



CATARINA MENDES ÁVILA  
BSc in Electrical and Computer Engineering

**DESIGN, SIMULATION AND OPTIMIZATION  
OF MICROSTRIP PATCH ANTENNAS  
FOR WI-FI 7**

SOME THOUGHTS ON THE LIFE, THE UNIVERSE,  
AND EVERYTHING ELSE

MASTER IN ELECTRICAL AND COMPUTER ENGINEERING  
SPECIALIZATION IN SPECIALIZATION NAME

NOVA University Lisbon  
*Draft: January 27, 2026*



# DESIGN, SIMULATION AND OPTIMIZATION OF MICROSTRIP PATCH ANTENNAS FOR WI-FI 7

SOME THOUGHTS ON THE LIFE, THE UNIVERSE,  
AND EVERYTHING ELSE

CATARINA MENDES ÁVILA

BSc in Electrical and Computer Engineering

**Adviser:** Mary Doe Adviser Name  
*Full Professor, NOVA University Lisbon*

**Co-advisers:** John Doe Co-Adviser Name  
*Associate Professor, NOVA University Lisbon*  
John Doe other Co-Adviser Name  
*Full Professor, NOVA University Lisbon*

MASTER IN ELECTRICAL AND COMPUTER ENGINEERING  
SPECIALIZATION IN SPECIALIZATION NAME

NOVA University Lisbon

*Draft: January 27, 2026*

# ABSTRACT

Regardless of the language in which the dissertation is written, usually there are at least two abstracts: one abstract in the same language as the main text, and another abstract in some other language.

The abstracts' order varies with the school. If your school has specific regulations concerning the abstracts' order, the NOVAthesis L<sup>A</sup>T<sub>E</sub>X (`novathesis`) (L<sup>A</sup>T<sub>E</sub>X) template will respect them. Otherwise, the default rule in the `novathesis` template is to have in first place the abstract in *the same language as main text*, and then the abstract in *the other language*. For example, if the dissertation is written in Portuguese, the abstracts' order will be first Portuguese and then English, followed by the main text in Portuguese. If the dissertation is written in English, the abstracts' order will be first English and then Portuguese, followed by the main text in English. However, this order can be customized by adding one of the following to the file `5_packages.tex`.

```
\ntsetup{abstractorder={<LANG_1>, ..., <LANG_N>}}  
\ntsetup{abstractorder={<MAIN_LANG>={<LANG_1>, ..., <LANG_N>}}}
```

For example, for a main document written in German with abstracts written in German, English and Italian (by this order) use:

```
\ntsetup{abstractorder={de={de,en,it}}}
```

Concerning its contents, the abstracts should not exceed one page and may answer the following questions (it is essential to adapt to the usual practices of your scientific area):

1. What is the problem?
2. Why is this problem interesting/challenging?
3. What is the proposed approach/solution/contribution?
4. What results (implications/consequences) from the solution?

**Keywords:** One keyword · Another keyword · Yet another keyword · One keyword more · The last keyword

# RESUMO

Independentemente da língua em que a dissertação está escrita, geralmente esta contém pelo menos dois resumos: um resumo na mesma língua do texto principal e outro resumo numa outra língua.

A ordem dos resumos varia de acordo com a escola. Se a sua escola tiver regulamentos específicos sobre a ordem dos resumos, o template (L<sup>A</sup>T<sub>E</sub>X) **novathesis** irá respeitá-los. Caso contrário, a regra padrão no template **novathesis** é ter em primeiro lugar o resumo *no mesmo idioma do texto principal* e depois o resumo *no outro idioma*. Por exemplo, se a dissertação for escrita em português, a ordem dos resumos será primeiro o português e depois o inglês, seguido do texto principal em português. Se a dissertação for escrita em inglês, a ordem dos resumos será primeiro em inglês e depois em português, seguida do texto principal em inglês. No entanto, esse pedido pode ser personalizado adicionando um dos seguintes ao arquivo `5_packages.tex`.

```
\abstractorder(<MAIN_LANG>) := {<LANG_1>, ..., <LANG_N>}
```

Por exemplo, para um documento escrito em Alemão com resumos em Alemão, Inglês e Italiano (por esta ordem), pode usar-se:

```
\ntsetup{abstractorder={de={de,en,it}}}
```

Relativamente ao seu conteúdo, os resumos não devem ultrapassar uma página e frequentemente tentam responder às seguintes questões (é imprescindível a adaptação às práticas habituais da sua área científica):

1. Qual é o problema?
2. Porque é que é um problema interessante/desafiante?
3. Qual é a proposta de abordagem/solução?
4. Quais são as consequências/resultados da solução proposta?

