

Observer Design, Kalman Filter

by Sergei Savin

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Observer Design

Observer as a controllable LTI

- From the previous lecture we remember that we wanted the estimate of the state to have the same dynamics as the actual state.
- We also remember that we want to take into account the error between the estimated and measured output y

Assume the measurement is perfect: $\hat{y} = y$. Then we can propose observer as the following dynamical system:

$$\dot{\hat{x}} = A\hat{x} + Bu + L(y - \hat{y})$$

Observer Design

Estimation error dynamics

Remember that state estimation error is $\epsilon = \hat{x} - x$ and the actual dynamics of the system is:

$$\begin{cases} \dot{x} = Ax + Bu \\ y = Cx \end{cases}$$

Then, we can write the state estimation error dynamics:

$$\dot{\hat{x}} - \dot{x} = A\hat{x} - Ax + Bu - Bu + L(y - \hat{y})$$

or:

$$\dot{\epsilon} = A\epsilon + L(Cx - C\hat{x})$$

or:

$$\dot{\epsilon} = A\epsilon - LC\epsilon$$

Observer Design

State estimation error stability

Thus, with the proposed observer, state estimation error is:

$$\dot{\epsilon} = (A - LC)\epsilon$$

From which immediately follows that the observer is *stable* (i.e., the state estimation error tends to zero), as long as the following matrix is negative-definite:

$$A - LC < 0$$

We need to find L .

Observer Design

Observer Design vs Controller design

Let us observe the key difference between observer design and controller design.

Controller design: find such K that:

$$A - BK < 0$$

Observer design: find such L that:

$$A - LC < 0$$

We have instruments for finding K , what about L ?

Observer Design

Special case

Assume that C is identity I . Then the observer design problem becomes:

- find such L that $A - LI < 0$.

Which is equivalent to the problem:

- find such L that $A - BL < 0$, where $B = I$.

And in that formulation, it is equivalent to the controller design problem, which we know how to solve.

Observer Design

General case: design via Riccati eq.

In general, we can observe that if $A - LC$ is negative-definite, then $(A - LC)^\top$ is negative-definite too (by definition of the negative-definiteness).

Therefore, we can solve the following *dual problem*:

- find such L that $A^\top - C^\top L^\top < 0$.

The dual problem is *equivalent* to the control design problem. We can solve it by producing and solving algebraic Riccati equation, as in the LQR formulation. In pseudo-code it can be represented the following way:

$$L^\top = \text{lqr}(A^\top, C^\top, Q, R).$$

where Q and R are weight matrices, determining the "sensitivity" or "aggressiveness" of the observer.

Kalman Filter

System with noise and disturbances

Assume we have a discrete linear system with disturbance and noise:

$$\begin{cases} \mathbf{x}_{i+1} = \mathbf{A}\mathbf{x}_i + \mathbf{B}\mathbf{u}_i + \mathbf{w} \\ \mathbf{y}_i = \mathbf{C}\mathbf{x}_i + \mathbf{v} \end{cases}$$

where \mathbf{w} and \mathbf{v} are disturbance and noise, both are random processes. Covariance matrix of \mathbf{w} is \mathbf{W} , covariance matrix of \mathbf{v} is \mathbf{V} .

Our estimation $\hat{\mathbf{x}}$ is now done in two steps: after dynamics update, and after sensor update. We will use the following notation for it:

- $\hat{\mathbf{x}}_{i|i-1}$ is the i -th step estimate after dynamics update; also called an *a priori* estimate.
- $\hat{\mathbf{x}}_{i|i}$ is the estimate after sensor update; also called an *a posteriori* estimate.

Kalman Filter

Estimates covariances

Both estimates $\hat{x}_{i|i-1}$ and $\hat{x}_{i|i}$ have associated state estimation errors:

- $\epsilon_{i|i-1} = x_i - \hat{x}_{i|i-1}$
- $\epsilon_{i|i} = x_i - \hat{x}_{i|i}$

Those state estimation errors have their covariance matrices:

- $P_{i|i-1} = \text{cov}(\epsilon_{i|i-1})$,
- $P_{i|i} = \text{cov}(\epsilon_{i|i})$

We can consider the output estimation error $\varepsilon_i = y_i - C\hat{x}_{i|i-1}$ and its covariance matrix Y_i :

$$Y_i = \text{cov}(y_i - C\hat{x}_{i|i-1})$$

Kalman Filter

Estimate updates

We know that $\hat{x}_{i|i-1}$ can be found as follows:

$$\hat{x}_{i|i-1} = A\hat{x}_{i-1|i-1} + Bu_i$$

where $\hat{x}_{i-1|i-1}$ is the estimate on the previous time step.

We will search for the update law for the a posteriori estimate $\hat{x}_{i|i}$ in the following form:

$$\hat{x}_{i|i} = \hat{x}_{i|i-1} + K_i(y_i - C\hat{x}_{i|i-1})$$

We need to find K_i , which is the Kalman filter gain.

Kalman Filter

Knows and unknowns

We know or measure:

- A , B and C
- W and V
- y_i

We need to find:

- $\hat{x}_{i|i-1}$ and $\hat{x}_{i|i}$
- K_i
- Y_i , $P_{i|i-1}$ and $P_{i|i}$

Kalman Filter

Covariance update and Kalman gain

There is a formula for the a priori estimation covariance update:

$$P_{i|i-1} = AP_{i-1|i-1}A^T + W$$

And a formula for the measurement covariance:

$$Y_i = CP_{i|i-1}C^T + V$$

With those two, we can find the Kalman filter gain:

$$K_i = P_{i|i-1}C^TY_i^{-1}$$

Now we can find the a posteriori covariance:

$$P_{i|i} = (I - K_iC)P_{i|i-1}$$

Lecture slides are available via Moodle.

You can help improve these slides at:

<https://github.com/SergeiSa/Linear-Control-Slides-Spring-2020>

Check Moodle for additional links, videos, textbook suggestions.