
Parameter Estimation and Control of Multirotors Using Integral Concurrent Learning

Cheng-Cheng Yang and Teng-Hu Cheng

Speaker : Cheng-Cheng Yang

Advisor : Dr. Teng-Hu Cheng

Date : 2021/08/02

Outlines

- Motivation
- Problem Formulation
- Controller Design
- Stability Analysis
- Experiments
- Conclusion

Motivation

- Knowledge of the geometric and inertia parameters is essential to achieving good control performance.
- The payload or sensors attaching to multirotors may change the geometric and inertia parameters.
- Some geometric and inertia parameters like moment of inertia can not be measured through instrument.
- Existing adaptive control method can only guarantee the stability of multirotors system, can not ensure the parameters converge.

Motivation

Problem
Formulation

Controller
Design

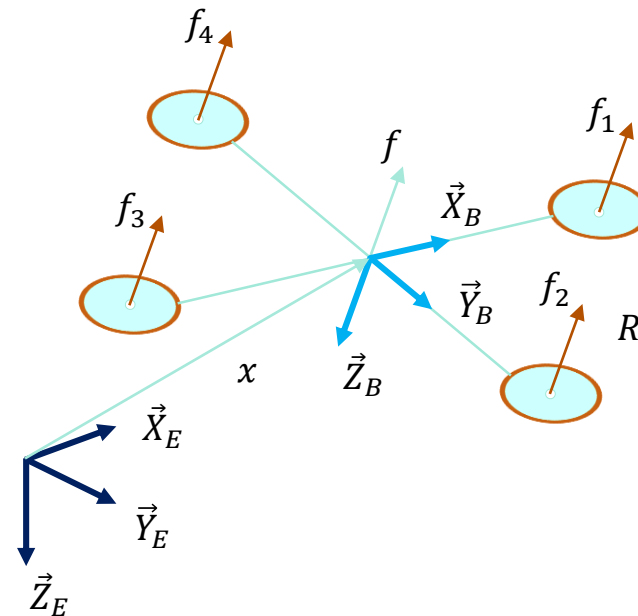
Stability
Analysis

Experiments

Conclusion

Problem Formulation - Definition of Symbols

| Symbol | Description |
|----------|---|
| x | Position of the multirotor |
| v | Velocity of the multirotor |
| R | Rotation matrix from the body-fixed frame to the inertial frame |
| Ω | Angular velocity in the body-fixed frame |
| f | Net thrust control input |
| M | Moment control input |
| m | Mass of the multirotor |
| J | Moment of inertia of the multirotor |



Body-fixed frame : $\{\vec{X}_B, \vec{Y}_B, \vec{Z}_B\}$

Inertial frame: $\{\vec{X}_E, \vec{Y}_E, \vec{Z}_E\}$

Motivation

Problem Formulation

Controller Design

Stability Analysis

Experiments

Conclusion

Problem Formulation - Dynamics of the Multirotor

- The multirotor is described by both translational and rotational dynamics.
- The translational dynamics considers forces such as the effects of gravity, thrusts, and the external force.
- The rotational dynamics takes the moment control input, rotational speed, and moment of inertia into account.

$$\begin{aligned}
 \dot{x} &= v \\
 m\dot{v} &= mg e_3 - f R e_3 \\
 \dot{R} &= R \hat{\Omega} \\
 J\dot{\Omega} + \Omega \times J\Omega &= M
 \end{aligned}
 \left. \begin{array}{l} \\ \\ \\ \end{array} \right\} \begin{array}{l} \text{Translational dynamics} \\ \\ \text{Rotational dynamics} \end{array}
 \quad J = \begin{bmatrix} J_{xx} & 0 & 0 \\ 0 & J_{yy} & 0 \\ 0 & 0 & J_{zz} \end{bmatrix}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Problem Formulation – Tracking Errors and Estimate Errors

- Position and velocity tracking errors

$$e_x \triangleq x - x_d$$

$$e_v \triangleq v - v_d$$

- Attitude error function on SO(3) based on [Geometric Tracking Control](#)

$$\Psi(R, R_d) \triangleq \frac{1}{2} \text{tr}[I - R_d^T R]$$

- Attitude tracking error and the angular velocity tracking error

$$e_R \triangleq \frac{1}{2} (R_d^T R - R^T R_d)^v$$

$$e_\Omega \triangleq \Omega - R^T R_d \Omega_d$$

- Estimate error of mass

$$\tilde{\theta}_m \triangleq \theta_m - \hat{\theta}_m, \theta_m = m \text{ (mass of the multirotor)}$$

- Estimate error of moment of inertia

$$\tilde{\theta}_J \triangleq \theta_J - \hat{\theta}_J, \theta_J = [J_{xx} \quad J_{yy} \quad J_{zz}]^T \text{ (moment of inertia of the multirotor)}$$



Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

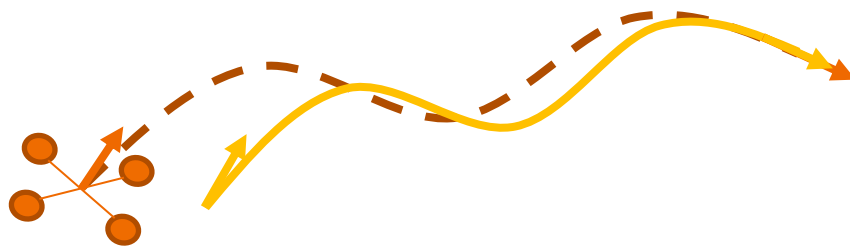
Experiments

Conclusion

Problem Formulation – Control Objectives

- Track a desired 3D trajectory
- Track a desired yaw angle
- Estimate the mass of the multirotor
- Estimate the moment of inertia of the multirotor

$$\begin{cases} e_x & \rightarrow 0 \\ e_v & \rightarrow 0 \\ e_R & \rightarrow 0 \\ e_\Omega & \rightarrow 0 \\ \tilde{\theta}_m & \rightarrow 0 \\ \tilde{\theta}_J & \rightarrow 0 \end{cases} \quad \text{as } t \rightarrow \infty$$



- 3D position
- 3D trajectory
- yaw angle
- desired yaw angle

Motivation

Problem Formulation

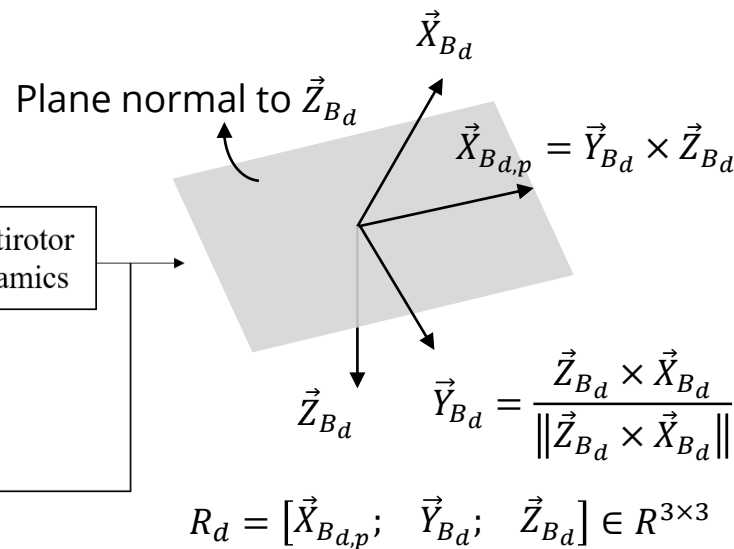
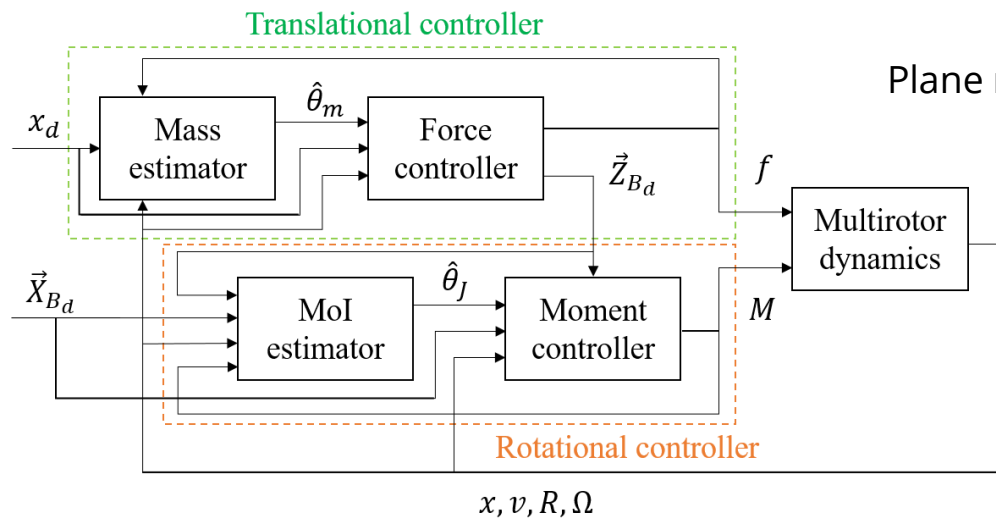
Controller Design

