

- 1) Two sensors  $S_1$ ,  $S_2$  measuring quantity  $x$ .  
Independent and  $\text{var}(S_1) = \sigma_1^2$ ,  $\text{var}(S_2) = \sigma_2^2$   
Best estimate for  $x$  and variance given  $S_1$ ,  $S_2$ .

Use a linear merge for this where  $x = S_{\text{new}}$ :

$$S_{\text{new}} = K S_1 + (1-K) S_2 \quad \text{where } 0 \leq K \leq 1 \quad (1)$$

$$\text{var}(S_{\text{new}}) = \text{var}(K S_1 + (1-K) S_2)$$

$$= \text{var}(K S_1) + \text{var}((1-K) S_2)$$

$$= K^2 \text{var}(S_1) + (1-K)^2 \text{var}(S_2) + \cancel{2 \text{covar}(K S_1, (1-K) S_2)} \quad \begin{matrix} S_1, S_2 \text{ independent} & 0 \end{matrix}$$

$$= K^2 \sigma_1^2 + (1-K)^2 \sigma_2^2 \quad (2)$$

Also want to minimize the variance of  $S_{\text{new}}$ :

$$\min_K \text{var}(S_{\text{new}}) = \min_K K^2 \sigma_1^2 + (1-K)^2 \sigma_2^2$$

$$\frac{d}{dk} (K^2 \sigma_1^2 + (1-K)^2 \sigma_2^2) = 0$$

$$0 = 2K \sigma_1^2 - 2(1-K) \sigma_2^2$$

$$0 = K \sigma_1^2 - \sigma_2^2 + K \sigma_2^2 \quad K = \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2}$$

We can substitute K back into our equations:

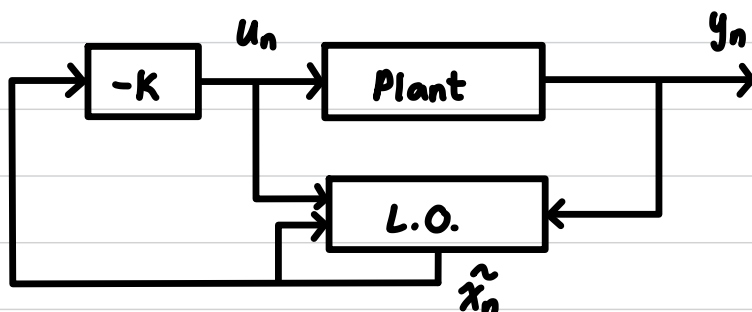
$$\textcircled{1} \quad J_{\text{new}} = \left( \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \right) J_1 + \left( 1 - \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \right) J_2$$

$$J_{\text{new}} = \left( \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \right) J_1 + \left( \frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2} \right) J_2$$

$$\textcircled{2} \quad \text{var}(J_{\text{new}}) = \left( \frac{\sigma_2^2}{\sigma_1^2 + \sigma_2^2} \right)^2 \sigma_1^2 + \left( \frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2} \right)^2 \sigma_2^2$$

$$\text{var}(J_{\text{new}}) = \frac{\sigma_2^4 \sigma_1^2 + \sigma_1^4 \sigma_2^2}{(\sigma_1^2 + \sigma_2^2)^2} = \frac{\sigma_1^2 \sigma_2^2}{\sigma_1^2 + \sigma_2^2}$$

2) Lüneberger Observer for L.T.O. System  
 Show error of observation can be controlled to zero.



$$e_{n+1} = x_{n+1} - \tilde{x}_{n+1} \quad (\text{control this to zero})$$

$$\textcircled{1} \quad x_{n+1} = A_n x_n + B_n u_n \quad y_n = C_n x_n \quad (\text{state})$$

$$\textcircled{2} \quad \tilde{x}_{n+1} = A_n \tilde{x}_n + B_n u_n + L_n (y_n - \tilde{y}_n) \quad (\text{estimation})$$

