

# Control Theory Homework 3

Selina Varouqa

April 1, 2020

## 1 Variant

Name: Selina Varouqa

Email: s.varouqa@innopolis.university

The variant for this assignment is A

## 2 Python Calculations

### 2.1 PD-controller Design

$\ddot{x} + \mu\dot{x} + kx = u$  is the time dependent second order linear ordinary differential equation, for my variant it is  $\ddot{x} + 3\dot{x} + 45x = u$ . Using PD Control:  $u = k_p\dot{e} + k_de$  and having the error  $e = x^* - x$  we substitute and get  $u = k_p(x^* - x) + k_d(\dot{x}^* - \dot{x})$ . This is how my system looks like without control:

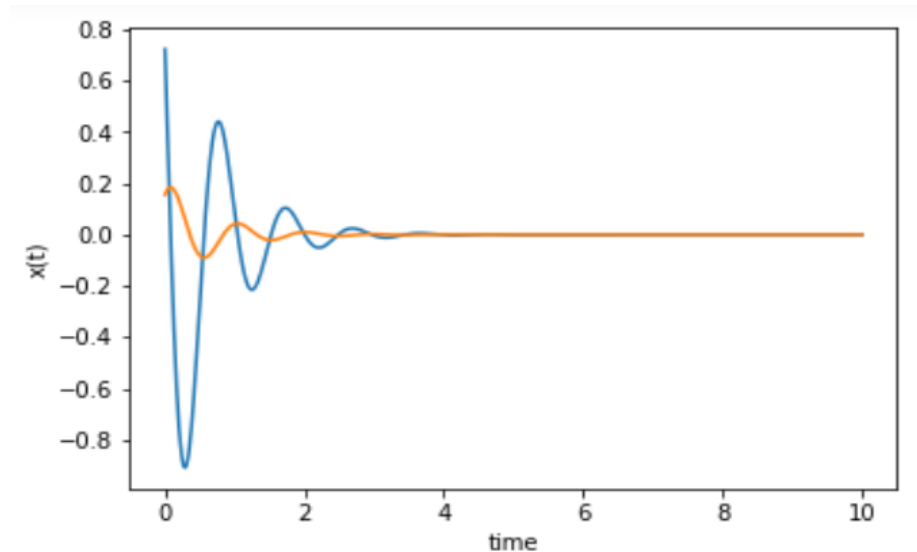


Figure 1: system without control

The first control applied to the system was as following:  $desiredx = 2\sin(2t)$  and  $desiredxdot = 4\cos(2t)$  with tuning  $k_p = 1000$  and  $k_d = 100$  we get the control as the graph shows, the orange is the  $x(t)$  and the blue is  $x'(t)$ :

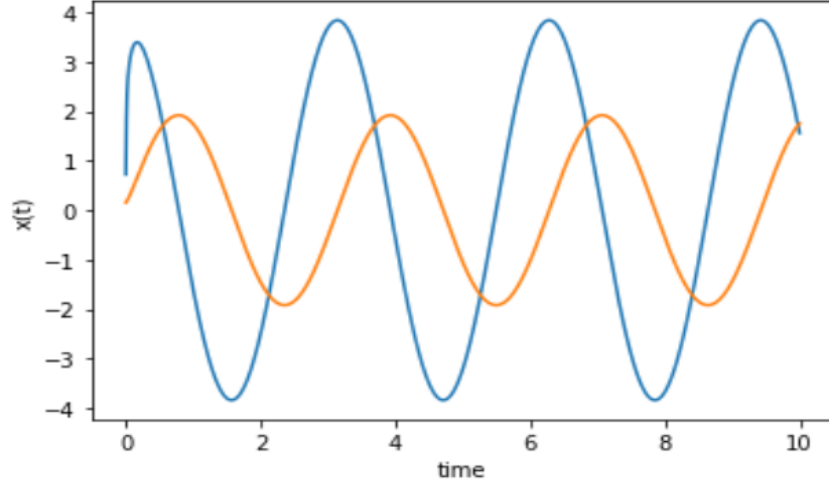


Figure 2: system with control with x desired as a sin function

The second control applied to the system was as following:  $desiredx = 5 * t$  and  $desiredxdot = 5$  along with tuning with  $k_p = 1000$  and  $k_d = 200$  we get the following control as the graph shows, orange is the  $x(t)$  and blue is  $x'(t)$ :