Stability Analysis

Experiments

Conclusion

Controller Design – Control Architecture



Motivation

Problem
Formulation

Controller
Design

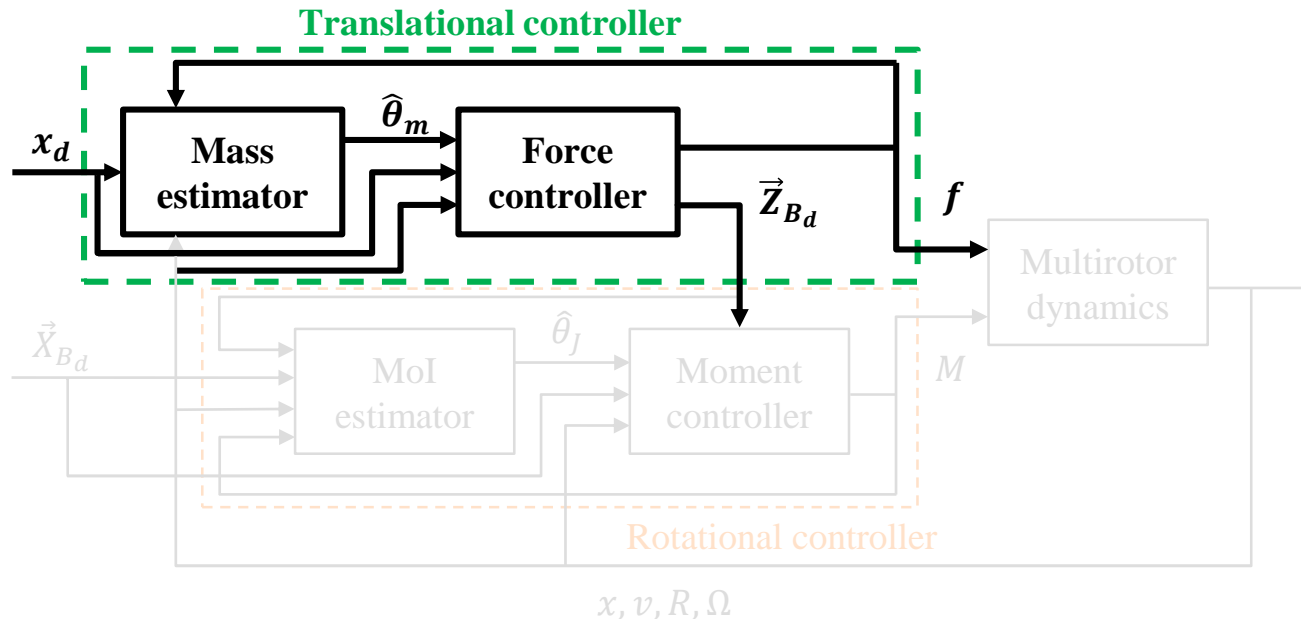
Stability
Analysis

Experiments

Conclusion

Controller Design – Translational Controller

- Translational controller



Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Controller Design – Translational Controller

- Translational controller

$$f = \underbrace{(k_x e_x + k_v e_v)}_{\text{feedback term}} + \underbrace{Y_m \hat{\theta}_m}_{\text{adaptive term}} \cdot Re_3, \quad Y_m = \begin{bmatrix} -\ddot{x}_{d_1} \\ -\ddot{x}_{d_2} \\ g - \ddot{x}_{d_3} \end{bmatrix} \text{ is a regression matrix}$$

- Integral CL-based adaptive control update law $\dot{\hat{\theta}}_m$

$$\dot{\hat{\theta}}_m = \underbrace{\Gamma_m Y_m^T (e_v + C_1 e_x)}_{\text{adaptive term}} + \underbrace{k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T (F(t_j) - y_m^{cl}(t_j) \hat{\theta}_m)}_{\text{ICL - based term}}$$

$$y_m^{cl}(t_j) \triangleq \begin{cases} 0_{n \times 1} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t Y_m^{cl}(\tau) d\tau & t > \Delta t \end{cases}, \quad F(t_j) \triangleq \begin{cases} 0_{n \times 1} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t f R e_3(\tau) d\tau & t > \Delta t \end{cases}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Controller Design – Translational Controller

- Y_m^{cl} defined as follows contains acceleration which is not implementable

$$fRe_3 = mge_3 - m\dot{v} = Y_m^{cl}\theta_m, \quad Y_m^{cl} = \begin{bmatrix} -\ddot{x}_1 \\ -\ddot{x}_2 \\ g - \ddot{x}_3 \end{bmatrix}$$

- By integrating Y_m^{cl} to be y_m^{cl} as defined in last page, y_m^{cl} becomes implementable
- Integrating both sides of the translational dynamics $fRe_3 = Y_m^{cl}\theta_m$ yields

$$\int_{t-\Delta t}^t fRe_3(\tau) d\tau = \int_{t-\Delta t}^t Y_m^{cl}(\tau)\theta_m d\tau \Rightarrow \int \mathbf{fRe}_3(\tau) \Big|_{\tau=t} - \int \mathbf{fRe}_3(\tau) \Big|_{\tau=t-\Delta t} = y_m^{cl}\theta_m$$

$$\begin{aligned} \dot{\hat{\theta}}_m &= \Gamma_m Y_m^T(e_v + C_1 e_x) + k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T \left(\mathbf{F}(t_j) - y_m^{cl}(t_j) \hat{\theta}_m \right) \\ &= \Gamma_m Y_m^T(e_v + C_1 e_x) + k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T y_m^{cl}(t_j) \tilde{\theta}_m \end{aligned}$$

Motivation

Problem
Formulation

Controller
Design

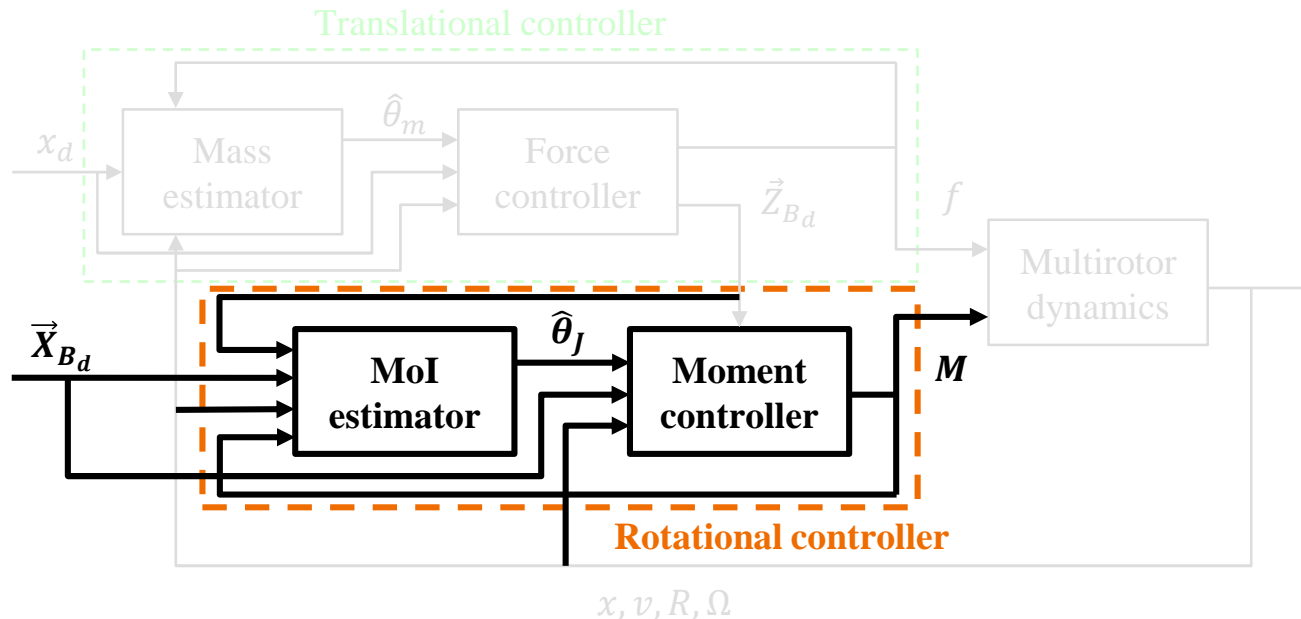
Stability
Analysis

Experiments

Conclusion

Controller Design – Rotational Controller

- Rotational controller



Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Controller Design – Rotational Controller

- Rotational controller

$$M = \underbrace{-k_R e_R - k_\Omega e_\Omega}_{\text{feedback term}} - \underbrace{Y_J \hat{\theta}_J}_{\text{adaptive term}}, \quad Y_J = \begin{bmatrix} \bar{\Omega}_1 & \Omega_2 \cdot \Omega_3 & -\Omega_2 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_3 & \bar{\Omega}_2 & \Omega_1 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_2 & -\Omega_1 \cdot \Omega_2 & \bar{\Omega}_3 \end{bmatrix}$$

- Integral CL-based adaptive control update law $\hat{\theta}_J$

$$\dot{\hat{\theta}}_J = \underbrace{\Gamma_J Y_J^T (e_\Omega + C_2 e_R)}_{\text{adaptive term}} + \underbrace{k_J^{cl} \Gamma_J \sum_{j=1}^{N_J} \left(y_j^{cl}(t_j) \right)^T (\bar{M}(t_j) - y_j^{cl}(t_j) \hat{\theta}_J)}_{\text{ICL-based term}}$$

$$y_j^{cl}(t_j) \triangleq \begin{cases} 0_{n \times m} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t Y_J^{cl}(\tau) d\tau & t > \Delta t \end{cases}, \quad \bar{M}(t_j) \triangleq \begin{cases} 0_{n \times m} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t M(\tau) d\tau & t > \Delta t \end{cases}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Controller Design – Rotational Controller

