

1.

a.

$$G(s) = \frac{1}{s(s+2)(s+4)}; G(j\omega) = \frac{1}{-6\omega^2 + i(8\omega - \omega^3)}$$

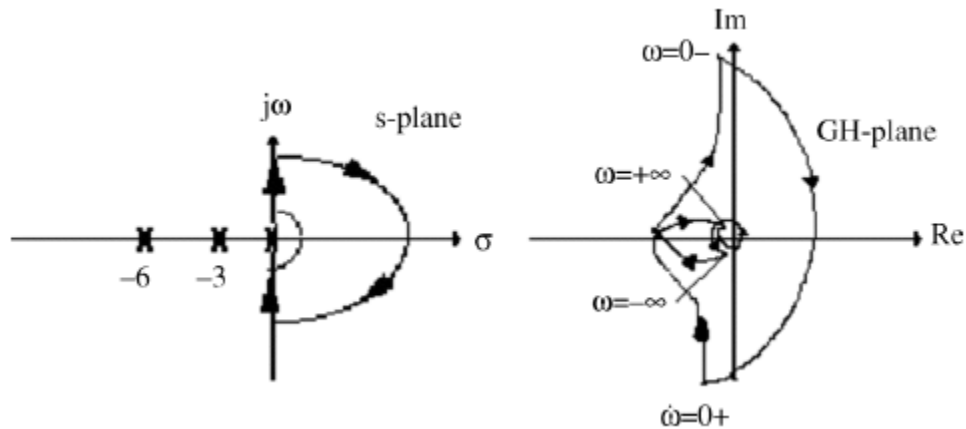
$$M(\omega) = \frac{1}{\sqrt{(8\omega - \omega^3)^2 + (6\omega^2)^2}}; \Phi(\omega) = -\left(\pi + \arctan\left[\frac{8 - \omega^2}{-6\omega}\right]\right)$$

b.

$$G(s) = \frac{(s+5)}{(s+2)(s+4)}; G(j\omega) = \frac{(\omega^2 + 40) - i(\omega^2 + 22)\omega}{\omega^4 + 20\omega^2 + 64}$$

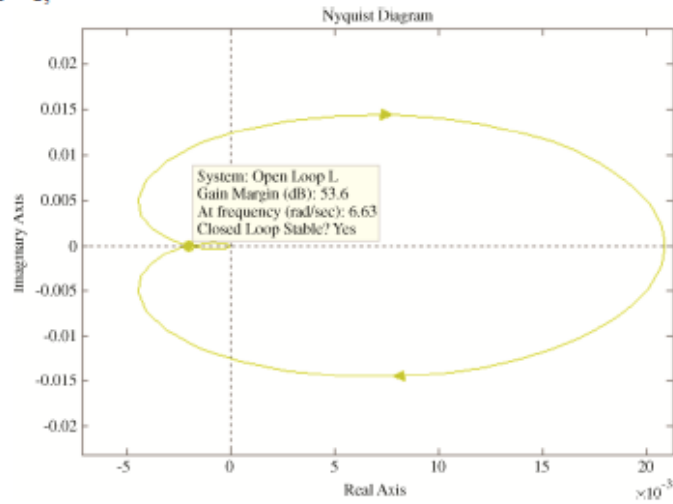
$$M(\omega) = \frac{\sqrt{(\omega^2 + 40)^2 + \omega^2(\omega^2 + 22)^2}}{\omega^4 + 20\omega^2 + 64}; \Phi(\omega) = \arctan\left(\frac{-[\omega^2 + 22]\omega}{\omega^2 + 40}\right)$$

5.



10.

