

## Lecture 15: The Transmission-Line Equations for Multiconductor Lines

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## Multiconductor transmission-line (MTL)

A multiconductor transmission line (MTL) is a wiring structure composed of a set of  $(n+1)$  parallel conductor line. It consists of  $n$  conductors and a reference conductor (denoted as the zeroth conductor) to which the  $n$  line voltages will be referenced. This choice of the reference conductor is not unique.

Example ( $N = 6$ ):



## MTL equation derivation from the integral form of Maxwell's equation



Name	Differential form	Integral form
Gauss's law	$\nabla \cdot \mathbf{E} = \frac{\rho}{\epsilon_0}$	$\oiint_{\partial V} \mathbf{E} \cdot d\mathbf{A} = \frac{Q(V)}{\epsilon_0}$
Gauss's law for magnetism	$\nabla \cdot \mathbf{B} = 0$	$\oiint_{\partial V} \mathbf{B} \cdot d\mathbf{A} = 0$
Maxwell-Faraday equation (Faraday's law of induction)	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	$\oint_{\partial S} \mathbf{E} \cdot d\mathbf{l} = -\frac{\partial \Phi_{B,S}}{\partial t}$
Ampère's circuital law (with Maxwell's correction)	$\nabla \times \mathbf{B} = \mu_0 \mathbf{J} + \mu_0 \epsilon_0 \frac{\partial \mathbf{E}}{\partial t}$	$\oint_{\partial S} \mathbf{B} \cdot d\mathbf{l} = \mu_0 I_S + \mu_0 \epsilon_0 \frac{\partial \Phi_{E,S}}{\partial t}$

- **Gauss's law:** [Electric chctic field](#). The [electric flux](#) across a closed surface is proportion
- **Gauss's law for magnetism:** There are no [magnetic monopoles](#). The [magnetic flux](#) across a closed surface is zero.
- **Faraday's law:** Time-varying [magnetic fields](#) produce an electric field.
- **Ampère's law:** Steady currents and time-varying electric fields (the latter due to Maxwell's correction) produce a magnetic field.



## Derivation of the 1<sup>st</sup> MTL equation

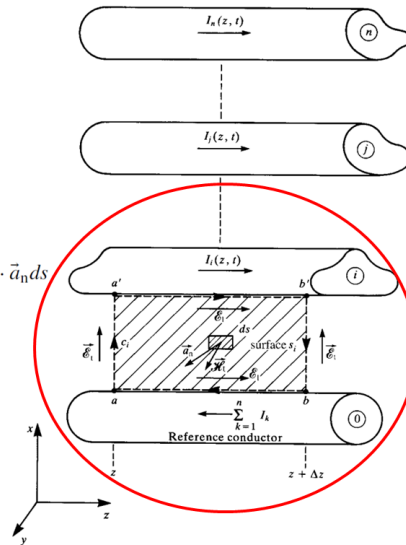
$$\oint_c \vec{\mathcal{E}} \cdot d\vec{l} = -\mu \frac{d}{dt} \int_s \vec{\mathcal{H}} \cdot d\vec{s}$$

Applying **Faraday's law** to the contour  $c_i$  that encloses surface  $s_i$  shown between the reference conductor and the  $i$ th conductor gives

$$\int_a^{a'} \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_{a'}^{b'} \vec{\mathcal{E}}_1 \cdot d\vec{l} + \int_{b'}^b \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_b^a \vec{\mathcal{E}}_1 \cdot d\vec{l} = \mu \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds$$

$\vec{\mathcal{E}}_t$  - the transverse electric field (in the  $x$ - $y$  cross-sectional plane)

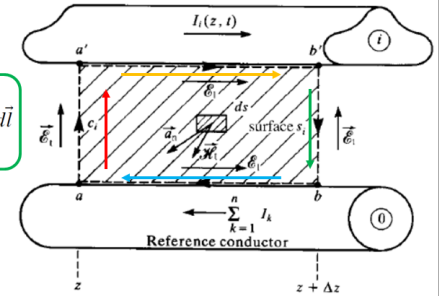
$\vec{\mathcal{E}}_1$  - the longitudinal or  $z$ -directed electric field (along the surfaces of the conductors)