**Palavras-chave:** Primeira palavra-chave · Outra palavra-chave · Mais uma palavra-chave · A última palavra-chave

# CONTENTS

<b>Acronyms</b>	vii
<b>1 Introduction</b>	1
1.1 Problem Statement and Motivation . . . . .	2
1.2 Objectives . . . . .	2
<b>2 State of the Art</b>	3
2.1 Fundamentals of Microstrip Patch Antennas . . . . .	3
2.1.1 Campos Radiados MPA Retangular . . . . .	3
2.1.2 Figures of Merit . . . . .	3
2.2 Sizing Methods . . . . .	6
2.2.1 Transmission Line Model . . . . .	6
2.2.2 Cavity Model . . . . .	6
2.2.3 Full Wave Model . . . . .	7
2.2.4 Exemplo MPA retangular (matemático?) . . . . .	7
2.3 Implementation State of the Art . . . . .	7
2.3.1 Geometry . . . . .	7
2.3.2 Bandwidth Enhancement Techniques . . . . .	8
2.3.3 Dual and Triple Band Methods . . . . .	10
2.3.4 Feed Methods . . . . .	13
2.3.5 AI Optimization . . . . .	15
<b>3 Work Plan</b>	16
3.0.1 Tasks . . . . .	16
3.0.2 Timeline . . . . .	16
3.0.3 Risk Management . . . . .	16
<b>4 Conclusions</b>	17
<b>Appendices</b>	

<b>A NOVAthesis covers showcase</b>	<b>18</b>
<b>B Appendix 2 Lorem Ipsum</b>	<b>19</b>
<b>Annexes</b>	
<b>I Annex 1 Lorem Ipsum</b>	<b>20</b>

## LIST OF FIGURES

## LIST OF TABLES

## ACRONYMS

**novathesis** NOVAthesis L<sup>A</sup>T<sub>E</sub>X (*pp. i, ii*)

## INTRODUCTION

Wireless communication systems are technologies that transfer information through the air without any physical wires [1]. This way of communicating can be achieved through the use of a wireless local area network, also known as WLAN [2]. A LAN is a network of computers linked together in a limited geographic area whilst a WLAN is the equivalent without any wire constraints. A WLAN is built on a cellular architecture that divides a system into cells with each being administered by a base station [3]. The benefits of using a WLAN are the mobility, flexibility, ease of management and maintenance and lowered cost [2]. There are multiple standards to a WLAN, also called Wi-Fi standards, some of them being 802.11n (Wi-Fi 4), 802.11ac (Wi-Fi 5), 802.11ax (Wi-Fi 6 and Wi-Fi 6E) and 802.11be (Wi-Fi 7), with these last two being the latest. 802.11ax offered greater throughput, better energy efficiency and enhanced performance in crowded environments [4] but with the increased demand for faster speeds and enhanced reliability in wireless communications [4] the Wi-Fi 7 is the most recent solution developed. The main improvements in Wi-Fi 7 that make it stand out from its predecessor are a channel bandwidth of up to 320 MHz that allows a more efficient and high-speed data transfer, the use of 4096-QAM modulation that lets the data that is transferred to be packed with more information in each radio wave and the introduction of the Multi-Link Operation (MLO), that grants users the ability to connect to multiple frequency bands, like 2.4 GHz, 5 GHz, 6 GHz at the same time, giving the user a stable connection with reduced delay[5]. One of the most common antenna technologies for Wi-Fi 7 and also the one this thesis will be focusing on is the Microstrip Patch Antenna also called patch antennas or MPAs. These are a type of antenna that consists of a metallic patch on a dielectric substrate with a ground plane on the other side. According to [6] the microstrip patch antenna was conceptualized in 1953 but only became practical by the 1970's when low loss substrate materials became more available. At that time patch antennas were being mostly used in spaceborne applications [7], in current times the MPAs are still used for aerospace applications and low-profile communications [8].

## 1.1 Problem Statement and Motivation

TO DO

## 1.2 Objectives

TO DO

## STATE OF THE ART

### 2.1 Fundamentals of Microstrip Patch Antennas

#### 2.1.1 Campos Radiados MPA Retangular

VANTAGENS E DESVANTAGENS DAS MPAS?

