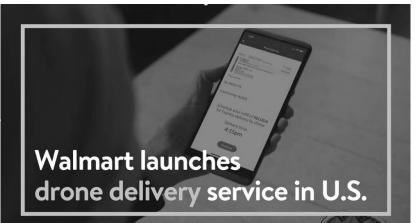


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Motivation – from Walmart's drone delivery



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Motivation – from movies Angel Has Fallen



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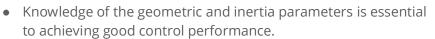
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- The payload or sensors attaching to multirotors may change the geometric and inertia parameters.
- Some geometric and inertia parameters like moment of inertia can not be measured through instrument.
- Existing adaptive control method can only guarantee the stability of multirotors system, can not ensure the parameters converge.

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Problem Formulation - Definition of Symbols

mbol Description					
Position of the multirotor					
Velocity of the multirotor					
Rotation matrix from the body- fixed frame to the inertial frame					
Angular velocity in the body- fixed frame					
Net thrust control input					
Moment control input					
Mass of the multirotor					
Moment of inertia of the multirotor					

f_4	Motivation
f_{3_4} f_{3_4} f_{3_6}	Problem Formulation
\vec{Y}_B f_2	Controller Design
\vec{X}_E \vec{X}_B	Stability Analysis
$ec{Z}_E$	Simulation and Experiments
Body-fixed frame : $\{\vec{X}_B, \vec{Y}_B, \vec{Z}_B\}$	Conclusion
Inertial frame: $\{ec{X}_E, ec{Y}_E, ec{Z}_E\}$	

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Problem Formulation - Tracking Errors and Estimate Errors

Position and velocity tracking errors

$$e_x \triangleq x - x_d$$

 $e_v \triangleq v - v_d$

• Attitude error function on SO(3) based on Geometric Tracking Control

$$\Psi(R, R_d) \triangleq \frac{1}{2} tr \left[I - R_d^T R \right]$$

Attitude tracking error and the angular velocity tracking error

$$e_R \triangleq \frac{1}{2} \left(R_d^T R - R^T R_d \right)^{\vee}$$

Estimate error of mass

 $e_{\Omega} \triangleq \Omega - R^T R_d \Omega_d$

 $\tilde{\theta}_m \triangleq \theta_m - \hat{\theta}_m$, $\theta_m = m$ (mass of the multirotor)

• Estimate error of moment of inertia



 $\tilde{\theta}_{diag} \triangleq \theta_{diag} - \hat{\theta}_{diag}$, $\theta_{diag} = [J_{xx} \quad J_{yy} \quad J_{zz}]^T$ (moment of inertia of the multirotor)

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Problem Formulation - Dynamics of the Multirotor



• The multirotor is described by both translational and rotational dynamics.

The translational dynamics considers forces such as the effects of gravity, thrusts, and the external force.

The rotational dynamics takes the moment of the control input, rotational speed, and moment of inertia into account.

$$\dot{x} = v$$

$$m\dot{v} = mge_3 - fRe_3$$

$$\dot{R} = R\hat{\Omega}$$

$$J\dot{\Omega} + \Omega \times J\Omega = M$$

Translational dynamics

Rotational dynamics

$$J = \begin{bmatrix} J_{xx} & 0 & 0 \\ 0 & J_{yy} & 0 \\ 0 & 0 & J_{zz} \end{bmatrix}$$

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Problem Formulation - Control Objectives

Track a desired 3D trajectory

Track a desired yaw angle

Estimate the mass of the multirotor

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Estimate the moment of inertia of the multirotor

$$\begin{array}{lll} (e_x & \rightarrow 0 \\ e_v & \rightarrow 0 \\ e_R & \rightarrow 0 \\ e_\Omega & \rightarrow 0 \\ \tilde{\theta}_m & \rightarrow 0 \\ \tilde{\theta}_{diag} & \rightarrow 0 \\ \end{array}$$

Design Stability

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yaw angle

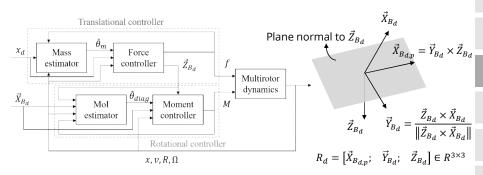
3D position

3D trajectory

desired yaw angle



Controller Design - Control Architecture



Motivation

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Controller Design - Translational Controller

