Digital Signal Processing

Finite Length Discrete Transform (DFT)

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Outline

- Discrete Fourier Series (DFS)
- Discrete Fourier Transform (DFT)
- Relation between DTFT and DFT
- Circular Convolution
- Linear Convolution Using DFT

Discrete Fourier Series

Periodic Sequence
$$\widetilde{x}[n] = \widetilde{x}[n+rN]$$
 We have

$$\widetilde{x}[n] = rac{1}{N} \sum_{k=0}^{N-1} \widetilde{X}[k] e^{j(2\pi/N)kn}$$
 Fourier Series $\widetilde{X}[k] = \sum_{n=0}^{N-1} \widetilde{x}[n] e^{-j(2\pi/N)kn}$

To show it

$$egin{aligned} \widetilde{X}[k] &= \sum_{n=0}^{N-1} \widetilde{x}[n] e^{-j(2\pi/N)kn} = \sum_{n=0}^{N-1} rac{1}{N} \sum_{r=0}^{N-1} \widetilde{X}[r] e^{j(2\pi/N)rn} e^{-j(2\pi/N)kn} \ &= \sum_{n=0}^{N-1} \widetilde{X}[r] rac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)(r-k)n} & rac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)(r-k)n} = 1 ext{ if } k-r = mN \end{aligned}$$

$$=\sum_{j=1}^{N-1}\widetilde{X}[r] \quad ext{if} \ k-r=mN$$

DFS is periodic

$$X[k+N] = \sum_{n=0}^{N-1} \widetilde{x}[n] e^{-j(2\pi/N)\,(k+N)n} = \sum_{n=0}^{N-1} \widetilde{x}[n] e^{-j(2\pi/N)kn} e^{j2\pi} = \widetilde{X}[k]$$

Example: DFS of
$$\widetilde{x}[n] = \sum_{r=-\infty}^{\infty} \delta[n-rN]$$

$$\mathsf{DFS} \quad \widetilde{X}[k] = \sum_{n=0}^{N-1} \widetilde{x}[n] W_N^{kn} = \sum_{n=0}^{N-1} \delta[n] W_N^{kn} = 1 \quad W_N^{kn} = e^{-j(2\pi/N)kn}$$

We can write
$$\widetilde{x}[n] = \sum_{r=-\infty}^{\infty} \delta[n-rN] = \frac{1}{N} \sum_{k=0}^{N-1} \widetilde{X}[k] W_N^{-kn}$$

Example: DFS of $\tilde{x}[n] = 1$

$$\widetilde{X}[k] = \sum_{n=0}^{N-1} \widetilde{x}[n] W_N^{kn} = \sum_{r=-\infty}^{\infty} N \delta[k-rN]$$

Properties of DFS:

Linearity

$$a\widetilde{x}_1[n] + b\widetilde{x}_2[n] \overset{ ilde{DFS}}{\longleftrightarrow} a\widetilde{X}_1[k] + b\widetilde{X}_2[k]$$

Shift

$$\widetilde{x}[n-m] \stackrel{DFS}{\longleftrightarrow} W_N^{km} \widetilde{X}[k]$$

$$W_N^{-nl}\widetilde{x}[n] \overset{ extit{DFS}}{\longleftrightarrow} \widetilde{X}[k-l]$$

Duality

$$\widetilde{x}[n] \stackrel{DFS}{\longleftrightarrow} \widetilde{X}[k] \longrightarrow \widetilde{X}[n] \stackrel{DFS}{\longleftrightarrow} N\widetilde{x}[-k]$$

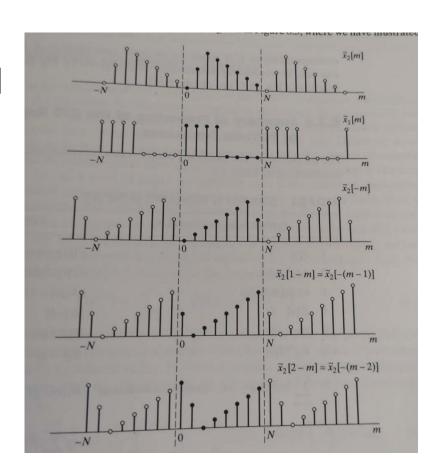
$$\widetilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \widetilde{X}[k] W_N^{-kn}$$
 $N\widetilde{x}[-n] = \sum_{k=0}^{N-1} \widetilde{X}[k] W_N^{kn}$