#### 2.1.2 Figures of Merit

The performance of an antenna, such as the MPA, can be characterized by numerous figures of metric although it isn't required to analyse them all to get a thorough description of the antenna as the importance of each parameter depends on the application of the antenna. It is important to note that some physical characteristics of the antennas can also influence these metrics such as the size of the antenna that according to [7] increasing the height of the antenna's substrate improves both the bandwidth and efficiency however there is an inherent trade-off with these changes as a taller antenna produces more surface waves which in turn degrade both the radiation pattern and polarization. The metrics that will be considered in this thesis will be described next in this chapter.

##### 2.1.2.1 Bandwidth

The bandwidth according to [7] is the range of frequencies where the antenna characteristics still have an acceptable value within the center frequency. There are two general ways we can classify an antenna based on its bandwidth: broadband and narrowband. Broadband antennas have a greater range of frequencies acceptable to the center frequency. Narrowband antennas have a smaller range of frequencies allowed on the bounds of the center frequency. A distinction is also made between the reasoning behind these frequencies limits as the antenna characteristics vary for different reasons, one of them being impedance and the other pattern. An impedance antenna is related to input impedance and the radiation efficiency while a pattern antenna is linked to the gain and the radiation pattern. Small antennas encounter problems related to their impedance bandwidth while

larger antennas face difficulties with their pattern bandwidth. If the antenna is neither large nor small, it can be influenced by either problem depending on its application.

### 2.1.2.2 Gain

The gain of the antenna according to [7] is the ratio of the intensity of the antenna in one direction by the intensity of the antenna if it was a perfect isotropic antenna. It combines both the antenna's efficiency and its directional capabilities. Usually, the relative gain is what is used. The relative gain has the same definition as the gain, but the values of the isotropic antenna are changed for the ones of a reference antenna whose gain is already known, keeping in mind that both antennas must have the same power input. In [7] two gains are defined: the gain and the realized gain. These can have the same value if the antenna is matched to the transmission line as the realized gain considers mismatch losses.

### 2.1.2.3 Directivity

The directivity is defined as the ratio between the radiation intensity in a specific direction and the radiation intensity averaged in all directions according to [7]. If there is no direction specified, the directivity refers to the direction where the radiation intensity is maximum. The directivity is relative as it compares the directional properties of the antenna to an isotropic one whose directivity is one. [7] mentions that the total directivity for antennas with orthogonal polarization components is the sum of partial directivities of any two orthogonal polarizations. These partial directivities are the radiation intensity in the specific polarization divided by the total average radiation over all directions. Different antennas can have different patterns, and each pattern has a way to approximate its directivity. For directional antennas, who have a single major lobe and insignificant minor lobes, the directivity is based on the Half-power beamwidth (HPBW) in two perpendicular planes. For omnidirectional antennas there are two different ways to approximate the directivity: the McDonald function and the Pozar function. The McDonald function is usually more accurate for patterns with minor lobes. The Pozar function is more accurate for patterns with little to none minor lobes. However, these approximations won't work well if there is more than one major lobe or if there are any significant minor lobes.

### 2.1.2.4 Efficiency

The radiation efficiency, or just efficiency, is defined by [12] as the ratio of the radiated power by the power received by the antenna. In [7] the radiation efficiency is called  $e_{cd}$  and it considers the conduction and dielectric losses a single measurement called loss resistance, and the ratio is of the power sent to radiation resistance by the power sent to both the radiation resistance and the loss resistance.

### 2.1.2.5 Radiation Pattern

The radiation pattern is defined by [7] as a graphical representation in space coordinates of different radiation properties such as directivity, intensity and polarization. It is usually determined in the far-field region in directional coordinates. [7] tells us there are three ways to visualize the signal: field pattern (linear scale), power pattern (linear scale) and power pattern (dB). The field pattern represents the strength of the electric or magnetic field in relation to the angular space. The power pattern in linear scale characterizes the square of the electric or magnetic field also in relation to the angular space and the power pattern in dBs depicts the electric or magnetic field in dBs in relation to the angular space.

The radiation pattern is represented by lobes that are separated by regions of low radiation intensity. As stated in [7] these lobes can be subclassified in major/main lobe, minor lobe, side lobe and back lobe. The main lobe is the lobe that contains the maximum signal strength. It is possible to have more than one major lobe, [7] mentions split-beam antennas as an example of this. The minor lobe is all the other lobes not considered major and represents the radiation in any other direction as the intended one. The side lobes are a subcategory of these are the lobes right next to the main ones. The back lobe is the minor lobe that is exactly  $180^\circ$  away from the main lobe. All minor lobes should be minimized.

