

Hybrid Beamforming for mmWave Massive MIMO

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1 Hybrid Beamforming for mmWave massive MIMO

2 Receiverside Design

3 Spectral Efficiency and Energy Efficiency Analysis

Hybrid Beamforming for mmWave massive MIMO

Hybrid Analog and Digital Beamforming Structure

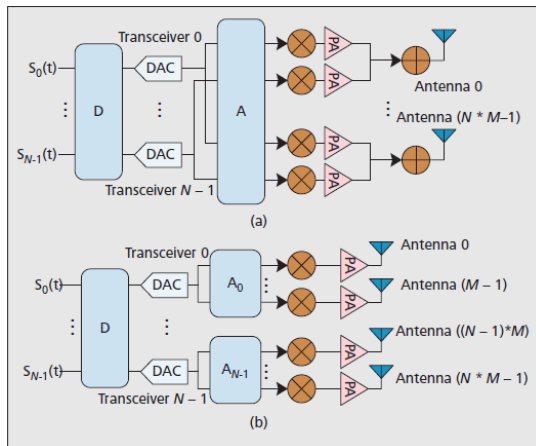


Figure: Hybrid Beamforming for LSAS system

Hybrid Beamforming Design

Prerequisites

$$N_s \leq N_t^{RF} \leq N_t$$

Transmitted signal

$$\mathbf{x} = \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{s}, \|\mathbf{F}_{RF} \mathbf{F}_{BB}\|_F^2 = N_s$$

Received Signal

$$\mathbf{y} = \sqrt{\rho} \mathbf{H} \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{s} + \mathbf{n}$$

Processed Received Signal

$$\tilde{\mathbf{y}} = \sqrt{\rho} \mathbf{W}_{BB}^* \mathbf{W}_{RF}^* \mathbf{H} \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{s} + \mathbf{W}_{BB}^* \mathbf{W}_{RF}^* \mathbf{n}$$

Rate Boundary from Shannon's Theory

$$R = \log_2 |\mathbf{I}_{N_s} + \frac{\rho}{N_s} \mathbf{R}_n^{-1} \mathbf{W}_{BB}^* \mathbf{W}_{RF}^* \mathbf{H} \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{F}_{BB}^* \mathbf{F}_{RF}^* \mathbf{H}^* \mathbf{W}_{RF} \mathbf{W}_{BB}|$$

$$\mathbf{R}_n = \mathbf{W}_{BB}^* \mathbf{W}_{RF}^* \mathbf{W}_{RF} \mathbf{W}_{BB}$$

$$(\mathbf{F}_{BB}^{opt}, \mathbf{F}_{RF}^{opt}) = \arg \max \log_2 |\mathbf{I}_{N_s} + \frac{\rho}{N_s \sigma_n^2} \mathbf{H} \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{F}_{BB}^* \mathbf{F}_{RF}^* \mathbf{H}^*|$$

$$\text{s.t. } \mathbf{F}_{RF} \in \mathcal{F}_{RF}, \|\mathbf{F}_{RF} \mathbf{F}_{BB}\|_F^2 = N_s$$

After Approximations: Objective Functions

$$\mathcal{R}(\mathbf{F}_{BB}, \mathbf{F}_{RF}) = \log_2 |\mathbf{I}_{N_s} + \frac{\rho}{N_s \sigma_n^2} \Sigma_1^2| - (N_s - \|\mathbf{V}_1^* \mathbf{F}_{RF} \mathbf{F}_{BB}\|_F^2)$$

Alternate Optimization Problem

$$(\mathbf{F}_{BB}^{opt}, \mathbf{F}_{RF}^{opt}) = \arg \min \|\mathbf{F}_{opt} - \mathbf{F}_{RF} \mathbf{F}_{BB}\|_F^2$$

$$\text{s.t. } \mathbf{F}_{RF} \in \mathcal{F}_{RF}, \|\mathbf{F}_{RF} \mathbf{F}_{BB}\|_F^2 = N_s, \mathbf{F}_{opt} = \mathbf{V}_1$$

