

## Nyquist Stability Criteria

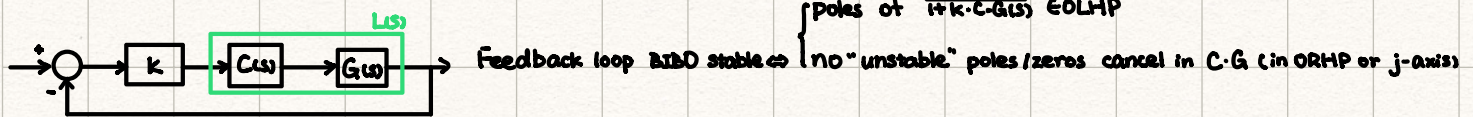
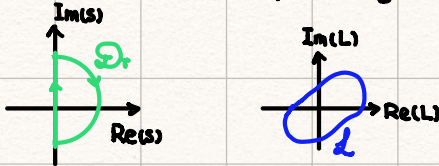


Fig. 1  $k$ -variable gain (const.) For  $k=1$ , unit feedback loop

want a graphical test for this

We'll use the Principle of Arg. for a special TF and for a special contour  $\mathcal{D}_r$  called Nyquist contour. (all in ORHP &  $j$ -axis)



Let:  $L(s) = C \cdot G(s)$  and  $\mathcal{L} = \{L(s) \mid s \in \mathcal{D}_r\}$  be the Nyquist plot of  $L(s)$  (image of  $\mathcal{D}_r$  under  $L(s)$ )

**THM:** Suppose  $L(s) = C \cdot G(s)$  a strictly proper TF (rational) which has NO poles on  $j$ -axis and has  $n$  poles in ORHP

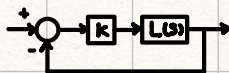
(in  $\{Re(s) > 0\}$ ) and suppose  $C \cdot G$  has no "unstable" pole-zero cancellation (in  $\{Re(s) > 0\}$ ) (ORHP,  $j$ -axis)

Then the feedback loop (or the closed loop (c.l.) system) is BIBO stable if and only if the Nyquist plot  $\mathcal{L}$  of  $L(s)$ :

(i) does NOT pass through the point  $-\frac{1}{k}$  ("critical point")

(ii) it encircles  $-\frac{1}{k}$  a # of times  $N = n$  c.w.

**Note:** feedback loop  $\Leftrightarrow$   $\left\{ \begin{array}{l} \text{poles of } 1/(1+K \cdot C \cdot G(s)) \in \text{OLHP} \\ \text{no "unstable" poles/zeros cancel in } C \cdot G \text{ (in ORHP or } j\text{-axis)} \end{array} \right.$



c.l. poles  $\equiv$  zeros of  $(1 + k \cdot L(s)) = \tilde{L}_k(s) \in \text{OLHP}$

"O.L. poles"  $\equiv$  poles of  $\tilde{L}_k(s) \equiv$  poles of  $L(s)$

**Note:** if  $\mathcal{L}$  passes through  $-\frac{1}{k} \Rightarrow \exists s \in \mathcal{D}_r$  s.t.  $L(s) = -\frac{1}{k} \Rightarrow 1 + kL(s) = 0 = \tilde{L}_k(s)$

$s \in \mathcal{D}_r \Rightarrow \left\{ \begin{array}{l} j\text{-axis} \\ \text{big semi-circles in ORHP} \end{array} \right. \Rightarrow \text{c.l. poles} \in s \Rightarrow \text{ILLEGAL}$

Apply Pr. of Arg to  $\tilde{L}_k(s)$ :

$N = \#$  of encircle of the origin in  $\tilde{L}$ -plane  $= \# \text{poles inside } \mathcal{D}_r - \# \text{zeros inside } \mathcal{D}_r \rightarrow \text{should be } 0$   
(ORHP) (ORHP)

$N = n$  c.w

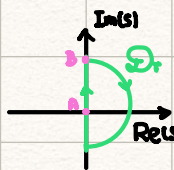
But:  $\tilde{L}_k(s) = 1 + kL(s) \Rightarrow \frac{1}{k} \tilde{L}_k(s) = \frac{1}{k} + L(s) \Rightarrow -\frac{1}{k} + \frac{1}{k} \tilde{L}_k(s) = L(s) \Rightarrow \text{the origin in } \tilde{L} \text{ plane} \equiv -\frac{1}{k} \text{ in the } L\text{-plane}$   
(they are translated / scaled)

Example:  $L(s) = \frac{1}{s-1}$   $\mathcal{D}_r = \{ \text{portions} \} : \textcircled{1} j\text{-axis} \quad \textcircled{2} \text{big semi-circle}$

$\textcircled{1} s = j\omega, \omega: [0, +\infty)$  (A,B)

$\textcircled{2} \text{big semi-circle}$



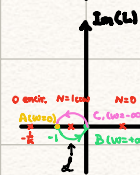


$$L(s) = \frac{1}{s(s-1)} = \frac{1}{(s-0)(s-1)} = \frac{1}{s(s-1)} = \frac{1}{s^2 - s} = \frac{1}{s^2} \cdot \frac{1}{1 - \frac{1}{s}} = \frac{1}{s^2} \cdot \left( 1 + \frac{1}{s} + \frac{1}{s^2} + \dots \right)$$

Re[L(jw)] Im[L(jw)]



For  $s = jw, w: (-\infty, 0]$



$$L(-jw) = \frac{1}{-jw-1} = \frac{1}{-(jw+1)} = -\frac{1}{jw+1} = -\frac{1}{1+jw} = -\frac{1}{1+jw} \cdot \frac{1-jw}{1-jw} = -\frac{1-jw}{1-w^2} = \frac{jw-1}{1-w^2}$$

$\Rightarrow$  by reflection



$$\odot s = r \cdot e^{j\theta}, r \gg 1; L(s) = \frac{1}{s(s-1)} \approx \frac{1}{s^2} = \frac{1}{r^2} e^{-j2\theta} \approx 0$$

$n=1 \Rightarrow$  By Nyquist.  $-1 < -\frac{1}{k} < 0 \Rightarrow k > 1$