- Y_f^{cl} defined as follows contains angular acceleration which is not implementable

$$M = J\dot{\Omega} + \Omega \times J\Omega = Y_f^{cl}\theta_J, \quad Y_f^{cl} = \begin{bmatrix} \dot{\Omega}_1 & -\Omega_2 \cdot \Omega_3 & \Omega_2 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_3 & \dot{\Omega}_2 & -\Omega_1 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_2 & \Omega_1 \cdot \Omega_2 & \dot{\Omega}_3 \end{bmatrix}$$

- By integrating Y_f^{cl} to be y_f^{cl} as defined in last page, y_f^{cl} becomes implementable
- Integrating both sides of the translational dynamics $M = Y_f^{cl}\theta_J$ yields

$$\int_{t-\Delta t}^t M(\tau) d\tau = \int_{t-\Delta t}^t Y_f^{cl}(\tau)\theta_J d\tau \Rightarrow \int \mathbf{M}(\boldsymbol{\tau}) \Big|_{\tau=t} - \int \mathbf{M}(\boldsymbol{\tau}) \Big|_{\tau=t-\Delta t} = y_f^{cl}\theta_J$$

$$\begin{aligned} \dot{\hat{\theta}}_J &= \Gamma_J Y_J^T (e_\Omega + C_2 e_R) + k_J^{cl} \Gamma_J \sum_{j=1}^{N_J} \left(y_J^{cl}(t_j) \right)^T \left(\bar{\mathbf{M}}(\mathbf{t}_j) - y_J^{cl}(t_j) \hat{\theta}_J \right) \\ &= \Gamma_J Y_J^T (e_\Omega + C_2 e_R) + k_J^{cl} \Gamma_J \sum_{j=1}^{N_J} \left(y_J^{cl}(t_j) \right)^T y_J^{cl}(t_j) \tilde{\theta}_J \end{aligned}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Stability Analysis – Closed-Loop Error Systems

- Taking the time derivative of error dynamics e_x, e_v defined in [Problem Formulation](#)

$$\dot{e}_x = \dot{e}_v$$

$$m\dot{e}_v = mge_3 - fRe_3 - m\ddot{x}_d$$

$$= Y_m\theta_m - fRe_3$$

$$= -k_x e_x - k_v e_v + Y_m \tilde{\theta}_m - X \quad , X = \frac{f}{e_3^T R_d^T R e_3} \left((e_3^T R_d^T R e_3) R e_3 - R_d e_3 \right)$$

- Taking the time derivative of error dynamics e_R, e_Ω defined in [Problem Formulation](#)

$$\dot{e}_R = \frac{1}{2} (R_d^T R \hat{e}_\Omega + \hat{e}_\Omega R^T R_d)^V$$

$$= \frac{1}{2} (tr[R^T R_d] I - R^T R_d) \equiv C(R_d^T R) e_\Omega$$

$$J\dot{e}_\Omega = J\dot{\Omega} + J(\hat{\Omega} R^T R_d \Omega_d - R^T R_d \dot{\Omega}_d)$$

$$= J\dot{\Omega} + J\bar{\Omega} = M + Y_J \theta_J = -k_R e_R - k_\Omega e_\Omega + Y_J \tilde{\theta}_J$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion



Stability Analysis – Translational Dynamics

- Let Lyapunov function V_1 defined as

$$V_1 = \frac{1}{2}k_x e_x^T e_x + \frac{1}{2}m e_v^T e_v + C_1 m e_x \cdot e_v + \frac{1}{2}\tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m$$

- V_1 is P.D. and it can be lower and upper bounded by

$$z_1^T M_{11} z_1 + \frac{1}{2}\tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m \leq V_1 \leq z_1^T M_{12} z_1 + \frac{1}{2}\tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m$$

$$z_1 \triangleq [\|e_x\|, \|e_v\|]^T$$

$$M_{11} = \frac{1}{2} \begin{bmatrix} k_x & -C_1 m \\ -C_1 m & m \end{bmatrix}$$

$$M_{12} = \frac{1}{2} \begin{bmatrix} k_x & C_1 m \\ C_1 m & m \end{bmatrix}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Stability Analysis – Translational Dynamics

- Taking the time derivative of V_1 yields

$$\dot{V}_1 = k_x e_x \cdot \dot{e}_x + e_v \cdot m \dot{e}_v + C_1 m \dot{e}_x \cdot e_v + C_1 e_x \cdot m \dot{e}_v - \tilde{\theta}_m^T \Gamma_m^{-1} \dot{\tilde{\theta}}_m$$

- Substitute \dot{e}_x and $m \dot{e}_v$ defined in [the previous page](#) into \dot{V}_1

$$\dot{V}_1 \leq -z_1^T W_1 z_1 + z_1^T W_{12} z_2 - k_m^{cl} \tilde{\theta}_m^T \left(\sum_{j=1}^{N_m} (y_m^{cl}(t_j))^T y_m^{cl}(t_j) \right) \tilde{\theta}_m$$

$$W_1 = \begin{bmatrix} k_x C_1 (1 - \alpha) & -\frac{1}{2} C_1 k_v (1 + \alpha) \\ -\frac{1}{2} C_1 k_v (1 + \alpha) & k_v (1 - \alpha) - C_1 m \end{bmatrix}, \quad W_{12} = \begin{bmatrix} k_x e_{v, \max} + C_1 B & 0 \\ B & 0 \end{bmatrix}$$

- M_{11} , M_{12} , W_1 in V_1 and \dot{V}_1 are positive-definite matrices if C_1 satisfies

$$C_1 < \min \left\{ \sqrt{\frac{k_x}{m}}, \frac{k_v (1 - \alpha)}{m}, \frac{4 k_x k_v (1 - \alpha)^2}{k_v^2 (1 + \alpha)^2 + 4 m k_x (1 - \alpha)} \right\}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Stability Analysis – Rotational Dynamics

- Let Lyapunov function V_2 defined as

$$V_2 = \frac{1}{2} e_\Omega \cdot J e_\Omega + k_R \Psi(R, R_d) + J C_2 e_R \cdot e_\Omega + \frac{1}{2} \tilde{\theta}_J^T \Gamma_J^{-1} \tilde{\theta}_J$$

- V_2 is P.D. and it can be lower and upper bounded by

$$z_2^T M_{21} z_2 + \frac{1}{2} \tilde{\theta}_J^T \Gamma_J^{-1} \tilde{\theta}_J \leq V_2 \leq z_2^T M_{22} z_2 + \frac{1}{2} \tilde{\theta}_J^T \Gamma_J^{-1} \tilde{\theta}_J$$

$$z_2 \triangleq [\|e_R\|, \|e_\Omega\|]^T$$

$$M_{21} = \frac{1}{2} \begin{bmatrix} k_R & -C_2 \lambda_{\max}(J) \\ -C_2 \lambda_{\max}(J) & \lambda_{\min}(J) \end{bmatrix}$$

$$M_{22} = \frac{1}{2} \begin{bmatrix} \frac{2k_R}{2 - \psi_2} & -C_2 \lambda_{\max}(J) \\ -C_2 \lambda_{\max}(J) & \lambda_{\min}(J) \end{bmatrix}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Stability Analysis – Rotational Dynamics

- Taking the time derivative of V_2 yields

$$\dot{V}_2 = (e_\Omega + C_2 e_R) \cdot (J \dot{e}_\Omega) + k_R e_\Omega \cdot e_R + J C_2 \dot{e}_R \cdot e_\Omega - \tilde{\theta}_J^T \Gamma_J^{-1} \dot{\tilde{\theta}}_J$$

- Substitute \dot{e}_R and $J \dot{e}_\Omega$ defined in [the previous page](#) into \dot{V}_2

$$\dot{V}_2 \leq -z_2^T W_2 z_2 - k_J^{cl} \tilde{\theta}_J^T \left(\sum_{j=1}^{N_J} (y_J^{cl}(t_j))^T y_J^{cl}(t_j) \right) \tilde{\theta}_J$$

$$W_2 = \begin{bmatrix} C_2 k_R & -\frac{C_2 k_\Omega}{2} \\ -\frac{C_2 k_\Omega}{2} & k_\Omega - C_2 \lambda_{\max}(J) \end{bmatrix}$$

- M_{21} , M_{22} , W_2 in V_2 and \dot{V}_2 are positive-definite matrices if C_2 satisfies

$$C_2 < \min \left\{ \frac{k_\Omega}{\lambda_{\max}(J)}, \frac{4k_\Omega k_R}{k_\Omega^2 + 4k_R \lambda_{\max}(J)}, \sqrt{\frac{k_R \lambda_{\min}(J)}{\lambda_{\max}(J)^2}} \right\}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Stability Analysis – Overall System

- Let $V = V_1 + V_2$ be a Lyapunov function for the system containing rotational and translational dynamics

$$\begin{aligned}
 V &= V_1 + V_2 \\
 &= \frac{1}{2} k_x e_x^T e_x + \frac{1}{2} m e_v^T e_v + C_1 m e_x \cdot e_v + \frac{1}{2} \tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m \\
 &\quad + \frac{1}{2} e_\Omega \cdot J e_\Omega + k_R \Psi(R, R_d) + J C_2 e_R \cdot e_\Omega + \frac{1}{2} \tilde{\theta}_J^T \Gamma_J^{-1} \tilde{\theta}_J \quad \dots \text{P.D.}
 \end{aligned}$$

