

a) Sea  $x_1(t) = A e^{j\omega_0 t}$ ,  $x_2(t) = B e^{j5\omega_0 t}$ ,  $\omega_0 = \frac{2\pi}{T}$   
 $T, A, B \in \mathbb{R}^+$

Determinar la distancia entre las dos señales.

Para encontrar la distancia resolvemos  $\lim_{T \rightarrow \infty} \frac{1}{T} \int_T |x_1(t) - x_2(t)|^2 dt$

Expandimos  $|x_1(t) - x_2(t)|^2 = x_1^2(t) - 2x_1(t)x_2(t) + x_2^2(t)$

$\lim_{T \rightarrow \infty} \frac{1}{T} \int_T (|x_1(t)|^2 - 2x_1(t)x_2(t) + |x_2(t)|^2) dt$ , repartimos la integral

$\lim_{T \rightarrow \infty} \frac{1}{T} \left[ \int_T |x_1(t)|^2 dt - 2 \int_T x_1(t)x_2(t) dt + \int_T |x_2(t)|^2 dt \right]$

$\lim_{T \rightarrow \infty} P_{x_1} - \frac{2}{T} \int_T x_1(t)x_2(t) dt - P_{x_2}$

donde  $P_{x_1} = \frac{1}{T} \int_T |x_1(t)|^2 dt = \frac{1}{T} \int_T (A e^{j\omega_0 t})^2 dt$

$\frac{1}{T} \int_T A^2 (e^{j\omega_0 t})^2 dt$ ,  $(e^{j\omega_0 t})^2 = e^{j\omega_0 t} (e^{j\omega_0 t})^*$

$\frac{A^2}{T} \int_T e^{j\omega_0 t} (e^{-j\omega_0 t}) dt$ ,  $e^a e^b = e^{a+b}$

$\frac{A^2}{T} \int_T e^{j\omega_0 t - j\omega_0 t} dt = \frac{A^2}{T} \int_T e^0 dt$  evaluamos desde  $-T/2$  a  $T/2$

$= \frac{A^2}{T} t \Big|_{-T/2}^{T/2} = \frac{A^2}{T} (T/2 - (-T/2)) = \frac{A^2}{T} T$

Por lo tanto  $P_{x_1} = A^2$

Ahora  $P_{x_2} = \frac{1}{T} \int_T |x_2(t)|^2 dt = \frac{1}{T} \int_T |B e^{j5\omega_0 t}|^2 dt$

$$\frac{1}{T} \int_T B^2 (e^{j\omega_0 t})^2 dt, (e^{j\omega_0 t})^2 = e^{j\omega_0 t} (e^{j\omega_0 t})^*$$

$$\frac{B^2}{T} \int_T e^{j\omega_0 t} e^{-j\omega_0 t} dt = \frac{B^2}{T} \int_T dt = \frac{B^2}{T} t \Big|_{-T/2}^{T/2} = \frac{B^2}{T} (T/2 - (-T/2))$$

$$P_{x_2} = B^2$$

Ahora:  $-\frac{2}{T} \int x_1(t) x_2(t) dt = -\frac{2}{T} \int_T A e^{j\omega_0 t} B e^{j\omega_0 t} dt$

$$-\frac{2AB}{T} \int_T e^{j\omega_0 t + j\omega_0 t} dt = -\frac{2AB}{T} \int_T e^{j6\omega_0 t} dt$$

realizamos la sustitución  $u = j6\omega_0 t$ ,  $du = j6\omega_0 dt$

$dt = \frac{du}{j6\omega_0}$ , Calculamos los nuevos límites de integración

teniendo en cuenta que los originales son  $\int_{-T/2}^{T/2}$

Si  $t = -T/2$ ,  $u = j6\omega_0(-T/2)$  , Si  $t = T/2$ ,  $u = j6\omega_0(T/2)$  reemplazamos  $\omega_0 = \frac{2\pi}{T}$

$$u = j6 \frac{2\pi}{T} \left(-\frac{T}{2}\right) \quad u = j6 \frac{2\pi}{T} \left(\frac{T}{2}\right)$$

$$u = -j6\pi$$

$$u = j6\pi$$

$$-\frac{2AB}{T} \int_{-j6\pi}^{j6\pi} \frac{e^u}{j6\frac{2\pi}{T}} du = -\frac{2AB}{T} \frac{j6\pi}{j6\frac{2\pi}{T}} [e^{j6\pi} - e^{-j6\pi}]$$