• Y_m^{cl} defined as follows contains acceleration terms which is not implementable

$$fRe_3 = mge_3 - m\dot{v} = Y_m^{cl}\theta_m, \quad Y_m^{cl} = \begin{bmatrix} -\ddot{x}_1 \\ -\ddot{x}_2 \\ g - \ddot{x}_3 \end{bmatrix}$$

- By integrating Y_m^{cl} to be y_m^{cl} as defined in last page, y_m^{cl} becomes implementable
- Integrating both sides of the translational dynamics $fRe_3 = Y_m^{cl}\theta_m$ yields

$$\begin{split} &\int_{t-\Delta t}^{t} fRe_3\left(\tau\right) d\tau = \int_{t-\Delta t}^{t} Y_m^{cl}(\tau) \theta_m d\tau \Rightarrow \int fRe_3(\tau) \Big|_{\tau=t} - \int fRe_3(\tau) \Big|_{\tau=t-\Delta t} = y_m^{cl} \theta_m \\ &\hat{\theta}_m = \Gamma_m Y_m^T (e_v + C_1 e_x) + k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T \left(F(t_j) - y_m^{cl}(t_j) \, \hat{\theta}_m \right) \\ &= \Gamma_m Y_m^T (e_v + C_1 e_x) + k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T y_m^{cl}(t_j) \tilde{\theta}_m \end{split}$$

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Controller Design - Translational Controller

Translational controller

$$f = \underbrace{\begin{pmatrix} k_x e_x + k_v e_v + Y_m \hat{\theta}_m \end{pmatrix} \cdot Re_3}_{= feedback term + adaptive term}, Y_m = \begin{bmatrix} -\ddot{x}_{d_1} \\ -\ddot{x}_{d_2} \\ g - \ddot{x}_{d_3} \end{bmatrix} \text{ is a regression matrix}$$

• Integral CL-based adaptive control update law $\hat{\theta}_m$

$$\dot{\hat{\theta}}_m = \Gamma_m Y_m^T(e_v + C_1 e_x) + k_m^{cl} \Gamma_m \sum_{j=1}^{N_m} \left(y_m^{cl}(t_j) \right)^T \left(F(t_j) - y_m^{cl}(t_j) \, \hat{\theta}_m \right)$$

= adaptive term + ICL - based term

$$y_m^{cl}(t_j) \triangleq \begin{cases} 0_{n \times 1} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t Y_m^{cl}(\tau) d\tau & t > \Delta t \end{cases}, \quad F(t_j) \triangleq \begin{cases} 0_{n \times 1} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^t f Re_3(\tau) d\tau & t > \Delta t \end{cases}$$

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Controller Design - Rotational Controller

Rotational controller

$$M = -k_R e_R - k_\Omega e_\Omega - Y_{diag} \hat{\theta}_{diag} \qquad , Y_{diag} = \begin{bmatrix} \Omega_1 & \Omega_2 \cdot \Omega_3 & -\Omega_2 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_3 & \overline{\Omega}_2 & \Omega_1 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_2 & -\Omega_1 \cdot \Omega_2 & \overline{\Omega}_3 \end{bmatrix}$$

$$= feedback term + adaptive term$$

• Integral CL-based adaptive control update law $\hat{ heta}_{diag}$

$$\dot{\theta}_{diag} = \Gamma_{diag} Y_{diag}^{T}(e_{\Omega} + C_{2}e_{R}) + k_{diag}^{cl} \Gamma_{diag} \sum_{j=1}^{N} \left(y_{diag}^{cl}(t_{j}) \right)^{T} \left(\overline{M}(t_{j}) - y_{diag}^{cl}(t_{j}) \hat{\theta}_{diag} \right)$$

= adaptive term + ICL - based term

$$y_{diag}^{cl}(t_j) \triangleq \begin{cases} 0_{n \times m} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^{t} Y_{diag}^{cl}(\tau) d\tau & t > \Delta t \end{cases}, \quad \overline{M}(t_j) \triangleq \begin{cases} 0_{n \times m} & t \in [0, \Delta t] \\ \int_{t-\Delta t}^{t} M(\tau) d\tau & t > \Delta t \end{cases}$$