## Derivation of the 1<sup>st</sup> MTL equation

$$\int_a^{a'} \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_{a'}^{b'} \vec{\mathcal{E}}_1 \cdot d\vec{l} + \int_{b'}^b \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_b^a \vec{\mathcal{E}}_1 \cdot d\vec{l} = \mu \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds$$

$$\begin{cases} V_i(z, t) = - \int_a^{a'} \vec{\mathcal{E}}_t(x, y, z, t) \cdot d\vec{l} \\ V_i(z + \Delta z, t) = - \int_b^b \vec{\mathcal{E}}_t(x, y, z + \Delta z, t) \cdot d\vec{l} \\ - \int_{a'}^{b'} \vec{\mathcal{E}}_1 \cdot d\vec{l} = -r_i \Delta z I_i(z, t) \\ - \int_b^a \vec{\mathcal{E}}_1 \cdot d\vec{l} = -r_0 \Delta z \sum_{k=1}^n I_k(z, t) \end{cases}$$



## Derivation of the 1<sup>st</sup> MTL equation

$$\int_a^{a'} \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_{a'}^{b'} \vec{\mathcal{E}}_1 \cdot d\vec{l} - \int_{b'}^b \vec{\mathcal{E}}_t \cdot d\vec{l} + \int_b^a \vec{\mathcal{E}}_1 \cdot d\vec{l} = \mu \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds$$

$$-V_i(z, t) + r_i \Delta z I_i(z, t) + V_i(z + \Delta z, t) + r_0 \Delta z \sum_{k=1}^n I_k(z, t) = \mu \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds$$

Dividing both sides by  $\Delta z$  and rearranging gives

$$\begin{aligned} \frac{V_i(z + \Delta z, t) - V_i(z, t)}{\Delta z} &= -r_0 I_1 - r_0 I_2 - \dots - (r_0 + r_i) I_i - \dots - r_0 I_n \\ &\quad + \mu \frac{1}{\Delta z} \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds \end{aligned}$$

## Derivation of the 1<sup>st</sup> MTL equation

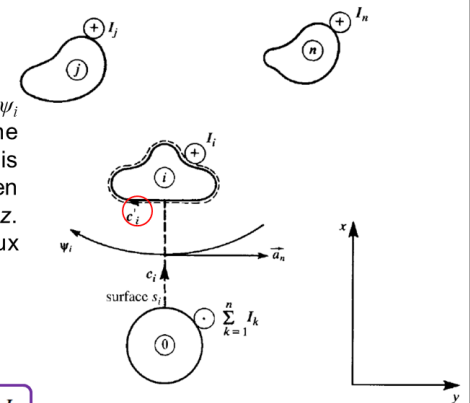
Applying **Ampere's law** to the contour  $c'_i$ , the current of the  $i$ th conductor is

$$I_i(z, t) = \oint_{c'_i} \vec{\mathcal{H}}_t \cdot d\vec{l}'$$

The per-unit-length magnetic flux  $\psi_i$  penetrating the surface  $s_i$  between the reference conductor and the  $i$ th conductor is defined to be in this clockwise direction when looking in the direction of increasing  $z$ . Therefore, this per-unit-length magnetic flux penetrating surface  $s_i$  can be written as

$$\begin{aligned} \psi_i &= -\mu \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds \\ &= l_{i1} I_1 + l_{i2} I_2 + \dots + l_{ii} I_i + \dots + l_{in} I_n \end{aligned}$$

The total magnetic flux is a linear combination of the fluxes due to the currents on all the conductors



cross-sectional view of the line looking in the direction of increasing  $z$

## Derivation of the 1<sup>st</sup> MTL equation

$$\left\{ \begin{aligned} \frac{V_i(z + \Delta z, t) - V_i(z, t)}{\Delta z} &= -r_0 I_1 - r_0 I_2 - \dots - (r_0 + r_i) I_i - \dots - r_0 I_n \\ &\quad + \mu \frac{1}{\Delta z} \frac{d}{dt} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds \\ \psi_i &= -\mu \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \int_{s_i} \vec{\mathcal{H}}_t \cdot \vec{a}_n ds \\ &= l_{i1} I_1 + l_{i2} I_2 + \dots + l_{ii} I_i + \dots + l_{in} I_n \end{aligned} \right.$$

$$\boxed{\Delta z \rightarrow 0} \quad \frac{\partial V_i(z, t)}{\partial z} = -r_0 I_1(z, t) - r_0 I_2(z, t) - \dots - (r_0 + r_i) I_i(z, t) - \dots - r_0 I_n(z, t) - l_{i1} \frac{\partial I_1(z, t)}{\partial t} - l_{i2} \frac{\partial I_2(z, t)}{\partial t} - \dots - l_{ii} \frac{\partial I_i(z, t)}{\partial t} - \dots - l_{in} \frac{\partial I_n(z, t)}{\partial t}$$