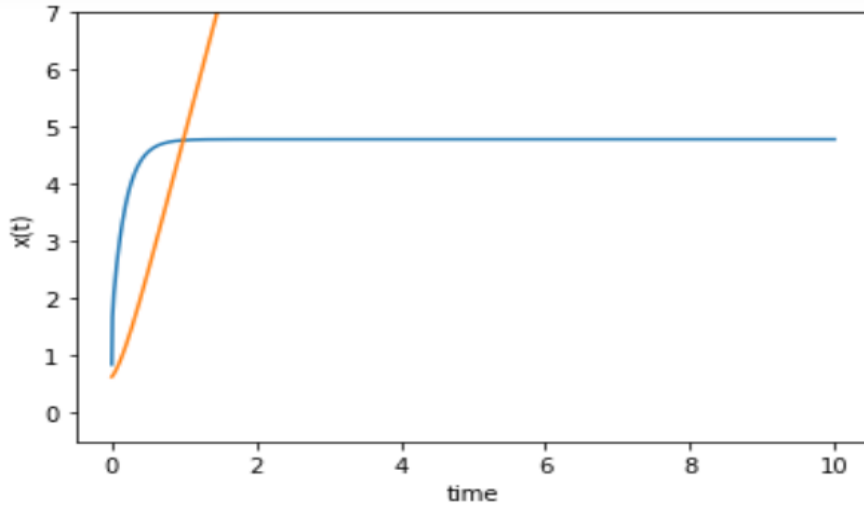


Figure 3: system with control with x desired as a linear function

## 2.2 Step function

I have tuned my  $k_p$  and  $k_d$  to be 50000 for each of them, it is a big number but from the observation of how the graph changes, with increasing  $k_p$  it makes the  $x$  desired approach the step function but the  $\dot{x}$  desired overshoots, that was solved by increasing  $k_d$  as well. It was observed that the less difference I had between  $k_p$  and  $k_d$  the more that  $k_d$  approached my desired function, it ended up being a very high number so I can get them both to approach the desired function, below is the resulting graph (green is step function, orange is the  $x(t)$  and the blue is  $\dot{x}(t)$ ):

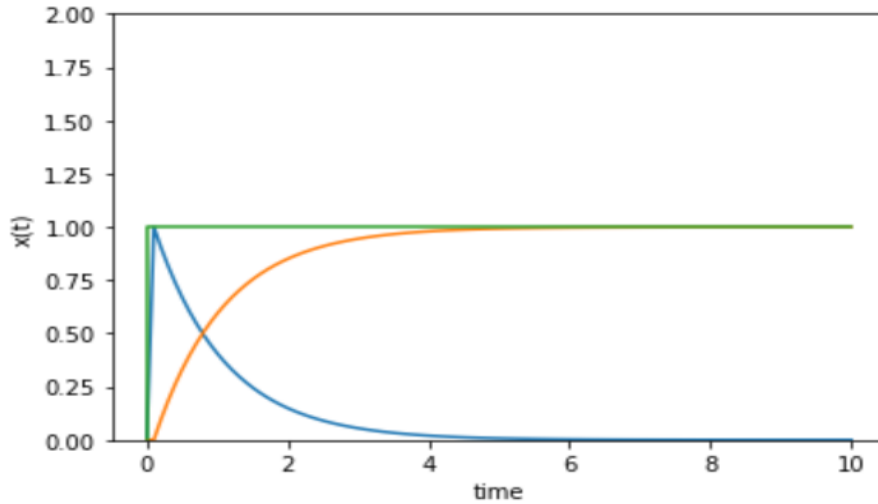


Figure 4: system with control approaching step input

## 2.3 Proving Stability

Proving stability for  $\ddot{x} + \mu\dot{x} + kx = u$  for my choice of  $k_p$  and  $k_d$  which are 50000 and 50000.  $\mu = 3, k = 45$  for my variant.