Solutions: Spatially Sparse Precoding via Orthogonal Matching Pursuit

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Structure

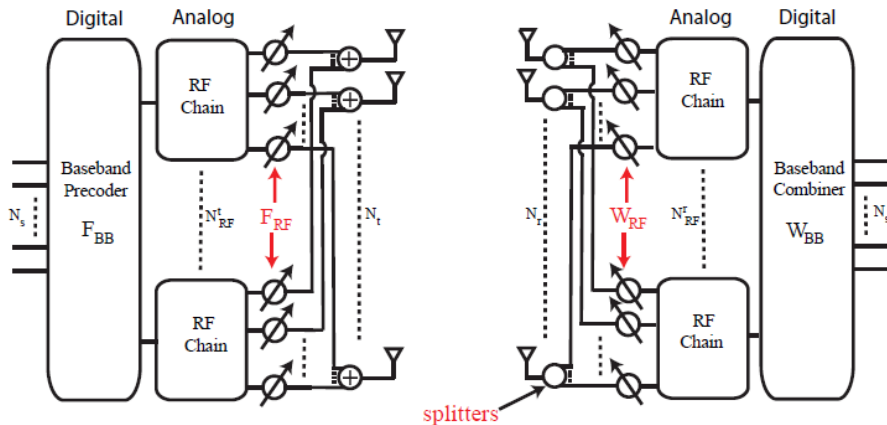


Figure: Block Diagram of Transmitter and Receiver Side

Receiver Design Criterion

$$(\mathbf{W}_{RF}^*, \mathbf{W}_{BB}^*) = \arg \min \mathbb{E}(\|\mathbf{s} - \mathbf{W}_{RF} \mathbf{W}_{BB} \mathbf{x}\|), \text{ s.t. } \mathbf{W}_{RF} \in \mathcal{W}_{RF}$$

Objective Function Reformulation

$$\mathcal{J}(\mathbf{W}_{RF}, \mathbf{W}_{BB}) = \|\mathbb{E}[\mathbf{y}\mathbf{y}^{*1/2}](\mathbf{W}_{MMSE} - \mathbf{W}_{RF} \mathbf{W}_{BB})\|_F^2$$

$$\mathbf{W}_{MMSE} = \mathbb{E}[\mathbf{s}\mathbf{y}^*]\mathbb{E}[\mathbf{y}\mathbf{y}^*]^{-1}$$

Solutions: Spatially Sparse MMSE Combining via Orthogonal Matching Pursuit

Similarities: $\|\cdot\|_F$ on Grassman manifolds, optimization for equivalent problem, Greedy matrix formulation

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$$C = WN \log_2(1 + MP\eta_{PA}/WN_0)$$

$$\eta_{SE} = \frac{C}{W} = N \log_2(1 + MP\eta_{PA}/WN_0)$$

$$\eta_{EE} = \frac{C}{P_{total}} = \frac{\eta_{SE}}{(2^{\frac{\eta_{SE}}{N}} - 1) \frac{N_0}{\eta_{PA}} \frac{N}{M} + \frac{NP_0 + P_{common} + NMP_{rfcircuit}}{W}}$$

Relationship between green points EE&SE

$$\eta_{EE} = \left(\frac{n_0 N 2^{\frac{\eta_{SE}^*}{N}} \ln 2}{L \eta_{PA}} \right)^{-1}$$

Uniqueness: For given L , η_{SE} , optimal N for η_{EE}^* is unique.

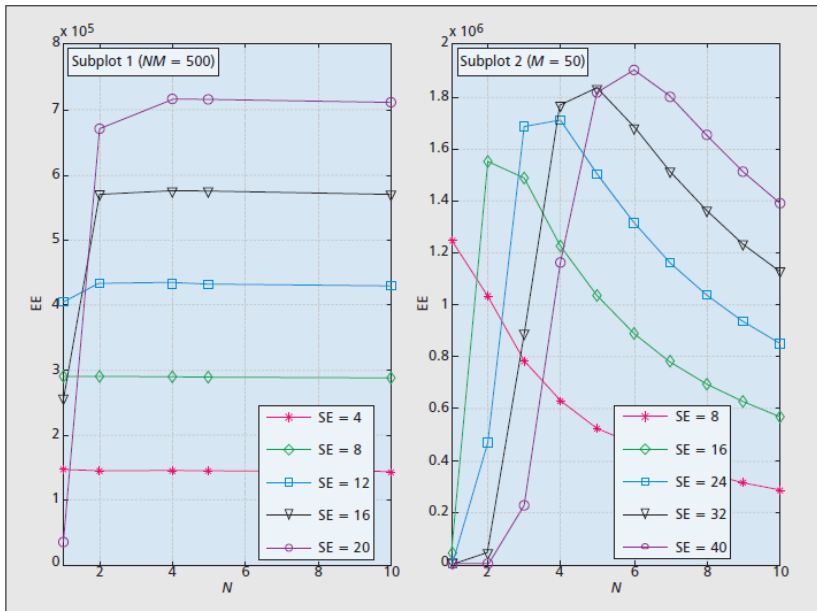




Figure: N-EE for different SE

-  S. Han, C. I I, Z. Xu, and C. Rowell, Large-scale antenna systems with hybrid analog and digital beamforming for millimeter wave 5G, *IEEE Communications Magazine*, vol. 53, no. 1, pp. 186C194, Jan. 2015.
-  O. E. Ayach, S. Rajagopal, S. Abu-Surra, Z. Pi, and R. W. Heath, Spatially Sparse Precoding in Millimeter Wave MIMO Systems, *IEEE Transactions on Wireless Communications*, vol. 13, no. 3, pp. 1499C1513, Mar. 2014.