1

Digital Signal Processing

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Abstract—This document provides the solution of Assignment 1.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://github.com/Charanyash/EE3900—
Digital_Signal_Processing/blob/main/
Assignment-1/Codes/
filter_codes_Sound_Noise.wav

2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.

2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

```
from scipy.fft import fftfreq
import soundfile as sf
from scipy import signal
#read .wav file
input signal,fs= sf.read("
    filter codes Sound Noise.wav")
#sampling frequency of Input signal
sampl freq = fs
# order of the filter
order = 4
#cutoff frequency 4kHz
cutoff freq = 4000
#digital frequency
Wn = 2*cutoff freq/sampl_freq
#b and a are numerator and denominator
   polynomials respectively.
b,a = signal.butter(order,Wn,'low')
#filter the input signal with butterworth filter.
output signal = signal.filtfilt(b,a,input signal
\#output \quad signal = signal.lfilter(b,a,
    input signal)
#write the output signal into .wav file.
sf.write('Sound_With_ReducedNoise.wav',
    output signal, fs)
```

2.4 The output of the python script Problem 2.3 is the audio file Sound With ReducedNoise.wav.Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as

background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

Solution: The plot of x(n) is given in 3.2

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 3.2.

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/main/ Assignment-1/Codes/xnyn.py

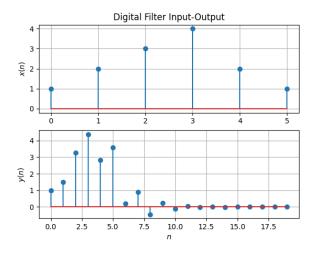


Fig. 3.2

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: Given that,

$$X(z) = \mathcal{Z}\{x(n)\}\tag{4.4}$$

$$=\sum_{n=-\infty}^{\infty}x(n)z^{-n} \tag{4.5}$$

So.

$$\mathcal{Z}\{x(n-1)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$
 (4.6)

Take k = n - 1.

$$= \sum_{k=-\infty}^{\infty} x(k) z^{-(k+1)}$$
 (4.7)

$$= z^{-1} \sum_{k=-\infty}^{\infty} x(k) z^{-k}$$
 (4.8)

$$= z^{-1} \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$
 (4.9)

$$= z^{-1}X(z) (4.10)$$

resulting in (4.2) and similarly following the above steps you will get,

$$Z\{x(n-k)\} = z^{-k}X(n)$$
 (4.11)

Hence proved.

4.2 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.12}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.11) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.13)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.14}$$

Solution: Now we will rewrite (3.2),

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2)$$
 (4.15)

Now since Z-transform is a linear operator we can write that,

$$Y(n) + \frac{1}{2}Y(n-1) = X(n) + X(n-2) \quad (4.16)$$

From (4.11),

$$Y(n) + \frac{z^{-1}}{2}Y(n) = X(n) + z^{-2}X(n)$$
 (4.17)

$$\implies \frac{Y(n)}{X(n)} = \frac{1 + z^{-2}}{1 + \frac{z^{-1}}{2}} \tag{4.18}$$

4.3 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.19)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.20)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.21}$$

Solution: The Z-transform of δn is,

$$\mathcal{Z}\left\{\delta n\right\} = \sum_{n=-\infty}^{\infty} \delta\left(n\right) z^{-n} \tag{4.22}$$

$$=\delta(0)z^{0}+0$$
 (Using (4.19)) (4.23)

$$=1 \tag{4.24}$$

and the Z-transform of unit-step function u(n) is,

$$U(n) = \sum_{n = -\infty}^{\infty} u(n) z^{-n}$$
 (4.25)

$$= 0 + \sum_{n=0}^{\infty} 1.z^{-n} \tag{4.26}$$

$$= 1 + z^{-1} + z^{-2} + \dots (4.27)$$

Above is a infinite geometric series with z^{-1} as common ratio, so we can write it as

$$U(n) = \frac{1}{1 - z^{-1}} : |z| > 1$$
 (4.28)

4.4 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a|$$
 (4.29)

Solution: The *Z*- transform will be

$$Z\{a^{n}u(n)\} = \sum_{n=0}^{\infty} a^{n}z^{-n}$$
 (4.30)

$$=1+\frac{a}{z}+\left(\frac{a}{z}\right)^2+\dots$$
 (4.31)

Above is a infinite geometric series with first

term 1 and common ratio as $\frac{a}{z}$ and it can be written as,

$$Z\{a^n u(n)\} = \frac{1}{1 - \frac{a}{z}} : |a| < |z|$$
 (4.32)

Therefore,

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a|$$
 (4.33)

4.5 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \tag{4.34}$$

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of x(n).