$\textcircled{1} - \textcircled{2}$

$$e_{n+1} = A_n x_n + B_n u_n - A_n \tilde{x}_n - B_n u_n - L_n (y_n - \tilde{y}_n)$$

$$e_{n+1} = A_n x_n - A_n \tilde{x}_n - L_n (y_n - \tilde{y}_n) \quad y_n = C_n x_n$$

$$e_{n+1} = A_n (x_n - \tilde{x}_n) - L_n C_n (x_n - \tilde{x}_n)$$

$$e_{n+1} = (A_n - L_n C_n)(x_n - \tilde{x}_n)$$

$$e_n = x_n - \tilde{x}_n$$

$$e_{n+1} = (A_n - L_n C_n) e_n$$

To control this error of  $e_{n+1}$  to zero,  $e_n$  must be multiplied by a factor less than 1, for all  $n$ . Doing this will shrink the error to zero as time approaches infinity.

$$\|A_n - L_n C_n\| < 1 \quad \text{for all } n$$

3) Iterative learning scheme to estimate  $y = \lambda_1 x + \lambda_2$  online as samples  $(x_i, y_i)$  arrive. Update process of the form  $\lambda_{n+1} = \lambda_n - u f(x_n, y_n)$ , give  $u$  and  $f$ .

Can use Stochastic Gradient Descent (SGD)

$$Q(w) = \sum_{i=1}^n Q_i(w) = \sum_{i=1}^n (y_i - \tilde{y}_i)^2 = \sum_{i=1}^n (y_i - \lambda_1 x_i - \lambda_2)^2$$

Find partial derivatives of  $\lambda_1$  and  $\lambda_2$ :

$$\frac{d}{d\lambda_1} (y_i - \lambda_1 x_i - \lambda_2)^2 = -2x_i (y_i - \lambda_1 x_i - \lambda_2)$$

$$\frac{d}{d\lambda_2} (y_i - \lambda_1 x_i - \lambda_2)^2 = -2(y_i - \lambda_1 x_i - \lambda_2)$$

Update functions:

$$\lambda_{n+1} = \lambda_n - u \begin{pmatrix} -2x_i (y_i - \lambda_1 x_i - \lambda_2) \\ -2(y_i - \lambda_1 x_i - \lambda_2) \end{pmatrix}$$

$$f_1(x_i, y_i) = -2x_i(y_i - \lambda_1 x_i - \lambda_2)$$

$$f_2(x_i, y_i) = -2(y_i - \lambda_1 x_i - \lambda_2)$$

$\mu$ : learning rate, hyperparameter

needs to be properly configured

too large may cause divergent behaviours

too small may never reach the minimum point

$$\begin{pmatrix} \lambda_1 \\ \lambda_2 \end{pmatrix}_{n+1} = \begin{pmatrix} \lambda_1 \\ \lambda_2 \end{pmatrix}_n - \mu \begin{pmatrix} -2x_i(y_i - \lambda_1 x_i - \lambda_2) \\ -2(y_i - \lambda_1 x_i - \lambda_2) \end{pmatrix}$$

4) Discrete System ABC and state  $x^*$ ,  $x(0)=0$   
 Find a control input  $u(0), u(1), \dots, u(N-1)$  such  
 that  $x(N) = x^*$  (set up matrix and determine solution)

$$x_{n+1} = Ax_n + Bu_n \quad y_n = Cx_n$$

Setting up the reachability matrix.

$$x_0 = 0$$

$$x_1 = Ax_0 + Bu_0 = Bu_0$$

$$x_2 = Ax_1 + Bu_1 = ABu_0 + Bu_1$$

$$x_3 = Ax_2 + Bu_2 = A^2Bu_0 + ABu_1 + Bu_2$$

$\vdots$

$$x_N = Ax_{N-1} + Bu_{N-1} = A^{N-1}Bu_0 + A^{N-2}Bu_1 + \dots + Bu_{N-1}$$

$$x_N = \begin{pmatrix} A^{N-1}B & A^{N-2}B & \dots & B \end{pmatrix} \begin{pmatrix} u_0 \\ u_1 \\ \vdots \\ u_{N-1} \end{pmatrix}$$

$\uparrow$   
 $R_N$  matrix for  $N$  time steps

$u$

To find if an input sequence from  $x_0 = 0$  to  $x_n = x^*$  we need to solve the following:

$$x^* = R_N u = (A^{N-1}B \quad A^{N-2}B \quad \dots \quad B) \begin{pmatrix} u_0 \\ u_1 \\ \vdots \\ u_{N-1} \end{pmatrix}$$

When does a solution exist?