System 1: For  $K = 1$ ,

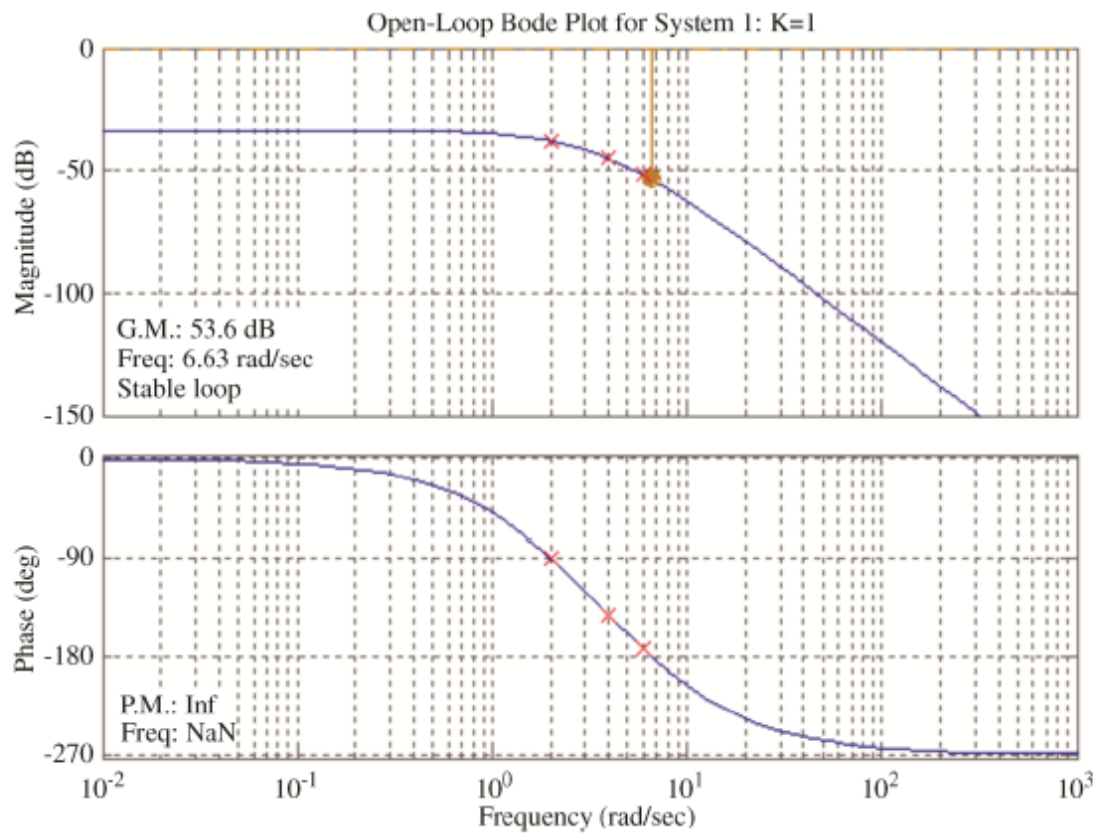


The Nyquist diagram intersects the real axis at -0.0021. Thus  $K$  can be increased to 478.63 before there are encirclements of -1. There are no poles encircled by the contour. Thus  $P = 0$ . Hence,  $Z = P - N$ ,  $Z = 0 + 0$  if  $K < 478.63$ ;  $Z = 0 - (-2)$  if  $K > 478.63$ . Therefore stability if  $0 < K < 478.63$ .

11.

Note: All results for this problem are based upon a non-asymptotic frequency response.

System 1: Plotting Bode plots for  $K = 1$  yields the following Bode plot,



$K = 1000$ :

For  $K = 1$ , phase response is  $180^\circ$  at  $\omega = 6.63$  rad/s. Magnitude response is -53.6 dB at this frequency.

For  $K = 1000$ , magnitude curve is raised by 60 dB yielding +6.4 dB at 6.63 rad/s. Thus, the gain margin is

-6.4 dB.

Phase margin: Raising the magnitude curve by 60 dB yields 0 dB at 9.07 rad/s, where the phase curve is  $200.3^\circ$ . Hence, the phase margin is  $180^\circ - 200.3^\circ = -20.3^\circ$ .

$K = 100$ :

For  $K = 1$ , phase response is  $180^\circ$  at  $\omega = 6.63$  rad/s. Magnitude response is -53.6 dB at this frequency.

For  $K = 100$ , magnitude curve is raised by 40 dB yielding -13.6 dB at 6.63 rad/s. Thus, the gain margin is 13.6 dB.

Phase margin: Raising the magnitude curve by 40 dB yields 0 dB at 2.54 rad/s, where the phase curve is  $107.3^\circ$ . Hence, the phase margin is  $180^\circ - 107.3^\circ = 72.7^\circ$ .

$K = 0.1$ :

For  $K = 1$ , phase response is  $180^\circ$  at  $\omega = 6.63$  rad/s. Magnitude response is -53.6 dB at this frequency.

For  $K = 0.1$ , magnitude curve is lowered by 20 dB yielding -73.6 dB at 6.63 rad/s. Thus, the gain margin is 73.6 dB.

23.

**System 1:** Using non-asymptotic frequency response plots, the zero dB crossing is at 9.7 rad/s at a phase of  $-163.2^\circ$ . Therefore the phase margin is  $180^\circ - 163.2^\circ = 16.8^\circ$ .  $|G(j\omega)|$  is down 7 dB at 14.75 rad/s. Therefore the bandwidth is 14.75 rad/s. Using Eq. (10.73),  $\zeta = 0.15$ . Using Eq. (4.38), %OS = 62.09%. Eq. (10.55) yields  $T_s = 2.76$  s, and Eq. (10.56) yields  $T_p = 0.329$  s.

27.

