# 离散时间系统的Z域分析

## 离散时间系统的Z域分析

❖连续时间系统通过Laplace变换将微分方程转 换成代数方程,离散时间系统通过Z变换将差 分方程转换成代数方程。

Z变换反Z变换→ 財域

时域

## 内容提要

- ❖Z变换的定义和收敛区
- ❖Z变换的性质
- ❖反Z变换
- ❖Z变换与Laplace变换
- ❖离散时间系统的z 域分析法

#### 重点与难点

- ❖z平面与s平面的对应关系
- ❖Z变换
  - ◆正反变换
  - ◆收敛区
  - ◆ 性质
- ❖离散时间系统的Z域分析法
  - ◆ 零输入响应
  - ◆ 零状态响应
  - ◆ 全响应
- ❖离散时间系统的稳定性

# Z变换

# 从Laplace变换到Z变换

#### ❖冲激取样信号

$$f_{\delta}(t) = f(t)\delta_{T}(t) = \sum_{k=-\infty}^{\infty} f(kT)\delta(t-kT)$$

双边拉普拉斯变换

$$F_{Sd}(s) = \sum_{k=-\infty}^{\infty} f(kT) \int_{-\infty}^{\infty} \delta(t - kT) e^{-st} dt = \sum_{k=-\infty}^{\infty} f(kT) e^{-skT}$$

令引入复变量  $z=e^{sT}$ ,f(kT)写成 f(k)

双边Z变换 
$$F(z) = \sum_{k=-\infty}^{\infty} f(k)z^{-k}$$

# Z变换的定义

$$f(k) \leftrightarrow F(z)$$
 z为复变量  $z = e^{sT} = e^{\sigma T} e^{j\omega T}$ 

双边Z变换 
$$F(z) = \sum_{k=-\infty}^{\infty} f(k)z^{-k}$$

单边Z变换 
$$F(z) = \sum_{k=0}^{\infty} f(k)z^{-k}$$
 有始序列

$$= f(0) + \frac{f(1)}{z} + \frac{f(2)}{z^2} + \dots$$

Z变换是z-1的幂级数,级数的系数即f(k)。 幂级数收敛时,Z变换才存在 – 收敛区。

# Z变换的收敛区

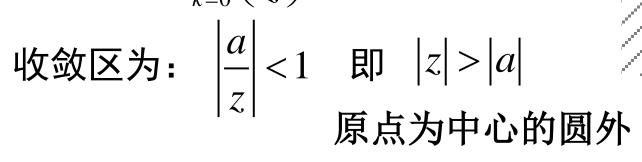
❖收敛区: 使F(z)存在并有限的z的取值范围。

单边Z变换

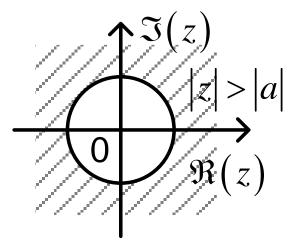
$$F(z) = \sum_{k=0}^{\infty} f(k)z^{-k} = f(0) + \frac{f(1)}{z} + \frac{f(2)}{z^{2}} + \dots$$

若 
$$f(k) = a^k \varepsilon(k)$$

则 
$$F(z) = \sum_{k=0}^{\infty} \left(\frac{a}{z}\right)^k$$
 等比级数求和



即 
$$|z| > |a|$$



### 收敛区练习1

❖有限长序列 f(k)=[1 2 3 2 1], k=[-2 -1 0 1 2]。

双边Z变换 
$$F(z) = \sum_{k=-\infty}^{\infty} f(k)z^{-k} = z^2 + 2z + 3 + 2z^{-1} + z^{-2}$$
  $0 < |z| < \infty$ 

单边Z变换 
$$F(z) = \sum_{k=0}^{\infty} f(k)z^{-k} = 3 + 2z^{-1} + z^{-2}$$
  $|z| > 0$ 

只要有限长序列各项存在且有限,则其z变换一定存在。 当包含z的正幂项时,收敛区不包含∞点。 当包含z的负幂项时,收敛区不包含0点。

## 收敛区练习2

#### ❖右边序列

$$f(k) = \begin{cases} a^k & k \ge k_1 \\ 0 & k < k_1 \end{cases}$$

双边Z变换 
$$F(z) = \sum_{k=k_1}^{\infty} f(k)z^{-k} = \frac{z}{z-a}$$

若  $k_1 < 0 \Rightarrow |a| < |z| < \infty$  原点为中心的圆外,但不包括无穷大点

若  $k_1 \ge 0 \Rightarrow |z| > |a|$  原点为中心的圆外

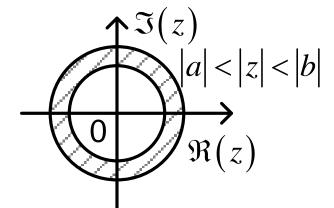
单边Z变换 
$$F(z) = \sum_{k=\max(0,k_1)}^{\infty} f(k)z^{-k} = \frac{z}{z-a} |z| > |a|$$

# 收敛区练习3

\*双边序列
$$f(k) = \begin{cases} a^k & k \ge 0 \\ b^k & k < 0 \end{cases} = a^k \varepsilon(k) + b^k \varepsilon(-k-1)$$

双边Z变换 
$$F(z) = \sum_{k=0}^{\infty} a^k z^{-k} + \sum_{k=-\infty}^{-1} b^k z^{-k} = \frac{z}{z-a} - \frac{z}{z-b}$$

若  $|a| < |b| \Rightarrow |a| < |z| < |b|$ 



单边Z变换 
$$F(z) = \sum_{k=0}^{\infty} a^k z^{-k} = \frac{z}{z-a}$$

# 双边Z变换与双边Laplace变换

$$f(k) = f_a(k)\varepsilon(k) + f_b(k)\varepsilon(-k-1)$$
  
双边Z变换  $|a| < |z| < |b|$   
 $F(z) = F_a(z) - F_b(z)$   
 $F_b(z) \leftrightarrow f_b(k)\varepsilon(k)$ 

双边Laplace变换

$$F(s) = F_a(s) - F_b(s)$$
  $\sigma_a < \sigma < \sigma_b$ 

$$F_b(s) \leftrightarrow f_b(t) \varepsilon(t)$$

所有极点都在收敛区外

表8-1 pp.277

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# 常见右边序列的Z变换

❖见书pp.275

$$\delta(k) \leftrightarrow 1$$
 整个z平面

$$\frac{|z| > |\gamma|}{\gamma^k \varepsilon(k) \longleftrightarrow \frac{z}{z - \gamma}}$$

$$\gamma = 1$$

$$\varepsilon(k) \longleftrightarrow \frac{z}{z - 1}$$

$$\gamma = e^{\lambda T}$$

$$e^{\lambda kT} \varepsilon(k) \leftrightarrow \frac{z}{z - e^{\lambda T}}$$

$$\gamma^{k-1}\varepsilon(k-1) \longleftrightarrow \frac{1}{z-\gamma} \qquad \varepsilon(k-1) \longleftrightarrow \frac{1}{z-1} \qquad e^{\lambda(k-1)T}\varepsilon(k-1) \longleftrightarrow \frac{1}{z-e^{\lambda T}}$$

$$\varepsilon(k-1) \longleftrightarrow \frac{1}{z-1}$$

$$e^{\lambda(k-1)T}\varepsilon(k-1) \leftrightarrow \frac{1}{z-e^{\lambda T}}$$

$$k\gamma^{k-1}\varepsilon(k) \longleftrightarrow \frac{z}{(z-\gamma)^2} \left| k\varepsilon(k) \longleftrightarrow \frac{z}{(z-1)^2} \right| ke^{\lambda(k-1)T}\varepsilon(k) \longleftrightarrow \frac{z}{(z-e^{\lambda T})^2}$$

$$k\varepsilon(k) \leftrightarrow \frac{z}{(z-1)^2}$$

$$ke^{\lambda(k-1)T}\varepsilon(k) \leftrightarrow \frac{z}{(z-e^{\lambda T})^2}$$

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#### Z变换练习1

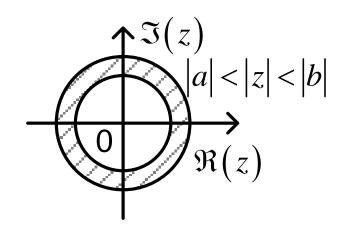
\* 双边序列
$$f(k) = \begin{cases} a^k & k \ge 0 \\ b^k & k < 0 \end{cases} = a^k \varepsilon(k) + b^k \varepsilon(-k-1)$$

方法一: 双边Z变换定义式:

$$F(z) = \sum_{k=0}^{\infty} a^k z^{-k} + \sum_{k=-\infty}^{-1} b^k z^{-k}$$

$$= \sum_{k=0}^{\infty} a^k z^{-k} + \sum_{j=1}^{\infty} b^{-j} z^j$$

$$= \frac{1}{1 - a/z} + \frac{z/b}{1 - z/b} = \frac{z}{z - a} + \frac{z}{b - z}$$



#### Z变换练习1

方法二:分别对左边和右边序列求Z变换

右边序列 
$$a^k \varepsilon(k) \leftrightarrow \frac{z}{z-a}$$

左边序列:三步

$$b^{k}\varepsilon(-k-1) \xrightarrow{k=-n} b^{-n}\varepsilon(n-1) \xrightarrow{Z\mathfrak{B}} F(\omega) \xrightarrow{\omega=z^{-1}} F(z)$$

$$b^{-n}\varepsilon(n-1) = \frac{1}{b}\frac{1}{b}^{n-1}\varepsilon(n-1) \longleftrightarrow F(\omega) = \frac{1/b}{\omega - 1/b}$$

$$F(z) \leftrightarrow \frac{-z}{z-b}$$

## Z变换练习1

方法三:

$$f(k)\varepsilon(k) \leftrightarrow F(z) \qquad |z| > |a|$$
$$f(k)\varepsilon(-k-1) \leftrightarrow -F(z) \quad |z| < |a|$$