This **first MTL equation** can be written in a compact form using matrix notation as

$$\boxed{\frac{\partial}{\partial z} \mathbf{V}(z, t) = -\mathbf{R} \mathbf{I}(z, t) - \mathbf{L} \frac{\partial}{\partial t} \mathbf{I}(z, t)}$$



## Derivation of the 1<sup>st</sup> MTL equation

The first MTL equation can be written in a compact form using matrix notation as

$$\boxed{\frac{\partial}{\partial z} \mathbf{V}(z, t) = -\mathbf{R} \mathbf{I}(z, t) - \mathbf{L} \frac{\partial}{\partial t} \mathbf{I}(z, t)}$$

where the  $n \times 1$  voltage and current vectors are defined as

$$\mathbf{V}(z, t) = \begin{bmatrix} V_1(z, t) \\ \vdots \\ V_i(z, t) \\ \vdots \\ V_n(z, t) \end{bmatrix} \quad \mathbf{I}(z, t) = \begin{bmatrix} I_1(z, t) \\ \vdots \\ I_i(z, t) \\ \vdots \\ I_n(z, t) \end{bmatrix}$$

The **per-unit-length inductance matrix**  $\mathbf{L}$  contains the individual per-unit-length self-inductances,  $l_{ii}$ , of the circuits and the per-unit-length mutual inductances between the circuits,  $l_{ij}$ .  $\mathbf{L}$  can be shown to be a symmetric matrix. Similarly, we define the **per-unit-length resistance matrix** as  $\mathbf{R}$ .

□

$$\mathbf{L} = \begin{bmatrix} l_{11} & l_{12} & \dots & l_{1n} \\ l_{12} & l_{22} & \dots & l_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ l_{1n} & l_{2n} & \dots & l_{nn} \end{bmatrix} \quad \mathbf{R} = \begin{bmatrix} (r_1 + r_0) & r_0 & \dots & r_0 \\ r_0 & (r_2 + r_0) & \dots & r_0 \\ \vdots & \vdots & \ddots & \vdots \\ r_0 & r_0 & \dots & (r_n + r_0) \end{bmatrix}$$

$$= \begin{bmatrix} r_1 & 0 & \dots & 0 \\ 0 & r_2 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & r_n \end{bmatrix} + \begin{bmatrix} r_0 & r_0 & \dots & r_0 \\ r_0 & r_0 & \dots & r_0 \\ \vdots & \vdots & \ddots & \vdots \\ r_0 & r_0 & \dots & r_0 \end{bmatrix}$$



## Derivation of the 2<sup>nd</sup> MTL equation

To place a closed surface  $s'$  around the  $i$ th conductor. The portion of the surface over the end caps is denoted as  $s'_e$ , whereas the portion over the sides is denoted as  $s'_s$ .

$$\oint_{s'} \vec{\mathcal{J}} \cdot d\vec{s}' = -\frac{d}{dt} Q_{\text{enc}} \quad \text{continuity equation or equation of conservation of charge}$$

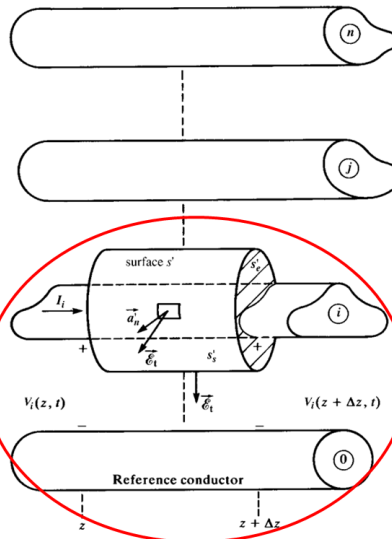
Over the end caps, we have

$$\iint_{s'_e} \vec{\mathcal{J}} \cdot d\vec{s}' = I_i(z + \Delta z, t) - I_i(z, t)$$

