

Frequency-domain Operators

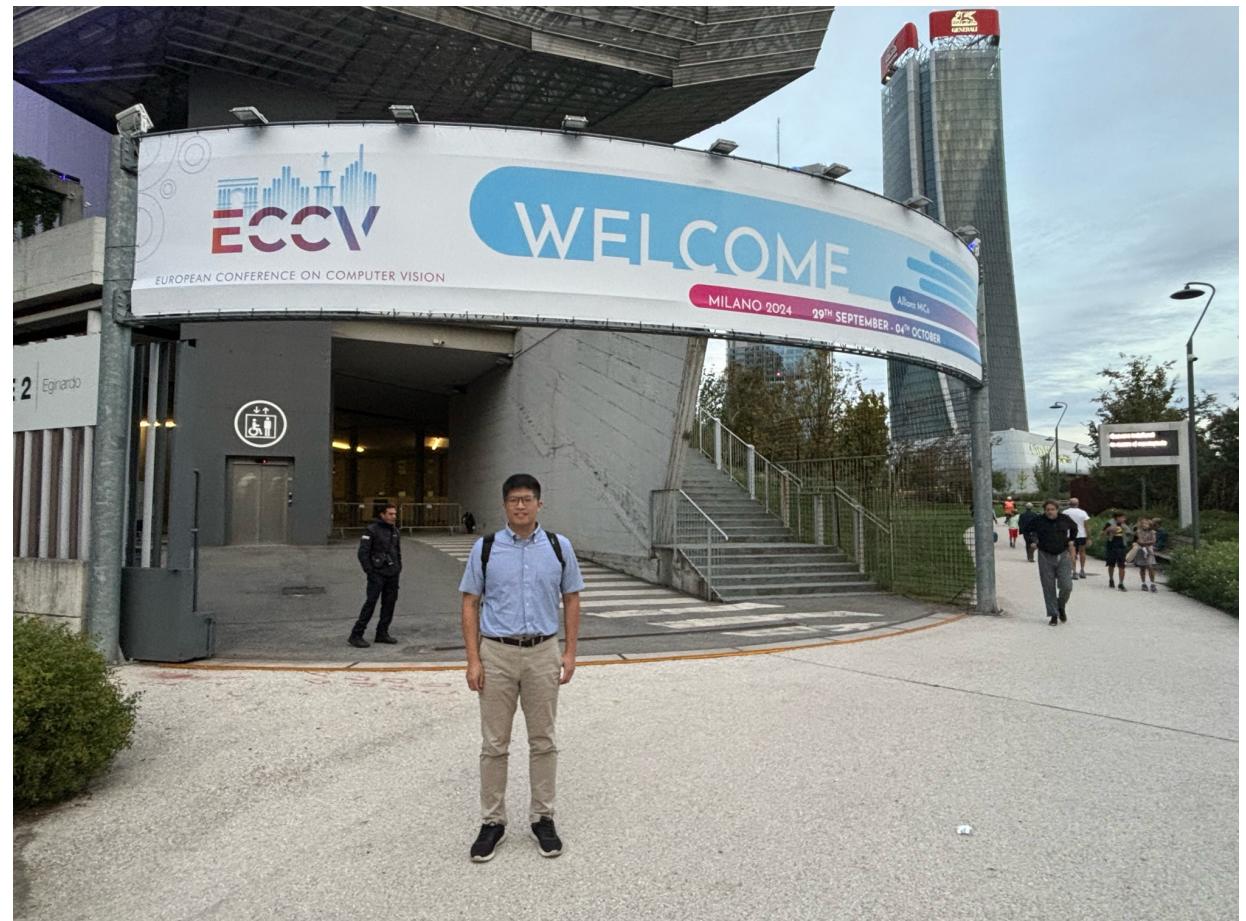
Fall 2024

Yi-Ting Chen

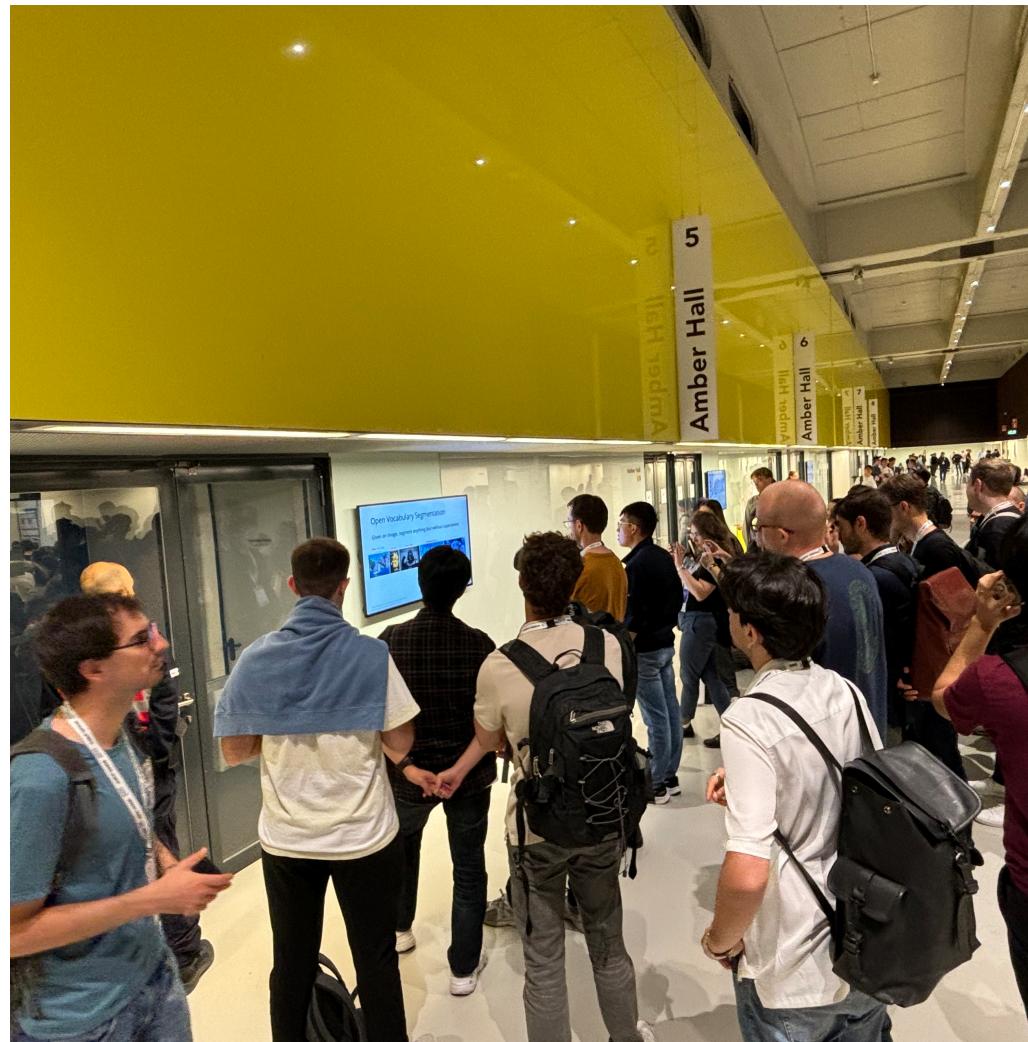
ECCV: European Conference on Computer Vision

Computer Vision and
Pattern Recognition
(CVPR)

International
Conference on
Computer Vision (ICCV)

