

1                   Simulation of the time resolution of a 50  $\mu\text{m}$  low-gain  
2                   avalanche detector.

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7    **Abstract**

In this paper we report simulation results on the timing resolution of a 50  $\mu\text{m}$  low-gain avalanche detector (LGAD). The simulation includes: sensor fluctuations, front-end electronics, and quantization. Comparisons on the performance for different front-end electronics (FEE) bandwidths (BW) are presented, as well as the dependance on singal-to-noise ratio (SNR). Two approaches to measure the timestamp are presented: leading edge (LE) and constant-fraction-discrimination (CFD). Additionally, the time resolution is studied as function of the irradiation of the sensor. Simulated LGAD pulses before irradiation, and after neutron fluences of  $5 \times 10^{14} \text{ n/cm}^2$  and  $1 \times 10^{15} \text{ n/cm}^2$ , are studied. The time resolution a 50  $\mu\text{m}$  LGADs was found to be 30 ps for FE electronics BWs larger than 350 MHz and SNRs larger than 30. The time resolution at a SNR of 30 for fluences of  $5 \times 10^{14} \text{ n/cm}^2$  and  $1 \times 10^{15} \text{ n/cm}^2$  were found to be 30 ps and 40 ps, respectively.

8    *Key words:*

9    Silicon, Timing, LGAD

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10   **Contents**

11	<b>1 Introduction</b>	<b>2</b>
12	<b>2 Simulation Framework</b>	<b>3</b>
13	2.1 LGAD pulse library and simulation . . . . .	3
14	2.2 Fron-end Electronics simulation and noise injection . . . . .	4
15	2.2.1 front-end implementation . . . . .	4
16	2.2.2 noise injection . . . . .	4
17	<b>3 Timing Reconstruction and Analysis</b>	<b>5</b>
18	3.1 Leading edge and constant fraction discriminators . . . . .	5
19	3.2 Time-walk correction and time over threshold . . . . .	5

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Preprint submitted to Nucl. Instrum. Meth. A

February 28, 2019

20	<b>4 LGAD Front-end Electronics Performance</b>	<b>5</b>
21	4.1 Front-end electronics shaping time studies . . . . .	5
22	4.2 Timing Performace as a function of signal-to-noise ratio . . . . .	5
23	4.3 Timing Performace as a function of irradiation . . . . .	5
24	<b>5 Conclusion</b>	<b>5</b>

## 25 1. Introduction

26 Future colliders, including the high luminosity upgrade of the Large Hadron Collider  
27 (HL-LHC) at CERN, will operate with an order of magnitude higher instantaneous lu-  
28 minosity compared to what has been achieved at the large hadron collider (LHC) so far.  
29 With the increased instantaneous luminosity, the rate of simultaneous interactions per  
30 bunch crossing (pileup) is projected to reach an average of 140 to 200. Pileup increases  
31 the difficulties in separating particles from the hard scattering interaction with those pro-  
32 duced in different pileup interactions. In particular, the ability to discriminate between  
33 jets produced in the events of interests, especially those associated with vector boson fu-  
34 sion processes, and jets produced by pileup interactions will be degraded. Additionally,  
35 the efficiency to identify high  $p_T$  isolated electrons and muons will be severely reduced  
36 due to the high density of pileup particles in their vicinity. The missing transverse energy  
37 resolution will also deteriorate, and several other physics objects performance metrics will  
38 suffer the detrimental effects of pileup.

39 One way to mitigate the pileup effects mentioned above, complementary to precision  
40 tracking methods, is to perform a time of arrival measurement associated with each par-  
41 ticle. Such a measurement with a precision of about 30-40 ps, will reduce the effective  
42 amount of pileup by a factor of 10, given that the spread in collision time of the pileup  
43 interactions at HL-LHC is foreseen to be approximately 200 ps. It has been previously  
44 shown that a precision of better than 20 ps can be achieved for electromagnetic showers  
45 measured with silicon sampling calorimeters [1–3] using traditional planar silicon de-  
46 tectors while precision of 30 ps can be achived for minimum ionizing particles (MIPs)  
47 measured with low-gain avalanches detectors (LGADs) [4–6].

48 LGADs are envisioned to be used in the CMS and ATLAS experiment upgrades for  
49 HL-LHC in order to overcome the event reconstruction challenges posed by the high rate  
50 of concurrent collisions per beam crossing. The implemented regions of pseudorapidity  
51 ( $\eta$ ) are:  $|\eta| > 1.5$ , and  $2.4 < |\eta| < 4.2$  for CMS and ATLAS, respectively. In order  
52 to achieve the desired timing precision across a large area of the detectors, the sensors  
53 will need to provide high uniformity of signal response and timing resolution. Beam test  
54 measurements have provided encouraging results towards achieving such detectors [4].