- Taking the time derivative of V and substituting \dot{V}_1 and \dot{V}_2 yields

$$\begin{aligned}
 \dot{V} = \dot{V}_1 + \dot{V}_2 &\leq -z_1^T W_1 z_1 + z_1^T W_{12} z_2 - k_m^{cl} \tilde{\theta}_m^T \left(\sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T y_m^{cl}(t_j) \right) \tilde{\theta}_m \\
 &\quad - z_2^T W_2 z_2 - k_J^{cl} \tilde{\theta}_J^T \left(\sum_{j=1}^{N_J} \left(y_J^{cl}(t_j) \right)^T y_J^{cl}(t_j) \right) \tilde{\theta}_J \quad \dots \text{N.D.}
 \end{aligned}$$

$$, \text{ where } \lambda_{\min}(W_2) > 4 \|W_{12}\|^2 / \lambda_{\min}(W_1)$$

Motivation

Problem
Formulation

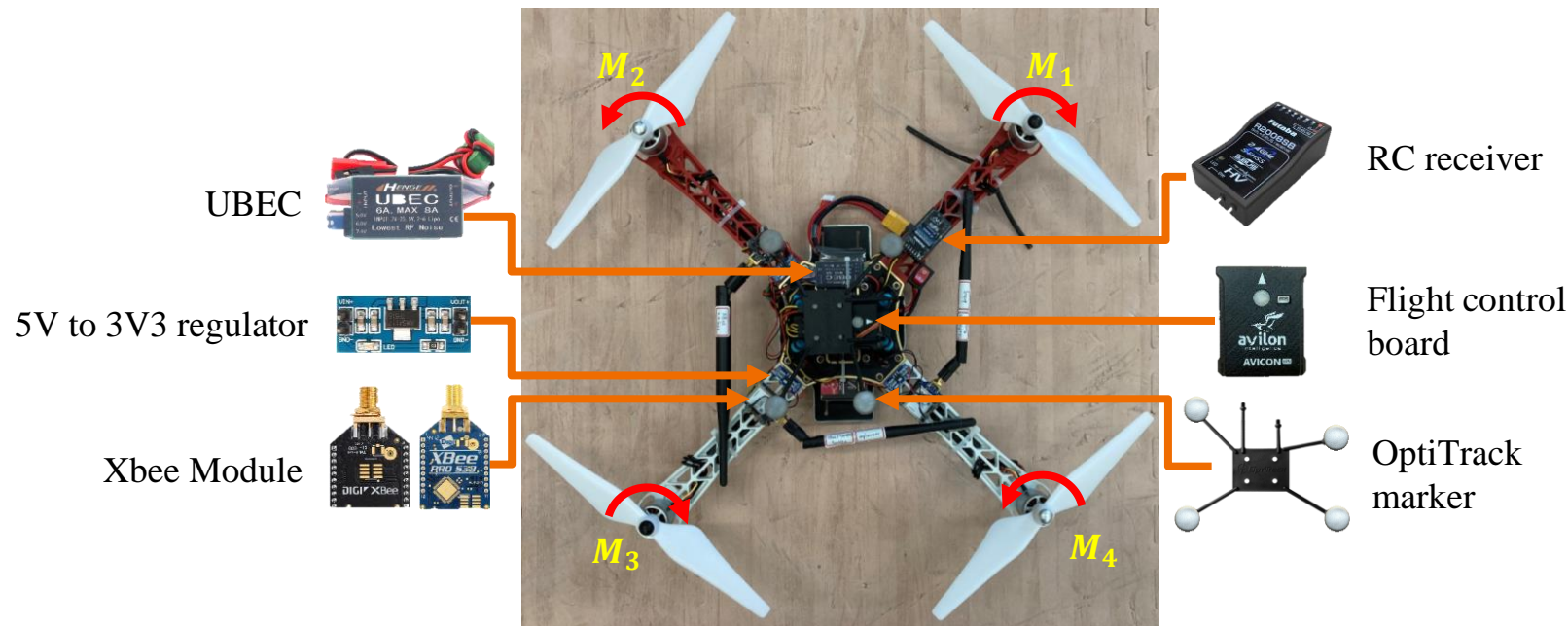
Controller
Design

Stability
Analysis

Experiments

Conclusion

Experiments – Hardware Architecture



Motivation

Problem
Formulation

Controller
Design

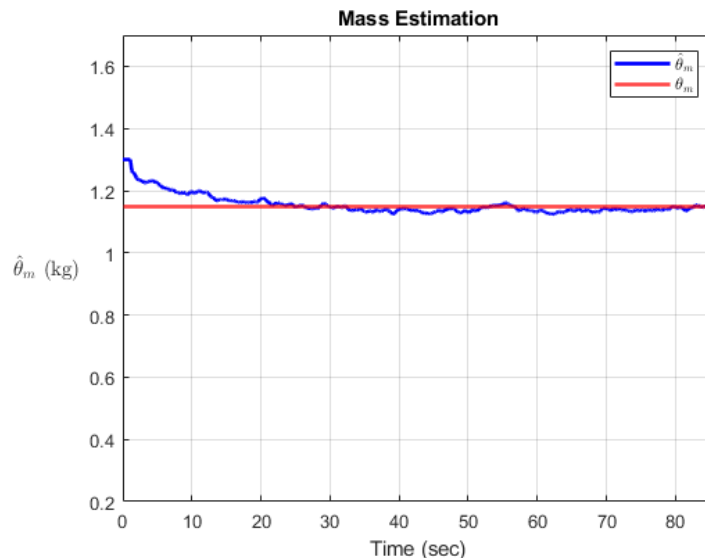
Stability
Analysis

Experiments

Conclusion

Experiments – Mass Estimation

- The mass estimation of the multirotor with ICL controller converged to 1.15kg and has 1% error with ground truth



| Mass estimation | Mass ground truth |
|-----------------|-------------------|
| 1.15 (kg) | 1.16 (kg) |

Motivation

Problem
Formulation

Controller
Design

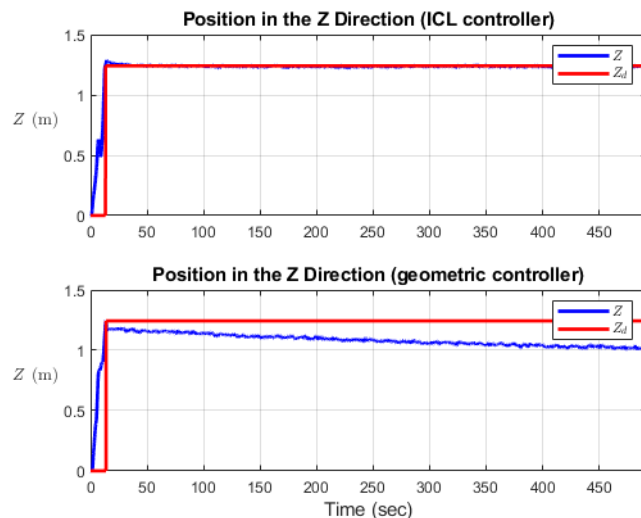
Stability
Analysis

Experiments

Conclusion

Experiments – Mass Estimation

- The accurate mass estimation also make the multirotor have better tracking performance in the z direction



Steady-state error in the z direction

ICL controller

Geometric controller

<0.5 (cm)

>10 (cm)

ICL controller : $f = (k_x e_x + k_v e_v + Y_m \hat{\theta}_m) \cdot Re_3$

Geometric controller : $f = (k_x e_x + k_v e_v + m g e_3 - m \ddot{x}_d) \cdot Re_3$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Experiments – Moment of Inertia Estimation

- The moment of inertia estimation converge to $[0.0033, 0.0032, 0.0033] (kg \cdot m^2)$ which is smaller than the ground truth

Motivation

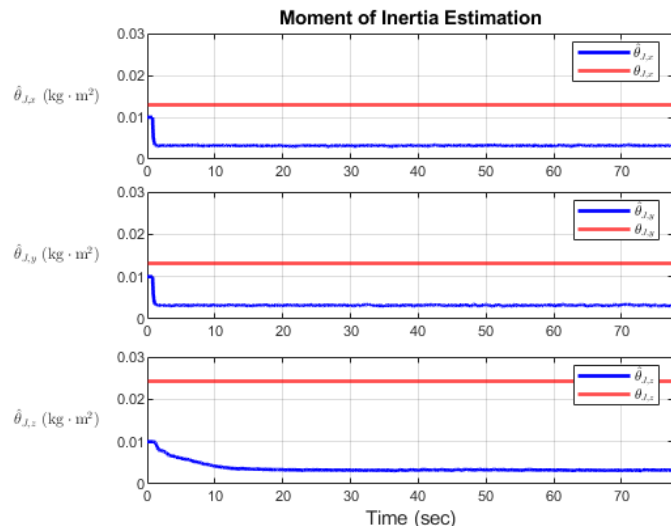
Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion



| Moment of inertia estimation ($kg \cdot m^2$) | Moment of inertia ground truth ($kg \cdot m^2$) |
|--|---|
| $\begin{bmatrix} 0.0033 & 0 & 0 \\ 0 & 0.0032 & 0 \\ 0 & 0 & 0.0033 \end{bmatrix}$ | $\begin{bmatrix} 0.013 & 0 & 0 \\ 0 & 0.013 & 0 \\ 0 & 0 & 0.024 \end{bmatrix}$ |