To properly analyze the radiation pattern, we must define planes to view it. These planes are the E-plane that is parallel to the electric-field vector and the H-plane that is parallel to the magnetic-field vector. [7] explains that with most antennas it is common to align one of the principal plane patterns with one of the geometrical principal planes. An example mentioned in [7] is of the omnidirectional pattern that as an infinite number of principal E-planes and only one principal H-plane.

According to [7] there are three regions around the antenna: the reactive near-field, the radiating near-field and the far-field. The reactive near-field is the region immediately touching the antenna where the reactive field predominates and the angular field distribution is evenly spread with minor variations. The radiating near-field is the region in between where the reactive field still predominates and depending on the distance from the antenna the pattern may begin to form. For antennas that are roughly the same size as the wavelength and antennas that are very small in size compared to the wavelength this region may not be present. The far-field, where we observe the radiation pattern graph, the angular distribution is stable, and we are able to see the different lobes.

To measure the radiation patterns size we need to determine its angles both in the plane field and in the solid field. The radian lets us calculate the width of a beam in the plane field and the steradian lets us calculate the beam's solid angle.

An antenna can have different classifications based on its radiation pattern, [7] describes three. The first one being an isotropic antenna, it is an antenna who emits radiation equally in every direction with no losses. It is only a theoretical concept, but it is used

as a reference for real antennas to see how directive they are. The next is a directional antenna, it is an antenna that has one direction where its directivity is much greater than in any other direction, the reference point for these antennas are the half-wave dipole antennas. The last antenna is the omnidirectional antenna whose pattern on the azimuth plane (H-plane) is nondirectional and on the elevation plane (E-plane) is directional, its shape is commonly known as a donut.

#### 2.1.2.6 Impedance

TO DO

## 2.2 Sizing Methods

Sizing methods according to [7], [8] and [10] are analytical models used as a first step in simulations to determine the physical parameters of the antenna such as the length, width, size of the patch and substrate parameters by evaluating input impedance, pattern, and bandwidth. The most popular sizing methods are the transmission line model, the cavity model and the full wave model.

### 2.2.1 Transmission Line Model

The transmission line model, as stated by [7], is the easiest one and it gives good physical insight, but it is less accurate and lacks versatility. [8] states that this method models the interior region of the patch antenna as a section of the transmission line and determines the characteristic impedance and the propagation constrain by the patch size and substrate parameters. [8] defines the rectangular patch by four edges where the variation of the field defines if the edge is a radiating type or not. It also uses the mode  $TM_{10}$  as an example explaining that the edges at  $x = 0$  and  $L$  are radiating types because along them the electric field is uniform and the walls at  $y = 0$  and  $W$  are non-radiating types because along them due to half-wave variation. For this reason, [8] states that the patch antenna's radiation patterns are treated as an array of two narrow slots separated by the path length and the input admittance is obtained by transforming these edges admittances to the feed point.

### 2.2.2 Cavity Model

The cavity model as opposed to the transmission line model is not restricted to rectangular patches and one-dimensional variations. In this model, according to [7], the region between the patch and the ground plane is treated as a cavity bounded electric conductors above and below and magnetic walls around. [8] tells us that this model relies on the assumption that because the substrates is very thin the fields in the interior region do not vary with  $z$ . To accurately determine these fields [8] tells us a good assumption to

make is that the wall all around the periphery of the patch is magnetic. For thin low dielectric constant this wall is placed at a distance approximately equal to its thickness. The problem with this model is that even though the interior fields are easy to determine this is only correct for the first order because the effect of the exterior fields has been excluded from the interior field determination. Despite this, it is still easy to determine the interior electric field distribution in terms of the cavity's eigenfunctions.

### 2.2.3 Full Wave Model

The full wave model is that by [8] to be able to overcome the limitations of the other two models. This model includes the effects of dielectric loss, conductor loss, space wave radiation, surface waves and external coupling and removing the simplifying assumptions made by the other models. It maintains accuracy by enforcing the boundary conditions at the air-dielectric interface using Green's functions or differential forms of Maxwell's equations. This method is very versatile as it provides the most accurate solution for the impedance and radiation characteristics, it can size arbitrary shapes, arrays and stacked elements. However, as it is numerically intensive and complex it requires a big computational cost and provides less immediate results. Common popular full-wave techniques, according to [8] are spectral-domain full-wave solution, mixed-potential electric field integral equation approach and finite-difference time-domain (FDTD).