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Controller Design – Rotational Controller

ullet Y_{diag}^{cl} defined as follows contains angular acceleration which is not implementable

$$M = J\dot{\Omega} + \Omega \times J\Omega = Y_{diag}^{cl}\theta_{diag}, \quad Y_{diag}^{cl} = \begin{bmatrix} \dot{\Omega}_1 & -\Omega_2 \cdot \Omega_3 & \Omega_2 \cdot \Omega_3 \\ \Omega_1 \cdot \Omega_3 & \dot{\Omega}_2 & -\Omega_1 \cdot \Omega_3 \\ -\Omega_1 \cdot \Omega_2 & \Omega_1 \cdot \Omega_2 & \dot{\Omega}_3 \end{bmatrix}$$

- By integrating Y_{diag}^{cl} to be y_{diag}^{cl} as defined in last page, y_m^{cl} becomes implementable
- Integrating both sides of the translational dynamics $M = Y_{diag}^{cl} \theta_{diag}$ yields

$$\int_{t-\Delta t}^{t} M(\tau) d\tau = \int_{t-\Delta t}^{t} Y_{diag}^{cl}(\tau) \theta_{diag} d\tau \Rightarrow \int M(\tau) \Big|_{\tau=t} - \int M(\tau) \Big|_{\tau=t-\Delta t} = y_{diag}^{cl} \theta_{diag}$$

$$\begin{split} \dot{\hat{\theta}}_{diag} &= \Gamma_{diag} Y_{diag}^T (e_{\Omega} + C_2 e_R) + k_{diag}^{cl} \Gamma_{diag} \sum_{j=1}^{N} \left(y_{diag}^{cl} (t_j) \right)^T \left(\overline{M}(t_j) - y_{diag}^{cl} (t_j) \hat{\theta}_{diag} \right) \\ &= \Gamma_{diag} Y_{diag}^T (e_{\Omega} + C_2 e_R) + k_{diag}^{cl} \Gamma_{diag} \sum_{i=1}^{N} \left(y_{diag}^{cl} (t_j) \right)^T y_{diag}^{cl} (t_j) \tilde{\theta}_{diag} \end{split}$$

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Stability Analysis – Translational Dynamics

• Let Lyapunov function V₁ defined as

$$V_1 = \frac{1}{2}k_x e_x^T e_x + \frac{1}{2}m e_v^T e_v + C_1 m e_x \cdot e_v + \frac{1}{2}\tilde{\theta}_m^T \Gamma_m^{-1}\tilde{\theta}_m$$

• V_1 is P.D. and it can be lower and upper bounded by

$$z_1^T M_{11} z_1 + \frac{1}{2} \tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m \le V_1 \le z_1^T M_{12} z_1 + \frac{1}{2} \tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m$$

$$z_1 \triangleq [||e_x||, \quad ||e_v||]^T$$

$$M_{11} = \frac{1}{2} \begin{bmatrix} k_x & -C_1 m \\ -C_1 m & m \end{bmatrix}$$

$$M_{12} = \frac{1}{2} \begin{bmatrix} k_x & C_1 m \\ C_1 m & m \end{bmatrix}$$



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Stability Analysis – Closed-Loop Error Systems

Taking the time derivative of error dynamics e_x , e_y defined in Problem Formulation

$$e_x = \dot{e}_v$$

$$m\dot{e}_v = mge_3 - fRe_3 - m\ddot{x}_d$$

$$= Y_m \theta_m - f R e_3$$

$$= -k_x e_x - k_v e_v + Y_m \tilde{\theta}_m - X \quad , X = \frac{f}{e_3^T R_d^T R e_3} \Big(\big(e_3^T R_d^T R e_3 \big) R e_3 - R_d e_3 \Big)$$