Experiments – Moment of Inertia Estimation

- We suppose the incorrect convergence is resulted from the measurement noise from gyroscope
- To present the experimental scene more realistically, white Gaussian noise is applied to the simulation
- Moreover, we designed an estimator to eliminate the influence of noise on the system

$$M = -k_R e_R - k_\Omega e_\Omega - \mathbf{Y}_J \hat{\theta}_J, \quad \mathbf{Y}_J = \begin{bmatrix} \bar{\Omega}_1 & \Omega_2 \cdot \Omega_3 & -\Omega_2 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_3 & \bar{\Omega}_2 & \Omega_1 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_2 & -\Omega_1 \cdot \Omega_2 & \bar{\Omega}_3 \end{bmatrix}$$

$$\dot{\hat{\theta}}_J = \Gamma_J \mathbf{Y}_J^T (e_\Omega + C_2 e_R) + k_J^{cl} \Gamma_J \sum_{j=1}^{N_J} \left(\mathbf{y}_J^{cl}(t_j) \right)^T (\bar{M}(t_j) - \mathbf{y}_J^{cl}(t_j) \hat{\theta}_J), \quad \mathbf{Y}_J^{cl} = \begin{bmatrix} \dot{\Omega}_1 & -\Omega_2 \cdot \Omega_3 & \Omega_2 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_3 & \dot{\Omega}_2 & -\Omega_1 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_2 & \Omega_1 \cdot \Omega_2 & \dot{\Omega}_3 \end{bmatrix}$$

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Experiments – Estimator of the Angular Velocity

- The rotational dynamics of the multirotor can be written as

$$\dot{\Omega} = J^{-1}M - J^{-1}\Omega \times J\Omega$$

$$\hat{\Omega}_k^- = \hat{\Omega}_{k-1} + \dot{\Omega}_{k-1}\Delta t$$

$$= \begin{bmatrix} \frac{M_1}{\hat{J}_{xx}} + \frac{\hat{J}_{yy}}{\hat{J}_{xx}}\Omega_2\Omega_3 - \frac{\hat{J}_{zz}}{\hat{J}_{xx}}\Omega_2\Omega_3 \\ \frac{M_2}{\hat{J}_{yy}} + \frac{\hat{J}_{xx}}{\hat{J}_{yy}}\Omega_1\Omega_3 - \frac{\hat{J}_{zz}}{\hat{J}_{yy}}\Omega_1\Omega_3 \\ \frac{M_3}{\hat{J}_{zz}} + \frac{\hat{J}_{xx}}{\hat{J}_{zz}}\Omega_1\Omega_2 - \frac{\hat{J}_{yy}}{\hat{J}_{zz}}\Omega_1\Omega_2 \end{bmatrix}$$

$$= \begin{bmatrix} \hat{\Omega}_{k-1,1} + \left(\frac{M_1}{\hat{J}_{xx}} + \frac{\hat{J}_{yy}}{\hat{J}_{xx}}\Omega_2\Omega_3 - \frac{\hat{J}_{zz}}{\hat{J}_{xx}}\Omega_2\Omega_3 \right) \cdot \Delta t \\ \hat{\Omega}_{k-1,2} + \left(\frac{M_2}{\hat{J}_{yy}} + \frac{\hat{J}_{xx}}{\hat{J}_{yy}}\Omega_1\Omega_3 - \frac{\hat{J}_{zz}}{\hat{J}_{yy}}\Omega_1\Omega_3 \right) \cdot \Delta t \\ \hat{\Omega}_{k-1,3} + \left(\frac{M_3}{\hat{J}_{zz}} + \frac{\hat{J}_{xx}}{\hat{J}_{zz}}\Omega_1\Omega_2 - \frac{\hat{J}_{yy}}{\hat{J}_{zz}}\Omega_1\Omega_2 \right) \cdot \Delta t \end{bmatrix}$$

- The estimated angular velocity can be generated as

$$\hat{\Omega}_k = \hat{\Omega}_k^- + K_{\Omega}(\Omega - \hat{\Omega}_k^-) , K_{\Omega} \text{ is a estimator gain}$$

Motivation

Problem
Formulation

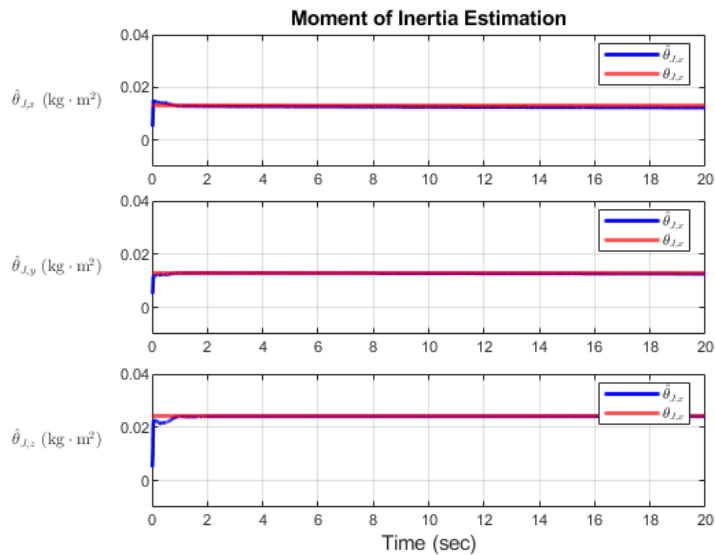
Controller
Design

Stability
Analysis

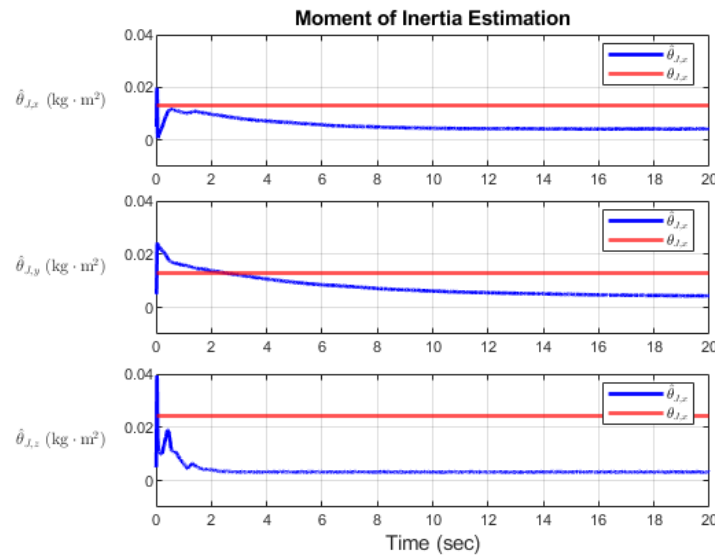
Experiments

Conclusion

Experiments – Moment of Inertia Estimation



Moment of inertia estimation **with noise and estimator** in the simulation



Moment of inertia estimation **with noise, without estimator** in the simulation

Motivation

Problem Formulation

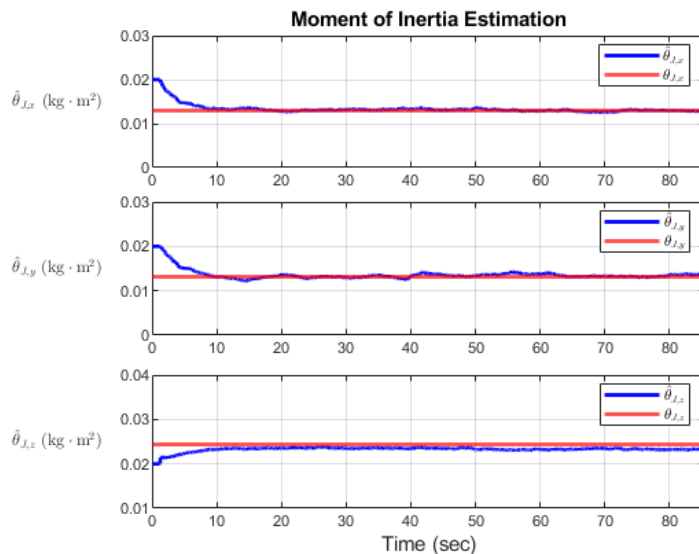
Controller Design

Stability Analysis

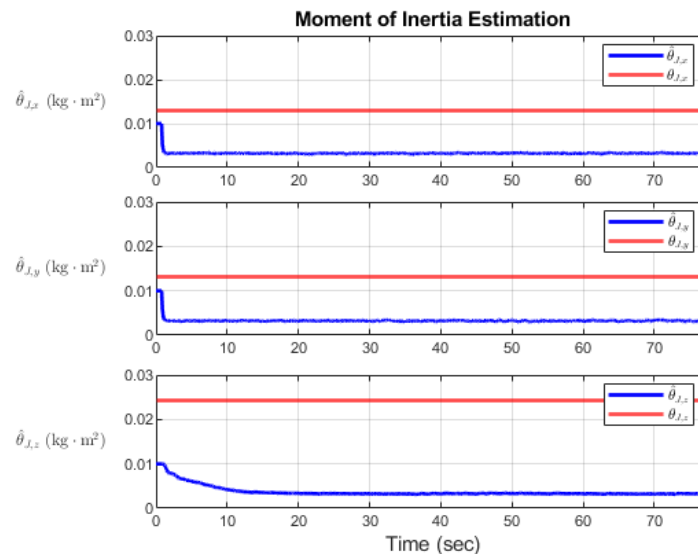
Experiments

Conclusion

Experiments – Moment of Inertia Estimation



Moment of inertia estimation **with noise and estimator** in the experiments



Moment of inertia estimation **with noise, without estimator** in the experiments