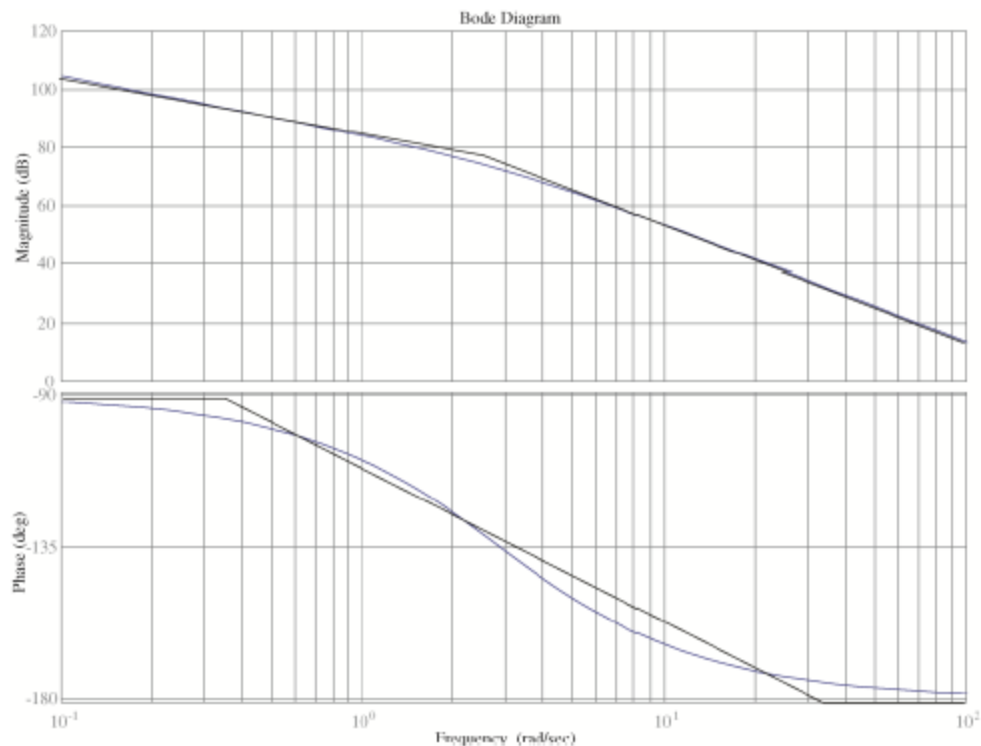
The phase margin of the given system is  $20^\circ$ . Using Eq. (10.73),  $\zeta = 0.176$ . Eq. (4.38) yields 57% overshoot. The system is Type 1 since the initial slope is -20 dB/dec. Continuing the initial slope down to the 0 dB line yields  $K_v = 4$ . Thus, steady-state error for a unit step input is zero; steady state error for a unit ramp input is  $\frac{1}{K_v} = 0.25$ ; steady-state error for a parabolic input is infinite.

36.

The exact Bode plot and the asymptotic approximations are shown on the following figure. The magnitude asymptotes are obtained by noting that when  $\omega \rightarrow 0$ ,  $P(s) \approx \frac{16782}{s}$ . So when

$\omega = 0.1$ ,  $|P(j0.1)| = 167820 = 104.5\text{db}$  and the slope of the line is  $-6\text{db/oct}$ . This means that  $|P(j0.2)| = 104.5\text{db} - 6\text{db} = 98.5\text{db}$ . So a line is drawn between these two points until  $\omega = 2.89$  is reached. For higher frequencies the slope is  $-12\text{db/oct}$ , so the line is continued.

To plot the phase asymptote at very low frequencies the phase is  $-180^\circ$  due to the integrator until  $2.89/10=0.289\text{rad/sec}$ . At very high frequencies from  $2.89*10=28.9\text{ rad/sec}$  and up the phase will be  $-270^\circ$  due to the plant's pole contribution. A line is drawn between  $0.289$  and  $28.9\text{ rad/sec}$  with  $-135^\circ$  at  $2.89\text{ rad/sec}$ .



# Problem 40

The Bode plot is shown below. The phase response is  $180^\circ$  at  $\omega = 3.87$  rad/s, where the gain is  $-6.59$  dB. Thus, the gain margin is  $6.59$  dB. Unity gain is at  $\omega = 2.52$  rad/s, where the phase is  $-76^\circ$  and at  $\omega = 3.12$  rad/s, where the phase is  $-133.1^\circ$ . Hence the phase margin is measured at  $\omega = 3.12$  rad/s and is  $180^\circ - 133.1^\circ = 46.9^\circ$ .

The margin data displayed in the command window after the M-file was run was virtually identical:

Gm(dB)	Pm(deg.)	180 deg. freq.(r/s)	0 dB freq. (r/s)
6.589 dB	46.87 <sup>o</sup>	3.874 r/s	3.121 r/s