$$\ddot{x} + \mu\dot{x} + kx = u$$

$$x = \begin{bmatrix} x_0 \\ x_1 \end{bmatrix} \text{ and } \dot{x} = \begin{bmatrix} \dot{x}_0 \\ \dot{x}_1 \end{bmatrix}$$

$$x_0 = x, x_1 = \dot{x}$$

$$\dot{x}_0 = \dot{x}, \dot{x}_1 = \ddot{x}$$

so, we have:  $\ddot{x} = -\mu\dot{x} - kx + u = \dot{x}_1$  state space representation would

be:  $\begin{bmatrix} \dot{x}_0 \\ \dot{x}_1 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -k & -\mu \end{bmatrix} \begin{bmatrix} x_0 \\ x_1 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$  let's say that matrix A is  $\begin{bmatrix} 0 & 1 \\ -k & -\mu \end{bmatrix}$  and matrix B is  $\begin{bmatrix} 0 \\ 1 \end{bmatrix}$

Assume there is a  $x^* = x^*(t)$  that  $\dot{x}^* = Ax^*$  then our control error  $e$  would be  $Ae - Bu$  so  $u = Ke$  when  $K = \begin{bmatrix} Kd & Kp \end{bmatrix}$

$$C = A - KB = \begin{bmatrix} 0 & 1 \\ -k & -\mu \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} Kd & Kp \end{bmatrix}$$

which is  $\begin{bmatrix} 0 & 1 \\ -k - k_d & -\mu - k_p \end{bmatrix}$  so we understand that for the system to be stable we have to have the eigenvalues to be negative. which means we can use that  $\lambda_1 + \lambda_2 = -3 - k_p < 0$  and  $\lambda_1 \cdot \lambda_2 = 45 + k_d > 0$

so  $k_p$  and  $k_d$  stabilizes the system such that  $k_d > -45, k_p > -3$  and my choice of  $k_p$  and  $k_d$  matches that.

## 2.4 PD control for a MIMO

this is a multiple input multiple output system. We basically will have a state matrix and an input matrix. this is how the system looks before control:

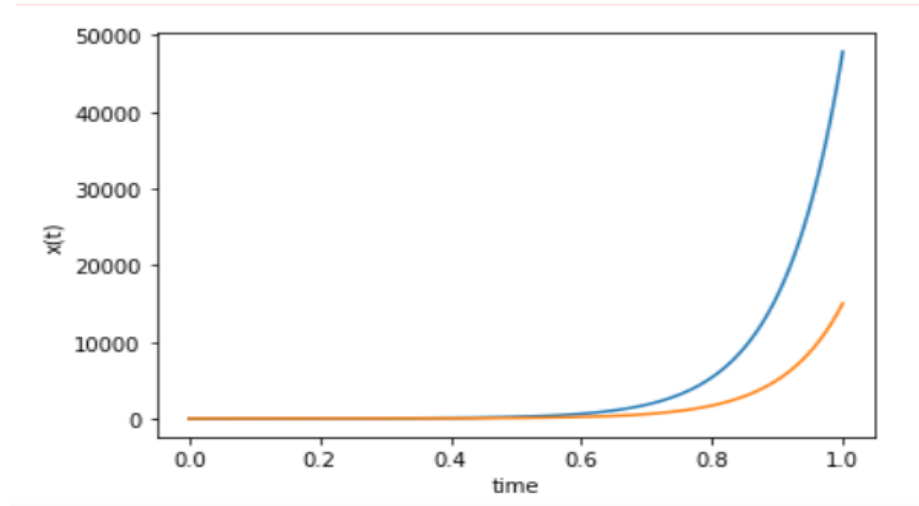


Figure 5: system before control

this is how the system looks after control:

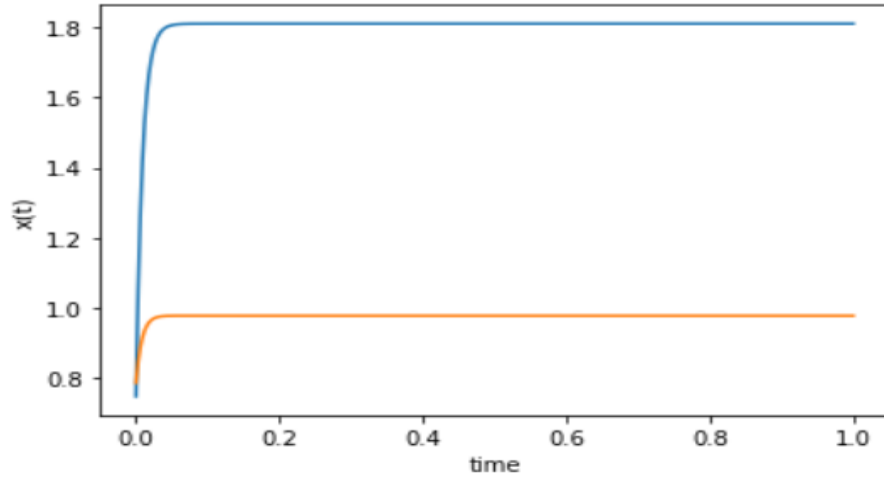


Figure 6: system after control

## 2.5 PI/PID controller

I implemented a PID control for the system for  $\ddot{x} + 3\dot{x} + 45x + 9.8 = u$

$$u = k_p e(t) + k_i \int_0^t e(t') dt' + k_d \frac{de(t)}{dt}$$

so my  $x$  desired  $= 5t$  ,  $\dot{x}$  desired  $= 5$  ,  $x$  int desired  $= (5/2) * t^2$

the values of for my choice of  $k_p$  and  $k_d$  and  $k_i$  are 500 and 30 and 10 for stabilizing the system.

this is how the system looks before PID control (green is integral component, orange is the  $x(t)$  and the blue is derivative component):

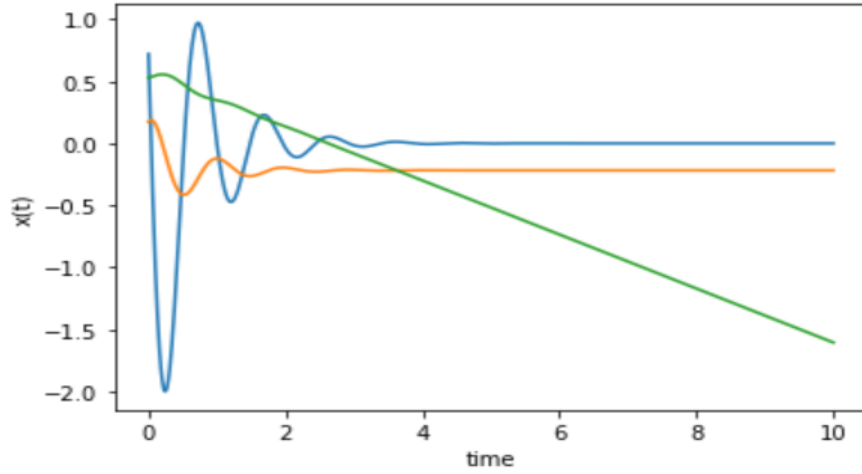


Figure 7: system before PID control

and this is how the system looks after PID control (green is integral component, orange is the  $x(t)$  and the blue is derivative component):