- $R$  is full rank (rank  $N$ )
- $R$  is invertible

Determinant of  $R$  is non-zero

A matrix with full rank indicates the system is fully controllable and any state can be achieved with a set of control inputs.



5) 3 points  $(x_i, y_i)$   $i=1, 2, 3$  and model  $y=ax+b$ , best estimate for  $a, b$  and justify

$$y_i = f(x_i) = ax_i + b = f_i$$

$$\min_{a,b} \sum_{i=1}^3 (y_i - f_i)^2 = \min_{a,b} \sum_{i=1}^n v_i^2 \quad v_i = y_i - f_i$$

$$f_i = ax_i + b = (x_i \ 1) \begin{pmatrix} a \\ b \end{pmatrix} \quad F = A\Theta$$

$$F = \begin{pmatrix} ax_1 + b \\ ax_2 + b \\ ax_3 + b \end{pmatrix} \quad A = \begin{pmatrix} x_1 & 1 \\ x_2 & 1 \\ x_3 & 1 \end{pmatrix} \quad \Theta = \begin{pmatrix} a \\ b \end{pmatrix} \quad Y = \begin{pmatrix} y_1 \\ y_2 \\ y_3 \end{pmatrix}$$

$$\min_{a,b} \sum_{i=1}^3 (y_i - f_i)^2 = \min_{a,b} V^T V \quad V = Y - F$$

$$= \min_{a,b} (Y - F)^T (Y - F)$$

$$= \min_{\Theta} (Y - A\Theta)^T (Y - A\Theta)$$

$$= \min_{\theta} (Y^T - (A\theta)^T)(Y - A\theta)$$

$$= \min_{\theta} Y^T Y - Y^T A \theta - (A\theta)^T Y - (A\theta)^T A \theta$$

$$= \min_{\theta} Y^T Y - 2Y^T A \theta - (A\theta)^T A \theta$$

Take derivative and set to 0 (minimize):

$$\frac{d}{d\theta} (Y^T Y - 2Y^T A \theta - (A\theta)^T A \theta) = 0$$

$$0 = 0 - 2A^T Y - 2A^T A \theta$$

$$\theta = (A^T A)^{-1} A^T Y \quad (\text{solve using MATLAB})$$

$$\theta = \begin{pmatrix} a \\ b \end{pmatrix} = \begin{pmatrix} \frac{y_1(2x_1 - x_2 - x_3) + y_2(2x_2 - x_1 - x_3) + y_3(2x_3 - x_1 - x_2)}{2x_1^2 - 2x_1x_2 - 2x_1x_3 + 2x_2^2 + 2x_3^2 - 2x_2x_3} \\ \frac{y_1(-x_1x_2 - x_1x_3 + x_2^2 + x_3^2) + y_2(-x_1x_2 - x_2x_3 + x_1^2 + x_3^2) + y_3(-x_1x_3 - x_2x_3 + x_1^2 + x_2^2)}{2x_1^2 - 2x_1x_2 - 2x_1x_3 + 2x_2^2 + 2x_3^2 - 2x_2x_3} \end{pmatrix}$$

(Solution 1)

6) State Space Equation of  $G$ , define new system  $\tilde{A} \tilde{B} \tilde{C}$  that has an integrator in the input path  $1/s G(s)$ .

$$A = \begin{pmatrix} 1 & 0.5 \\ 0 & 1 \end{pmatrix} \quad B = \begin{pmatrix} 0 \\ 1 \end{pmatrix} \quad C = (0 \ 1)$$