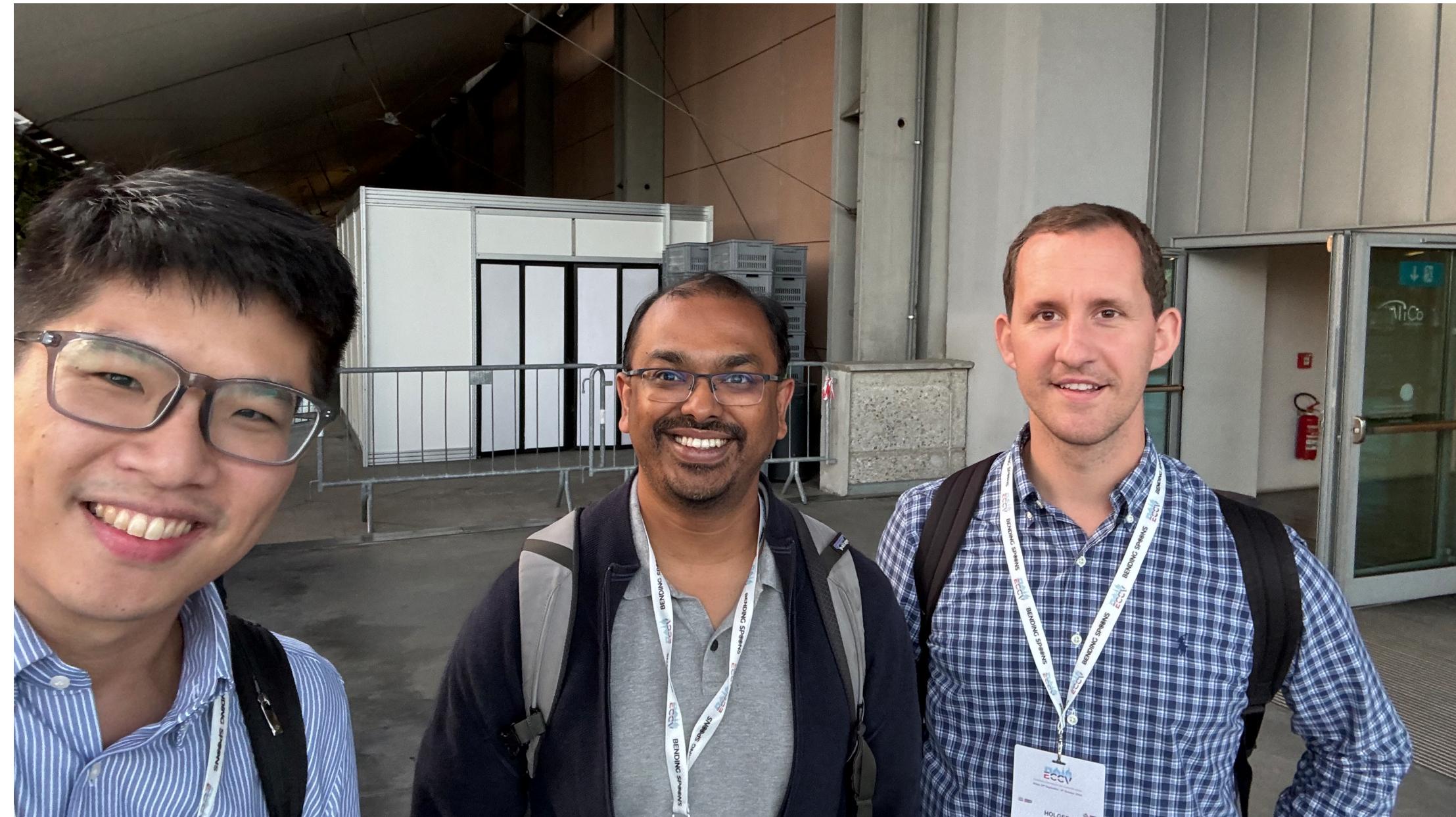


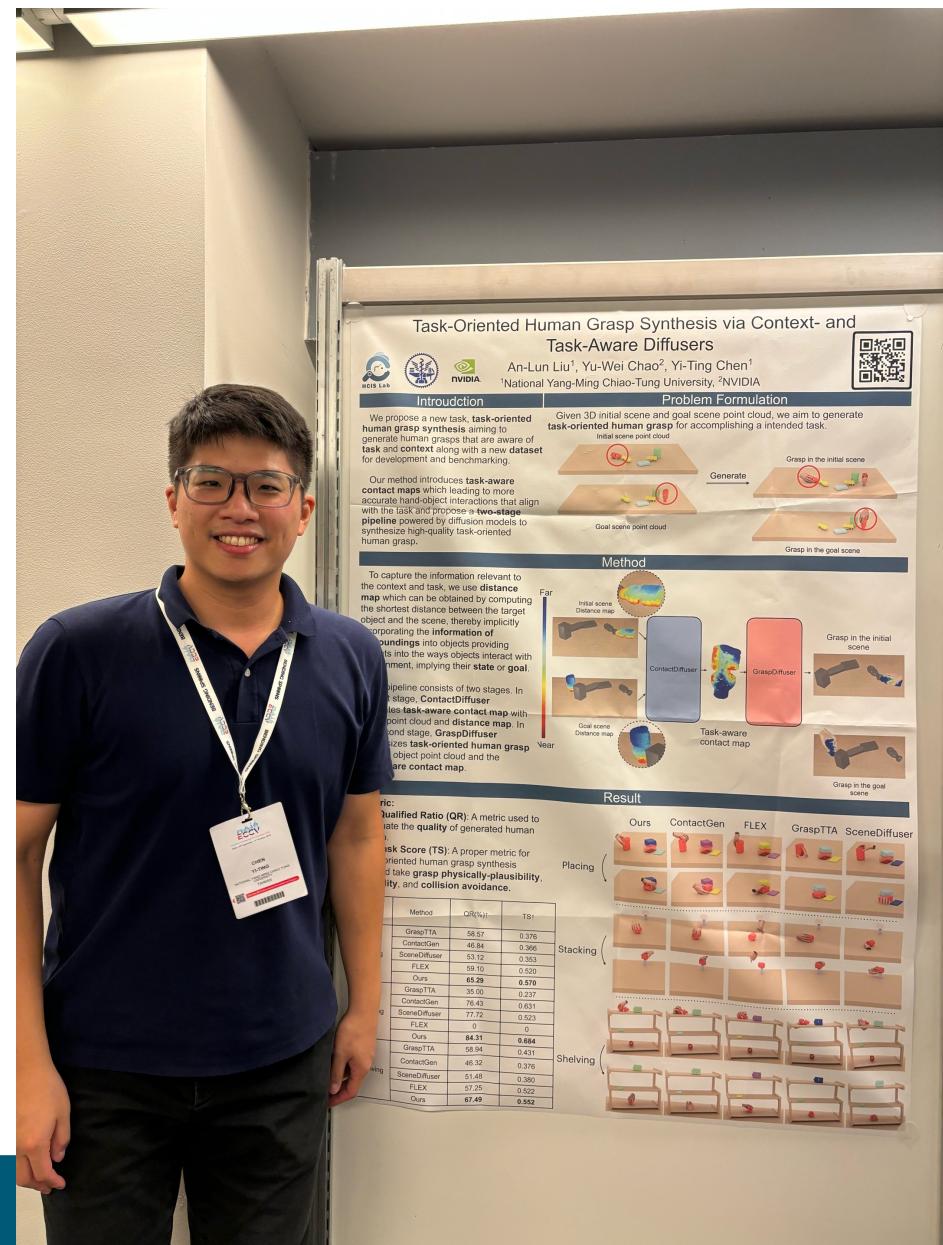












085

NeuroNCAP: Photorealistic Closed-loop Safety Testing for Autonomous Driving

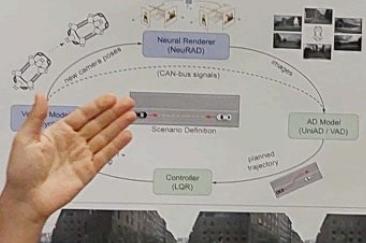
William Ljungbergh*, Adam Tonderski*, Joakim Johander, Holger Caesar, Kalle Åström, Michael Felsberg, Christoffer Petersson

TLDR

We leverage NeRFs for autonomous driving to **realistically simulate safety-critical scenarios** from a sequence of real-world data and evaluate end-to-end AD systems in closed-loop.