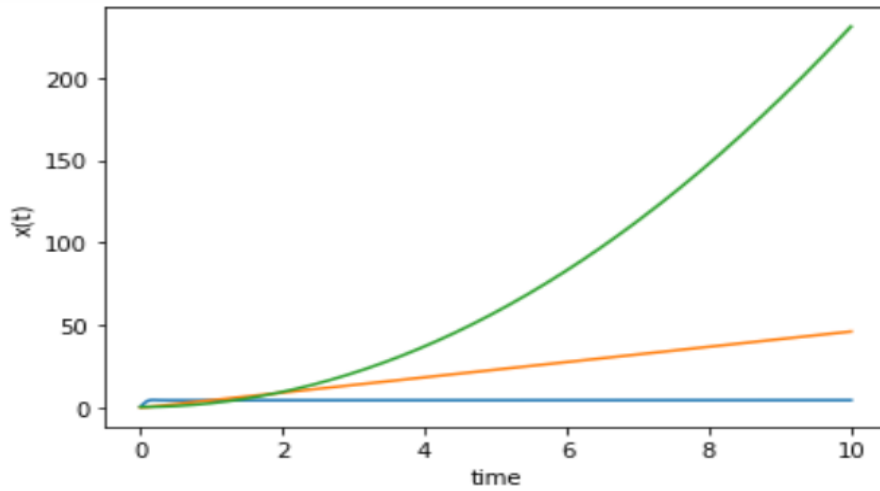


Figure 8: system after PID control

### 3 PID Controller Scheme

Using simulink, this is the scheme of the system before the control:

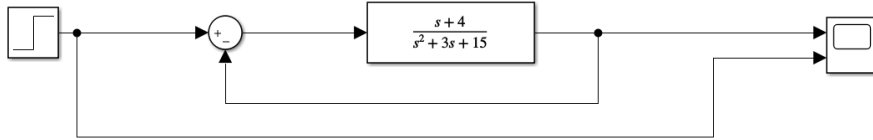


Figure 9: system scheme before PID control

and this is the graph that is in the scope before the control:

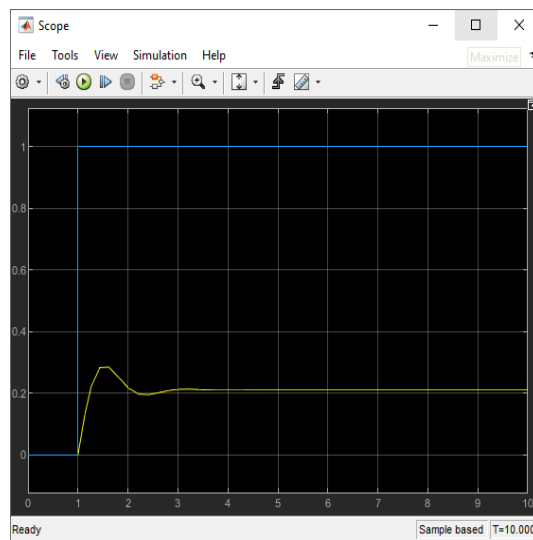


Figure 10: system graph before PID control

and below is the scheme of the system after adding PID controller:

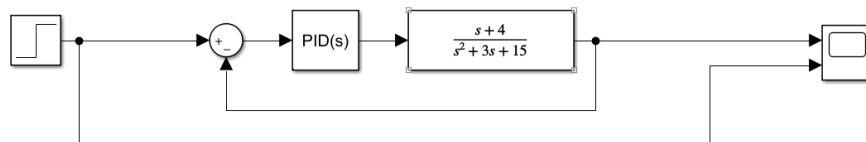


Figure 11: system scheme after PID control

First, I have tuned my  $K_p$ ,  $K_i$ , and  $K_d$  for 100, 50, and 1 (respectively) and I got the result of the following graph (before using auto-tune):

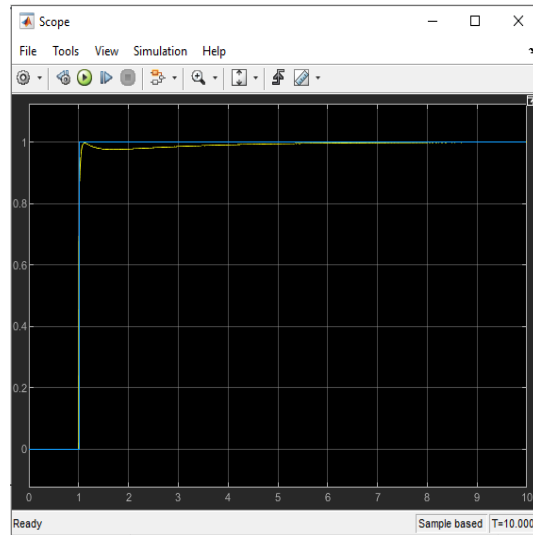


Figure 12: system graph after PID control

After using opening to auto-tune, I actually looked at the following table which compared the one I have and the one proposed:

Controller Parameters		
	Tuned	Block
P	8.2985	100
I	28.7197	50
D	0	1
N	100	100
Performance and Robustness		
	Tuned	Block
Rise time	0.155 seconds	0.0247 seconds
Settling time	1.33 seconds	NaN seconds
Overshoot	5.37 %	0 %
Peak	1.05	0.997
Gain margin	Inf dB @ NaN rad/s	Inf dB @ NaN rad/s
Phase margin	69 deg @ 10.3 rad/s	103 deg @ 182 rad/s
Closed-loop stability	Stable	Stable

Figure 13: table



## 4 Compensator Scheme

This is how the scheme of the system looks before the compensator:

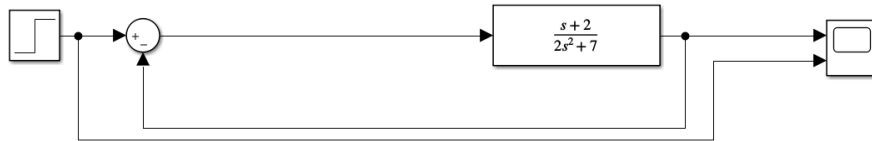


Figure 14: before compensator

and this is how the graph of the system before the compensator:

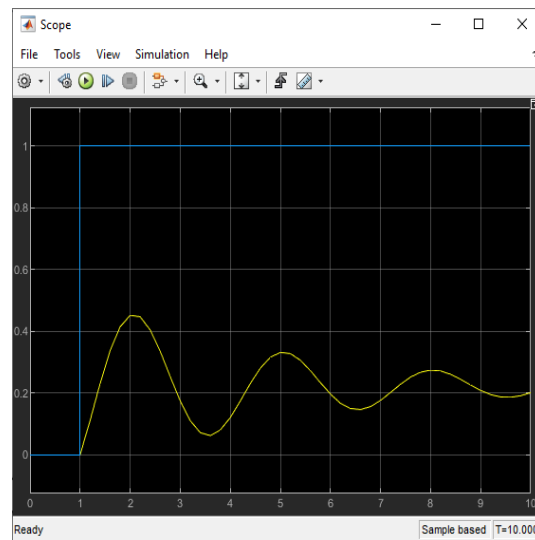


Figure 15: before compensator

after adding the compensator and but before changing coefficients and gain, it looks like this:

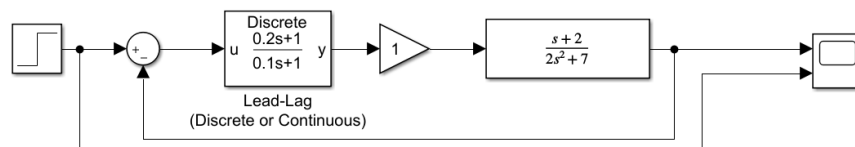


Figure 16: compensator with system scheme

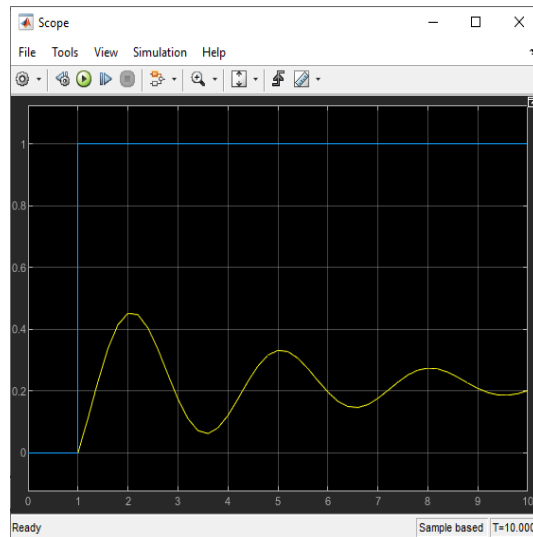


Figure 17: compensator w/o manual tuning

but, then I tuned the coefficients and the gain manually until I got this graph:

