{ \iint \mathbf{b}_a(\mathbf{x}_a) \left[\sum \widetilde{\mathfrak{g}}^{(m)}(q,\rho) \mathbf{\Lambda}^{(m)}(\theta_\rho) \right] \mathbf{b}_b(\mathbf{x}_b) \, d\mathbf{x}_b \, d\mathbf{x}_a \right\} dq$$
(28)

⁷I refer to 4-dimensional integrals like (26) as "panel-panel integrals" because they are a sum of contributions of integrals over pairs of flat triangular panels.

$$\mathcal{G}_{ij}^{ ext{ iny EE}} = \delta_{ij}$$

4 Metal-on-Insulator geometries

$$\mathbf{E} = iw\mathbf{A} - \nabla\phi$$
$$= iw\mu G_0 \star \mathbf{J} - \frac{1}{i\omega\epsilon} \nabla G_0 \star \boldsymbol{\rho}$$

$$\begin{split} \widetilde{G}^{\mbox{\tiny App}} &= \frac{1}{2\pi} q J_0(q\rho) \zeta^{\mbox{\tiny App}} \\ \widetilde{G}^{\mbox{\tiny Apz}} &= -\frac{1}{2\pi} (\epsilon_r - 1) q^2 J_1(q\rho) \zeta^{\mbox{\tiny Apz}} \\ \widetilde{G}^{\mbox{\tiny \Phi}} &= \frac{1}{2\pi} \rho J_0(q\rho) \zeta^{\mbox{\tiny \Phi}} \end{split}$$

$$\zeta^{\text{App}} = \frac{1}{D^{\text{TE}}} \times \begin{cases} e^{-u_0 z} \\ \frac{\sinh u(z+h)}{\sinh uh} \end{cases}$$

$$\zeta^{\text{Apz}} = \frac{1}{D^{\text{TE}}D^{\text{TM}}} \times \begin{cases} e^{-u_0 z} \\ \frac{\cosh u(z+h)}{\cosh uh} \end{cases}$$

$$\zeta^{\Phi} = \frac{N}{D^{\text{TE}}D^{\text{TM}}} \times \begin{cases} e^{-u_0 z} \\ \frac{\sinh u(z+h)}{\sinh uh} \end{cases}$$

$$\frac{N}{D_{\text{TE}}D_{\text{TM}}} \xrightarrow{u_0 \to u} \frac{1 - e^{-2u_0 h}}{u_0(\epsilon + 1)} \sum_{n=0}^{\infty} \left[- \eta e^{-2u_0 h} \right]^n$$

5 Unit-test framework

The LIBSUBSTRATE standalone library comes with a unit-test suite to test core functionality related to calculation of substrate DGFs. Separately, the unit-test suite for LIBSCUFF includes tests to check the integration of LIBSUBSTRATE into LIBSCUFF.

5.1 LIBSUBSTRATE unit tests

5.1.1 tGTwiddle

The unit-test code tGTwiddle.cc tests that the full Fourier-space DGF $\widetilde{\Gamma}(\mathbf{q}, z_{\text{D}}, z_{\text{S}})$ satisfies the appropriate boundary conditions at each layer of the layered substrate, namely

$$C^{+}(\mathbf{P}, i, \ell)\widetilde{\Gamma}_{ij}^{\mathrm{PQ}}(\mathbf{q}, z_{\ell} + \eta, z_{\mathrm{S}})C^{-}(\mathbf{P}, i, \ell)\widetilde{\Gamma}_{ij}^{\mathrm{PQ}}(\mathbf{q}, z_{\ell} - \eta, z_{\mathrm{S}})$$
(29)

where

$$C^{\pm}(P, i, \ell) = \begin{cases} 1, & i \in \{x, y\} \\ \epsilon_{\ell}^{\pm}, & i = z, P = E \\ \mu_{\ell}^{\pm}, & i = z, P = H \end{cases}$$

where $\{\epsilon,\mu\}_{\ell}^{\pm}$ are the material properties for the layer above/below z_{ℓ} , i.e. (Figure $\ref{eq:condition}$)

$$\{\epsilon_{\ell}, \mu_{\ell}\}^{+} = \{\epsilon_{\ell}, \mu_{\ell}\}, \qquad \{\epsilon_{\ell}, \mu_{\ell}\}^{-} = \{\epsilon_{\ell+1}, \mu_{\ell+1}\}.$$