右边序列 
$$a^k \varepsilon(k) \leftrightarrow \frac{z}{z-a}$$
  $|z| > |a|$ 

左边序列 
$$b^k \varepsilon(-k-1) \leftrightarrow \frac{-z}{z-b}$$
  $|z| < |b|$ 

$$F(z) = \frac{z}{z-a} - \frac{z}{z-b} \qquad |a| < |z| < |b|$$

# Z变换的性质

#### Z变换的性质 - 线性特性

❖(单边)Z变换定义式

$$f(k) \longleftrightarrow F(z)$$
  $F(z) = \sum_{k=0}^{\infty} f(k)z^{-k}$ 

❖线性特性  $f_1(k) \leftrightarrow F_1(z), f_2(k) \leftrightarrow F_2(z)$ 

$$a_1 f_1(k) + a_2 f_2(k) \longleftrightarrow a_1 F_1(z) + a_2 F_2(z)$$

一般情况下,收敛区为两个序列收敛区的公共部分。 某些特殊情况下,收敛区有扩大的可能。

例如 
$$\varepsilon(k) - \varepsilon(k-1) = \delta(k)$$

#### Z变换的性质 – 移序特性

❖移序特性  $f(k) \leftrightarrow F(z)$ 

$$f(k+1) \leftrightarrow z[F(z)-f(0)]$$

单边Z变换

$$f(k+n) \leftrightarrow z^{n} \left[ F(z) - \sum_{i=0}^{n-1} f(i) z^{-i} \right] \qquad n > 0$$

$$= z^{n} \left[ F(z) - f(0) - z^{-1} f(1) - z^{-2} f(2) - \dots \right]$$

 $f(k-n)\varepsilon(k-n) \longleftrightarrow z^{-n}F(z)$ 

一般情况下,收敛区不变。某些特殊情况下, 收敛区有变化的可能。

$$\delta(k) \leftrightarrow 1$$
 整个z平面  $\delta(k+n) \leftrightarrow z^n \quad |z| < \infty$   $\delta(k-n) \leftrightarrow z^{-n} \quad |z| > 0$  信号与线性系统电子讲

## 有始序列移序特性

❖对双边Z变换  $f(k)\varepsilon(k)\leftrightarrow F(z)$ 

$$f(k+n)\varepsilon(k+n) \leftrightarrow z^n F(z)$$
  $n>0$ 

$$f(k-n)\varepsilon(k-n) \longleftrightarrow z^{-n}F(z)$$

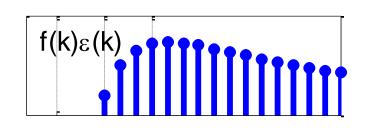
#### ❖对单边Z变换

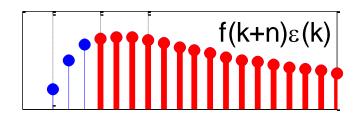
$$f(k+n)\varepsilon(k+n) \qquad n>0$$

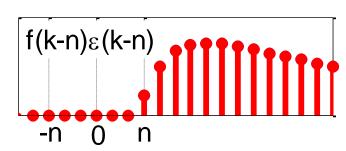
相当于  $f(k+n)\varepsilon(k)$ 

$$\leftrightarrow z^n \left[ F(z) - \sum_{i=0}^{n-1} f(i) z^{-i} \right]$$

$$f(k-n)\varepsilon(k-n) \longleftrightarrow z^{-n}F(z)$$







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#### 双边序列移序特性

❖对双边Z变换

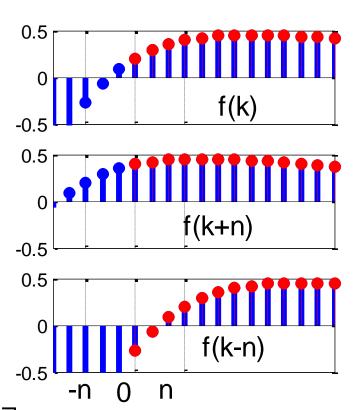
$$f(k) \leftrightarrow F(z)$$

$$f(k+n) \leftrightarrow z^{n} F(z) \qquad n > 0$$
$$f(k-n) \leftrightarrow z^{-n} F(z)$$

❖对单边Z变换 n>0

$$f(k+n) \leftrightarrow z^n \left[ F(z) - \sum_{i=0}^{n-1} f(i) z^{-i} \right]_{0.5}^{0.5}$$

$$f(k-n) \leftrightarrow z^{-n} \left[ F(z) + \sum_{i=-n}^{-1} f(i) z^{-i} \right]$$



## Z变换的性质 - z域尺度变换

❖z域尺度变换  $f(k) \leftrightarrow F(z)$   $\gamma_1 < |z| < \gamma_2$ 

$$a^k f(k) \leftrightarrow F\left(\frac{z}{a}\right) \qquad a\gamma_1 < |z| < a\gamma_2$$

推导

$$\sum_{k=0}^{\infty} a^k f(k) z^{-k} = \sum_{k=0}^{\infty} f(k) \left(\frac{z}{a}\right)^{-k} = F\left(\frac{z}{a}\right)$$

$$|\gamma_1| < \left|\frac{z}{a}\right| < \gamma_2$$

#### ❖求(单边)Z变换

$$f(k) = a^{k} \varepsilon(k) - a^{k} \varepsilon(k-1)$$

$$f_1(k) = a^k \varepsilon(k) \longleftrightarrow F_1(z) = \frac{z}{z - a} \qquad |z| > |a|$$

$$f_2(k) = a^k \varepsilon(k-1) \longleftrightarrow F_2(z) = \frac{a}{z-a} \quad |z| > |a|$$

$$F(z) = \frac{z - a}{z - a} = 1$$
 整个z平面 收敛区扩大

或 
$$f(k) = a^k \lceil \varepsilon(k) - \varepsilon(k-1) \rceil = \delta(k) \leftrightarrow F(z) = 1$$

$$\gamma^{k} \varepsilon(k) \longleftrightarrow \frac{z}{z-\gamma}, |z| > |\gamma|$$

\*求(単边)Z变換
$$\cos k\omega_0 \varepsilon(k) = \frac{e^{jk\omega_0}\varepsilon(k) + e^{-jk\omega_0}\varepsilon(k)}{2} \qquad |z| > 1$$

$$\leftrightarrow \frac{1}{2} \left[ \frac{z}{z - e^{j\omega_0}} + \frac{z}{z - e^{-j\omega_0}} \right] = \frac{z(z - \cos \omega_0)}{z^2 - 2z\cos \omega_0 + 1}$$

$$\sin k\omega_0 \varepsilon(k) = \frac{e^{jk\omega_0}\varepsilon(k) - e^{-jk\omega_0}\varepsilon(k)}{2}$$

$$\leftrightarrow \frac{1}{2} \left[ \frac{z}{z - e^{j\omega_0}} - \frac{z}{z - e^{-j\omega_0}} \right] = \frac{z \sin \omega_0}{z^2 - 2z \cos \omega_0 + 1}$$

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\*求(单边)Z变换 $f(k) = \beta^k \cos k\omega_0 \varepsilon(k)$ 

$$\cos k\omega_0 \varepsilon(k) \leftrightarrow F(z) = \frac{z(z - \cos \omega_0)}{z^2 - 2z\cos \omega_0 + 1} \qquad |z| > 1$$

$$F(\frac{z}{\beta}) = \frac{\frac{z}{\beta} \left(\frac{z}{\beta} - \cos \omega_0\right)}{\left(\frac{z}{\beta}\right)^2 - 2\frac{z}{\beta} \cos \omega_0 + 1} = \frac{1 - \beta z^{-1} \cos \omega_0}{1 - 2\beta z^{-1} \cos \omega_0 + \beta^2 z^{-2}}$$
$$\frac{\left|\frac{z}{\beta}\right| > 1 \Rightarrow |z| > |\beta|}{\left|\frac{z}{\beta}\right| > 1 \Rightarrow |z| > |\beta|}$$

## Z变换的性质 - z域微分

**❖z**域微分 
$$f(k)$$
 ↔  $F(z)$ 

$$kf(k) \leftrightarrow -z \frac{d}{dz} F(z)$$

推导

$$F(z) = \sum_{k=0}^{\infty} f(k)z^{-k}$$

$$\frac{d}{dz}z^{-k} = -kz^{-k-1}$$

$$\frac{d}{dz}F(z) = \sum_{k=0}^{\infty} f(k) \frac{d}{dz} z^{-k} = -z^{-1} \sum_{k=0}^{\infty} k f(k) z^{-k}$$

$$\sum_{k=0}^{\infty} kf(k)z^{-k} = \mathscr{Z}\left\{kf(k)\right\}$$

# Z变换的性质 – 卷积定理

\*卷积定理  $f_1(k) \leftrightarrow F_1(z), f_2(k) \leftrightarrow F_2(z)$   $f_1(k) * f_2(k) \leftrightarrow F_1(z) F_2(z)$ 

#### 推导

$$F(z) = \sum_{k=-\infty}^{\infty} \left[ \sum_{j=-\infty}^{\infty} f_1(j) f_2(k-j) \right] z^{-k}$$

$$= \sum_{j=-\infty}^{\infty} f_1(j) \left[ \sum_{k=-\infty}^{\infty} f_2(k-j) z^{-k} \right]$$

$$= \sum_{j=-\infty}^{\infty} f_1(j) z^{-j} F_2(z) = F_1(z) F_2(z)$$

# Z变换的性质 – 初值终值定理

❖初值和终值定理  $f(k) \leftrightarrow F(z)$ 

$$f(0) = \lim_{z \to \infty} F(z) \qquad F(z) = f(0) + \sum_{k=1}^{\infty} f(k)z^{-k}$$
$$f(\infty) = \lim_{z \to 1} (z-1)F(z)$$

$$\lim_{z \to 1} \sum_{k=0}^{\infty} \left[ f(k+1) - f(k) \right] z^{-k} = \sum_{k=0}^{\infty} \left[ f(k+1) - f(k) \right] = f(\infty) - f(0)$$