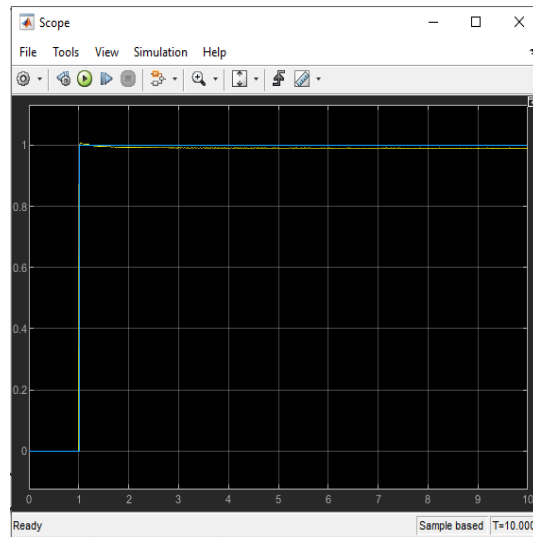


Figure 18: compensator with manual tuning

and the scheme looked like this:

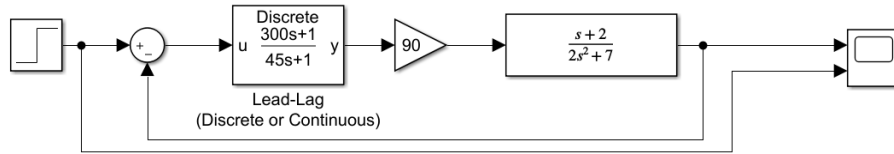


Figure 19: compensator with manual tuning

and last but not least, when I used the control system designer with this code

```
controlSystemDesigner(tf([1,2],[2,0,7]))
```

and it showed that my system (the one I tuned manually) was pretty stable according to the following graphs:

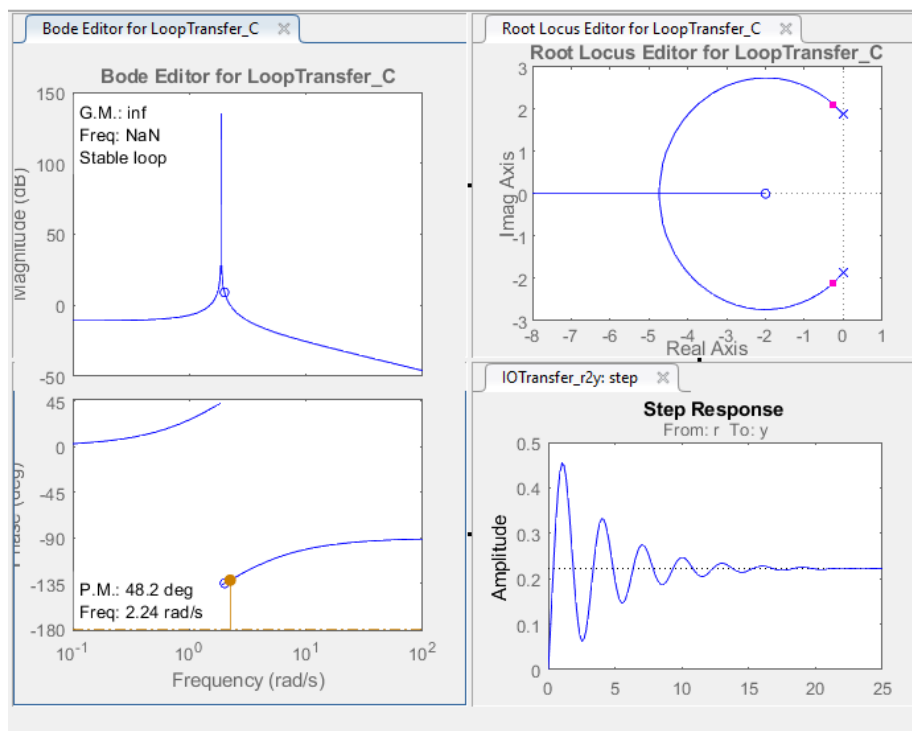


Figure 20: compensator with manual tuning