If a ground plane is present, we have the additional condition

$$\widetilde{\Gamma}_{ij}^{PQ}(q, z_{GP}, z_{S}) = 0 \quad \text{for} \quad i \in \{x, y\}.$$
(30)

Conditions (??) and (??) must hold independently of the indices $Q \in \{E, H\}$ and $j \in \{1, 2, 3\}$ and of the values of \mathbf{q} and z_s .

A Symbols and indices used in this document

A.1 Symbols

Symbol	Arguments	Description
\mathcal{F}	r, geometry	Field six-vector $\mathcal{F} = \begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix}$
С	r, geometry	Current six-vector $C = {J \choose M}$ or $C = {K \choose N}$
Г	$\rho, z_{\scriptscriptstyle \mathrm{D}}, z_{\scriptscriptstyle \mathrm{S}}, \omega, \mathrm{geometry}$	Full (bare+scattered) 6×6 dyadic Green's function, $\mathcal{F} = \Gamma \star \mathcal{C}$
$oldsymbol{\Gamma}^{0r}$	$oldsymbol{ ho}, z_{ exttt{ iny D}}, z_{ exttt{ iny S}}, \omega, \epsilon^r, \mu^r$	Bare (homogeneous) 6×6 dyadic Green's function in region r
\mathcal{G}	$\rho, z_{\scriptscriptstyle \mathrm{D}}, z_{\scriptscriptstyle \mathrm{S}}, \omega, \mathrm{geometry}$	Scattering contribution to Γ $(\Gamma = \Gamma^{0r} + \mathcal{G})$
\mathcal{P}	r, geometry	$ ext{Potential eight-vector } oldsymbol{\mathcal{P}} = \left(egin{array}{c} \mathbf{A}^{ ext{E}} \ \mathbf{\Phi}^{ ext{E}} \ \mathbf{A}^{ ext{M}} \ \mathbf{\Phi}^{ ext{M}} \end{array} ight)$
S	r, geometry	$\text{Source eight-vector } \boldsymbol{\mathcal{S}} = \left(\begin{array}{c} \mathbf{J} \\ \rho^{\text{E}} \\ \mathbf{M} \\ \rho^{\text{M}} \end{array} \right)$
Λ	$ ho, z_{\scriptscriptstyle m D}, z_{\scriptscriptstyle m S}, \omega, { m geometry}$	Full (bare+scattered) 8 × 8 dyadic Green's function, $\mathcal{P} = \Pi \star \mathcal{S}$
$oldsymbol{\Lambda}^{0r}$	$\rho, z_{\scriptscriptstyle \mathrm{D}}, z_{\scriptscriptstyle \mathrm{S}}, \omega, \mathrm{geometry}$	Bare (homogeneous) 8×8 dyadic Green's function for region r
L	$\rho, z_{\scriptscriptstyle \mathrm{D}}, z_{\scriptscriptstyle \mathrm{S}}, \omega, \mathrm{geometry}$	Scattering contribution to Λ ($\Lambda = \Lambda^{0r} + \mathcal{L}$)

A.2 Indices

Index	Range	Significance
i, j	$\{1, 2, 3\}$	Cartesian directions x, y, z
I,J	$\{1, 2, 3, 4, 5, 6\}$	Electric/magnetic field/current components
μ, ν	$\{1, 2, 3, 4, 5, 6, 7, 8\}$	Electric/magnetic potential/source components $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$

B 8×8 Dyadic Green's Functions

The usual 6×6 dyadic Green's function Γ operates on a six-vector of currents to yield a six-vector of fields. It is convenient to consider a slightly different object that operates on an *eight*-vector of sources to yield an *eight*-vector of potentials.