Over the sides of the surface,  $\frac{I_i(z, t)}{\text{transverse current}} = \frac{I_g(z, t)}{\text{conduction current}} + \frac{I_c(z, t)}{\text{displacement current}}$

• **conduction current**  $\vec{\mathcal{J}}_c = \sigma \vec{\mathcal{E}}_t$

• **displacement current**  $\vec{\mathcal{J}}_d = \epsilon \frac{\partial \vec{\mathcal{E}}_t}{\partial t}$

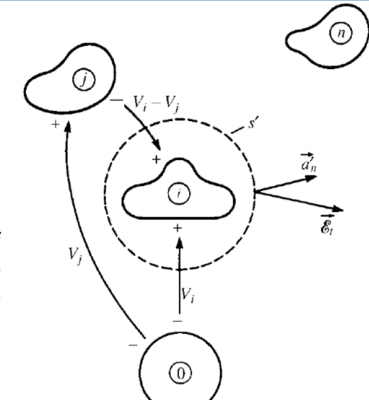


## Derivation of the 2<sup>nd</sup> MTL equation

• **conduction current**  $\vec{\mathcal{J}}_c = \sigma \vec{\mathcal{E}}_t$

$$\iint_{s'_s} \vec{\mathcal{J}}_c \cdot d\vec{s}' = \sigma \iint_{s'_s} \vec{\mathcal{E}}_t \cdot d\vec{s}'$$

Defining **per-unit-length conductance**  $g_{ij}$  S/m between each pair of conductors as the ratio of conduction current flowing between the two conductors in the transverse plane to the voltage between the two conductors.



$$\sigma \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \iint_{s'_s} \vec{\mathcal{E}}_t \cdot d\vec{s}' = g_{i1} (V_i - V_1) + \dots + g_{ii} V_i + \dots + g_{in} (V_i - V_n)$$

$$= -g_{i1} V_1(z, t) - g_{i2} V_2(z, t) - \dots + \sum_{k=1}^n g_{ik} V_k(z, t) - \dots - g_{in} V_n(z, t)$$



## Derivation of the 2<sup>nd</sup> MTL equation

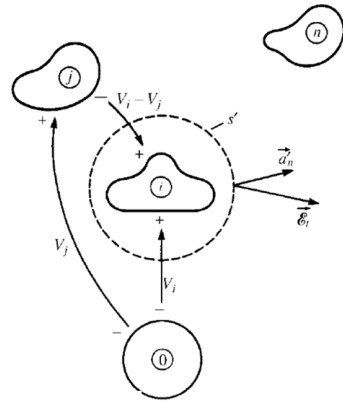
- displacement current  $\vec{J}_d = \varepsilon \frac{\partial \vec{E}_t}{\partial t}$

The charge enclosed by the surface (residing on the conductor surface) is, by Gauss' law,

$$Q_{enc} = \varepsilon \iint_{s'_s} \vec{E}_t \cdot d\vec{s}'$$

The charge per unit of line length can be defined in terms of the per-unit-length capacitances  $c_{ij}$  between each pair of conductors.

$$\begin{aligned} \varepsilon \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}' &= c_{i1}(V_i - V_1) + \dots + c_{in}V_i + \dots + c_{in}(V_i - V_n) \\ &= -c_{i1}V_1(z, t) - c_{i2}V_2(z, t) - \dots + \sum_{k=1}^n c_{ik}V_i(z, t) \\ &\quad - \dots - c_{in}V_n(z, t) \end{aligned}$$



## Derivation of the 2<sup>nd</sup> MTL equation

$$\oiint_{s'} \vec{J} \cdot d\vec{s}' = -\frac{d}{dt} Q_{enc}$$

$$\iint_{s'_c} \vec{J} \cdot d\vec{s}' = I_i(z + \Delta z, t) - I_i(z, t)$$

$$\iint_{s'_s} \vec{J}_c \cdot d\vec{s}' = \sigma \iint_{s'_s} \vec{E}_t \cdot d\vec{s}'$$

$$Q_{enc} = \varepsilon \iint_{s'_s} \vec{E}_t \cdot d\vec{s}'$$

$$\frac{I_i(z + \Delta z, t) - I_i(z, t)}{\Delta z} + \sigma \frac{1}{\Delta z} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}' = -\varepsilon \frac{1}{\Delta z} \frac{d}{dt} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}'$$