Periodic Convolution

 $\widetilde{x}_1[n]$ $\widetilde{x}_2[n]$ With Period of N

$$\widetilde{x}_3[n] = \sum_{m=0}^{N-1} \widetilde{x}_1[m] \widetilde{x}_2[n-m]$$

- Summation only in one period interval
- Computation needs:
 - Flip
 - Circular shift
 - Sum



Periodic Convolution

$$\widetilde{x}_1[n] \overset{DFS}{\longleftrightarrow} \widetilde{X}_1[k]$$

$$\widetilde{x}_1[n] \overset{DFS}{\longleftrightarrow} \widetilde{X}_1[k]$$
 $\widetilde{x}_2[n] \overset{DFS}{\longleftrightarrow} \widetilde{X}_2[k]$ Period of N

What is DFS for

$$\widetilde{x}_3[n] = \sum_{m=0}^{N-1} \widetilde{x}_1[m] \widetilde{x}_2[n-m]$$

$$\widetilde{X}_{3}[k] = \sum_{n=0}^{N-1} \widetilde{x}_{3}[n]W_{N}^{kn} = \sum_{n=0}^{N-1} \sum_{m=0}^{N-1} \widetilde{x}_{1}[m]\widetilde{x}_{2}[n-m]W_{N}^{kn}$$

$$=\sum_{m=0}^{N-1}\widetilde{x}_{1}[m]\sum_{n=0}^{N-1}\widetilde{x}_{2}[n-m]W_{N}^{kn}=\sum_{m=0}^{N-1}\widetilde{x}_{1}[m]\widetilde{X}_{2}[k]W_{N}^{km}=\widetilde{X}_{1}[k]\widetilde{X}_{2}[k]$$

$$egin{aligned} \widetilde{x}_3[n] &= \sum_{m=0}^{N-1} \widetilde{x}_1[m] \widetilde{x}_2[n-m] & \stackrel{DFS}{\longleftrightarrow} \widetilde{X}_1[k] \widetilde{X}_2[k] \end{aligned}$$

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Discrete Fourier Transform

Fourier Transform of Periodic signal

$$\text{Using DFS} \quad \widetilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \widetilde{X}[k] W_N^{-kn} \quad W_N^{-kn} \overset{DTFT}{\longleftrightarrow} \sum_{k=-\infty}^{\infty} 2\pi \delta \Big(w - \frac{2\pi}{N} k \Big)$$

$$\widetilde{X}(e^{j\omega}) = rac{2\pi}{N} \sum_{k=-\infty}^{\infty} \widetilde{X}[k] \delta\!\left(w - rac{2\pi}{N} k
ight)$$

Example: DFT of a periodic impulse train

$$\widetilde{p}[n] = \sum_{r=-\infty}^{\infty} \delta[n-rN]$$

$$\widetilde{P}[k] = 1$$
 for all k



$$\widetilde{P}[k] = 1 \; ext{for all} \; k \qquad \qquad \widetilde{P}(e^{j\omega}) = rac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta \Big(w - rac{2\pi}{N} k \Big)$$

Discrete Fourier Transform

Periodic signal can be expressed as

$$\widetilde{x}[n] = x[n] \circledast \widetilde{p}[n] = x[n] \circledast \sum_{r=-\infty}^{\infty} \delta[n-rN]$$

x[n] is the finite length signal defined in one period $0 \le n \le N-1$

Property of DTFT
$$\ \widetilde{X}(e^{j\omega}) = X(e^{j\omega})\widetilde{P}(e^{j\omega})$$

$$= \frac{2\pi}{N}X(e^{j\omega})\sum_{k=-\infty}^{\infty}\delta\Big(w-\frac{2\pi k}{N}\Big)$$