### 2.2.4 Exemplo MPA retangular (matemático?)

TO DO

## 2.3 Implementation State of the Art

### 2.3.1 Geometry

An important factor in the design of a patch antenna is its geometry. As said before, MPAs are composed by a metallic patch and a dielectric substrate, for the substrates choice as stated in [7] what is sought is a thick one with a dielectric constrain in the lower end of the usual range ( $2.2 \leq \epsilon_r \leq 12$ ) as for the patch there are a variety of different shapes to pick from the most basic shapes such as rectangular, circular and triangular to some other unique shapes like flowers, trees or butterflies as seen in [13]. The rectangular patch is considered by [8] as the most basic because it is easy to fabricate, simple to analyze and can be used for a wide range of applications. The circular patch can be seen also by [8] as an alternative to the basic rectangular shape in certain applications, namely arrays, as it is slightly smaller than it. The triangular shape is stated by [14] to being an alternative to both the rectangular and circular geometries when the priority miniaturization as it occupies less metalized area on the substrate than the previous. However, at the present time these simpler designs don't satisfy the need for multi-band operation to accommodate

multiple services [7], to get this we need modified antenna shapes. Some methods with modified shapes that let multiple frequencies resonate are slotted patches, that allow for dual-band when a U-slot is cut in one of the patches and for triple-band when another U-slot is cut into either patch, as described in [15]. Stacked microstrip patch antennas which, as mentioned in [13], consist of different layers of dielectric material and patches that allow for a broader bandwidth and gain, however as mentioned before the increase in height causes degradation of the radiation pattern. Defected ground structure (DGS), as it is explained in [13], is used on the ground plane changing its paths length, as well as its impedance and capacity values resulting in a larger bandwidth. Fractal geometry, described in [7], as a recursively generated geometry that uses an iterative process which leads to structures that are self-similar and self-affine that provide multiple current paths of different lengths leading to multiple resonant frequencies.

### **2.3.2 Bandwidth Enhancement Techniques**

The microstrip patch antenna, though widely used, still has the major disadvantages of a narrow bandwidth and low gain. A lot of research has been done to improve these problems while keeping their compact structure, such as in [20]. In this section we will talk about some of these techniques.

#### **2.3.2.1 Slot loading**

In the slot technique slots of various shapes such as U or L [16] or of irregular shapes [18], are implanted on the antenna's patch and strategically alter the current path length and impedance improving its matching and allowing for a wider bandwidth without increasing the antenna's size. Adding more slots adds new resonate frequencies and therefore the bandwidth is enhanced [18]. The slot method however can introduce structural instability, surface wave loss [16] and decreased the radiation efficiency [17]. This method can show a decrease [17] in gain but depending on the implementation can also there can also be an increase [19] in it.

#### **2.3.2.2 Defected Ground Structure**

In the defective ground structure technique, a defect is engraved on the ground plane of the beneath the patch, this engraving disrupts the surface wave propagation across the substrate layer. The defects can be simple or more complex depending on the desired performance [18] [19]. This method improves the bandwidth and enhances the radiation efficiency [16] and it can be used in transmission lines, power amplifiers, oscillators and more mentioned in [18] and [19]. The trade-off of this technique is stated in [16] and is the weakening of the ground plane's robustness and it adds complexity to the fabrication process.

### 2.3.2.3 Advanced Material Integrations

Another bandwidth enhancement method is advanced material integrations [16] that with its integration improve gain and enhance bandwidth. [16] provides us with two examples, the first one being the frequency selective surface (FSS) layers that introduce bandwidth filtering and the other which is also mentioned in [19] is metamaterials that improve antenna radiation performance as it minimizes the side lobe ratio and improves directivity. However, according to [16] this method increases the antenna's thickness and fabrication complexity as well as the cost.

### 2.3.2.4 Feeding Techniques

The next method for improving the bandwidth is the feeding technique. This technique has multiple implementations, and they are Coplanar Waveguide-fed, multiple feeding technique and dual feed technique. The CPW-fed uses a center conductor between two ground planes to feed the antenna. Its advantages and disadvantages are mentioned in [16]: it makes fabrication simple and keeps the antenna compact but without a design optimization it can have radiation losses and lack of impedance match. The multiple feeding technique is as stated in the name the use of multiple feed lines to feed the antenna in multiple points. According to [17] this causes an increase in the radiation efficiency and impedance match. It enhances the bandwidth because it creates multiple resonance frequencies close to one another. The dual feed structure is mentioned in both [18] and [19] and it works by having two feed points in one or two patches and it is used to control the vertical mode and prevent other modes from being excited. It prevents polarization degradation, helps with impedance matching and improves gain. Some feeding methods not included in this category that also increase bandwidth will be explained further in the feeding methods part of this thesis.