## Derivation of the 2<sup>nd</sup> MTL equation

$$\begin{aligned} \left\{ \begin{aligned} \sigma \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}' &= g_{i1}(V_i - V_1) + \dots + g_{in}V_i + \dots + g_{in}(V_i - V_n) \\ &= -g_{i1}V_1(z, t) - g_{i2}V_2(z, t) - \dots + \sum_{k=1}^n g_{ik}V_i(z, t) \\ &\quad - \dots - g_{in}V_n(z, t) \\ \varepsilon \lim_{\Delta z \rightarrow 0} \frac{1}{\Delta z} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}' &= c_{i1}(V_i - V_1) + \dots + c_{in}V_i + \dots + c_{in}(V_i - V_n) \\ &= -c_{i1}V_1(z, t) - c_{i2}V_2(z, t) - \dots + \sum_{k=1}^n c_{ik}V_i(z, t) \\ &\quad - \dots - c_{in}V_n(z, t) \end{aligned} \right. \end{aligned}$$

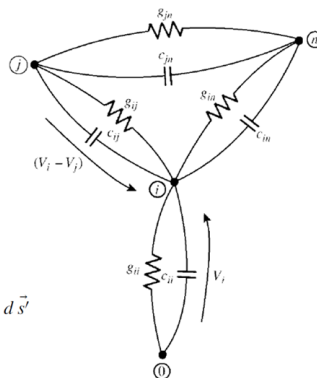
$$\frac{I_i(z + \Delta z, t) - I_i(z, t)}{\Delta z} + \sigma \frac{1}{\Delta z} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}' = -\varepsilon \frac{1}{\Delta z} \frac{d}{dt} \iint_{s'_s} \vec{E}_t \cdot d\vec{s}'$$

$\Delta z \rightarrow 0$

$$\begin{aligned} \frac{\partial I_i(z, t)}{\partial z} &= g_{i1}V_1(z, t) + g_{i2}V_2(z, t) + \dots - \sum_{k=1}^n g_{ik}V_i(z, t) + \dots \\ &\quad + g_{in}V_n(z, t) + c_{i1} \frac{\partial}{\partial t} V_1(z, t) + c_{i2} \frac{\partial}{\partial t} V_2(z, t) + \dots \\ &\quad - \sum_{k=1}^n c_{ik} \frac{\partial}{\partial t} V_i(z, t) + \dots + c_{in} \frac{\partial}{\partial t} V_n(z, t) \end{aligned}$$

This **second MTL equation** can be written in a compact form using matrix notation as

$$\frac{\partial}{\partial z} \mathbf{I}(z, t) = -\mathbf{G} \mathbf{V}(z, t) - \mathbf{C} \frac{\partial}{\partial t} \mathbf{V}(z, t)$$



## Derivation of the 2<sup>nd</sup> MTL equation

The second MTL equation can be written in a compact form using matrix notation as

$$\frac{\partial}{\partial z} \mathbf{I}(z, t) = -\mathbf{G} \mathbf{V}(z, t) - \mathbf{C} \frac{\partial}{\partial t} \mathbf{V}(z, t)$$

where the  $n \times 1$  voltage and current vectors are defined as

$$\mathbf{V}(z, t) = \begin{bmatrix} V_1(z, t) \\ \vdots \\ V_i(z, t) \\ \vdots \\ V_n(z, t) \end{bmatrix} \quad \mathbf{I}(z, t) = \begin{bmatrix} I_1(z, t) \\ \vdots \\ I_i(z, t) \\ \vdots \\ I_n(z, t) \end{bmatrix}$$

The per-unit-length conductance matrix  $\mathbf{G}$  represents the conduction current flowing between the conductors in the transverse plane and is defined as

$$\mathbf{G} = \begin{bmatrix} \sum_{k=1}^n g_{1k} & -g_{12} & \dots & -g_{1n} \\ -g_{12} & \sum_{k=1}^n g_{2k} & \dots & -g_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ -g_{1n} & -g_{2n} & \dots & \sum_{k=1}^n g_{nk} \end{bmatrix}$$