$$= \frac{2\pi}{N}\sum_{k=-\infty}^{\infty}X(e^{j(2\pi/N)k})\delta\Big(w-\frac{2\pi k}{N}\Big)$$
 Compare with $\ \widetilde{X}(e^{j\omega}) = \frac{2\pi}{N}\sum_{k=-\infty}^{\infty}\widetilde{X}[k]\delta\Big(w-\frac{2\pi}{N}k\Big)$

$$\widetilde{X}[k] = X(e^{j(2\pi/N)k}) = X(e^{j\omega})|_{\omega = (2\pi/N)k}$$

DFS is the uniform samples of the DTFT of finite length Sequence

Discrete Fourier Transform

$$\widetilde{x}[n] = rac{1}{N} \sum_{k=0}^{N-1} \widetilde{X}[k] W_N^{-kn}$$

Periodic sequence and finite length sequence

$$\widetilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n-rN] = x[(n)_N]$$

DFS: Discrete, Finite length, Periodic

$$\widetilde{X}[k] = X[k-rN] = X[(k)_N]$$

For one period

$$X[k] = \sum_{k=0}^{N-1} x[n]W_N^{kn} \qquad x[n] = rac{1}{N}\sum_{k=0}^{N-1} X[k]W_N^{-kn}$$

Properties of Discrete Fourier Transform

Linearity

$$ax_1[n] + bx_2[n] \stackrel{DFT}{\longleftrightarrow} aX_1[k] + bX_2[k]$$

Duality

$$x[n] \stackrel{DFT}{\longleftrightarrow} X[k]$$
 $X[n] \stackrel{DFT}{\longleftrightarrow} Nx[(-k)_N]$

Shift

$$x[(n-m)_N] \stackrel{DFT}{\longleftrightarrow} W_N^{km} X[k]$$

Define the periodic sequence

$$egin{align} \widetilde{x}_1[n] &= x[(n)_N] & \widetilde{x}_1[n] & \stackrel{DFS}{\longleftrightarrow} \widetilde{X}_1[k] \ \widetilde{x}_2[n] &= \widetilde{x}_1[n-m] &= x[(n-m)_N] & \widetilde{x}_2[n] & \stackrel{DFS}{\longleftrightarrow} W_N^{km} \widetilde{X}_1[k] \ \end{pmatrix} \end{split}$$

For one period
$$x[(n-m)_N] \stackrel{DFT}{\longleftrightarrow} W_N^{km} X[k]$$

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Relation Between DTFT and DFT

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} = \sum_{n=0}^{N-1} x[n]e^{-j\omega n}$$

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-jrac{2\pi n}{N}k} = X(e^{j\omega})|_{\omega = rac{2\pi}{N}k}$$

DFT is the uniform sample of DTFT

$$X(z) = \sum_{n=0}^{N-1} x[n] z^{-n}$$
 $X[k] = X(z)|_{z=e^{-jrac{2\pi}{N}k}}$

DFT is the uniform sample of X(z) on unit circle

Relation Between DTFT and DFT

DTFT from DFT

$$\begin{split} X(e^{j\omega}) &= \sum_{n=0}^{N-1} x [n] e^{-j\omega n} \ = \sum_{n=0}^{N-1} \left(\frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-kn} \right) e^{-j\omega n} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] \sum_{n=0}^{N-1} W_N^{-kn} e^{-j\omega n} \quad \text{Change the order of summation} \\ &\sum_{n=0}^{N-1} W_N^{-kn} e^{-j\omega n} = \frac{1 - e^{-j(\omega N - 2\pi k)}}{1 - e^{-j[\omega - (2\pi k)/N]}} = \varPhi \left(\omega - \frac{2\pi k}{N} \right) \\ &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] \varPhi \left(\omega - \frac{2\pi k}{N} \right) \qquad \qquad \varPhi (\omega) = \frac{\sin(\omega N/2)}{N \sin(\omega/2)} e^{-j\omega[(N-1)/2]} \end{split}$$

