

# Singular Value Decomposition

- General case of  $[K]\{u\}=b$  when  $[K]_{m \times n}$   
 $m > n$   
 can we define  $[K]^{-1}$  such that  $\{u\}=[K]^{-1}\{b\}$ ?

• Any  $[K]$  can be factored as

$$[K]=[U][\text{diag}(w)][V^T]$$

such that

-  $[K]$  and  $[U]$  are  $m \times n$

-  $[\text{diag}(w)]$  and  $[V]$  are  $n \times n$

- Columns of  $[U]$  are orthogonal,  $U^T U = I$

- Columns of  $[V]$  are orthogonal,  $V^T V = I$

also  $V V^T = I$  since square

- IF  $m=n$ ,  $U U^T = I$

- All  $w_i \geq 0$

e.g.

$$[K] = \begin{bmatrix} | & | & & | \\ u_1 & u_2 & \dots & u_n \\ | & | & & | \end{bmatrix} \begin{bmatrix} w_1 & & & 0 \\ & w_2 & & \\ & & \ddots & \\ 0 & & & w_n \end{bmatrix} \begin{bmatrix} - & - & V_1 & - & - \\ - & - & V_2 & - & - \\ & & \vdots & & \\ - & - & V_n & - & - \end{bmatrix}$$

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$\{U\}_i = \text{Left singular vectors of } [K]$

$\{V\}_i = \text{Right singular vectors of } [K]$

$\omega_i = \text{Singular values (typically ordered largest to smallest such that } K = \omega_i / \omega_n)$

$$[K][V] = [U][\text{diag}(\omega)] \quad \text{since } V^T V = I$$

or for each  $i$

$$[K]\{V\}_i = \omega_i \{U\}_i$$

i.e.  $[K]$  maps  $\{V\}_i$  onto  $\{U\}_i$  with scaling  $\omega_i$

○ A. Square, nonsingular case ( $m=n$ )

$$[K]^{-1} = [V][\text{diag}(1/\omega)][U]^T$$

so  $\{U\} = [V][\text{diag}(1/\omega)][U]^T \{b\}$

• Columns of  $[V]$  constitute the natural basis for  $\{U\}$

i.e.  $\{U\} = [V]\{c\} \Leftrightarrow \{c\} = [V]^T \{U\}$

• Columns of  $[U]$  are natural basis for  $\{b\}$

i.e.  $\{b\} = [U]\{d\} \Leftrightarrow \{d\} = [U]^T \{b\}$

• Inversion projects  $\{b\}$  onto  $U$ -space, multiplies by  $(1/\omega)$  and reassembles  $\{U\}$  in  $V$ -space

$$\{u\} = \sum_{i=1}^N \{V\}_i \frac{\{u\}_i \cdot \{b\}}{\omega_i} \Leftrightarrow c_i = \frac{d_i}{\omega_i} \left. \vphantom{\sum_{i=1}^N} \right\} \begin{array}{l} \text{when } [K] \\ \text{symmetric} \\ [U] = [V] \\ \omega \rightarrow \lambda \end{array} \quad (3)$$

- Relative to noise in  $\{b\}$

$$\begin{aligned} [\text{Cov}(u)] &= [K^{-1}] [\text{Cov}(b)] [K^{-T}] \\ &= [V] [\text{diag}(1/\omega)] [U]^T [\text{Cov}(b)] [U] [\text{diag}(1/\omega)] [V]^T \end{aligned}$$

$$[\text{Cov}(c)] = [\text{diag}(1/\omega)] [\text{Cov}(d)] [\text{diag}(1/\omega)]$$

$$\text{IF } [\text{Cov}(b)] = \sigma^2 I = [\text{Cov}(d)]$$

then we have

$$[\text{Cov}(u)] = \sigma^2 [V] [\text{diag}(1/\omega)^2] [V]^T$$

$$[\text{Cov}(c)] = \sigma^2 [\text{diag}(1/\omega)^2]$$

- Here  $\{b\}$  is projected onto  $[U]$  (rather than  $V$ )
- Noise present in  $\{u_i\}$  (from  $\{b\}$ ) will show up in  $\{u\}$  through  $\{V\}_i$  amplified by  $1/\omega_i$
- Small  $\omega_i$  are noise amplifiers relative to large ones making  $K$  the measure of noise distortion  
condition #

## B. Square Singular Case

IF a singular value is zero, i.e.  $\omega_N = 0$

• If  $\{b\}$  orthogonal to  $\{u\}_N$ , i.e.  $\{b\} \cdot \{u\}_N = 0$   
(or  $d_N = 0$ )

$$\text{then } \{u\} = \sum_{i=1}^{N-1} \{V\}_i \frac{\{u\}_i \cdot \{b\}}{\omega_i}$$

Same as non-singular case, except  $\{V\}$  left out (Not needed to satisfy  $[K]\{u\} = \{b\}$ )