• Taking the time derivative of error dynamics e_R , e_Ω defined in Problem Formulation

$$\begin{split} \dot{e}_R &= \frac{1}{2} \left(R_d^T R \hat{e}_\Omega + \hat{e}_\Omega R^T R_d \right)^\vee \\ &= \frac{1}{2} (tr[R^T R_d] I - R^T R_d) \equiv C \left(R_d^T R \right) e_\Omega \end{split}$$

$$J\dot{e}_{\Omega} = J\dot{\Omega} + J(\widehat{\Omega}R^{T}R_{d}\Omega_{d} - R^{T}R_{d}\dot{\Omega}_{d})$$

$$= J\dot{\Omega} + J\overline{\Omega} = M + Y_{diag}\theta_{diag} = -k_R e_R - k_\Omega e_\Omega + Y_{diag}\tilde{\theta}_{diag}$$





Stability Analysis – Translational Dynamics

• Taking the time derivative of V_1 yields

$$\dot{V}_1 = k_x e_x \cdot \dot{e}_x + e_v \cdot m \dot{e}_v + C_1 m \dot{e}_x \cdot e_v + C_1 e_x \cdot m \dot{e}_v - \tilde{\theta}_m^T \Gamma_m^{-1} \dot{\hat{\theta}}_m$$

• Substitute \dot{e}_r and $m\dot{e}_n$ defined in the previous page into \dot{V}_1

$$\dot{V}_{1} \leq -z_{1}^{T}W_{1}z_{1} + z_{1}^{T}W_{12}z_{2} - k_{m}^{cl}\tilde{\theta}_{m}^{T} \left(\sum_{j=1}^{N} \left(y_{m}^{cl}(t_{j})\right)^{T} y_{m}^{cl}(t_{j})\right)\tilde{\theta}_{m}$$

$$W_{1} = \begin{bmatrix} k_{x}C_{1}(1-\alpha) & -\frac{1}{2}C_{1}k_{v}(1+\alpha) \\ -\frac{1}{2}C_{1}k_{v}(1+\alpha) & k_{v}(1-\alpha) - C_{1}m \end{bmatrix}, \qquad W_{12} = \begin{bmatrix} k_{x}e_{v,max} + C_{1}B & 0 \\ B & 0 \end{bmatrix}$$

• M_{11} , M_{12} , W_1 in V_1 and \dot{V}_1 are positive-definite matrices if C_1 satisfies

$$C_1 < min \left\{ \sqrt{\frac{k_x}{m}}, \frac{k_v(1-\alpha)}{m}, \frac{4k_xk_v(1-\alpha)^2}{k_v^2(1+\alpha)^2 + 4mk_x(1-\alpha)} \right\}$$

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Stability Analysis - Rotational Dynamics

• Let Lyapunov function V_2 defined as

$$V_2 = \frac{1}{2} e_\Omega \cdot J e_\Omega + k_R \Psi(R,R_d) + J C_2 e_R \cdot e_\Omega + \frac{1}{2} \tilde{\theta}_{diag}^T \Gamma_{diag}^{-1} \tilde{\theta}_{diag}$$

• V_2 is P.D. and it can be lower and upper bounded by

$$\begin{split} z_2^T M_{21} z_2 + \frac{1}{2} \tilde{\theta}_{diag}^T \Gamma_{diag}^{-1} \tilde{\theta}_{diag} \leq V_2 \leq z_2^T M_{22} z_2 + \frac{1}{2} \tilde{\theta}_{diag}^T \Gamma_{diag}^{-1} \tilde{\theta}_{diag} \\ z_2 \triangleq [\|e_R\|, \quad \|e_\Omega\|]^T \end{split}$$

$$M_{21} = \frac{1}{2} \begin{bmatrix} k_R & -C_2 \lambda_{max}(J) \\ -C_2 \lambda_{max}(J) & \lambda_{min}(J) \end{bmatrix}$$

$$M_{22} = \frac{1}{2} \begin{bmatrix} \frac{2k_R}{2 - \psi_2} & -C_2 \lambda_{max}(J) \\ -C_2 \lambda_{max}(J) & \lambda_{min}(J) \end{bmatrix}$$

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Stability Analysis - Overall System