55 In this paper, we report simulation results on the timing resolution of a 50  $\mu\text{m}$  low-  
56 gain avalanche detector (LGAD) which includes the effects of the sensor fluctuations,  
57 front-end electronics (FEE), and quantization. Our results indicate that for FEE analog  
58 bandwidths (BW) larger than 350 MHz and signal-to-noise ratios (SNRs) larger than  
59 30, measured at the output of the FFE, time resolutions of 30 ps and 40 ps are obtained  
60 when using time-walk corrections based on time-over-threshold (ToT) measurements to  
61 both timestamping techniques: constant-fraction-discrimination (CFD) and leading-edge  
62 (LE), respectively. These results are compatible with previous measurements on LGAD  
63 timing resolutions carried out in laboratory and beam test conditions [4–6]. We study

the time resolution for four different FEE shaping times: 0.5 ps, 1.0 ps, 2.0 ps, and 4.0 ps; three SNR: 20, 30, 100; and three irradiation levels: pre-radiation,  $5 \times 10^{14}$  n/cm<sup>2</sup>, and  $1 \times 10^{15}$  n/cm<sup>2</sup>. For every point in this scan we evaluate the time resolution for LE and CFD.

The paper is organized as follows: the simulation is described in Sec. 2; algorithms used in the timing reconstruction and analysis are described in Sec. 3; simulation results are presented in Sec. 4, followed by the conclusion in Sec. 5.

## 2. Simulation Framework

The simulation framework is based on c++ programming language. The LGAD pulses are obtained from Weightfield2 (WF2), a 2-dimensional silicon simulator [?]. WF2 provides sets of 1000 LGAD pulses which models the response of the sensor to minimum ionizing particles (MIPs). We generated 3 sets of LGAD pulses for a 50  $\mu$ m LGAD: pre-irradiation, and after neutron fluences of  $5 \times 10^{14}$  n/cm<sup>2</sup> and  $1 \times 10^{15}$  n/cm<sup>2</sup>. The simulation framework takes the LGAD pulses (from WF2) and adds gaussian white noise (hereafter white noise). At this point the LGAD pulses with the added white noise are fed into the simulation of the FEE (see Fig. 1). The output of the FEE simulation is the convolution of the impulse response function and the input signal at the FEE. We consider four shaping constants for the impulse response of the FEE: 0.5, 1.0, 2.0, and 4.0 ns (the FEE simulation will be described in detail in Sec. 2.2). At the output of the FEE block we have a "realistic" LGAD pulse which includes the effects of sensor fluctuations, shaping of the FEE, and noise. A waveform analysis is performed with the pulses obtained at the output of the FEE block. We assign timestamps to each pulse by using algorithms that emulate an ideal LE discriminator and an ideal CFD. For each threshold we obtain a LE and CFD timestamps as well as the corresponding time-over-threshold (ToT) of the pulse. The SNR is defined as the ratio of the most probable value (MPV) of the amplitude distribution to the width of the amplitude distribution at a fixed sample (where only noise is present). We study 3 SNR scenarios: 20, 30, and 100. A schematic diagram of the simulation is shown in Fig. 1.

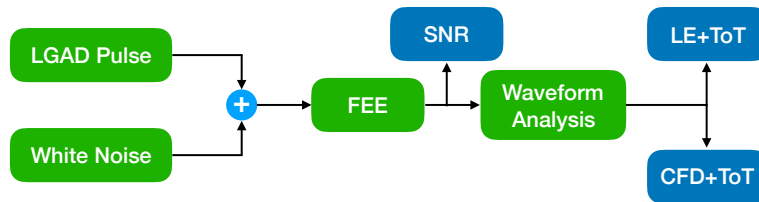


Figure 1: A schematic diagram of the simulation. Each simulation configurable block is shown in green. The most relevant outputs of the simulation are shown in blue.

### 2.1. LGAD pulse library and simulation

We need to ask Nicolo to send us a paragraph for the Weightfield2 (WF2)

## 2.2. Fron-end Electronics simulation and noise injection

The front-end simulation is implemented in c++ programing language. It combines analytical calculations when possible but it mostly relies on numerical methods. We implement most calculations in the time domain, while the frequency domain is mostly used to cross-check noise and the FEE expected response. Sections 2.2.1 and 2.2.2 detail the front-end and noise implementation in the simulation.