In the presence of magnetic currents, the usual (electric-current-sourced) vector and scalar potentials $\mathbf{A}^{\text{E}}, \Phi^{\text{E}}$, are joined by their magnetic-current-sourced counterparts $\mathbf{A}^{\text{M}}, \Phi^{\text{M}}$, which are related to the fields according to

$$\begin{split} \mathbf{E} &= i\omega\mu\mathbf{A}^{\mathrm{E}} - \frac{1}{i\omega\epsilon}\nabla\Phi^{\mathrm{E}} - \nabla\times\mathbf{A}^{\mathrm{M}} \\ \mathbf{M} &= \nabla\times\mathbf{A}^{\mathrm{E}} + i\omega\epsilon\mathbf{A}^{\mathrm{M}} - \frac{1}{i\omega\mu}\nabla\Phi^{\mathrm{M}}. \end{split}$$

In a homogeneous region, the potentials 8 produced by given source distributions $\{{\bf J},{\bf M}\}$ are

$$\begin{split} \mathbf{A}^{\scriptscriptstyle{\mathrm{E}}}(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}}) &= \int G_0(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}} - \mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \mathbf{J}(\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \, d\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}, \qquad \Phi^{\scriptscriptstyle{\mathrm{E}}}(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}}) = \int G_0(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}} - \mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \big[\nabla \cdot \mathbf{J} \big] \, d\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}} \\ \mathbf{A}^{\scriptscriptstyle{\mathrm{M}}}(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}}) &= \int G_0(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}} - \mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \mathbf{M}(\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \, d\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}, \qquad \Phi^{\scriptscriptstyle{\mathrm{M}}}(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}}) = \int G_0(\mathbf{x}_{\scriptscriptstyle{\mathrm{D}}} - \mathbf{x}_{\scriptscriptstyle{\mathrm{S}}}) \big[\nabla \cdot \mathbf{M} \big] \, d\mathbf{x}_{\scriptscriptstyle{\mathrm{S}}} \end{split}$$

where

$$G_0(\mathbf{r}) = \frac{e^{ik|\mathbf{r}|}}{4\pi|\mathbf{r}|}.$$

$$\begin{pmatrix} E_x \\ E_y \\ E_z \\ H_x \\ H_y \\ H_z \end{pmatrix} = \begin{pmatrix} i\omega\mu G_0 & 0 & 0 & -\frac{1}{i\omega\epsilon}\partial_x G_0 & 0 & \partial_z G_0 & -\partial_y G_0 & 0 \\ 0 & i\omega\mu G_0 & 0 & -\frac{1}{i\omega\epsilon}\partial_y G_0 & -\partial_z G_0 & 0 & \partial_x G_0 & 0 \\ 0 & 0 & i\omega\mu G_0 & -\frac{1}{i\omega\epsilon}\partial_z G_0 & \partial_y G_0 & -\partial_x G_0 & 0 & 0 \\ 0 & -\partial_z G_0 & \partial_y G_0 & 0 & i\omega\epsilon G_0 & 0 & 0 & -\frac{1}{i\omega\mu}\partial_x G_0 \\ \partial_z G_0 & 0 & \partial_x G_0 & 0 & 0 & i\omega\epsilon G_0 & 0 & -\frac{1}{i\omega\mu}\partial_y G_0 \\ -\partial_y G_0 & \partial_x G_0 & 0 & 0 & 0 & i\omega\epsilon G_0 & -\frac{1}{i\omega\mu}\partial_z G_0 \end{pmatrix} \star \begin{pmatrix} J_x \\ J_y \\ J_z \\ \nabla \cdot \mathbf{J} \\ M_x \\ M_y \\ M_z \\ \nabla \cdot \mathbf{M} \end{pmatrix}$$

⁸Note that my $\Phi^{E,M}$ are $i\omega$ times the actual scalar potentials due to the charge distributions associated with currents J, M.