又因为 
$$f(k+1)-f(k) \leftrightarrow z[F(z)-f(0)]-F(z)$$

$$\lim_{z \to 1} \sum_{k=0}^{\infty} \left[ f(k+1) - f(k) \right] z^{-k} = \lim_{z \to 1} (z-1) F(z) - f(0)$$

f(∞)存在的条件: F(z)的收敛区是单位圆外整个z平面,或极点在单位圆内,若在单位圆上则是单阶极点。

#### ❖求(单边)Z变换

$$k\varepsilon(k) \leftrightarrow -z\frac{d}{dz}\left(\frac{z}{z-1}\right) = \frac{z}{\left(z-1\right)^2} \qquad |z| > 1$$

$$k^{2}\varepsilon(k) \leftrightarrow -z\frac{d}{dz}\left(\frac{z}{(z-1)^{2}}\right) = \frac{z(z+1)}{(z-1)^{3}} \quad |z| > 1$$

$$\varepsilon(k) \leftrightarrow \frac{z}{z-1}$$
  $kf(k) \leftrightarrow -z\frac{d}{dz}F(z)$ 

\* 求巻积和
$$f_1(k) = a^k \varepsilon(k)$$
 $f_2(k) = b^k \varepsilon(k)$ 

$$F_1(z) = \frac{z}{z - a} \quad |z| > |a|$$

$$F_2(z) = \frac{z}{z - b} \quad |z| > |b|$$

$$|z| > \max(|a|,|b|)$$

$$F(z) = \frac{z}{z-a} \frac{z}{z-b} = \frac{1}{a-b} \left( \frac{az}{z-a} - \frac{bz}{z-b} \right)$$

$$f(k) = \frac{1}{a-b} \left( aa^k \varepsilon(k) - bb^k \varepsilon(k) \right) = \frac{a^{k+1} - b^{k+1}}{a-b} \varepsilon(k)$$

#### ❖利用Z变换解差分方程

$$r(k+1)-0.9r(k) = 0.05\varepsilon(k+1)$$
  $r(-1) = 0$ 

两边同时进行Z变换

$$R(z) - 0.9z^{-1}R(z) = 0.05\frac{z}{z-1}$$
  $|z| > 1$ 

$$R(z) = \frac{0.05z^2}{(z-0.9)(z-1)} = \frac{-0.45z}{z-0.9} + \frac{0.5z}{z-1}$$

$$f(k) = \left[0.5 - 0.45(0.9)^{k}\right] \varepsilon(k)$$

#### ❖求零状态响应

$$e(k) = \varepsilon(k)$$

$$h(k) = a^{k} \varepsilon(k) - a^{k-1} \varepsilon(k-1)$$

$$e(k) = \varepsilon(k) \longleftrightarrow E(z) = \frac{z}{z-1}$$

$$h(k) = a^{k} \varepsilon(k) - a^{k-1} \varepsilon(k-1)$$

$$\leftrightarrow H(z) = (1 - z^{-1}) \frac{z}{z - a} = \frac{z - 1}{z - a} \qquad |z| > |a|$$

$$R(z) = E(z)H(z) = \frac{z}{z-a} \iff r(k) = a^k \varepsilon(k)$$

# 反Z变换

# 单边Laplace反变换 – 复习

- \*定义式
- ❖查表+Laplace变换的性质
- ❖部分分式法(Heaviside展开法)
- ❖围线积分法(留数法)

n个单阶极点 
$$D(s) = (s-s_1)(s-s_2)...(s-s_n) = 0$$
 部分分式法 留数法

$$f(t) = \sum_{i=1}^{n} K_i e^{s_i t} \varepsilon(t)$$

$$f(t) = \sum_{i=1}^{n} \operatorname{Res}_i \varepsilon(t)$$

$$K_i = \left[ \left( s - s_i \right) \frac{N(s)}{D(s)} \right]_{s=s_i}$$
  $\operatorname{Res}_i = \left[ \left( s - s_i \right) F(s) e^{st} \right]_{s=s_i}$ 

#### 反Z变换

- ❖定义式 幂级数展开法
- ❖查表+Z变换的性质
- ❖部分分式展开法
- ❖留数法(围线积分法)

#### 幂级数展开法

❖Z变换定义式

$$F(z) = \sum_{k=0}^{\infty} f(k)z^{-k} \qquad |z| > |a|$$

❖幂级数展开 – 利用长除法得到幂级数系数

$$F(z) = \frac{N(z)}{D(z)} = A_0 + A_1 z^{-1} + A_2 z^{-2} + \dots$$

❖反Z变换

$$f(0) = A_0, f(1) = A_1, f(2) = A_2,...$$

### ❖求反变换

$$F(z) = \frac{z}{(z-1)^2} \qquad |z| > 1$$

$$z^{-1} + 2z^{-2} + 3z^{-3} + \dots$$

$$z^{2} - 2z + 1)z$$

$$\underline{z - 2 + z^{-1}}$$

$$2 - z^{-1}$$

$$\underline{2 - 4z^{-1} + 2z^{-2}}$$

$$3z^{-1} - 2z^{-2}$$

$$\underline{3z^{-1} - 6z^{-2} + 3z^{-3}}$$

$$4z^{-2} - 3z^{-3}$$

$$F(z) = 0 + z^{-1} + 2z^{-2} + 3z^{-3} + \dots$$
$$= \sum_{k=0}^{\infty} kz^{-k}$$

$$f(k) = k\varepsilon(k)$$

不易得到f(k)的数学表达式

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### 部分分式展开法

基本变换 - 分子包含z

$$\frac{z}{z-\gamma} \leftrightarrow \gamma^k \varepsilon(k) \qquad \frac{z}{(z-\gamma)^2} \leftrightarrow k \gamma^{k-1} \varepsilon(k)$$

一般对F(z)/z进行展开

F(z)/z引入一个等于零的极点

若全部为单根 
$$\frac{F(z)}{z} = \frac{K_0}{z} + \sum_{i=1}^n \frac{K_i}{z - \gamma_i}$$

即 
$$F(z) = K_0 + \sum_{i=1}^{n} \left[ K_i \frac{z}{z - \gamma_i} \right]$$
  $\gamma_i$ 为 $F(z)$ 的极点

$$\leftrightarrow f(k) = K_0 \delta(k) + \sum_{i=1}^{n} K_i \gamma_i^k \varepsilon(k)$$

# 部分分式法待定系数的确定

- ❖同Laplace反变换中待定系数的确定方法
  - ◆ 待定系数法
  - ◆ 系数计算公式(单根)

$$K_0 = [F(z)]_{z=0}$$
  $K_i = \left[ (z - \gamma_i) \frac{F(z)}{z} \right]_{z=\gamma_i}$ 

$$F(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$
  $|z| > 1$ 

$$\frac{F(z)}{z} = \frac{z}{z^2 - 1.5z + 0.5} = \frac{z}{(z - 0.5)(z - 1)} = \frac{K_1}{z - 0.5} + \frac{K_2}{z - 1}$$

$$K_1 = -1, K_2 = 2$$

$$F(z) = \frac{-z}{z - 0.5} + \frac{2z}{z - 1}$$

$$f(k) = (2 - 0.5^k) \varepsilon(k)$$

$$|z| > 1$$
 右边序列

$$\frac{z}{z-\gamma} \leftrightarrow \gamma^k \varepsilon(k)$$

**❖求反变换** 
$$F(z) = \frac{z^2}{z^2 + 16}$$

$$\frac{F(z)}{z} = \frac{z}{z^2 + 16} = \frac{K_1}{z - 4j} + \frac{K_2}{z + 4j}$$

共轭单阶极点  $\pm 2j$ 

$$K_1 = \frac{1}{2}, K_2 = K_1^* = \frac{1}{2}$$

$$K_1 = \frac{1}{2}, K_2 = K_1^* = \frac{1}{2}$$
  $f(k) = \frac{1}{2} [(4j)^k + (-4j)^k] \varepsilon(k)$ 

$$F(z) = \frac{1/2}{z - 4j} + \frac{1/2}{z + 4j}$$

$$F(z) = \frac{1/2}{z - 4j} + \frac{1/2}{z + 4j} = \frac{4^k}{2} \left[ \left( e^{j\frac{\pi}{2}} \right)^k + \left( e^{-j\frac{\pi}{2}} \right)^k \right] \varepsilon(k)$$

$$=4^k\cos\left(rac{\pi}{2}k
ight)arepsilon(k)$$
 信号与线性系统电子讲义

### ❖求对应的右边序列

$$F(z) = \frac{4z^3 + 7z^2 + 3z + 1}{z^3 + z^2 + z}$$

$$\frac{F(z)}{z} = \frac{4z^3 + 7z^2 + 3z + 1}{z^2(z^2 + z + 1)} = \frac{K_0}{z} + \frac{K_1}{z^2} + \frac{K_2z + K_3}{z^2 + z + 1}$$

$$K_0 = 2, K_1 = 1, K_2 = 2, K_3 = 4$$

$$F(z) = 2 + \frac{1}{z} + \frac{2z^2 + 4z}{z^2 + z + 1} = F_1(z)$$

$$0, -\frac{1}{2} \pm \frac{\sqrt{3}}{2} j = e^{\pm \frac{2}{3}\pi}$$

$$f(k) = 2\delta(k) + \delta(k-1) + f_1(k)$$

$$\begin{cases} \frac{z\left(\mathbf{z}-\cos\omega_{0}\right)}{z^{2}-2z\cos\omega_{0}+1} & \leftrightarrow \cos\omega_{0}k\varepsilon\left(k\right) \\ \frac{z\sin\omega_{0}}{z^{2}-2z\cos\omega_{0}+1} & \leftrightarrow \sin\omega_{0}k\varepsilon\left(k\right) \\ F_{1}\left(z\right) & = \frac{2z^{2}+4z}{z^{2}+z+1} = \frac{2z\left(z+\frac{1}{2}\right)}{z^{2}+z+1} + \frac{2\sqrt{3}\left(\frac{\sqrt{3}}{2}z\right)}{z^{2}+z+1} \\ f_{1}\left(k\right) & = 2\cos\left(\frac{2}{3}k\pi\right)\varepsilon\left(k\right) + 2\sqrt{3}\sin\left(\frac{2}{3}k\pi\right)\varepsilon\left(k\right) \\ f\left(k\right) & = 2\delta\left(k\right) + \delta\left(k-1\right) + 4\cos\left(\frac{2}{3}k\pi-\frac{\pi}{3}\right)\varepsilon\left(k\right) \end{cases}$$