Solution: Download the code for the plot 4.5 from the link below

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/main/ Assignment-1/Codes/dtft.py

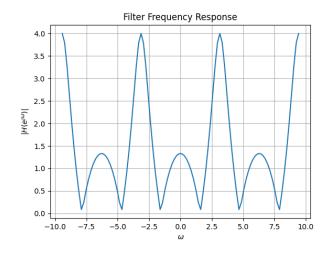


Fig. 4.5: $|H(e^{j\omega})|$

The graph is symmetric about origin and has a period of 6.25.

5 IMPULSE RESPONSE

5.1 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z)$$
 (5.1)

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse*

response of the system defined by (3.2).

Solution: The H(z) can be written as,

$$H(z) = \frac{1}{1 + \frac{z^{-1}}{2}} + \frac{z^{-2}}{1 + \frac{z^{-1}}{2}}$$
 (5.2)

From (4.29) we can write it as,

$$h(n) = \left(\frac{-1}{2}\right)^n u(n) + \left(\frac{-1}{2}\right)^{n-2} u(n-2) \quad (5.3)$$

5.2 Sketch h(n). Is it bounded? Convergent? **Solution:** Download the code for the plot 5.

Solution: Download the code for the plot 5.2 from the below link,

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/main/ Assignment-1/Codes/hn.py

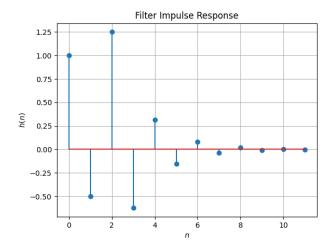


Fig. 5.2: h(n) as inverse of H(n)

From the plot 5.2, we can say that h(n) is bounded and converges to 0 as n increases. And theoritically we can the same using (5.3),

$$h(n) = \begin{cases} 0 & , n \le 0 \\ \left(\frac{-1}{2}\right)^n & , 0 < n < 2 \\ 5\left(\frac{1}{2}\right)^n & , n \ge 2 \end{cases}$$
 (5.4)

As $5\left(\frac{1}{2}\right)^n \to 0$ as $n \to \infty$ we can say h(n) is bounded and Convergent.

5.3 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.5}$$

Is the system defined by (3.2) stable for the impulse response in (5.1)?

Solution: From (5.3),

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(\left(\frac{-1}{2} \right)^n u(n) + \left(\frac{-1}{2} \right)^{n-2} u(n-2) \right)$$
(5.6)

$$=2\left(\frac{1}{1+\frac{1}{2}}\right)$$
 (5.7)

$$=\frac{4}{3}\tag{5.8}$$

: the system is stable.

5.4 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2),$$
 (5.9)

This is the definition of h(n).

Solution: Download the code for the plot 5.4 from the below link,

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/main/ Assignment-1/Codes/hndef.py

Note that this is same as 5.2.

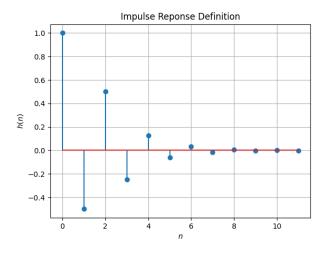


Fig. 5.4: From the definition of h(n)

5.5 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.10)

Comment. The operation in (5.10) is known as *convolution*.

Solution: Download the code for plot 5.5 from the below link

wget https://github.com/Charanyash/EE3900— Digital_Signal_Processing/blob/main/ Assignment-1/Codes/ynconv.py

Note that the plot is same that as in 3.2.

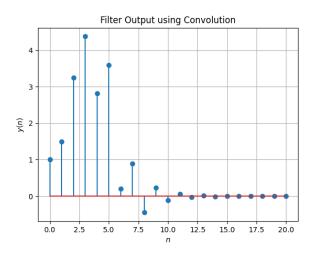


Fig. 5.5: y(n) using the convolution definition

5.6 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.11)

Solution: Substitute k := n - k in (5.10), we will get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.12)

$$=\sum_{n-k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.13)

$$=\sum_{k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.14)

6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. ??. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/yndft. py

- 6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.
- 6.5 Wherever possible, express all the above equations as matrix equations.

7 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m) y(n-m) = \sum_{k=0}^{N} b(k) x(n-k) \quad (7.1)$$

where the input signal is x(n) and the output signal is y(n) with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

- 7.2 Repeat all the exercises in the previous sections for the above a and b.
- 7.3 What is the sampling frequency of the input signal?

Solution: Sampling frequency(fs)=44.1kHZ.

7.4 What is type, order and cutoff-frequency of the above butterworth filter

Solution: The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

7.5 Modifying the code with different input parameters and to get the best possible output.