Motivation

Problem Formulation

Controller Design

Stability Analysis

Experiments

Conclusion

Experiments – Moment of Inertia Estimation

- The moment of inertia estimation converge to $[0.013, 0.014, 0.022]$ ($kg \cdot m^2$) and has 8% error with ground truth

Motivation

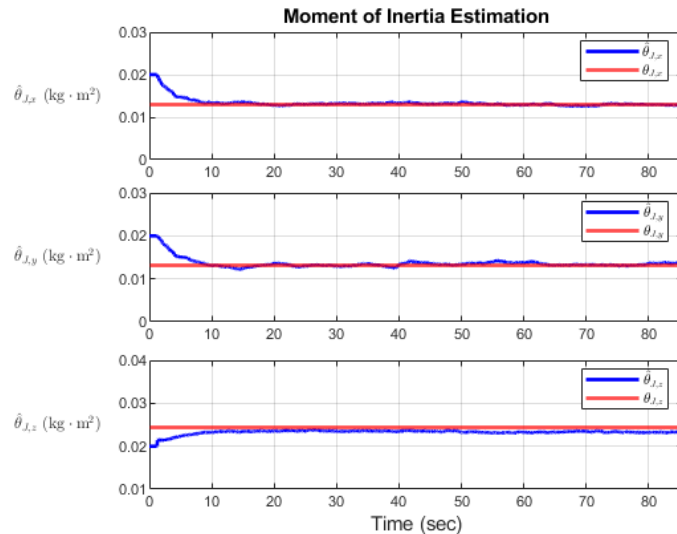
Problem Formulation

Controller Design

Stability Analysis

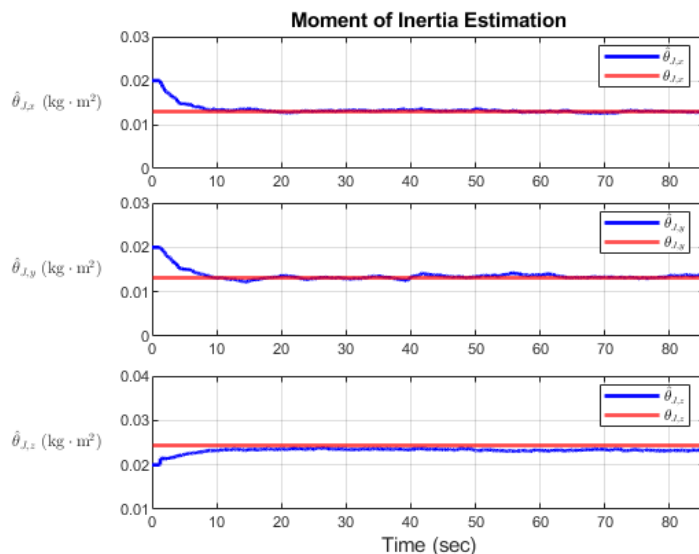
Experiments

Conclusion

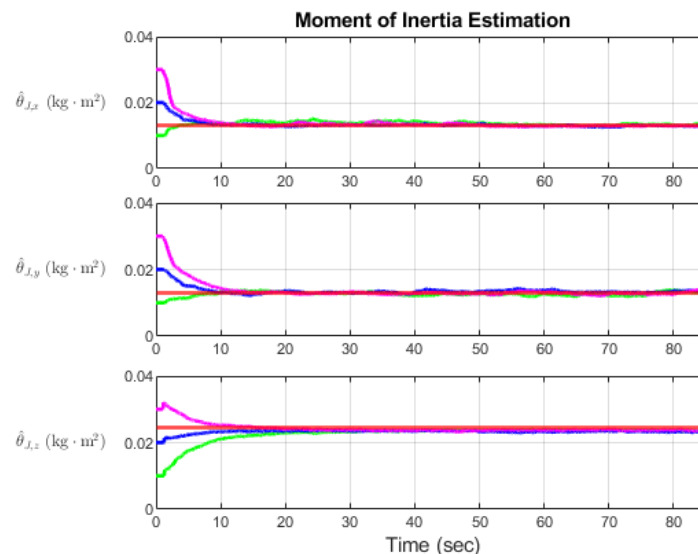


| Moment of inertia estimation ($kg \cdot m^2$) | Moment of inertia ground truth ($kg \cdot m^2$) |
|---|---|
| $\begin{bmatrix} 0.013 & 0 & 0 \\ 0 & 0.014 & 0 \\ 0 & 0 & 0.022 \end{bmatrix}$ | $\begin{bmatrix} 0.013 & 0 & 0 \\ 0 & 0.013 & 0 \\ 0 & 0 & 0.024 \end{bmatrix}$ |

Experiments – Moment of Inertia Estimation



Moment of inertia estimation **with noise and estimator** in the experiments



Moment of inertia estimation **with different initial values** in the experiments

Motivation

Problem Formulation

Controller Design

Stability Analysis

Experiments

Conclusion

Experiments – Trajectory Generation

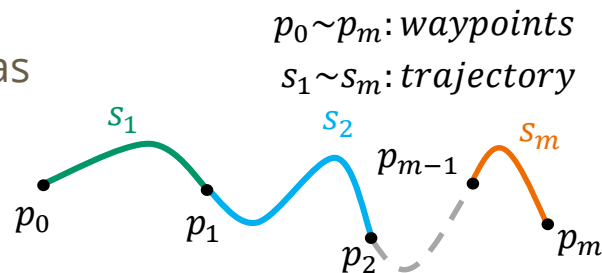
- Formulate the trajectory generation problem as a quadratic programming (QP) problem
- Write the trajectory passing through given waypoints as piecewise polynomial function of order n as

$$s_i(t) = \sum_{j=0}^n \sigma_{ij} t^j, t_{i-1} \leq t < t_i, i \in \{1, 2, \dots, m\},$$

with cost function and constraints defined as

$$\min \int_{t_0}^{t_m} \left\| \frac{d^4 s_i}{dt^4} \right\|^2 dt,$$

$$s. t. A\sigma = b$$



Motivation

Problem Formulation

Controller Design

Stability Analysis

Experiments

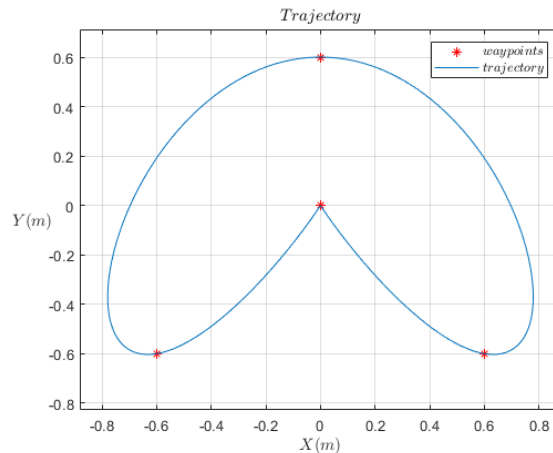
Conclusion

Experiments – Trajectory Generation

- The waypoints are given as

$$p_0 = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \quad p_1 = \begin{bmatrix} 0.6 \\ -0.6 \end{bmatrix} \quad p_2 = \begin{bmatrix} 0 \\ 0.6 \end{bmatrix} \quad p_3 = \begin{bmatrix} -0.6 \\ -0.6 \end{bmatrix} \quad p_4 = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