The per-unit-length capacitance matrix  $\mathbf{C}$  represents the displacement current flowing between the conductors in the transverse plane and is defined as

$$\mathbf{C} = \begin{bmatrix} \sum_{k=1}^n c_{1k} & -c_{12} & \dots & -c_{1n} \\ -c_{12} & \sum_{k=1}^n c_{2k} & \dots & -c_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ -c_{1n} & -c_{2n} & \dots & \sum_{k=1}^n c_{nk} \end{bmatrix}$$

## Derivation of the MTL equation and its simplify

$$\frac{\partial}{\partial z} \mathbf{V}(z, t) = -\mathbf{R}\mathbf{I}(z, t) - \mathbf{L} \frac{\partial}{\partial t} \mathbf{I}(z, t)$$

$$\frac{\partial}{\partial z} \mathbf{I}(z, t) = -\mathbf{G}\mathbf{V}(z, t) - \mathbf{C} \frac{\partial}{\partial t} \mathbf{V}(z, t)$$

If the conductors are perfect conductors, then  $\mathbf{R} = \mathbf{0}$ , whereas if the surrounding medium is lossless ( $\sigma = 0$ ), then  $\mathbf{G} = \mathbf{0}$ . The line is said to be **lossless** if both the conductors and the medium are lossless in which case the MTL equations simplify to

$$\frac{\partial}{\partial z} \begin{bmatrix} \mathbf{V}(z, t) \\ \mathbf{I}(z, t) \end{bmatrix} = - \begin{bmatrix} \mathbf{0} & \mathbf{L} \\ \mathbf{C} & \mathbf{0} \end{bmatrix} \frac{\partial}{\partial t} \begin{bmatrix} \mathbf{V}(z, t) \\ \mathbf{I}(z, t) \end{bmatrix}$$



## Derivation of the uncoupled MTL equation

$$\frac{\partial}{\partial z} \mathbf{V}(z, t) = -\mathbf{R}\mathbf{I}(z, t) - \mathbf{L} \frac{\partial}{\partial t} \mathbf{I}(z, t) \quad (a)$$

$$\frac{\partial}{\partial z} \mathbf{I}(z, t) = -\mathbf{G}\mathbf{V}(z, t) - \mathbf{C} \frac{\partial}{\partial t} \mathbf{V}(z, t) \quad (b)$$

The first-order, coupled forms can be placed in the form of **second-order, uncoupled equations** by differentiating (a) with respect to  $z$  and differentiating (b) with respect to  $t$  to yield

The uncoupled, second-order equations

$$\begin{cases} \frac{\partial^2}{\partial z^2} \mathbf{V}(z, t) = -\mathbf{R} \frac{\partial}{\partial z} \mathbf{I}(z, t) - \mathbf{L} \frac{\partial^2}{\partial z \partial t} \mathbf{I}(z, t) \\ \frac{\partial^2}{\partial z \partial t} \mathbf{I}(z, t) = -\mathbf{G} \frac{\partial}{\partial t} \mathbf{V}(z, t) - \mathbf{C} \frac{\partial^2}{\partial t^2} \mathbf{V}(z, t) \end{cases} \rightarrow \begin{cases} \frac{\partial^2}{\partial z^2} \mathbf{V}(z, t) = [\mathbf{R}\mathbf{G}] \mathbf{V}(z, t) + [\mathbf{R}\mathbf{C} + \mathbf{L}\mathbf{G}] \frac{\partial}{\partial t} \mathbf{V}(z, t) + \mathbf{L}\mathbf{C} \frac{\partial^2}{\partial t^2} \mathbf{V}(z, t) \\ \frac{\partial^2}{\partial z^2} \mathbf{I}(z, t) = [\mathbf{G}\mathbf{R}] \mathbf{I}(z, t) + [\mathbf{C}\mathbf{R} + \mathbf{G}\mathbf{L}] \frac{\partial}{\partial t} \mathbf{I}(z, t) + \mathbf{C}\mathbf{L} \frac{\partial^2}{\partial t^2} \mathbf{I}(z, t) \end{cases}$$



## MTL equation Derivation from the per-unit-length equivalent circuit

## Derivation of the 1<sup>st</sup> MTL equation

As an alternative method, we derive the MTL equations from the per-unit-length equivalent circuit shown in right figure.