**❖求反变换** 
$$F(z) = \frac{z(z+1)}{(z-3)(z-1)^2}$$
 |z|>3

$$\frac{F(z)}{z} = \frac{(z+1)}{(z-3)(z-1)^2} = \frac{K_1}{z-1} + \frac{K_2}{(z-1)^2} + \frac{K_3}{z-3}$$

$$K_1 = -1, K_2 = -1, K_3 = 1$$

$$|z|>3$$
 右边序列

$$F(z) = \frac{-z}{z-1} + \frac{-z}{(z-1)^2} + \frac{z}{z-3}$$

$$\frac{z}{z-\gamma} \longleftrightarrow \gamma^k \varepsilon(k)$$

$$f(k) = (3^k - k - 1)\varepsilon(k)$$

$$\frac{z}{\left(z-\gamma\right)^{2}} \leftrightarrow k\gamma^{k-1}\varepsilon(k)$$

### ❖p重根 – 待定系数

$$\frac{F_{i}(z)}{z} = \frac{K_{i1}}{(z - \gamma_{i})} + \frac{K_{i2}}{(z - \gamma_{i})^{2}} + \dots + \frac{K_{ip}}{(z - \gamma_{i})^{p}} = \sum_{k=1}^{p} \frac{K_{ik}}{(z - \gamma_{i})^{k}}$$

$$K_{ik} = \frac{1}{(p-k)!} \frac{d^{p-k}}{dz^{p-k}} \left[ \left( z - \gamma_i \right)^p \frac{F(z)}{z} \right]_{z=\gamma_i}$$

$$K_{ip} = \left[ \left( z - \gamma_i \right)^p \frac{F(z)}{z} \right]_{z = \gamma_i} \qquad K_{i(p-1)} = \frac{d}{dz} \left[ \left( z - \gamma_i \right)^p \frac{F(z)}{z} \right]_{z = \gamma_i}$$

$$f_i(t) = \left[ K_{i1} \gamma_i^k + K_{i2} k \gamma_i^{k-1} + \frac{K_{i3}}{2!} k (k-1) \gamma_i^{k-2} + \dots \right] \mathcal{E}(k)$$

#### ❖p重根

$$\frac{z}{\left(z-\gamma\right)^{p}} \leftrightarrow \frac{k(k-1)....(k-p+2)}{(p-1)!} \gamma^{k-p+1} \varepsilon(k)$$

$$\frac{z}{z-\gamma} \leftrightarrow \gamma^k \varepsilon(k)$$

两边同时对 γ 求导

$$\frac{z}{\left(z-\gamma\right)^2} \leftrightarrow k\gamma^{k-1}\varepsilon(k)$$

$$\frac{z}{\left(z-\gamma\right)^{3}} \leftrightarrow \frac{1}{2} k \left(k-1\right) \gamma^{k-2} \varepsilon \left(k\right)$$

**❖求反变换** 
$$F(z) = \frac{z^3 + z^2}{(z-1)^3}$$

$$\frac{F(z)}{z} = \frac{z^2 + z}{(z - 1)^3} = \frac{K_1}{(z - 1)^3} + \frac{K_2}{(z - 1)^2} + \frac{K_3}{z - 1}$$

$$K_1 = (z-1)^3 \frac{F(z)}{z} \Big|_{z=1} = 2$$
  $K_2 = \frac{d}{dz} \left[ (z-1)^3 \frac{F(z)}{z} \right]_{z=1} = 3$ 

$$K_3 = \frac{1}{2} \frac{d^2}{dz^2} \left[ (z-1)^3 \frac{F(z)}{z} \right]_{z=1} = 1$$

$$F(z) = \frac{2z}{(z-1)^3} + \frac{3z}{(z-1)^2} + \frac{z}{z-1}$$

$$F(z) = \frac{z^3 + z^2}{(z-1)^3} \qquad |z| > 1$$

$$F(z) = \frac{2z}{(z-1)^3} + \frac{3z}{(z-1)^2} + \frac{z}{z-1}$$

右边序列 
$$f(k) = [k(k-1)+3k+1]\varepsilon(k)$$

# 留数法(围线积分法)

### ❖C为收敛区内一包含原点的闭合曲线

$$f(k) = \sum \text{Res} \left[ \underline{F(z)} z^{k-1} \right]_{\text{ChWA}}$$

F(z)zk-1的极点

只考虑有始序列

$$f(k) = \sum \operatorname{Res}_{i}$$

<sub>上</sub> 所有极点均在C内

单阶极点

$$\operatorname{Res}_{i} = \left[ \left( z - \gamma_{i} \right) F(z) z^{k-1} \right]_{z = \gamma_{i}}$$

k=0时,出现原点处一阶极点。 k<0时,原点处出现极点阶数随k变化的极点。

需分别计算留数。

p阶极点

$$\operatorname{Res}_{i} = \frac{1}{(p-1)!} \frac{d^{p-1}}{dz^{p-1}} \left[ \left( z - \gamma_{i} \right)^{p} F(z) z^{k-1} \right]_{z=1}^{p}$$

 $J_{z=\gamma_{i}}$ 信号与线性系统电子讲义

# 留数法(围线积分法)

❖复变函数理论: 具有有限个极点的复变函数在 复平面内所有极点留数的和加上函数在无穷远

点留数的和等于零。
$$Res[F(z)z^{k-1}]_{Chown} = f(k)$$

$$+\operatorname{Res}\left[F(z)z^{k-1}\right]_{C$$
外极点  $+\operatorname{Res}\left[F(z)z^{k-1}\right]_{z=\infty}=0$ 

$$f(k) = -\operatorname{Res}\left[F(z)z^{k-1}\right]_{C \text{ Proposition } K} - \operatorname{Res}\left[F(z)z^{k-1}\right]_{z=\infty}$$

$$\operatorname{Res}\left[F(z)z^{k-1}\right]_{z=\infty} = -\operatorname{Res}\left[F(z^{-1})z^{-k+1}z^{-2}\right]_{z=0}$$

**❖求反变换** 
$$F(z) = \frac{z^3 + 2z^2 + 1}{z(z-1)(z-0.5)}$$
 |z|>1

$$F(z)z^{k-1} = \frac{z^3 + 2z^2 + 1}{z(z-1)(z-0.5)}z^{k-1} = \frac{z^3 + 2z^2 + 1}{(z-1)(z-0.5)}z^{k-2}$$

 $k \ge 2$  时,两个单阶极点  $z_1 = 1, z_2 = 0.5$ 

$$\operatorname{Res}_{1} = \left[ (z-1)F(z)z^{k-1} \right]_{z=1} = \left[ \frac{z^{3} + 2z^{2} + 1}{(z-0.5)}z^{k-2} \right]_{z=1} = 8$$

$$\operatorname{Res}_{2} = \left[ \left( z - 0.5 \right) F(z) z^{k-1} \right]_{z=0.5} = \left[ \frac{z^{3} + 2z^{2} + 1}{(z-1)} z^{k-2} \right]_{z=0.5} = -13(0.5)^{k}$$

$$k \ge 2$$
 时  $f(k) = 8 - 13(0.5)^k$ 

$$F(z)z^{k-1} = \frac{z^3 + 2z^2 + 1}{(z-1)(z-0.5)}z^{k-2}$$

$$k=1$$
 时,三个单阶极点  $z_1=1, z_2=0.5, z_3=0$ 

$$\operatorname{Res}_{3} = \left[ zF(z)z^{k-1} \right]_{z=0,k=1}$$

$$= \left[ \frac{z^3 + 2z^2 + 1}{(z-1)(z-0.5)} z^{k-1} \right]_{z=0,k=1} = 2$$

$$f(k) = 8-13(0.5)^{1} + 2 = 3.5$$

$$F(z)z^{k-1} = \frac{z^3 + 2z^2 + 1}{(z-1)(z-0.5)}z^{k-2}$$

$$k=0$$
 时,四个极点  $z_1=1, z_2=0.5, z_{3.4}=0$ 

$$\operatorname{Res}_{3,4} = \frac{1}{(2-1)!} \left\{ \frac{d^{2-1}}{dz^{2-1}} \left[ z^2 F(z) z^{k-1} \right] \right\}_{z=0,k=0}$$

$$= \frac{d}{dz} \left| \frac{z^3 + 2z^2 + 1}{(z - 0.5)(z - 1)} \right|_{z = 0} = 6$$

$$f(0) = 8-13(0.5)^{0}+6=1$$

$$f(k) = \begin{cases} 1 & k = 0 \\ 3.5 & k = 1 \\ 8 - 13(0.5)^k & k \ge 2 \end{cases}$$

### 双边序列

#### ❖判断极点归属

- ◆ 极点在收敛区内边界以内, 极点对应右边序列。
- ◆ 极点在收敛区外边界以外, 极点对应左边序列。
- ❖右边序列部分可由部分分式法或留数法得到。
- ❖左边序列部分分三步进行
  - $\bullet$  z=w<sup>-1</sup>, F(w); F(w) $\rightarrow$ f(n); n=-k, f(k).
- ❖左边序列部分也可以由下面关系得到

$$f(k)\varepsilon(k) \longleftrightarrow F(z) \qquad |z| > |a|$$
$$f(k)\varepsilon(-k-1) \longleftrightarrow -F(z) \qquad |z| < |a|$$