### 2.3.2.5 Parasitic Patch

The parasitic patch method is a technique that works by adding extra radiating patches that are not being fed near the main radiating patch, allowing them to couple with the main patch causing multiple frequencies to resonate close to each other enhancing the bandwidth [17]. It is used to improve gain according to [18] and [19] and it has better impedance matching and radiation efficiency according to [17]. This technique can have two configurations that are mentioned in both [18] and [19]: the coplanar technique and the stacked technique. In the coplanar technique multiple patches are coupled on a single plane above the dielectric substrate and one of them is given radiation, becoming the main patch. In the stacked technique the patches are stacked vertically with a dielectric layer in between which allows the patches to share common aperture area. In this last configuration the parasitic patches with lower dielectric constant are added on top of the radiating patch which gives us the advantages mentioned before while maintaining its

physical size.

#### 2.3.2.6 Air Gap

The air gap method is a method that utilizes air as the substrate in between the ground and the radiating patch. According to [17] it is used to overcome the limitations of high dielectric materials such as lower radiation efficiency and gain and narrower bandwidth. The use of air as a dielectric substrate, stated by [18] and [19] allows for an effective radiation pattern and low return loss. This happens because air has lower permittivity. According to [18] this technique creates more directive antennas. It is also explained in [18] that using this technique in a single patch and an aperture-coupled antenna allows for two frequencies to resonate in the structure. Both [18] and [19] state that increasing the height of the air gap causes the distance between the two frequencies to decrease which is useful for bandwidth enhancement and gain improvement.

#### 2.3.2.7 Shorting Pin

The shorting pin technique according to [18] and [19] works by adding a pin to the patch antenna which leads to a decrease in the resonating frequency while maintaining the antenna's size. Both references state that to increase high impedance matching the pin must be placed close to the antenna's feed point. [18] states that the shorting pin model allows for lower resonance frequency which in turn allow for a bigger degree of miniaturization for a fixed frequency. In addition to improving both bandwidth and gain this method helps decrease cross-polarization levels.

#### 2.3.2.8 Dielectric Substrate

According to [18] and [19] the primary function of the substrate is to provide strength to the antenna's structure the choice of material is essential as it directly impacts some critical parameters such as bandwidth, efficiency and radiation pattern. [19] states that low dielectric constants usually provide excellent performance in terms of higher bandwidth in comparison to high substrate constants. [18] and [19] say that the propagation of surface waves state can be reduced by choosing the correct dielectric substrate. Furthermore, [19] states that the bandwidth can be improved by using multilayer dielectrics.

### 2.3.3 Dual and Triple Band Methods

The need for multiband microstrip patch antennas was already mentioned in the geometry section, continuing that idea is [21] that also mentions it's need in compact communication devices and systems to allow for multiple wireless communication standards in a single antenna system. Different methods to achieve dual and tri-band antennas are through the geometry of the antenna, through feeding techniques, by adding multiple resonant elements, through loading and traps, through metamaterials or through hybrid techniques.

### 2.3.3.1 Geometry

The geometry techniques are slots, previously mentioned, such as U, E and H, which work by having secondary resonant paths allowing for dual-band if there is only one slot, and tri-band if another one is added this fact is stated in [15]. In [22] in addition to an U-shaped slot we can see I-shaped and T-shaped slots. The next geometry method was also mentioned before, and it is fractal geometry whose patterns allow for multiple paths of different sizes. In [23] they mention the most common fractal geometries which are: Hilbert curve, Sierpinski gasket, and Koch snowflake and in addition to those the paper focuses on an H-fractal antenna. In [24] Sierpinski gasket is the topic of study, and it also states that fractal antennas are inherently multiband. The last geometry method is the meandered fork-shapes which works as a combination of the action of meandering, meaning the arms of the fork are folded to achieve miniaturization, and a fork shaped antenna. The benefit of meandering stated in [25] is that it increases the electrical length of the antenna without increasing its size. Each arm of the fork-shaped antenna corresponds to a different resonance, and each stub corresponds to a specific frequency band, this is explained by [26]. The benefit of this fork shape is that each stub can be individually adjusted allowing to shift a frequency without compromising the rest.