Integrator state denoted by  $z$

$\dot{z} = u$       $u$ : input to system

$v = z$       $v$ : output of integrator

Augmented System:

$$\begin{pmatrix} \dot{x} \\ \dot{z} \end{pmatrix} = \begin{pmatrix} A & 0 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} x \\ z \end{pmatrix} + \begin{pmatrix} B \\ 1 \end{pmatrix} u \quad y = (C \ 0) \begin{pmatrix} x \\ z \end{pmatrix}$$

$$\tilde{A} = \begin{pmatrix} 1 & 0.5 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{pmatrix} \quad \tilde{B} = \begin{pmatrix} 0 \\ 1 \\ 1 \end{pmatrix} \quad \tilde{C} = (0 \ 1 \ 0)$$

## (Solution 2)

6) State Space Equation of  $G$ , define new system  $\hat{A} \hat{B} \hat{C}$  that has an integrator in the input path  $1/s G(s)$ .

$$A = \begin{pmatrix} 1 & 0.5 \\ 0 & 1 \end{pmatrix} \quad B = \begin{pmatrix} 0 \\ 1 \end{pmatrix} \quad C = (0 \ 1)$$

Laplace Transform:

$$sX(s) = AX(s) + BU(s) \quad Y(s) = CX(s)$$

$$(sI - A)X(s) = BU(s)$$

$$X(s) = (sI - A)^{-1}BU(s)$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{C(sI - A)^{-1}BU(s)}{U(s)} = C(sI - A)^{-1}B$$

$$sI - A = \begin{pmatrix} s-1 & 0.5 \\ 0 & s-1 \end{pmatrix}$$

$$(sI - A)^{-1} = \frac{1}{(s-1)^2} \begin{pmatrix} s-1 & -0.5 \\ 0 & s-1 \end{pmatrix}$$

$$C(sI - A)^{-1} = \left( 0 \quad \frac{s-1}{(s-1)^2} \right) = \left( 0 \quad \frac{1}{s-1} \right)$$

$$C(sI - A)^{-1} = \frac{1}{s-1} \cdot G(s) \quad (\text{add } 1/s \text{ integrator})$$

$$Y(s) = G(s)U(s)$$

$$s(s-1)Y(s) = U(s)$$

$$s^2Y(s) - sY(s) = U(s)$$

$$\ddot{y} - \dot{y} = u \quad \text{and} \quad \dot{y} = y + u$$

Let  $x_1 = y$  and  $x_2 = \dot{y}$

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u, \quad y = (1 \ 0) \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

$$A = \begin{pmatrix} 0 & 1 \\ 0 & 1 \end{pmatrix} \quad B = \begin{pmatrix} 0 \\ 1 \end{pmatrix} \quad C = (1 \ 0)$$

7) Variance of  $AX + BY$

$A, B$  matrices      $X, Y$  variables

$$\text{var}(AX + BY)$$

$$= \text{var}(AX) + \text{var}(BY) + 2\text{covar}(AX, BY)$$

Proof of  $\text{var}(AX) = A\text{var}(X)A^T$ :

$$= E[(AX - E(AX))(AX - E(AX))^T]$$

$$= E[(AX - AE(X))(AX - AE(X))^T]$$

$$= E[A(X - E(X))(X - E(X))^T A^T]$$

$$= AE[(X - E(X))(X - E(X))^T]A^T$$

$$= A\text{var}(X)A^T$$

Proof of  $\text{covar}(AX, BY) = A \text{covar}(X, Y) B^T$

$$= E[(AX - E(AX))(BY - E(BY))^T]$$

$$= E[(AX - AE(X))(BY - BE(Y))^T]$$

$$= E[A(X - E(X))(Y - E(Y))^T B^T]$$

$$= AE[(X - E(X))(Y - E(Y))^T] B^T$$

$$= A \text{covar}(X, Y) B^T$$

Subbing in value of proofs, we get:

$$= A \text{var}(X) A^T + B \text{var}(Y) B^T + 2A \text{covar}(X, Y) B^T$$

Note: Dimensions of  $A, B$  should conform with both  $X, Y$  vectors and covariance matrices.