Frequency Domain interpolation

Relation Between DTFT and DFT

Frequency Domain Sampling

$$egin{aligned} Y[k] &= X(e^{j\omega})|_{\omega = 2\pi k/N} = \sum_{l = \infty}^{\infty} x[l] e^{-jrac{2\pi}{N}kl} \ y[n] &= rac{1}{N} \sum_{k = 0}^{N-1} Y[k] W_N^{-kn} = rac{1}{N} \sum_{k = 0}^{N-1} \sum_{l = \infty}^{\infty} x[l] e^{-jrac{2\pi}{N}kl} W_N^{-kn} \ &= \sum_{l = -\infty}^{\infty} x[l] rac{1}{N} \sum_{k = 0}^{N-1} W_N^{-kl} W_N^{-kn} &rac{1}{N} \sum_{k = 0}^{N-1} W_N^{-kl} W_N^{-kn} = 1 \; ext{ for } l = n + mN \ &= \sum_{l = -\infty}^{\infty} x[n + lN] \end{aligned}$$

Time Domain Shift: Alias if x[n] is longer than N

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Properties of Discrete Fourier Transform

Circular Convolution

Fircular
$$x_3[n] = x_1[n] \odot x_2[n]$$
 $= \sum_{n=0}^{N-1} x_1[m] x_2[(n-m)_N]$ $= \sum_{m=0}^{N-1} x_1[m] x_2[(n-m)_N]$ $X_3[k] = \sum_{n=0}^{N-1} x_3[n] W_N^{kn} = \sum_{n=0}^{N-1} \sum_{m=0}^{N-1} x_1[m] x_2[(n-m)_N] W_N^{kn}$ $= \sum_{m=0}^{N-1} x_1[m] \sum_{n=0}^{N-1} x_2[(n-m)_N] W_N^{kn} = \sum_{m=0}^{N-1} x_1[m] W_N^{km} X_2[k]$ $= X_1[k] X_2[k]$

$$x_1[n] \circledcirc x_2[n] \overset{{\scriptscriptstyle DFT}}{\longleftrightarrow} X_1[k] X_2[k]$$

One period of DFS

 Let g[n] & h[n] be two finite-length sequences of length N and M

$$y_{\scriptscriptstyle L}[n] = g[n] \circledast h[n]$$

Y_L[n] of length L=N+M-1

Define
$$g_e[n] = \begin{cases} g[n], 0 \le n \le N-1 \\ 0, N \le n \le L-1 \end{cases}$$
 $h_e[n] = \begin{cases} h[n], 0 \le n \le M-1 \\ 0, M \le n \le L-1 \end{cases}$

$$y_{\scriptscriptstyle L}[n] = g[n] \circledast h[n] = g_{\scriptscriptstyle e}[n] \circledcirc h_{\scriptscriptstyle e}[n]$$

$$Y_L[k] = G_e[k]H_e[k]$$
 Of length L

Finite Length convolve with infinite length sequence

$$y[n] = h[n] \otimes x[n]$$

h[n] of length *M*

Break x
$$x[n] = \sum_{m=0}^{\infty} x_m [n-mN]$$
 $X_m[n]$ of length N

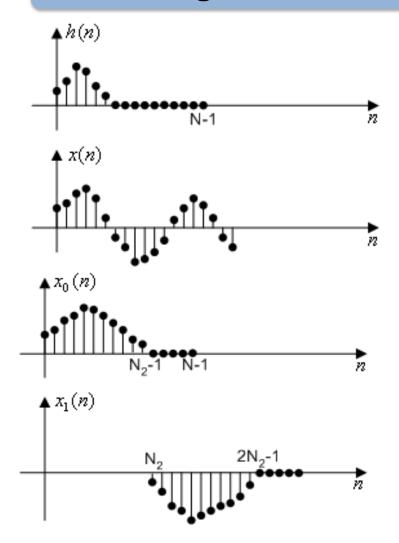
$$egin{aligned} y[n] &= h[n] \circledast x[n] = h[n] \circledast \sum_{m=0}^{\infty} x_m[n-mN] \ &= \sum_{m=0}^{\infty} h[n] \circledast x_m[n-mN] = \sum_{m=0}^{\infty} y_m[n-mN] \end{aligned}$$