- The desired trajectory is generation as



Motivation

Problem
Formulation

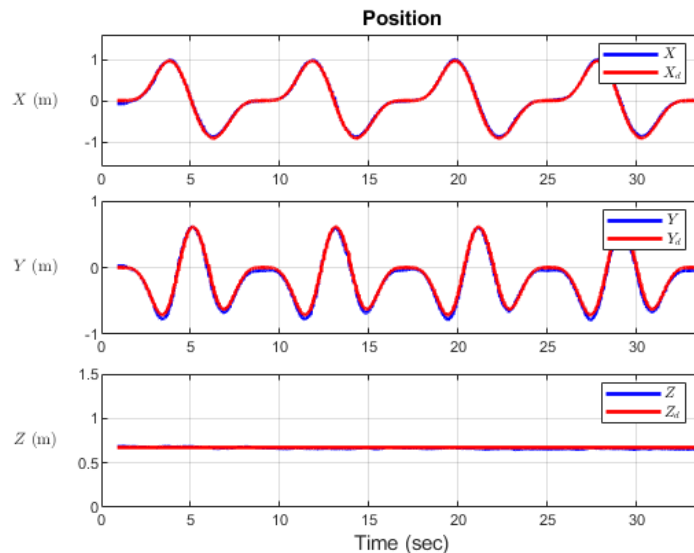
Controller
Design

Stability
Analysis

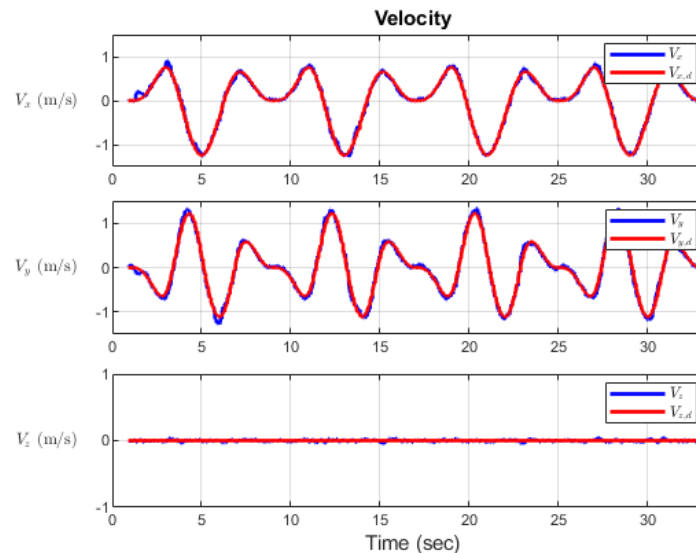
Experiments

Conclusion

Experiments – Tracking Performance



Position tracking performance of the multirotor using ICL controller



Velocity tracking performance of the multirotor using ICL controller

Motivation

Problem Formulation

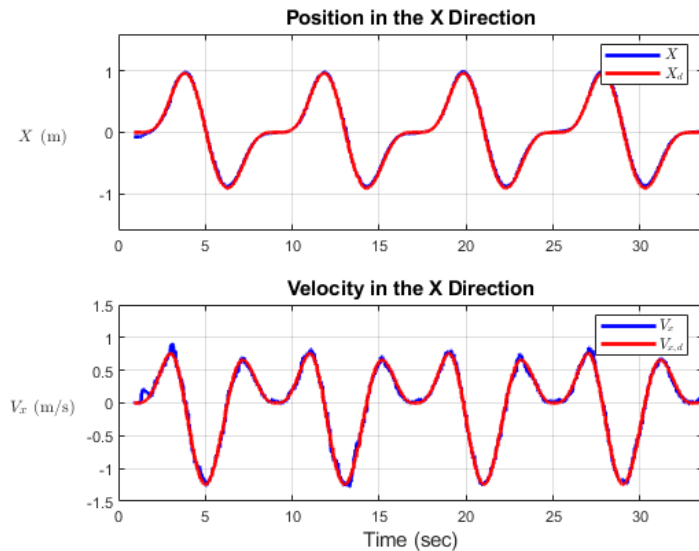
Controller Design

Stability Analysis

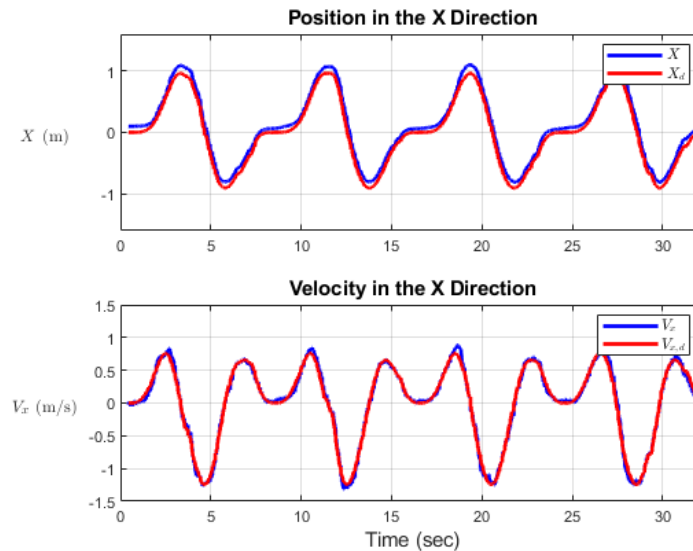
Experiments

Conclusion

Experiments – Comparison



Tracking performance of the multirotor using **ICL controller**



Tracking performance of the multirotor using **geometric controller**

Motivation

Problem Formulation

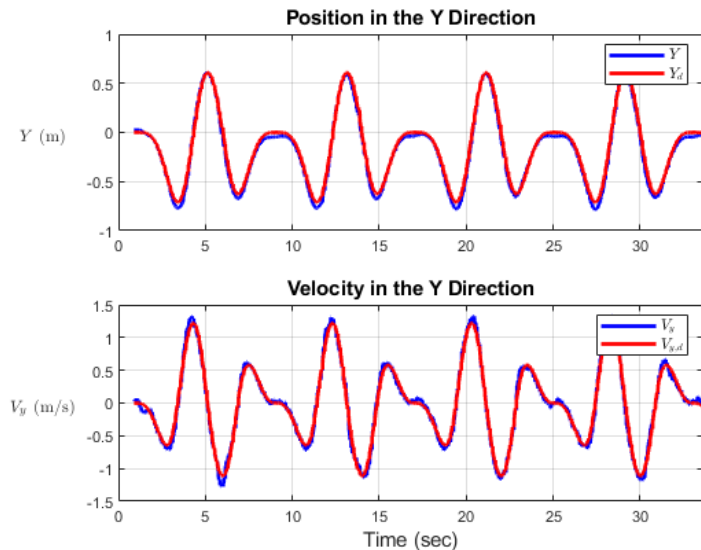
Controller Design

Stability Analysis

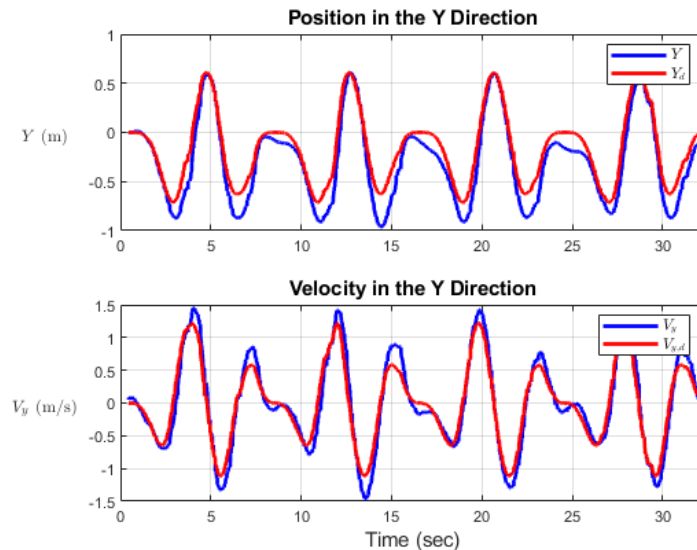
Experiments

Conclusion

Experiments – Comparison



Tracking performance of the multirotor using **ICL controller**



Tracking performance of the multirotor using **geometric controller**

Motivation

Problem Formulation

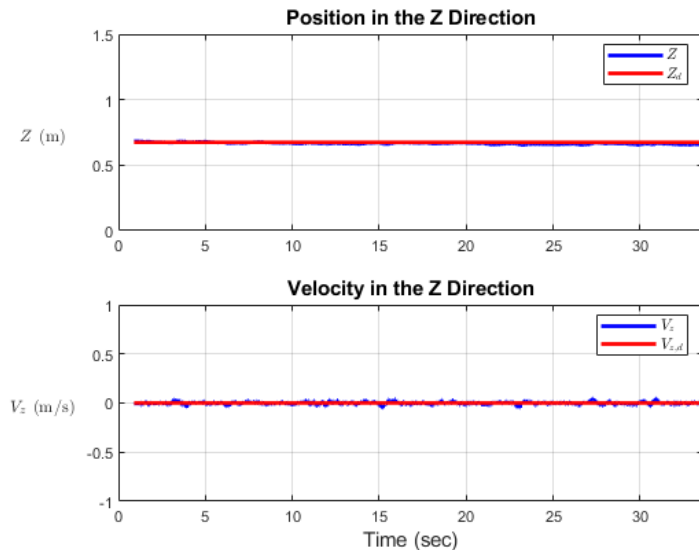
Controller Design

Stability Analysis

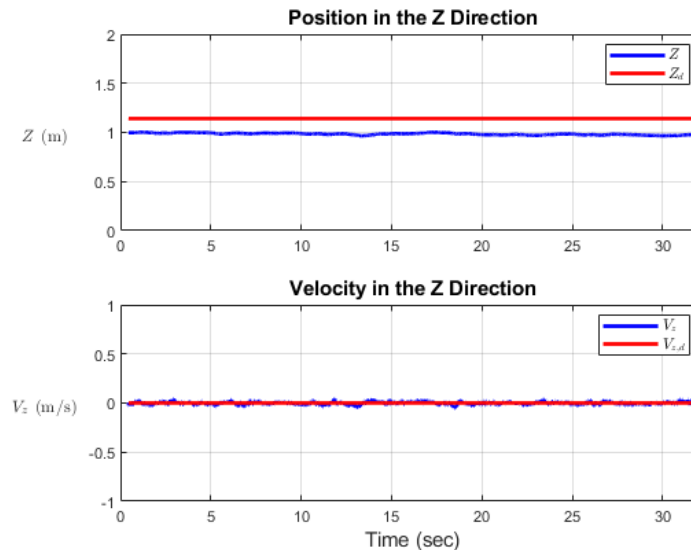
Experiments

Conclusion

Experiments – Comparison



Tracking performance of the multirotor using **ICL controller**



Tracking performance of the multirotor using **geometric controller**

Motivation