$$-\frac{2AB}{j12\pi} \left[ (\cos(6\pi) + j\sin(6\pi)) - (\cos(6\pi) - j\sin(6\pi)) \right]$$

$$\frac{-2AB}{j12M} \left[ (-1 + j0) - (1 - j0) \right]$$

$$\frac{-2AB}{j12M} \left[ 1 - 1 \right] = \frac{-2AB}{j12M} \left[ 0 \right] = 0$$

$$\frac{2}{T} \int_T x_1(t) x_2(t) dt = 0$$

Entonces  $\lim_{T \rightarrow \infty} \frac{1}{T} \int_T |x_1(t) - x_2(t)|^2 dt$

$$= \lim_{T \rightarrow \infty} P_{x_1} - \frac{2}{T} \int_T x_1(t) x_2(t) dt - P_{x_2} = A^2 - 0 + B^2$$

$$\overline{P_{x_1 - x_2}} = A^2 + B^2$$

$$b) X(t) = 3 \cos(1000\pi t) + 5 \sin(2000\pi t) + 10 \cos(11000\pi t) \quad F_s = 5000$$

Tomamos la frecuencia angular mayor entre las tres señales.

$$\omega_1 = 1000\pi, \quad \omega_2 = 2000\pi, \quad \omega_3 = 11000\pi \quad ; \quad \omega = 2\pi F$$

$$F_1 = \frac{1000\pi}{2\pi}, \quad F_2 = \frac{2000\pi}{2\pi}, \quad F_3 = \frac{11000\pi}{2\pi}$$

$$F_1 = 500, \quad F_2 = 1000, \quad F_3 = 5500$$

En este caso  $\omega_3 = 11000\pi$ , partiendo de  $\omega = 2\pi F = \frac{2\pi}{T}$  despejamos T.

$$T = \frac{2\pi}{\omega} = \frac{2\pi}{11000\pi} = \frac{1}{5500} \quad \text{ahora despejamos F.}$$

$$F = \frac{1}{T} = \frac{1}{\frac{1}{5500}} = 5500 \text{ Hz}$$

La relación de Nyquist nos dice:  $F_s \geq 2F$

$$5000 \geq 2(5500) \rightarrow 5000 \geq 11000 \quad \text{No se cumple}$$

La frecuencia de muestreo del conversor análogo digital no será suficiente para representar adecuadamente la señal.

### Discreticemos la señal

Como una computadora no puede recibir infinitos valores de tiempo continuos, realizamos el cambio  $t = nT_s$ , pero  $T_s = \frac{1}{F_s}$  entonces  $t = \frac{n}{F_s}$ , por el conversor análogo digital  $F_s = 5000 \text{ Hz}$

reemplazamos  $t = \frac{n}{5K}$  en la señal original

$$x\left[\frac{n}{F_s}\right] = 3 \cos\left(\frac{1K\pi}{5K}n\right) + 5 \sin\left(\frac{2K\pi}{5K}n\right) + 10 \cos\left(\frac{11K\pi}{5K}n\right)$$

$$x\left[\frac{n}{F_s}\right] = 3 \cos\left(\frac{\pi}{5}n\right) + 5 \sin\left(\frac{2\pi}{5}n\right) + 10 \cos\left(\frac{11\pi}{5}n\right)$$

la señal  $10 \cos\left(\frac{11\pi}{5}n\right)$  donde  $\Omega = \frac{11\pi}{5}$  es una copia porque  $\frac{11\pi}{5}$  está fuera del rango  $[-\pi, \pi]$  para llegar al original hacemos  $\Omega_{\text{original}} = \Omega_{\text{copia}} - 2\pi$

$$\Omega_{\text{original}} = \frac{11\pi}{5} - 2\pi = \frac{\pi}{5}$$

por lo tanto la señal discretizada es:

$$x\left[\frac{n}{F_s}\right] = 3 \cos\left(\frac{\pi}{5}n\right) + 5 \sin\left(\frac{2\pi}{5}n\right) + 10 \cos\left(\frac{\pi}{5}n\right)$$

$$x\left[\frac{n}{F_s}\right] = 13 \cos\left(\frac{\pi}{5}n\right) + 5 \sin\left(\frac{2\pi}{5}n\right)$$