### 2.3.3.2 Feed

The feeding techniques are three, two of them will be further explained in the feed methods chapter of this thesis and the third was already talked about in the bandwidth enhancement methods. Those are aperture coupling, proximity coupling and dual feed techniques. For the aperture coupling [27] explains that it is a non-contacting feeding method where the feed line and the patch aren't directly connected. In the same paper it is stated that with this feeding method an antenna can operate at dual frequency just by modifying. Finally on that same paper, it is said that dual frequency aperture coupled antennas can substitute large bandwidth planar antennas, as having two distinct bands allows for separating of transmit-receive functions. For the proximity coupling, also a non-contacting feeding method, where two substrates are used where one as the feeding line and the other the radiating patch. In [28] proximity coupling is used in an antenna with two spiral strips of different lengths which generate different frequencies based on its dimensions allowing for a tri-band antenna with omnidirectional radiation patterns. For the last method, the dual feed, as it was explained in a previous chapter, works by having two feed lines connected to the same patch in different points. In [29] this method is obtained by using a planar feed and a coupled feed to excite different frequencies, these frequencies don't interfere with each other as they are electrically different and physically separated.

### 2.3.3.3 Multiple Resonant Elements

For multiple resonant elements we will only talk about stacked patches and coplanar parasitic elements. Stacked patches work by stacking extra radiating patches that are not fed but get energy by coupling with the main patch. They were already talked about in the bandwidth enhancement methods chapter as they work for both broad banding and multi banding, more specifically for dual band as stated in [30] by having the two stacked patches have different relative dimensions. Coplanar parasitic elements work by placing other elements on the same layer as the patch which is the only one being fed. This works as the parasitic elements stated before in the bandwidth enhancement chapter. An example of this is in [31] where two U-shaped parasitic elements are placed around the patch. In this paper the two elements are responsible for the second and third frequency bands, as they have different sizes, they vibrate at different frequencies.

### 2.3.3.4 Loading/Traps

Loading techniques work by adding structures that “load” the antenna. This can be done through reactive loading, shorting pins or DGS, the last one was also mentioned in the bandwidth enhancement methods. With reactive loading we can achieve a dual-band antenna like in [32] and it works by loading the patch with a rectangular slot in a specific position and size this creates a new path of different electrical length allowing for a second frequency to resonate. With shorting pins, we can achieve a dual-band antenna like in [33] by placing physical metallic pins connecting the radiating patch and the ground plane. These pins are inserted in a specific location to tune the input impedance of the  $TM_{01}$  mode, and the feed is placed to provide the desired input impedance of the  $TM_{03}$  mode making the antenna support two different resonant modes.

### 2.3.3.5 Metamaterials

For metamaterials, which are layers placed above the antenna radiator used to redirect, focus or create new electromagnetic resonances, we have complementary split ring resonators and superstrates. Complimentary split ring resonators, or just CSRR, exhibit negative permittivity and permeability. In [34] the CSRR are embedded on the ground plane and act like an artificial resonator, which allows for multiband design as the new resonances are created by the excitation of the CSRR. In the same paper we can see the multiband designs, as it is shown that adding one CSRR introduces a new lower frequency to the patch making it dual band, and adding two CSRR of different sizes introduces two new unique frequencies making it tri-band. Superstrates such as the superstrate metal ring used in [35], work by “splitting” a single band into two bands. A resonating lower band because of the radius of the metal ring and a higher band due to the induced electrical fields between the patch and the superstrate.

### 2.3.3.6 Hybrid

The final method to achieve dual and triband antennas is a hybrid method. Hybrid methods are a combination of two or more of the methods stated, such as in paper [29] it is used dual feed to get two different resonant modes and to achieve high isolation in a small antenna a T-shaped slot is cut into the radiating patch and a L-shaped slot is cut into the ground patch. The combination of these two techniques allows for a good isolation from signal leaks onto each port resulting in high radiation efficiency of the dual band antenna. [31] can also be considered a hybrid method as it combines both parasitic elements and U-slots. DGS which were mentioned in the bandwidth enhancement chapter could also work for multi bands as it can introduce new resonant frequencies however as in [36] it usually doesn't work alone making it a hybrid method. In [37] aperture coupling is used in combination with air gap stacked patches which work for dual-band and possibly tri-band with the aperture coupled patch creating the lower band and the stacked patch creating the second, the air gap can tune the resonant frequencies.

### 2.3.3.7 Use Cases

TO DO

## 2.3.4 Feed Methods

For the feeding of the microstrip antenna there are many configurations that can be used, the four most popular ones according to [7] are the microstrip line, the coaxial probe, the aperture coupling and the proximity coupling. To choose the appropriate feeding technique we have to take into consideration some factors. [8] considers the efficiency of power transfer between the feed line and the antenna one of those factors, because if there is no impedance matching there will be surface wave loss and some unintended radiation leaking from the feed leading to bigger side lobes. A good choice of feeding method is one that minimizes this unintended radiation according to [8].