Photorealistic closed-loop simulation

1. The Neural Renderer generates high-quality sensor data. The renderer is trained on a sequence of real-world driving data.
2. The AD Model predicts a future ego-vehicle trajectory based on terminal camera input and the ego-vehicle state.
3. The Controller converts the planned trajectory to a set of acceleration and steering commands.
4. The Vehicle Model propagates the ego-state forward in time based on the control inputs.

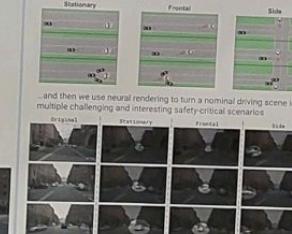


Contributions

- ★ Release open source framework for photorealistic closed-loop simulation for autonomous driving
- ★ Construct safety-critical scenarios, inspired by the industry standard Euro NCAP that cannot safely be collected in the real world.
- ★ Novel evaluation protocol that focuses on collisions in closed-loop rather than displacement metrics in open-loop.
- ★ Show that two SoTA end-to-end planners fail severely in our safety-critical scenarios despite accurately perceiving the environment.

Construct safety-critical scenarios in the wild

With inspiration from the industry standard EuroNCAP we define three safety-critical scenario types:



...and then we use neural rendering to turn a nominal driving scene into multiple challenging and interesting safety-critical scenarios



For more information, code, and examples, visit our project page at research.zenseact.com/publications/neuro-ncap