求反变换 
$$F(z) = \frac{1 + 2z^{-1}}{1 - 2z^{-1} + z^{-2}} \qquad \frac{z}{z - \gamma} \leftrightarrow \gamma^{k} \varepsilon(k)$$

$$F(z) = \frac{3z}{(z - 1)^{2}} + \frac{z}{z - 1} \qquad \frac{z}{(z - \gamma)^{2}} \leftrightarrow k\gamma^{k-1} \varepsilon(k)$$

$$|z| > 1$$
 时,右边序列  $f(k) = (3k+1)\varepsilon(k)$ 

$$|z|<1$$
 时,左边序列

$$F(w) = F(z)|_{z=w^{-1}} = \frac{-1}{w-1} + \frac{3w}{(w-1)^2}$$

$$F(w) \leftrightarrow f(n) = (3n-1)\varepsilon(n-1)$$

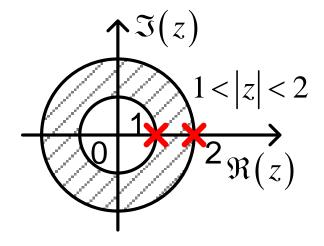
$$f(k) = f(n)|_{n-k} = -(3k+1)\varepsilon(-k-1)$$

$$F(z) = \frac{3z^2 - 5z}{(z-1)(z-2)}$$

$$F(z) = \frac{2z}{z-1} + \frac{z}{z-2}$$

对应的右边序列为

$$f_1(k) = 2\varepsilon(k) + 2^k \varepsilon(k)$$



$$\gamma_1 = 1$$
 对应右边序列  $f(k)\varepsilon(k)$ 

$$\gamma_2 = 2$$
 对应左边序列  $-f(k)\varepsilon(-k-1)$ 

$$f(k) = 2\varepsilon(k) - 2^{k}\varepsilon(-k-1)$$

(1) 
$$f(k) = 4^{-k} \left[ \varepsilon(k) - \varepsilon(k-2) \right]$$
  
 $F(z) = \sum_{k=0}^{\infty} f(k) z^{-k} = \sum_{k=0}^{1} 4^{-k} z^{-k} = 1 + \frac{1}{4} z^{-1}$ 

(2) 
$$f(k) = k(-1)^k \varepsilon(k)$$

$$F(z) = -z \frac{d}{dz} \left( \frac{z}{z+1} \right) = \frac{-z}{(z+1)^2}$$

(3) 
$$f(k) = \sum_{n=0}^{k} a^n = \sum_{n=0}^{\infty} a^n \varepsilon(k-n) = a^k \varepsilon(k) * \varepsilon(k)$$

$$F(z) = \frac{z}{z - a} \frac{z}{z - 1}$$

也可以利用定义式求 (3) 
$$f(k) = \sum_{n=0}^{\infty} a^n$$

$$F(z) = \sum_{k=0}^{\infty} f(k) z^{-k}$$

$$= \sum_{k=0}^{\infty} \left[ \sum_{n=0}^{k} a^{n} \right] z^{-k} = \sum_{k=0}^{\infty} \left[ \sum_{n=0}^{\infty} a^{n} \varepsilon(k-n) \right] z^{-k}$$

$$= \sum_{n=0}^{\infty} a^{n} \left[ \sum_{k=0}^{\infty} \varepsilon(k-n) z^{-k} \right]$$

$$= \sum_{n=0}^{\infty} a^{n} \left[ z^{-n} \frac{z}{z-1} \right] = \frac{z}{z-a} \frac{z}{z-1}$$

### ❖留数法

$$F(z) = \frac{3z^2 - 5z}{(z-2)(z-1)}$$

$$F(z)z^{k-1} = \frac{3z-5}{(z-2)(z-1)}z^{k}$$

两个单阶极点  $\gamma_1 = 1, \gamma_2 = 2$ 

只考虑|z|>2,右边序列的情况,即k≥0

Res<sub>1</sub> = 
$$\left[ (z-1)F(z)z^{k-1} \right]_{z=1} = \left[ \frac{3z-5}{(z-2)}z^k \right]_{z=1} = 2$$

$$\operatorname{Res}_{2} = \left[ (z-2)F(z)z^{k-1} \right]_{z=2} = \left[ \frac{3z-5}{(z-1)}z^{k} \right]_{z=2} = 2^{k}$$

$$f(k) = 2\varepsilon(k) + 2^{k}\varepsilon(k)$$
信号与线性系统电子

号与线性系统电子讲义

\* 求反变换 
$$F(z) = \frac{z^2}{(z+1)(z-2)} = \frac{\frac{1}{3}z}{z+1} + \frac{\frac{2}{3}z}{z-2}$$
(1)  $|z| > 2$  (2)  $|z| < 1$  (3)  $1 < |z| < 2$ 

(1) 
$$|z| > 2$$

(2) 
$$|z| < 1$$

(3) 
$$1 < |z| < 2$$

(1) 
$$|z| > 2$$
 右边序列  $f(k) = \left(\frac{1}{3}(-1)^k + \frac{2}{3}2^k\right)\varepsilon(k)$ 

(2) 
$$|z| < 1$$
 左边序列  $f(k) = -\left(\frac{1}{3}(-1)^k + \frac{2}{3}2^k\right)\varepsilon(-k-1)$ 

(3)1<|z|<2 双边序列

$$f(k) = \frac{1}{3}(-1)^k \varepsilon(k) - \frac{2}{3}2^k \varepsilon(-k-1)$$

# Z变换与Laplace变换的关系

# Z变换与Laplace变换

❖将F(s)所对应的连续时间函数进行取样得到的 离散时间信号的F(z)。

$$F(s)$$
 L反变换  $f(t)$  取样  $f(k)$  Z变换  $F(z)$ 

Laplace反变换 
$$f(t) = \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} F(s) e^{st} ds$$

以间隔T<sub>S</sub>取样 
$$f(kT_S) = \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} F(s) e^{skT_S} ds$$

Z变换 
$$F(z) = \sum_{k=0}^{\infty} f(kT_S) z^{-k}$$

# Z变换与Laplace变换

$$F(z) = \sum_{k=0}^{\infty} f(k) z^{-k}$$

$$= \sum_{k=0}^{\infty} \frac{1}{2\pi i} \left[ \int_{\sigma - j\infty}^{\sigma + j\infty} F(s) e^{skT_s} ds \right] z^{-k}$$

$$= \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} F(s) \left[ \sum_{k=0}^{\infty} e^{skT_s} z^{-k} \right] ds$$

$$=\frac{1}{2\pi j}\int_{\sigma-j\infty}^{\sigma+j\infty}F(s)\frac{z}{z-e^{sT_S}}ds$$

#### 留数形式

$$F(z) = \sum \text{Res} \left| \frac{zF(s)}{z - e^{sT_s}} \right|$$

# Z变换与Laplace变换

留数形式 
$$F(z) = \sum \operatorname{Res} \left[ \frac{zF(s)}{z - e^{sT_s}} \right]$$

❖F(s)的一阶极点 $s_1$ 对应F(z)的一阶极点  $\gamma_1$ 

Res<sub>1</sub> = 
$$\left[ (s - s_1) \frac{zF(s)}{z - e^{sT_s}} \right]_{s=s} = \frac{K_1 z}{z - e^{s_1 T_s}} = \frac{K_1 z}{z - \gamma_1}$$

$$K_1 = \left[ \left( s - s_1 \right) F(s) \right]_{s = s_1}$$

$$\gamma_1 = e^{s_1 T_S}$$

# Z变换与Laplace变换练习1

\*求Z变换
$$f(k) = \sin k\omega_0 T_S \varepsilon(kT_S)$$

(1) F(s) 
$$f(t) = \sin \omega_0 t \varepsilon(t) \leftrightarrow F(s) = \frac{\omega_0}{s^2 + \omega_0^2}$$
  
 $s_1 = j\omega_0, s_2 = -j\omega_0$   
(2) 求留数  $K_1 = \left[ \left( s - s_1 \right) F(s) \right]_{s=s_1} = -\frac{j}{2}$   
 $K_2 = \left[ \left( s - s_2 \right) F(s) \right]_{s=s_2} = \frac{j}{2}$ 

(3) 
$$F(z) = \frac{K_1 z}{z - e^{s_1 T_S}} + \frac{K_2 z}{z - e^{s_2 T_S}} = \frac{z \sin \omega_0 T_S}{z^2 - 2z \cos \omega_0 T_S + 1}$$

# Z变换与Laplace变换练习2

$$f(k) = e^{-akT_S} \varepsilon(kT_S)$$

(1) F(s) 
$$f(t) = e^{-at} \mathcal{E}(t) \longleftrightarrow F(s) = \frac{1}{s+a}$$
  
 $s_1 = -a$ 

(2) 求留数 
$$K_1 = \left[ (s+a)F(s) \right]_{s=-a} = 1$$

(3) 
$$F(z) = \frac{K_1 z}{z - e^{s_1 T_S}} = \frac{z}{z - e^{-aT_S}}$$

### z平面与s平面的映射关系

$$z = e^{sT_S}$$
 复频率  $s = \sigma + j\omega$ 
则  $z = e^{sT_S} = e^{(\sigma + j\omega)T_S}$ 

$$z = re^{j\theta}$$
  $r = e^{\sigma T_S}$   $\theta = \omega T_S$ 

s平面虚轴映射为z平面上单位圆  $\sigma=0$  r=1 s平面右半平面映射为单位圆的圆外  $\sigma>0$  r>1 s平面左半平面映射为单位圆的圆内  $\sigma<0$  r<1

H(s)极点位于左半平面,系统稳定。

对应H(z)极点位于单位圆内,系统稳定。

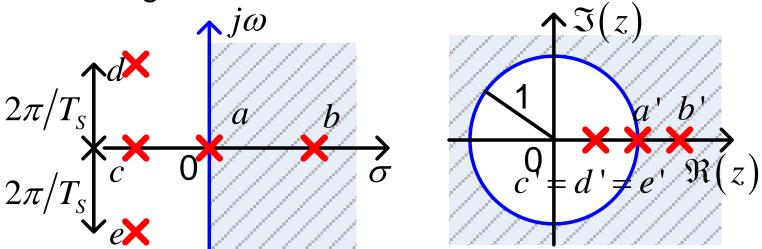
### z平面与s平面的映射关系

复频率  $s = \sigma + j\omega$ 

$$z = re^{j\theta} = e^{sT_S} = e^{(\sigma + j\omega)T_S}$$
  $r = e^{\sigma T_S}$   $\theta = \omega T_S$ 

s平面虚轴上相差 $2\pi/T_s$  的点映射为z平面上同一个点。

ω改变 $2\pi/T_S$  , θ旋转一圈。



s平面上高度为2π/T<sub>S</sub>的窄带映射整个z平面

## 离散时间系统的Z域分析法

时域到Z域 → Z域求解代数方程 → Z域到时域

# Z域求解系统响应

❖时域→Z域,得到Z域的输入激励信号。

$$e(k) \rightarrow E(z)$$

- ❖得到Z域系统方程 对差分方程进行Z变换
  - ◆零输入响应:包含初始条件,令e(k)=0
  - ◆零状态响应:不包含初始条件,e(k)→E(z)
  - ◆ 全响应:包含初始条件, e(k)→E(z)
- ❖由Z域系统方程求系统响应

$$R_{zi}(z) + R_{zs}(z)$$
  $R(z)$ 

 $H(z) = \frac{R_{zs}(z)}{E(z)}$ 

❖Z域→时域,得到时域的输出信号。

$$R_{zi}(z) + R_{zs}(z) \rightarrow r_{zi}(k) + r_{zs}(k)$$
  $R(z) \rightarrow r(k)$ 