• Let $V = V_1 + V_2$ be a Lyapunov function for the system containing rotational and translational dynamics

$$\begin{split} V &= V_1 + V_2 \\ &= \frac{1}{2} k_x e_x^T e_x + \frac{1}{2} m e_v^T e_v + C_1 m e_x \cdot e_v + \frac{1}{2} \tilde{\theta}_m^T \Gamma_m^{-1} \tilde{\theta}_m \\ &+ \frac{1}{2} e_\Omega \cdot J e_\Omega + k_R \Psi(R, R_d) + J C_2 e_R \cdot e_\Omega + \frac{1}{2} \tilde{\theta}_{diag}^T \Gamma_{diag}^{-1} \tilde{\theta}_{diag} \text{ ... P.D.} \end{split}$$

• Taking the time derivative of V and substituting \dot{V}_1 and \dot{V}_2 yields

$$\begin{split} \dot{V} &= \dot{V}_1 + \dot{V}_2 \leq -z_1^T W_1 z_1 + z_1^T W_{12} z_2 - k_m^{cl} \tilde{\theta}_m^T \left(\sum_{j=1}^N \left(y_m^{cl}(t_j) \right)^T y_m^{cl}(t_j) \right) \tilde{\theta}_m \\ &- z_2^T W_2 z_2 - k_{diag}^{cl} \tilde{\theta}_{diag}^T \left(\sum_{j=1}^N \left(y_{diag}^{cl}(t_j) \right)^T y_{diag}^{cl}(t_j) \right) \tilde{\theta}_{diag} \quad \dots \text{ N.D.} \\ &\text{, where } \lambda_{min}(W_2) > \frac{4\|W_{12}\|^2}{\lambda_{min}(W_1)} \end{split}$$

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Stability Analysis - Rotational Dynamics

• Taking the time derivative of V_2 yields

$$\dot{V}_2 = (e_{\Omega} + C_2 e_R) \cdot (J \dot{e}_{\Omega}) + k_R e_{\Omega} \cdot e_R + J C_2 \dot{e}_R \cdot e_{\Omega} - \tilde{\theta}_{diag}^T \Gamma_{diag}^{-1} \, \dot{\hat{\theta}}_{diag}$$

• Substitute \dot{e}_R and $J\dot{e}_\Omega$ defined in the previous page into \dot{V}_2

$$\dot{V}_2 \leq -z_2^T W_2 z_2 - k_{diag}^{cl} \tilde{\theta}_{diag}^T \left(\sum_{j=1}^N \left(y_{diag}^{cl} \left(t_j \right) \right)^T y_{diag}^{cl} \left(t_j \right) \right) \tilde{\theta}_{diag}$$

$$W_2 = \begin{bmatrix} C_2 k_R & -\frac{C_2 k_{\Omega}}{2} \\ -\frac{C_2 k_{\Omega}}{2} & k_{\Omega} - C_2 \lambda_{max}(I) \end{bmatrix}$$

• M_{21} , M_{22} , W_2 in V_2 and \dot{V}_2 are positive-definite matrices if C_2 satisfies

$$C_{2} < min\left\{\frac{k_{\Omega}}{\lambda_{max}(J)}, \frac{4k_{\Omega}k_{R}}{k_{\Omega}^{2} + 4k_{R}\lambda_{max}(J)}, \sqrt{\frac{k_{R}\lambda_{min}(J)}{\lambda_{max}(J)^{2}}}\right\}$$



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Simulation – Setup and Ground Truth

- A six-rotor multirotor was used as our model in ROS Gazebo
- The ground truth of moment of inertia I and m were unknown parameters to be estimated in the simulations, and were used for evaluating the estimate error but not for implementing the controller



Parameter	Value						
m	1.568(kg)						
J	$\begin{bmatrix} 0.035 & 0 & 0 \\ 0 & 0.046 & 0 \\ 0 & 0 & 0.0977 \end{bmatrix} (kg \cdot m^2)$						
d	0.215(m)						

Motivation

Problem Formulation

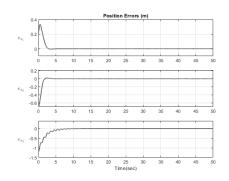
Controller Design

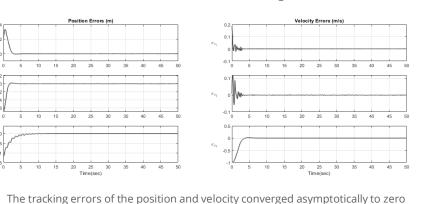
Stability Analysis

Simulation and Experiments

Simulation – Translational Error Tracking







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• The multirotor can track a desired 3D trajectory without the information of mass and moment of inertia