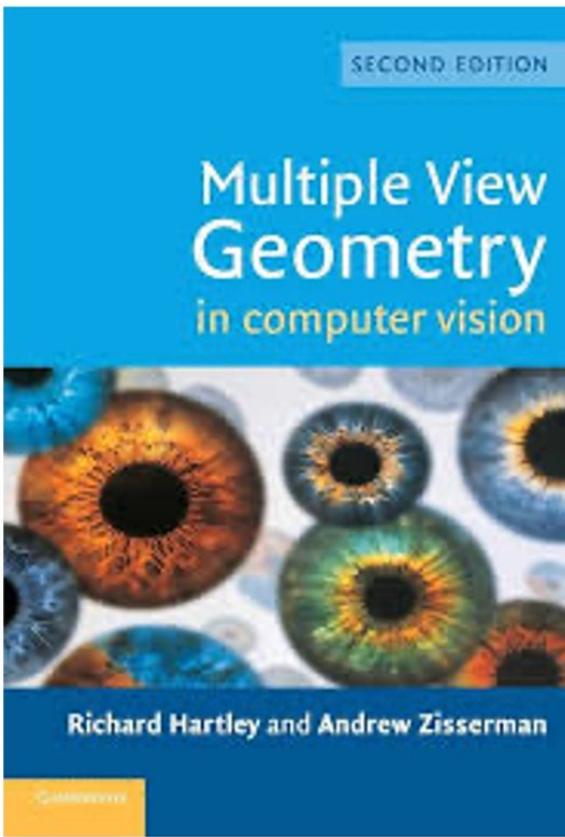
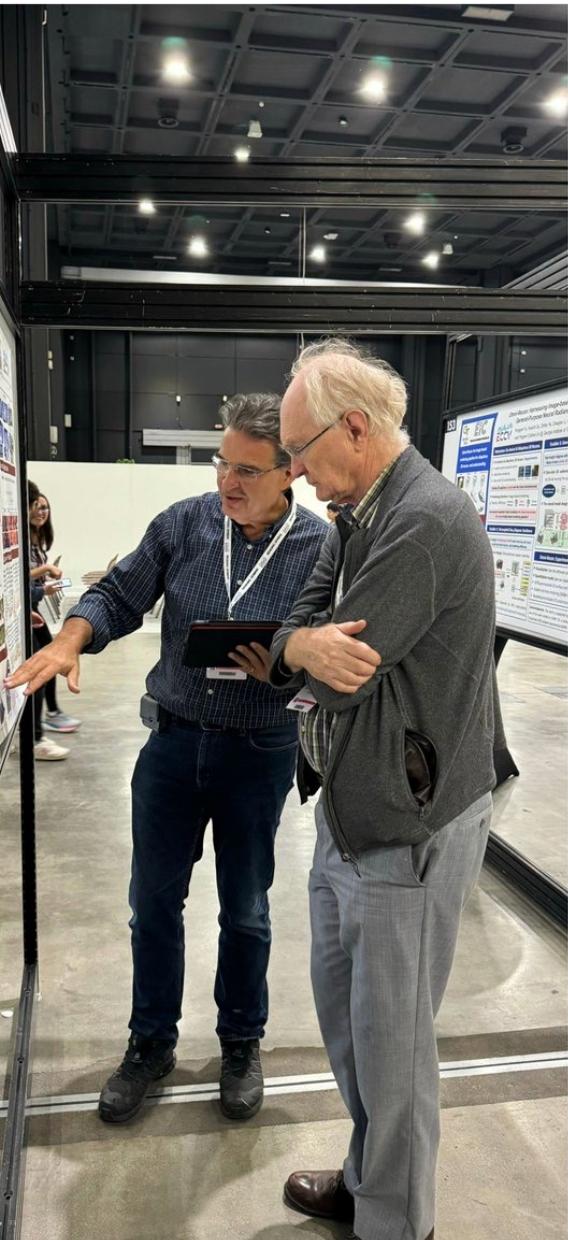
End-to-end planners fail severely in safety-critical scenarios

We evaluate end-to-end driving models and find that while they perceive the objects correctly they fail to take actions to avoid collision

$$\text{NNS} = \begin{cases} 0.0 & \text{if no collision} \\ 4.0 \cdot \max(0, 1 - v_t/v_{t_0}) & \text{otherwise} \end{cases}$$



9



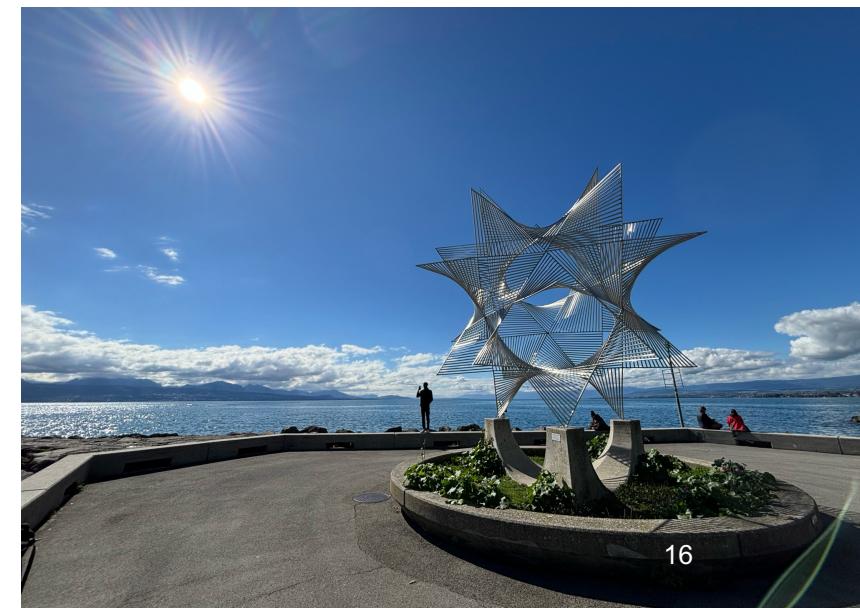
















Work Hard and Smart, Play Harder!!