But solution not unique ... can add any amount of  $\{V\}_N$  to solution without changing right hand side since by defn

$$[K]\{V\}_i = \omega_i \{u\}_i$$

and

$$[K]\{V\}_N = 0$$

$\therefore$  can have family of solutions

$$\begin{aligned} \{u\} &= \sum_{i=1}^{N-1} \{V\}_i \frac{\{u\}_i \cdot \{b\}}{\omega_i} + \alpha \{V\}_N \\ &= \sum_{i=1}^{N-1} \{V\}_i \frac{d_i}{\omega_i} + \alpha \{V\}_N \end{aligned}$$

all of which satisfy  $[K]\{u\} = \{b\}$

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Note: ... common to set  $\alpha = 0$  since this is the minimum variance solution

$$\text{Var}(u) = \sum_{i=1}^{N-1} \left( \frac{\{u\}_i \{b\}}{w_i} \right)^2 + \alpha^2 \left. \vphantom{\sum_{i=1}^{N-1}} \right\} \begin{array}{l} \text{operationally} \\ \text{same as} \\ \text{setting } \frac{1}{w_N} \rightarrow 0 \end{array}$$

- But when  $d_N \equiv \{u\}_N \cdot \{b\} \neq 0$  (more likely) especially due to R/O

have same options... either include an arbitrary amount of  $\{V\}_N$  in sol'n or remove it

$$\{u\} = \sum_{i=1}^{N-1} c_i \{V\}_i + \alpha \{V\}_N$$

$$[K]\{u\} = \{b\} = \sum_{i=1}^N d_i \{u\}_i$$

$$[K] \left( \sum_{i=1}^{N-1} c_i \{V\}_i + \alpha \{V\}_N \right) - \sum_{i=1}^N d_i \{u\}_i \equiv \{r\}$$

$$\sum_{i=1}^{N-1} c_i \underbrace{[K]\{V\}_i}_{w_i \{u\}_i} + \alpha \underbrace{[K]\{V\}_N}_{=0 \text{ since } w_N=0} - \sum_{i=1}^N d_i \{u\}_i$$

$$\begin{aligned} \{r\} &= \sum_{i=1}^{N-1} c_i w_i \{u\}_i - \sum_{i=1}^N d_i \{u\}_i \\ &= \sum_{i=1}^{N-1} (c_i w_i - d_i) u_i - d_N \{u\}_N \end{aligned}$$

$$\text{Var}(r) = \sum_{i=1}^{N-1} (c_i w_i - d_i)^2 + d_N^2$$

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the minimum variance sol'n is  $c_i = \frac{d_i}{w_i}$

doesn't matter what  $\alpha$  is, i.e.  $\alpha=0$  adds nothing to the residual

$\therefore$  Conclude if  $w_N=0$ , treat it as if infinite, i.e. eliminate  $N$ th term and will get minimum variance sol'n, i.e. keep  $\{V\}_N$  out of sol'n - it is undetermined and ignore the presence of  $\{U\}_N$  in  $\{b\}$ , there is no way to reduce its presence in the residual

### C. Square, Nearly-Singular case (i.e. $w_N \approx 0$ )

Options: include  $w_N$  in the computation or treat it as if zero

- IF retain, then have full  $N$  dimensional basis for  $\{b\}$  and  $\{U\}$ , but inversion becomes noise amplifier
- IF remove, then avoid noise amplification, but sol'n can't have any  $\{V\}_N$  content and any forcing in  $\{U\}_N$  is ignored. IF  $\{U\}_N$  an important mode of forcing, then in trouble because noise in its specification will dominate
- In practice ... consider all  $w_i=0$  by defining a cutoff based on condition number such that  $w_i$ 's below the cutoff are treated as zero

- Eliminates any  $\{V\}_i$  in solution for these  $W_i$  and ignores any  $\{U\}_i$  in the forcing (rather than amplifying it)

### Over-Determined Case $m > n$ (more eqn's than unknowns)

- will always have non zero residual  $\{r\} = [K]\{u\} - \{b\}$  but SVD gives minimum variance residual
- columns of  $[V]$  are complete, orthonormal basis for  $\{u\} \Rightarrow \{u\} = [V]\{c\} \Leftrightarrow \{c\} = [V]^T \{u\}$
- columns of  $[U]$  an incomplete basis for  $\{b\}$   
 $\{d\} = [U]^T \{b\} \Leftrightarrow [U]\{d\} = \{b\} - \{b'\}$   
where  $\{b'\}$  lies outside  $[U]$  space  $\Rightarrow [U]^T \{b'\} = 0$
- Since  $[K]\{V\}_i = \omega_i \{U\}_i$  and  $\{u\}$  is completely contained in  $[V]$ , then  $[K]\{u\}$  completely contained in  $[U]$ ,  $\therefore \{b'\}$  cannot be reached by any  $\{u\}$ .

then  $\{u\} = \sum_{i=1}^N c_i \{V\}_i$

$$\{r\} = \sum_{i=1}^N (c_i \omega_i - d_i) \{U\}_i - \{b'\}$$

$$\text{Var}(r) = \sum_{i=1}^N (c_i \omega_i - d_i)^2 + \text{Var}(b')$$

and minimum variance sol'n is  $c_i = \frac{d_i}{w_i}$

$$\{u\} = \sum_i \{V\}_i \frac{\{u\}_i \cdot \{b\}}{w_i}$$

### Under-Determined Case $m < n$

- add equations  $0 \cdot u = 0$  to make  $m = n$ , which generates  $n - m$  modes with  $w_i = 0$
- handle as above, eliminate the  $w_i = 0$

So in every case we have the same procedure

- eliminate all singular and nearly-singular modes ( $w_i \approx 0$ ) from the calculations
- results in a reduced-rank system which prevents noise amplification by leaving it uninvolved