Motivation

Problem

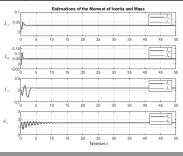
Formulation

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Simulation – Estimate Mass and Moment of Inertia NETWORKED CONTROL ROBOTICS LAB

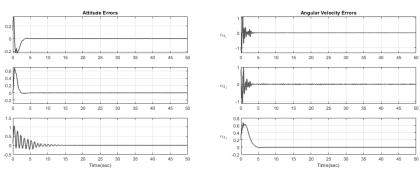
	Estimation value	Ground truth			
Mass	1.534	1.568			
Moment of inertia	$\begin{bmatrix} 0.033 & 0 & 0 \\ 0 & 0.051 & 0 \\ 0 & 0 & 0.091 \end{bmatrix}$	$\begin{bmatrix} 0.035 & 0 & 0 \\ 0 & 0.046 & 0 \\ 0 & 0 & 0.0977 \end{bmatrix}$			

- The estimates of the moment of inertia and mass during flight
- The normalized estimate errors of the moment of inertia and mass were 7.9% and 1.5%, respectively



	L	0		()	0.	.097	77]		Controlle
	Esti	mations	of the	Moment	of Inert	ia and M	lass			Design
								J_{ee} J_{ee}		,
5	10	15	20	25	30	35	40	45 50		Stability Analysis
_								J _{sx}		
5	10	15	20	25	30	35	40	45 50		Simulation
\/							_[J_{zz} J_{zz}		Experimer
V										
5	10	15	20	25	30	35	40	45 50		Conclusio
WW	~~~		-	-				θ _n		Conclusio
5	10	15	20	25	30	35	40	45 50		

Simulation – Rotational Error Tracking



• The tracking errors of the attitude and angular velocity converged asymptotically to zero

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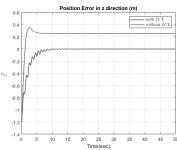
Controller Design

Stability Analysis

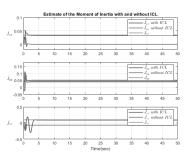
Simulation and Experiments

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Simulation - Adaptive Control v.s. Adaptive ICL Control



- The figure above compares the controller performance without and with the ICL controller
- This demonstrates the importance and robustness of our developed controller



- The figure above compares the estimated parameters when using the adaptive controller and the developed ICL controller
- Asymptotic convergence of the estimate errors converged asymptotically when using the ICL controller

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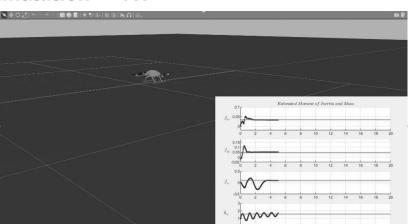
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Simulation - Video





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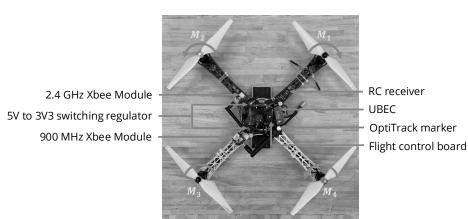
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Experiments – Hardware Architecture





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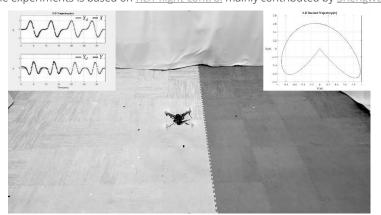
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Experiments - Video

• The experiments is based on <u>ncrl-flight-control</u> mainly contributed by <u>Shengwen</u>



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- An ICL controller has been developed for controlling a multirotor with an unknown mass and moment of inertia
- The control architecture can be applied to many types of multirotors of unknown mass, including hexacopters and octocopters
- The ICL controller ensures the steady-state errors resulted from the wrong parameters be eliminated
- The ICL controller can guarantee asymptotic convergence of the system parameters, while the adaptive controller cannot
- Future work can be estimate other parameters of the multirotor, such as off-diagonal elements in the inertia matrix and the center of mass.

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