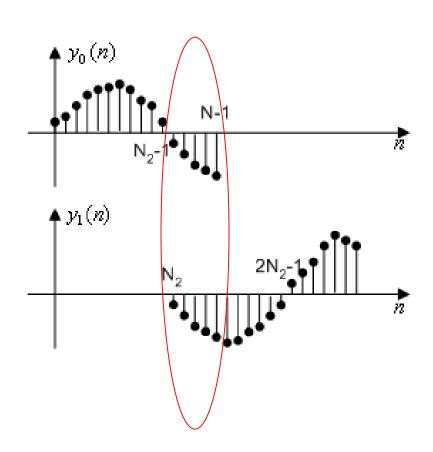
 $y_m[n]$ of length N+M-1

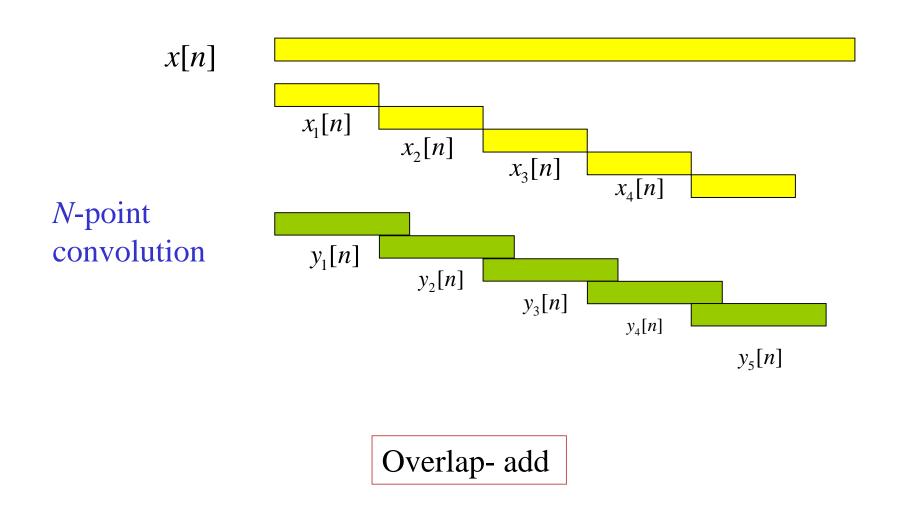


Overlap and Add

Finite Length convolve with infinite length sequence







Finite Length convolve with infinite length sequence

$$y[n] = h[n] \circledast x[n]$$

h[n] of length M

Break x
$$x[n] = \sum_{m=0}^{\infty} x_m [n - m(N - M + 1)]$$
 X_m[n] of length N

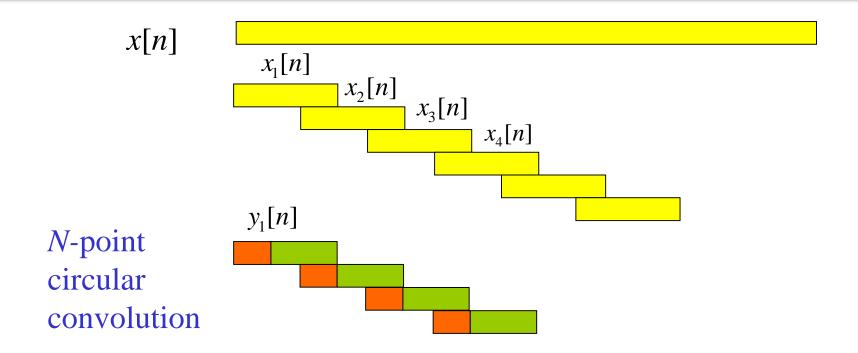
and overlaps

$$egin{align} y[n] &= h[n] \circledast x[n] = h[n] \circledast \sum_{m=0}^\infty x_m [n-m(N-M+1)] \ &= \sum_{m=0}^\infty y_m [n-mN] & y_m[n] = h[n] \circledcirc x_m[n] \end{aligned}$$

 $y_m[n]$ of length N+M-1



Overlap and Save



Overlap-save