Problem Formulation

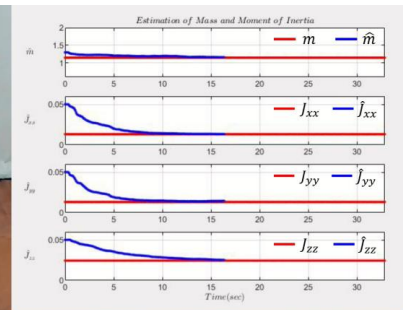
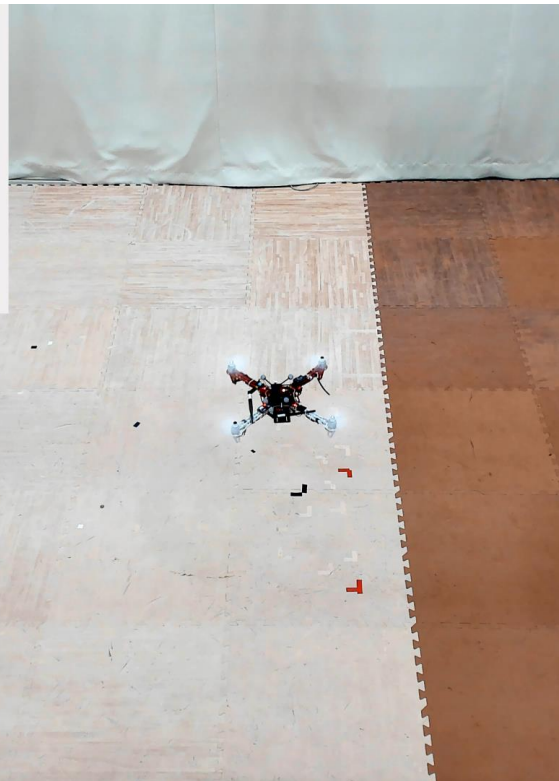
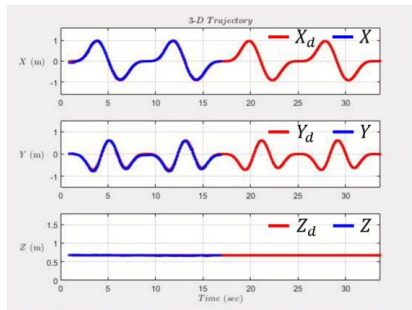
Controller Design

Stability Analysis

Experiments

Conclusion

Experiments – Video (ICL controller)



Motivation

Problem Formulation

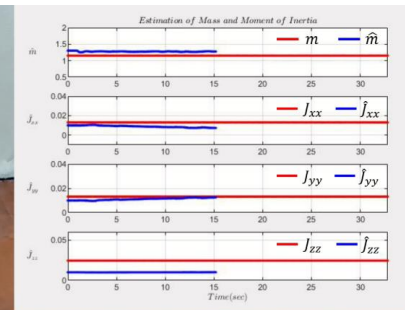
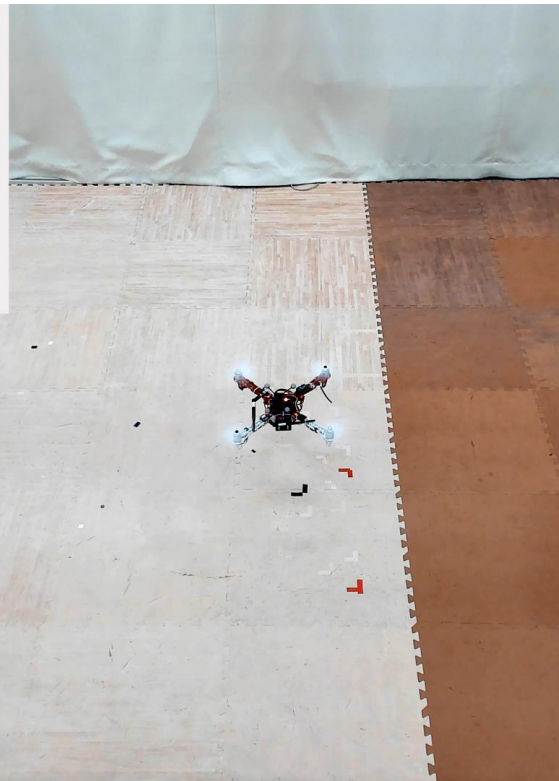
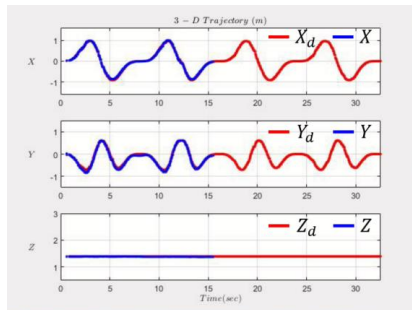
Controller Design

Stability Analysis

Experiments

Conclusion

Experiments – Video (adaptive controller)



Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Experiments – Video (geometric controller)

Motivation

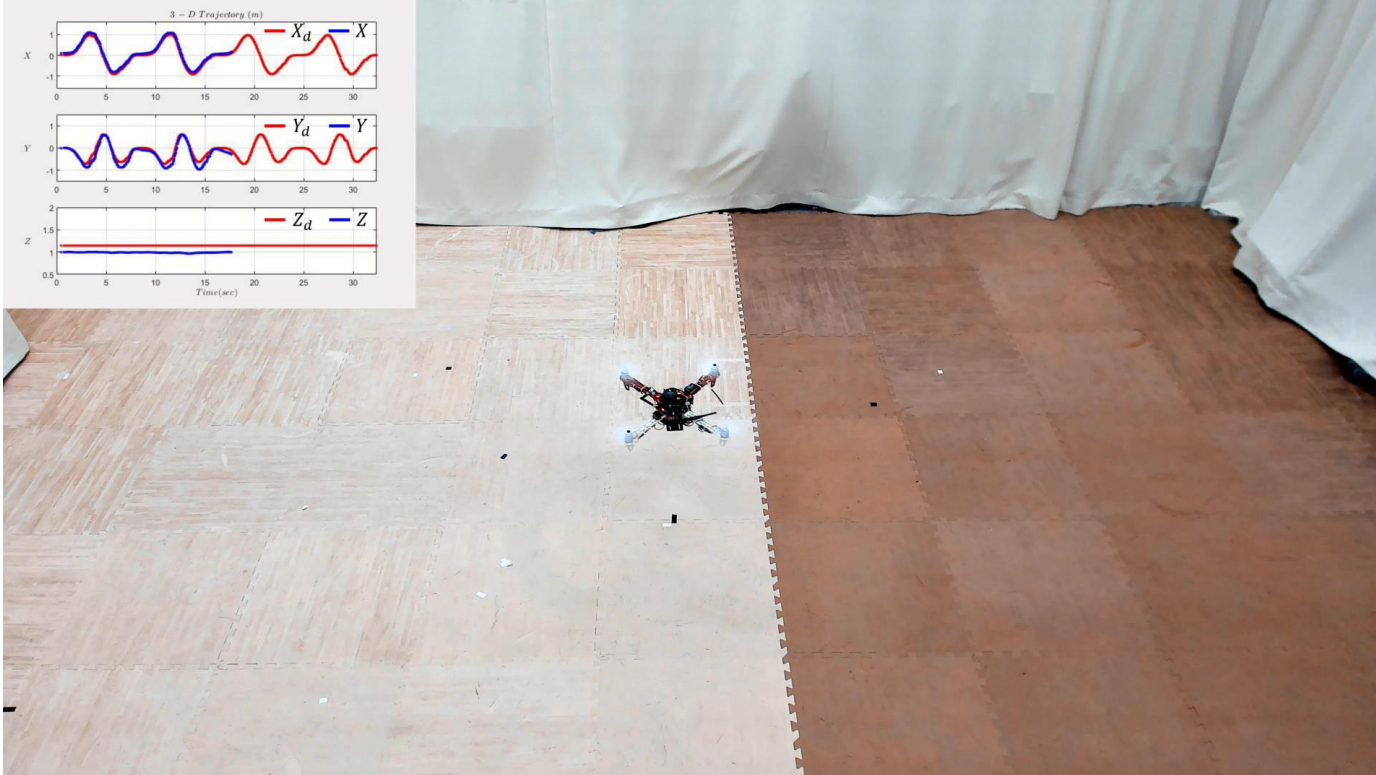
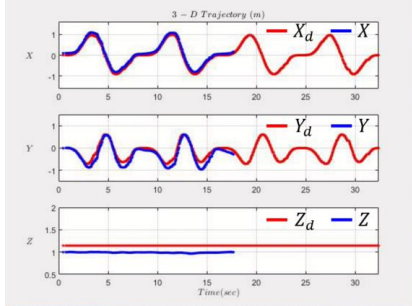
Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion



Conclusion

- An ICL controller has been developed for controlling a multirotor with an unknown mass and moment of inertia
- The control architecture can be applied to many types of multirotors of unknown mass
- The ICL controller ensures the steady-state errors resulted from the wrong parameters be eliminated
- The ICL controller can guarantee asymptotic convergence of the system parameters, while the adaptive controller cannot
- Future work can be estimate other parameters of the multirotor, such as off-diagonal elements in the inertia matrix and the center of mass.

Motivation

Problem
Formulation

Controller
Design

Stability
Analysis

Experiments

Conclusion

Thanks for listening!