### 2.2.1. front-end implementation

The fron-end simulation is based on a single amplification stage. We focus on the BW of such amplifier rather than variations thereof. The fron-end chose is a second order low-pass filter which transfer function and impulse response are given by equations 1 and 2, respectively.

$$H(S) = \frac{\frac{1}{\tau_s^2}}{(S + 1/\tau_s)^2} \quad (1) \quad h(t) = \frac{t}{\tau_s^2} e^{-t/\tau_s} \quad (2)$$

The output pulse of the FEE is the convolution (in time domain) of the pulse from the LGAD library and the FEE impulse response (see Eq. 2). The time base for the pulses and the convolution is 1 ps – this is the sampling time that we use for the whole signal simulation, noise is treated differently and covered in Sec. 2.2.2. As stated above we focus the study on the BW of the FEE, to that end we scan the  $\tau_s$  paremeter in Eq. 2 in the following set: {0.5, 1, 2, 4} ns, this parameter is hereafter referred to as shaping time (ST). Figure 2 (left) shows the comparison of the impulse and LGAD responses for a ST of 1 ns while Figure 2 (right) shows the LGAD response for all STs studied. We observe that the LGAD response is delayed with respect to the impulse response, and that pulse slew rate is decreased in the first nanosecond of the pulse. We also observe the expected behavior when comparing the LGAD responses for the different STs, pulse risetimes scale with the ST and the decay time is dominated by the ST.

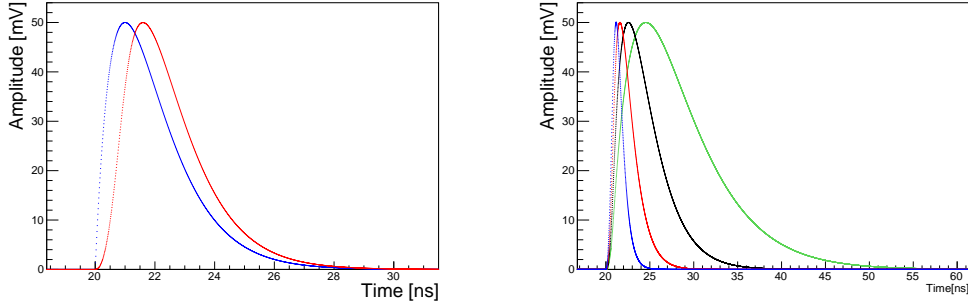


Figure 2: (Left) Comparison of impulse and LGAD reponses for the a shaping time (ST) of 1 ns. (Right) LGAD response for the four shaping times studied: {0.5, 1, 2, 4} ns. All pulses have been normalized to achive a peak amplitude of 50 mV. Legends for the shaping times are shown in the plots.

### 2.2.2. noise injection

blah blah blah

Shaping time (ns)	0.5	1.0	2.0	4.0
Risetime (ns)	$0.7 \pm xx$	$0.9 \pm xx$	$1.4 \pm xx$	$2.5 \pm xx$

Table 1: Measured risetime for all shaping times studied:  $\{0.5, 1, 2, 4\}$  ns. Risetime is the 10% – 90% time difference as measured by the CFD algorithm described in Sec. 3.1.

### 3. Timing Reconstruction and Analysis

*3.1. Leading edge and constant fraction discriminators*

*3.2. Time-walk correction and time over threshold*

### 4. LGAD Front-end Electronics Performance

We present a number of different studies of the LGAD sensors. such that they are above the noise levels listed for each board in Sec. ?? . All measurements other than those described in Sec. ?? and 4.3 were performed at room temperature.

*4.1. Front-end electronics shaping time studies*

*4.2. Timing Performace as a function of signal-to-noise ratio*

*4.3. Timing Performace as a function of irradiation*

### 5. Conclusion

### Acknowledgment

We thank the FTBF personnel and Fermilab accelerator’s team for very good beam conditions during our test beam time. We also appreciate the technical support of the Fermilab SiDet department for the rapid production of wire-bonded and packaged LGAD assemblies. We would like to thank Alan Prosser and Ryan Rivera for their critical help in setting up the DAQ and trigger chain. We thank Ned Spencer, Max Wilder, and Forest McKinney-Martinez for their technical assistance, and the CNM and HPK manufacturing team. We acknowledge the help of V. Cindro and I. Mandic with the neutron irradiations.

This document was prepared using the resources of the Fermi National Accelerator Laboratory (Fermilab), a U.S. Department of Energy, Office of Science, HEP User Facility. Fermilab is managed by Fermi Research Alliance, LLC (FRA), acting under Contract No. DE-AC02-07CH11359. Part of this work was performed within the framework of the CERN RD50 collaboration.

This work was supported by the Fermilab LDRD 2017.027; by the United States Department of Energy grant DE-FG02-04ER41286; by the California Institute of Technology High Energy Physics under Contract DE-SC0011925; by the European Union’s Horizon 2020 Research and Innovation funding program, under Grant Agreement no. 654168 (AIDA-2020) and Grant Agreement no. 669529 (ERC UFSD669529); by the Italian Ministero degli Affari Esteri and INFN Gruppo V; and by the Spanish Ministry of Economy, Industry and Competitiveness through the Particle Physics National Program (ref. FPA2014-55295-C3-2-R and FPA2015-69260-C3-3-R) co-financed with FEDER funds.

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