## 对差分方程进行Z变换

- ❖利用移序特性  $f(k) \leftrightarrow F(z)$
- ◆一般单边z变换 (k≥0) 教材pp.295-296

$$f(k+n) \longleftrightarrow z^{n} \left[ F(z) - \sum_{i=0}^{n-1} f(i) z^{-i} \right]$$

$$f(k-n) \longleftrightarrow z^{-n} \left[ F(z) + \sum_{i=-n}^{-1} f(i) z^{-i} \right]$$

#### ❖求响应

$$r(k+2)-0.7r(k+1)+0.1r(k) = 7e(k+2)-2e(k+1)$$
  
 $r_{zi}(0) = 2, r_{zi}(1) = 4$   $e(k) = \varepsilon(k)$ 

零输入响应 - 差分方程齐次解

$$r(k+2)-0.7r(k+1)+0.1r(k)=0$$

$$\gamma_1 = 0.5, \gamma_2 = 0.2$$

$$r_{zi}(k) = \left[C_1 \cdot 0.5^k + C_2 \cdot 0.2^k\right] \varepsilon(k)$$
$$= \left[12(0.5)^k - 10(0.2)^k\right] \varepsilon(k)$$

零输入响应 - 对差分方程进行Z变换

$$r(k+2)-0.7r(k+1)+0.1r(k)=0$$

$$\begin{bmatrix} z^{2}R_{zi}(z) - z^{2}r_{zi}(0) - zr_{zi}(1) \end{bmatrix} -0.7 \begin{bmatrix} zR_{zi}(z) - zr_{zi}(0) \end{bmatrix} + 0.1R_{zi}(z) = 0$$

$$R_{zi}(z) = \frac{2z^2 + 2.6z}{z^2 - 0.7z + 0.1} = \frac{12z}{z - 0.5} - \frac{10z}{z - 0.2}$$

$$r_{zi}(k) = \left\lceil 12(0.5)^k - 10(0.2)^k \right\rceil \varepsilon(k)$$

零状态响应 – 系统函数H(z)

$$\varepsilon(k) \leftrightarrow \frac{z}{z-1}$$

$$r(k+2)-0.7r(k+1)+0.1r(k)=7e(k+2)-2e(k+1)$$

$$H(S) = \frac{7S^2 - 2S}{S^2 - 0.7S + 0.1} \quad H(z) = \frac{7z^2 - 2z}{z^2 - 0.7z + 0.1}$$

$$R_{zs}(z) = \frac{7z^2 - 2z}{z^2 - 0.7z + 0.1} \frac{z}{z - 1} = \frac{12.5z}{z - 1} - \frac{5z}{z - 0.5} - \frac{0.5z}{z - 0.2}$$

$$r_{zs}(k) = \left[12.5 - 5(0.5)^k - 0.5(0.2)^k\right] \varepsilon(k)$$

$$r(k) = r_{zi}(k) + r_{zs}(k) = \left[12.5 + 7(0.5)^k - 10.5(0.2)^k\right] \varepsilon(k)$$

全响应 - 对差分方程进行Z变换 r(k+2)-0.7r(k+1)+0.1r(k)=7e(k+2)-e(k+1) $[z^2R(z)-z^2r(0)-zr(1)]-0.7[zR(z)-zr(0)]+0.1R(z)$  $=7\left[z^{2}E(z)-z^{2}e(0)-ze(1)\right]-2\left[zE(z)-ze(0)\right]$ 代入  $r(0) = r_{zi}(0) + r_{zs}(0)$   $r(1) = r_{zi}(1) + r_{zs}(1)$  $(z^2-0.7z+0.1)R(z)-(z^2-0.7z)r_{zi}(0)-zr_{zi}(1)$  $-(z^2-0.7z)r_{zs}(0)-zr_{zs}(1)$  两项相抵消

$$= \left(7z^2 - 2z\right)E\left(z\right) - \left(7z^2 - 2z\right)e\left(0\right) - 7ze\left(1\right)$$

全响应 - 对差分方程进行Z变换

$$r(k+2)-0.7r(k+1)+0.1r(k)=7e(k+2)-e(k+1)$$

$$[z^{2}R(z)-z^{2}r_{zi}(0)-zr_{zi}(1)]-0.7[zR(z)-zr_{zi}(0)]$$

$$+0.1R(z)=7z^{2}E(z)-zE(z)$$

$$R(z) = \frac{7z^2 - z}{z^2 - 0.7z + 0.1} E(z) + \frac{2z^2 + 2.6z}{z^2 - 0.7z + 0.1}$$

$$R_{zs}(z)$$

$$R_{zi}(z)$$

$$r(k) = r_{zi}(k) + r_{zs}(k) = \left[12.5 + 7(0.5)^k - 10.5(0.2)^k\right] \varepsilon(k)$$

信号与线性系统电子讲义

❖求单位函数响应,并写出差分方程。

$$e(k) = \left(-\frac{1}{2}\right)^{k} \varepsilon(k) \qquad r_{zs}(k) = \left[\frac{3}{2}\left(\frac{1}{2}\right)^{k} + 4\left(-\frac{1}{3}\right)^{k} - \frac{9}{2}\left(-\frac{1}{2}\right)^{k}\right] \varepsilon(k)$$

$$E(z) = \frac{z}{z + 1/2} \qquad R_{zs}(z) = \frac{3/2z}{z - 1/2} + \frac{4z}{z + 1/3} - \frac{9/2z}{z + 1/2}$$

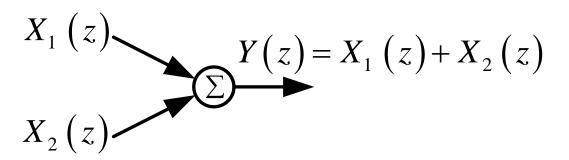
$$H(z) = \frac{R_{zs}(z)}{E(z)} = \frac{z^2 + 2z}{z^2 - \frac{1}{6}z - \frac{1}{6}} = \frac{3z}{z - \frac{1}{2}} + \frac{-2z}{z + \frac{1}{3}}$$
$$h(k) = \left[3\left(\frac{1}{2}\right)^k - 2\left(-\frac{1}{3}\right)^k\right] \varepsilon(k)$$

$$r(k+2) - \frac{1}{6}r(k+1) - \frac{1}{6}r(k) = e(k+2) + 2e(k+1)$$
 信号与线性系统电子讲义

### Z域框图

#### ❖基本元件

◆加法器



◆标量乘法器

$$X(z) = aX(z)$$

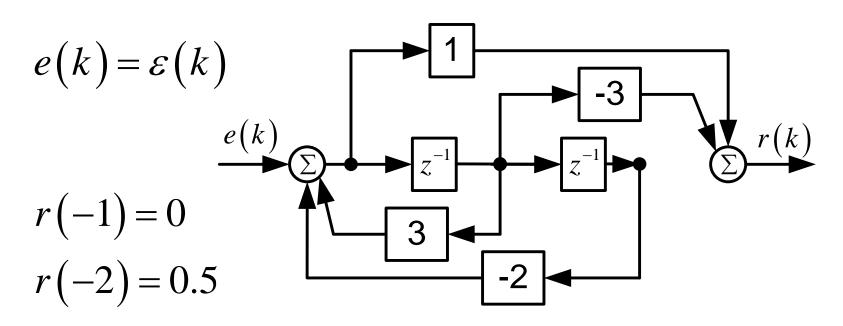
◆延时器

$$X\left(z\right) \longrightarrow z^{-1} Y\left(z\right) = z^{-1} X\left(z\right)$$

❖引入两个辅助方程作框图 时域

$$y(k+1)$$
  $y(k)$ 

❖求单位函数响应,零状态响应,零输入响应



$$\begin{cases} z^{2}Q - 3zQ + 2Q = E(z) \\ R(z) = z^{2}Q - 3zQ \end{cases} \qquad H(z) = \frac{z^{2} - 3z}{z^{2} - 3z + 2}$$

$$H(z) = \frac{z^2 - 3z}{z^2 - 3z + 2} = \frac{2z}{z - 1} - \frac{z}{z - 2}$$

$$h(k) = (2-2^k)\varepsilon(k)$$
  $e(k) = \varepsilon(k) \leftrightarrow \frac{z}{z-1}$ 

$$R_{zs}(z) = \frac{z^2 - 3z}{z^2 - 3z + 2} \frac{z}{z - 1} = \frac{z^2(z - 3)}{(z - 1)^2(z - 2)}$$
$$= \frac{2z}{(z - 1)^2} + \frac{3z}{z - 1} + \frac{-2z}{z - 2}$$

$$r_{zs}(k) = \left[2k + 3 - 2(2)^{k}\right] \varepsilon(k)$$

$$\gamma_1 = 1, \gamma_2 = 2$$

$$r_{zi}(k) = (C_1 + C_2 2^k) \varepsilon(k)$$

$$r_{zi}(-1) = r(-1) = 0$$
  $\Rightarrow \begin{cases} C_1 = 1 \\ C_2 = -2 \end{cases}$ 

$$r_{zi}(k) = \left[1 - 2(2)^{k}\right] \varepsilon(k)$$

$$r(k) = \left\lceil 2k + 4 - 4(2)^k \right\rceil \varepsilon(k)$$

### 离散时间系统的稳定性

#### ❖从极点分布进行判断:

- ◆ 极点在单位圆内,系统稳定。
- ◆单位圆上单阶极点,系统临界稳定。

$$H(z) = \frac{N(z)}{D(z)}$$
  $D(z) = 0$  →极点

#### ❖通过特征方程系数进行判断

◆利用双线性变换和Routh-Hurwitz判据

$$z = \frac{\lambda + 1}{\lambda - 1} \qquad D(z) = 0 \xrightarrow{z = \frac{\lambda + 1}{\lambda - 1}} G(\lambda) = 0$$