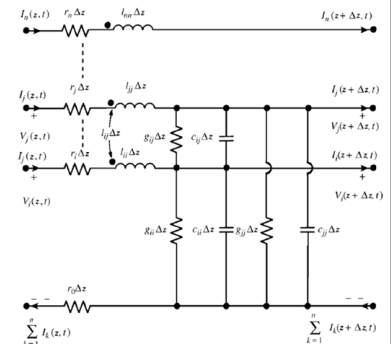
Writing Kirchhoff's voltage law around the  $i$ th circuit consisting of the  $i$ th conductor and the reference conductor yields.

$$-V_i(z, t) + r_i \Delta z I_i(z, t) + V_i(z + \Delta z, t) + r_0 \Delta z \sum_{k=1}^n I_k(z, t) = -l_{i1} \Delta z \frac{\partial I_1(z, t)}{\partial t}$$

$$-l_{i2} \Delta z \frac{\partial I_2(z, t)}{\partial t} - \dots - l_{in} \Delta z \frac{\partial I_n(z, t)}{\partial t}$$

$$\Delta z \rightarrow 0$$

$$\begin{aligned} \frac{\partial V_i(z, t)}{\partial z} &= -r_0 I_1(z, t) - r_0 I_2(z, t) - \dots - (r_0 + r_i) I_i(z, t) - \dots \\ &\quad - r_0 I_n(z, t) - l_{i1} \frac{\partial I_1(z, t)}{\partial t} - l_{i2} \frac{\partial I_2(z, t)}{\partial t} - \dots - l_{in} \frac{\partial I_n(z, t)}{\partial t} - \dots \\ &\quad - l_{in} \frac{\partial I_n(z, t)}{\partial t} \end{aligned}$$



$$\frac{\partial}{\partial z} \mathbf{V}(z, t) = -\mathbf{R}\mathbf{I}(z, t) - \mathbf{L} \frac{\partial}{\partial t} \mathbf{I}(z, t)$$



## Derivation of the 2<sup>nd</sup> MTL equation

Similarly, the second MTL equation can be obtained by applying Kirchhoff's current law to the  $i$ th conductor in the per-unit-length equivalent circuit.

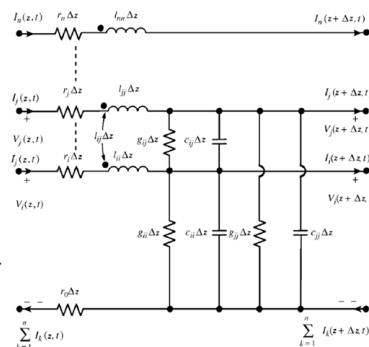
$$I_i(z + \Delta z, t) - I_i(z, t) = -g_{i1}\Delta z(V_i - V_1) - \dots - g_{in}\Delta z(V_i - V_n) - g_{im}\Delta z(V_i - V_n) - c_{i1}\Delta z \frac{\partial}{\partial t}(V_i - V_1) - \dots - c_{in}\Delta z \frac{\partial}{\partial t}(V_i - V_n) - c_{im}\Delta z \frac{\partial}{\partial t}(V_i - V_n)$$

$\Delta z \rightarrow 0$

$$\frac{\partial I_i(z, t)}{\partial z} = g_{i1}V_1(z, t) + g_{i2}V_2(z, t) + \dots - \sum_{k=1}^n g_{ik}V_k(z, t) + \dots + g_{in}V_n(z, t) + c_{i1} \frac{\partial}{\partial t} V_1(z, t) + c_{i2} \frac{\partial}{\partial t} V_2(z, t) + \dots - \sum_{k=1}^n c_{ik} \frac{\partial}{\partial t} V_k(z, t) + \dots + c_{in} \frac{\partial}{\partial t} V_n(z, t)$$



$$\frac{\partial}{\partial z} \mathbf{I}(z, t) = -\mathbf{G} \mathbf{V}(z, t) - \mathbf{C} \frac{\partial}{\partial t} \mathbf{V}(z, t)$$



Thank you!