7 countries this year ;-)

One work for many purposes

ROAD++ @ ECCV2024

Home Accepted Papers Challenge Dataset Speakers Venue Call for Papers Workshop Archive

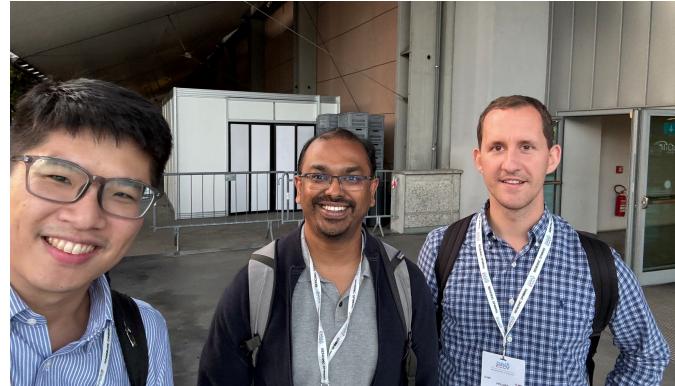
ROAD++: The Third Workshop & Challenge: Event Detection for Situation Awareness in Autonomous Driving

Co-hosted by ECCV 2024

Sep 29, MiCo Milano

Organization Committee

Prof. Fabio Cuzzolin Oxford Brookes University	Dr. Salman Khan Oxford Brookes University	Dr. Reza Javanmard Alitapchi University of Science and Technology of Mezhdunarodnaya	Dr. Eleonora Giunchiglia TU Wien	Izzeddin Teeti Oxford Brookes University
Mihaela Catalina Stolian University of Oxford	Dr. Andrew Bradley Oxford Brookes University	Dr. Gurkirt Singh Swiss Federal Institute of Technology in Zurich	Dr. Naoufel Weryghi Khalifa University	Dr. Nadya Abdel Majid Khalifa University
Chi-Hsi Kung National Yang Ming Chiao Tung University	Dr. Yi-Hsuan Tsai Google	Prof. Yi-Ting Chen National Yang Ming Chiao Tung University		