单位圆外极点→右半平面极点

Routh-Hurwitz判据

# 离散时间系统稳定性练习1

$$r(k+2)+0.1r(k+1)-0.2r(k)=e(k+2)+e(k+1)$$

$$H(z) = \frac{z^2 + z}{z^2 + 0.1z - 0.2} = \frac{z(z+1)}{(z-0.4)(z+0.5)}$$

$$\gamma_1 = 0.4, \gamma_2 = -0.5$$

两极点均在单位圆内 → 系统稳定

# 离散时间系统稳定性练习2

$$D(z) = z^3 - 0.5z^2 + 0.25z - 0.075$$

$$G(\lambda) = D\left(\frac{\lambda+1}{\lambda-1}\right) = \frac{0.675\lambda^3 + 2.475\lambda^2 + 3.025\lambda + 1.825}{(\lambda-1)^3}$$

#### 分子多项式系数同号且无缺项, Routh-Hurwitz阵列

$$\lambda^3$$
 0.675 3.025

$$\lambda^2$$
 2.475 1.825

$$\lambda$$
 2.527

1 1.825

$$G(\lambda)=0$$
 无右半平面根

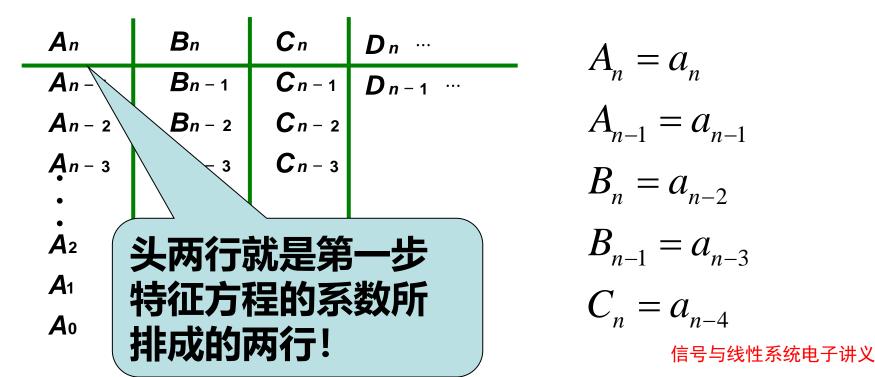
→ 系统稳定

# 构筑Routh-Hurwitz阵列的步骤:

第一步: 把G(A)分子所有系数按如下顺序排成两行。

$$a_n$$
  $a_{n-2}$   $a_{n-4}$   $a_{n-6}$  以此类推,排到  $a_0$  为止  $a_{n-1}$   $a_{n-3}$   $a_{n-5}$   $a_{n-7}$ 

#### 第二步:排列R-H阵列规则如下



# Routh-Hurwitz阵列计算公式:

#### 下面各行按下列公式计算:

$$A_{n-2} = -\frac{1}{A_{n-1}} \begin{vmatrix} A_n & B_n \\ A_{n-1} & B_{n-1} \end{vmatrix}$$

$$B_{n-2} = -\frac{1}{A_{n-1}} \begin{vmatrix} A_n & C_n \\ A_{n-1} & C_{n-1} \end{vmatrix}$$

$$A_{n-3} = -\frac{1}{A_{n-2}} \begin{vmatrix} A_{n-1} & B_{n-1} \\ A_{n-2} & B_{n-2} \end{vmatrix} \qquad B_{n-3} = -\frac{1}{A_{n-2}} \begin{vmatrix} A_{n-1} & C_{n-1} \\ A_{n-2} & C_{n-2} \end{vmatrix}$$

$$B_{n-2} = -\frac{1}{A_{n-1}} \begin{vmatrix} A_n & C_n \\ A_{n-1} & C_{n-1} \end{vmatrix} \qquad C_{n-2} = -\frac{1}{A_{n-1}} \begin{vmatrix} A_n & D_n \\ A_{n-1} & D_{n-1} \end{vmatrix}$$

$$B_{n-3} = -\frac{1}{A_{n-2}} \begin{vmatrix} A_{n-1} & C_{n-1} \\ A_{n-2} & C_{n-2} \end{vmatrix}$$

# 离散时间系统稳定性练习3

#### ❖使系统稳定的常数P的范围

$$D(z) = z^2 + 0.25z + P$$

$$G(\lambda) = D\left(\frac{\lambda + 1}{\lambda - 1}\right) = \frac{(5/4 + P)\lambda^{2} + (2 - 2P)\lambda + (3/4 + P)}{(\lambda - 1)^{3}}$$

分子多项式系数同号且无缺项, Routh-Hurwitz阵列

$$\lambda^{2} \cdot \frac{5}{4} + P \quad \frac{3}{4} + P \quad P > -\frac{3}{4}$$

$$\lambda \quad 2 - 2P \quad P > -\frac{5}{4}$$

$$1 \quad \frac{3}{4} + P \quad P < 1$$

# 离散时间系统Z域分析法练习1

\* 求零状态响应 
$$r(k+2)-5r(k+1)+6r(k)=e(k)$$
  $e(k)=\varepsilon(k)$  1

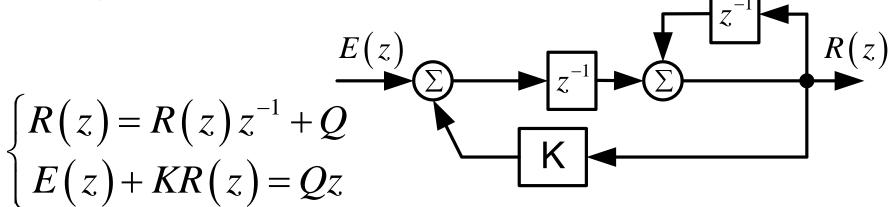
$$H(z) = \frac{1}{z^2 - 5z + 6}$$

$$Y_{zs}(z) = E(z)H(z) = \frac{1}{z^2 - 5z + 6} \frac{z}{z - 1}$$
$$= \frac{1}{2} \frac{z}{z - 1} - \frac{z}{z - 2} + \frac{1}{2} \frac{z}{z - 3}$$

$$r(k) = \left| \frac{1}{2} - 2^k + \frac{1}{2} 3^k \right| \varepsilon(k)$$

# 离散时间系统Z域分析法练习2

#### ❖系统稳定的K值范围



$$R(z) = R(z)z^{-1} + z^{-1}[E(z) + KR(z)]$$

$$H(z) = \frac{R(z)}{E(z)} = \frac{1}{z - (K+1)}$$

$$|K+1| \le 1 \qquad -2 \le K \le 0$$

系统稳定

# 离散时间系统的频率响应

# 离散序列的Fourier变换

❖s平面虚轴s=jω对应于z平面单位圆,单位圆上 Z变换即序列的Fourier变换。

$$F(z) = \sum_{k=-\infty}^{+\infty} f(k) z^{-k}$$

$$F(e^{j\omega}) = F(z)|_{z=e^{j\omega}} = \sum_{k=-\infty}^{+\infty} f(k)e^{-jk\omega}$$

$$f(k) = \frac{1}{2\pi i} \iint_{z=e^{jw}} F(z) z^{k-1} dz = \frac{1}{2\pi} \int_{-\pi}^{+\pi} F(e^{j\omega}) e^{jk\omega} d\omega$$

 $F(e^{j\omega})$ 具有周期性  $-\pi < \omega < \pi$ 

 $\omega$ 与 $T_S$ 无关,称归一化频率,  $\omega=2\pi$  相当于 $\omega_S$ 

# 离散序列的Fourier变换

#### ❖若考虑采样频率,则

$$F\left(e^{j\Omega T_S}\right) = \sum_{k=-\infty}^{+\infty} f\left(kT_S\right) e^{-jkT_S\Omega} \qquad \omega = \Omega T_S = 2\pi \frac{\Omega}{\omega_S}$$

$$f(kT_S) = \frac{1}{\omega_S} \int_{-\frac{\omega_S}{2}}^{+\frac{\omega_S}{2}} F(e^{j\Omega T_S}) e^{jkT_S\Omega} d\Omega - \frac{\omega_S}{2} < \Omega < \frac{\omega_S}{2}$$

$$F(e^{j\Omega T_S})$$

$$\left(-\frac{\omega_s}{2}, \frac{\omega_s}{2}\right) or(0, \omega_s)$$

#### ω 归一化角频率

$$F(e^{j\omega})$$

$$(-\pi,\pi)or(0,2\pi)$$

# 离散时间系统的频率特性

❖当Laplace变换收敛区包含虚轴时

$$H(j\omega) = H(s)|_{s=j\omega}$$

❖当Z变换收敛区包含单位圆时

$$H(e^{j\omega}) = H(z)|_{z=e^{j\omega}} = \sum_{k=0}^{\infty} h(k)e^{-jk\omega}$$

$$H\left(e^{j\omega}\right) = \left|H\left(e^{j\omega}\right)\right| e^{j\phi(\omega)}$$