### 2.3.4.1 Microstrip Line Feed

The microstrip feed line, also known as inset feed, is a narrow strip of metal, much thinner than the patch itself that connects directly to its edge. The inset feed advantages, mentioned in [7] are the ease of fabrication, simple to match (just adjust the inset position) and simple to model. The disadvantages are also mentioned in [7] where it is explained that with the thickness increase of the substrate there will be an increase in unintended radiation and surface waves, these factors will limit the usable bandwidth to a very narrow range.

### 2.3.4.2 Coaxial Probe Feed

The coaxial probe feed is also easy to fabricate and match. A description of the feed is presented in [8] where it is said that the center coaxial conductor passes through the substrate and is soldered to the antenna patch where the best impedance match is achieved and because of this it has low unintended radiation. Its disadvantages are mentioned in [7] them being also having a narrow bandwidth and being difficult to model. [7] mentions another problem shared by both the microstrip feed line and the coaxial probe feed: they possess inherent asymmetries because they connect to the patch at a single point or edge. This problem generates higher order modes which in turn produce cross-polarized radiation.

### 2.3.4.3 Aperture Coupled Feed

To solve the problems of narrow bandwidth, unintended radiation and cross polarization presented by both the microstrip feed line and the coaxial probe feed the aperture coupling microstrip feed was introduced. The aperture coupling feed is a non-contacting technique that according to [7] consists of two substrates separated by a ground plane. A microstrip feed line is located on the bottom side of the lower substrate and the radiation patch is on the top side of the upper substrate. The feed line coupled its energy to the patch through the slot between the substrates. The advantages of this feed can be taken from [7] and [8] and they are improved bandwidth, independent optimization of the feed mechanism and of the radiation element, it prevents united radiation and improves polarization purity because the ground plane creates a shielding effect. The disadvantages of this feed line can be found in [7] and they are the increased difficulty of fabrication and the added complexity to match.

### 2.3.4.4 Proximity Coupled Feed

The last feeding method mentioned was the proximity coupled microstrip feed, which is [8] also mentioned as electromagnetically coupled microstrip feed. This is also a non-contacting feed technique that uses two substrate layers where the lower layer has the feeding line, and the top layer has the radiating patch. In this case the energy is capacitively transferred through the substrate layers. The advantages of this method stated in [7] and [8] and they are larger bandwidth than the rest, reduced unintended radiation, as there is no soldering cross polarization and pattern degradation problems are reduced and it has flexibility in the design. The disadvantages are also stated in both books, and they are the difficulty in fabrication and the lack of the shielding effect created by the slot in the aperture coupling feed method.

### 2.3.5 AI Optimization

Usually, to find the right antenna dimensions for dual and tri-band antennas engineers can use mathematical formulas or even trial and error, the problem with these methods is the many variables an antenna has, making it hard and time consuming to find the perfect antenna combination. A solution to this problem is the use of AI and machine learning to help us find the best design for our needs.

#### 2.3.5.1 Genetic Algorithm

An AI method used for optimization is binary-coded genetic algorithm, an example of this is [38] where it is used to improve a tri-band antenna. The genetic algorithm (GA) works by making iterations including selection, reproduction and variation: it starts with random antenna designs and after testing each design, it selects the ones with the best tri-band response and combines them to get a new generation of antennas, it also adds random changes so maintain diversity. The binary-coded genetic algorithm in [38] works by adding or removing radiating cells from the patch allowing for non-intuitive shapes. This method reduces complexity and residual radiation and improves the impedance bandwidth, directivity and gain. However, this method needs significant computation resources and several interactions making the design cycles longer.

#### 2.3.5.2 Support Vector Regression

The method to solve this is stated in [39] and uses machine learning is the support vector regression (SVR), a supervised machine learning technique, predicts performance and optimizes the design parameters of the antenna more efficiently and it is best suited for complex and nonlinear relations between the antenna's physical dimensions and its performance parameters. SVR works by first running many simulations to create a dataset, after this training is done the model can instantly calculate the best dimensions to achieve any specific dual or tri-band antenna. The advantages of this method, according to [39] are the speeding of the optimization process and the minimizes the necessity for lengthy simulations.

## WORK PLAN

### **3.0.1 Tasks**

TO DO

### **3.0.2 Timeline**

TO DO

### **3.0.3 Risk Management**

TO DO

# 4

## CONCLUSIONS

TODO

A

*NOVATHESIS COVERS SHOWCASE*

B

## APPENDIX 2 LOREM IPSUM

I

## ANNEX 1 LOREM IPSUM