归一化频率响应函数

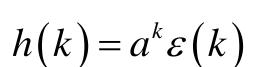
- ◆周期函数,周期为2π
- ◆幅度频谱为偶函数
- ◆相位频谱为奇函数

# 离散时间系统的频率特性练习1

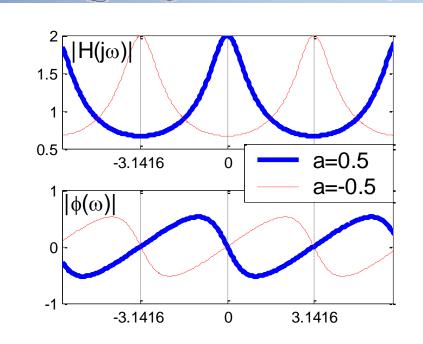
#### ❖求系统频率特性

$$r(k+1)-ar(k) = e(k+1)$$
$$0 < a < 1$$

$$H(z) = \frac{z}{z - a} \qquad |z| > a$$



$$H\left(e^{j\omega}\right) = \frac{e^{j\omega}}{e^{j\omega} - a}$$



$$\left| H\left(e^{j\omega}\right) \right| = \frac{1}{\sqrt{1 + a^2 - 2a\cos\omega}}$$

$$\phi(\omega) = -arctg \left( \frac{a \sin \omega}{1 - a \cos \omega} \right)$$
 信号与线性系统电子讲

# 离散时间系统的频率特性练习2

#### ❖求系统零状态响应

$$e(k) = e^{jk\omega} \varepsilon(k)$$

$$H(z) \leftrightarrow h(k)$$

$$r_{zs}(k) = e(k) * h(k) = \sum_{i=-\infty}^{+\infty} h(i)e(k-i)$$

$$=\sum_{i=-\infty}^{+\infty}h(i)e^{j(k-i)\omega}$$

$$\sum_{i=-\infty}^{+\infty} h(i) (e^{j\omega})^{-i} = H(z) \Big|_{z=e^{j\omega}}$$

$$=e^{jk\omega}\sum_{i=-\infty}^{+\infty}h(i)(e^{j\omega})^{-i}$$

$$r_{zs}(k) = e^{jk\omega}H(e^{j\omega})$$

离散指数序列的响应为同频率的离散指数序列

# 离散时间系统的频率特性练习2 100

$$e(k) = e^{jk\omega}\varepsilon(k) \qquad r_{zs}(k) = e^{jk\omega}H(e^{j\omega})$$

$$e(k) = \cos(k\pi)\varepsilon(k) = \frac{1}{2}e^{jk\pi}\varepsilon(k) + \frac{1}{2}e^{-jk\pi}\varepsilon(k)$$

$$r_{zs}(k) = \frac{1}{2}e^{jk\pi}H(e^{j\pi}) + \frac{1}{2}e^{-jk\pi}H(e^{-j\pi})$$

$$= \frac{1}{2}|H(e^{j\pi})|e^{jk\pi+\phi(\pi)} + \frac{1}{2}|H(e^{-j\pi})|e^{-j[k\pi+\phi(\pi)]}$$

$$= |H(e^{j\pi})|\cos[k\pi+\phi(\pi)]$$

# 第八章复习

#### ❖Z变换

- ◆定义式
- ◆收敛区
- ◆ 性质
- ◆ 反变换
- ❖Z变换与Laplace变换的关系
- ❖离散时间系统Z域分析法
- ❖离散时间系统的稳定性
- ❖离散时间系统的频率响应特性

# ❖求Z变换

$$f(k) = \left[ k(-1)^k \sum_{n=0}^k 2^n \right] \varepsilon(k)$$

$$f_1(k) = \left[\sum_{n=0}^k 2^n\right] \varepsilon(k) = \varepsilon(k) * 2^k \varepsilon(k)$$

$$\leftrightarrow F_1(z) = \frac{z^2}{(z-1)(z-2)} \qquad |z| > 2$$

$$a^k f(k) \longleftrightarrow F\left(\frac{z}{a}\right)$$

$$f_{2}(k) = (-1)^{k} f_{1}(k) \longleftrightarrow F_{2}(z) = \frac{z^{2}}{(z+1)(z+2)}$$

$$\frac{z^{2}}{(z+1)(z+2)}$$

$$kf(k) \leftrightarrow -z \frac{d}{dz} F(z)$$

$$f(k) = kf_2(k) \leftrightarrow F(z) = -z \frac{d}{dz} F_2(z)$$

$$F(z) = -z \frac{d}{dz} \frac{z^2}{(z+1)(z+2)} = \frac{-z^2(3z+4)}{(z+1)^2(z+2)^2} |z| > 2$$

❖求反Z变换 
$$F(z) = \ln\left(1 + \frac{a}{z}\right)$$
  $|z| > |a|$ 

$$\frac{d}{dz}F(z) = \frac{1}{1+\frac{a}{z^2}} \cdot \left(-\frac{a}{z^2}\right) = \sum_{k=0}^{\infty} \left(-\frac{a}{z}\right)^k \cdot \left(-\frac{a}{z^2}\right)$$

$$-z\frac{d}{dz}F(z) = \sum_{k=0}^{\infty} (-1)^k a^{k+1}z^{-k-1} = \sum_{n=1}^{\infty} (-1)^{n-1} a^n z^{-n}$$
$$k+1 = n$$

$$kf(k) = (-1)^{k-1} a^k \varepsilon(k-1)$$

$$f(k) = \left(-1\right)^{k-1} \frac{a^{k}}{k} \varepsilon(k-1)$$

- ❖画出模拟框图 (时域或Z域)
- ❖求系统函数,绘出极零图
- ❖判断系统稳定性
- ❖系统零状态响应

$$r(k+2)-\frac{3}{4}r(k+1)+\frac{1}{8}r(k)=e(k+2)+\frac{1}{3}e(k+1)$$

$$e(k) = \varepsilon(k) - \varepsilon(k-1)$$

信号与线性系统电子讲义

$$r(k+2) - \frac{3}{4}r(k+1) + \frac{1}{8}r(k) = e(k+2) + \frac{1}{3}e(k+1)$$

$$\begin{cases} S^2q - \frac{3}{4}Sq + \frac{1}{8}q = e(k) \\ r(k) = S^2q + \frac{1}{3}Sq & 1 \end{cases}$$
时域模拟框图
$$e(k) = \frac{1}{3}Sq + \frac{1}{3}Sq$$

$$r(k+2) - \frac{3}{4}r(k+1) + \frac{1}{8}r(k) = e(k+2) + \frac{1}{3}e(k+1)$$

$$\begin{cases} z^2Q - \frac{3}{4}zQ + \frac{1}{8}Q = E(z) \\ R(z) = z^2Q + \frac{1}{3}zQ & 1 \\ \hline E(z) & 2 \\ \hline 2 & 3/4 \\ \hline -1/8 & 6 \\ \hline 68 - 5 & 5 \\ \hline 68 - 5 & 5 \\ \hline 1/3 & 2 \\ \hline 2 & 1 \\ \hline 1/3 & 2 \\ \hline 2 & 1 \\ \hline 3/4 & -1/8 \\ \hline 68 - 5 & 5 & 6 \\ \hline 2 & 1 & 2 \\ \hline 3/4 & -1/8 & -1/8 \\ \hline 68 - 5 & 5 & 6 \\ \hline 68 -$$

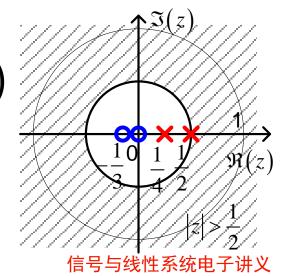
$$r(k+2) - \frac{3}{4}r(k+1) + \frac{1}{8}r(k) = e(k+2) + \frac{1}{3}e(k+1)$$

$$H(z) = \frac{z^2 + \frac{1}{3}z}{z^2 - \frac{3}{4}z + \frac{1}{8}} = \frac{-\frac{7}{3}z}{z - \frac{1}{4}} + \frac{\frac{10}{3}z}{z - \frac{1}{2}} \qquad |z| > \frac{1}{2}$$

$$h(k) = -\frac{7}{3} \left(\frac{1}{4}\right)^k \varepsilon(k) + \frac{10}{3} \left(\frac{1}{2}\right)^k \varepsilon(k)$$

两个极点: 1/4, 1/2 系统稳定

两个零点: 0, -1/3



$$e(k) = \varepsilon(k) - \varepsilon(k-1) = \delta(k) \leftrightarrow 1$$

$$|z| > \frac{1}{2}$$

$$R_{zs}(z) = H(z) = \frac{z^2 + \frac{1}{3}z}{z^2 - \frac{3}{4}z + \frac{1}{8}} = \frac{-\frac{7}{3}z}{z - \frac{1}{4}} + \frac{\frac{10}{3}z}{z - \frac{1}{2}}$$

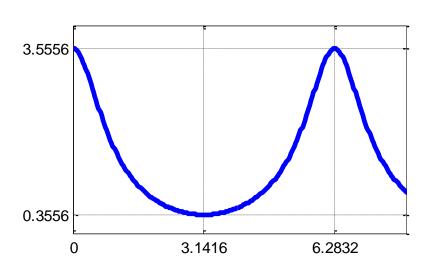
$$r_{zs}(k) = h(k) = -\frac{7}{3} \left(\frac{1}{4}\right)^k \varepsilon(k) + \frac{10}{3} \left(\frac{1}{2}\right)^k \varepsilon(k)$$

$$H\left(e^{j\omega}\right) = \frac{e^{j2\omega} + \frac{1}{3}e^{j\omega}}{e^{j2\omega} - \frac{3}{4}e^{j\omega} + \frac{1}{8}}$$

$$\omega = 0 \Longrightarrow H(e^{j\omega}) = \frac{32}{9}$$

$$\omega = \pi \Longrightarrow H(e^{j\omega}) = \frac{16}{45}$$

$$\omega = 2\pi \Rightarrow H(e^{j\omega}) = \frac{32}{9}$$



### 变换域分析法比较

#### ❖离散系统Z域分析法

- ◆ Z变换
- ◆ 差分方程→代数方程
- ◆ 圆外: |z|>a
- ◆单位圆内极点
- ◆ 单位圆上 → F(e<sup>jω</sup>)

#### ❖连续系统复频域分析法

- ◆ Laplace变换
- ◆ 微分方程→代数方程
- ◆ e<sup>st</sup>=e<sup>(a+bj)t</sup>
- ◆ 直线以右: σ >a
- ◆ 左半平面极点
- ◆ 虚轴上 → F(jω)

都可以用部分分式法和留数法求反变换 都有单双边变换 变换都可自动引入初始条件 都可把卷积运算转换成乘积运算