1-D Linear Time Invariant System



Linear System

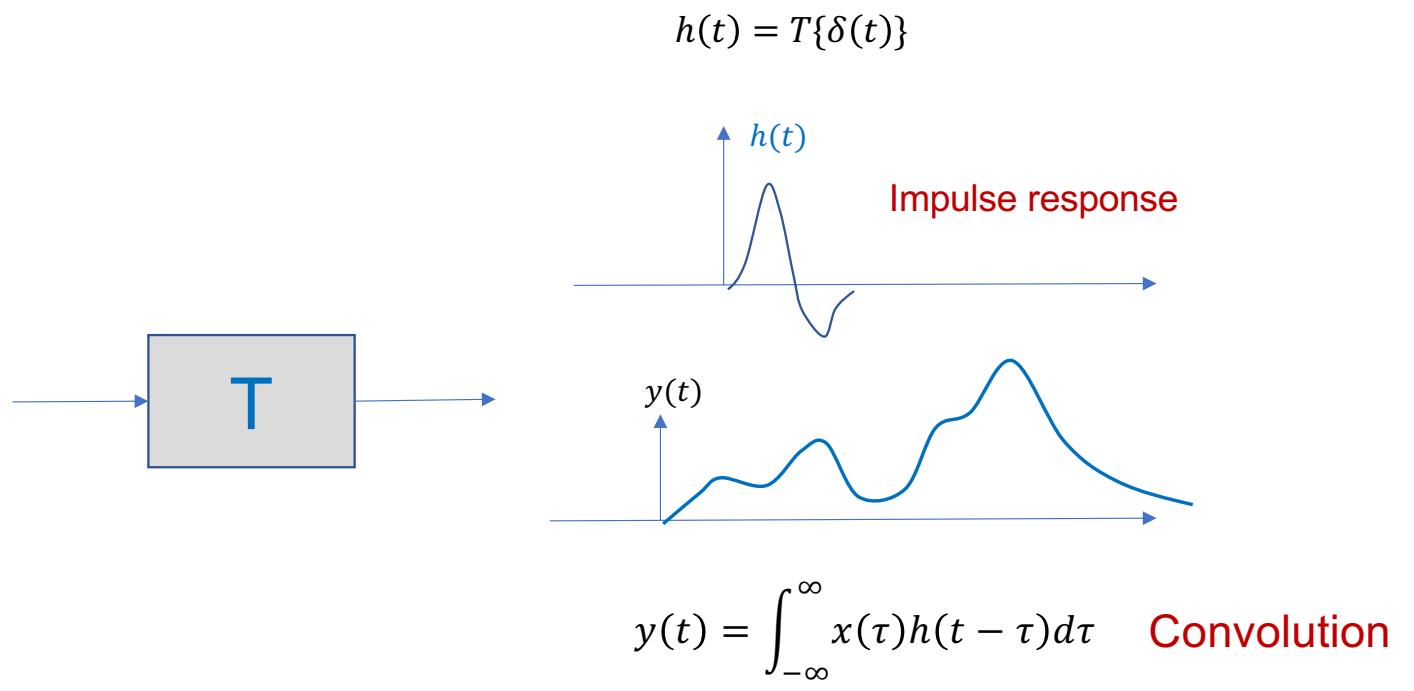
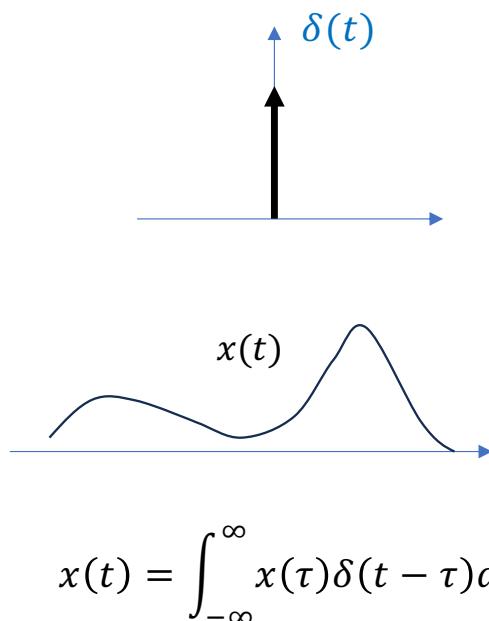
Additivity: $T\{x_1(t) + x_2(t)\} = T\{x_1(t)\} + T\{x_2(t)\} = y_1(t) + y_2(t)$

Homogeneity: $T\{ax(t)\} = aT\{x(t)\} = ay(t)$

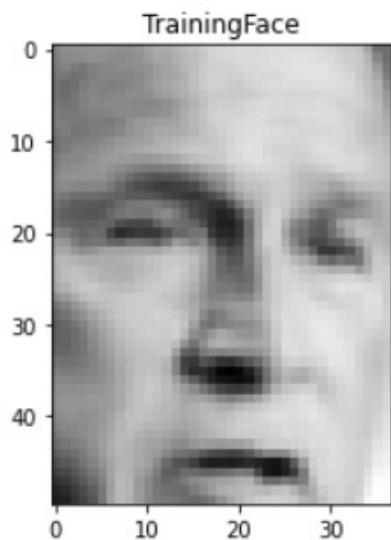
Time-Invariant System

If $y(t) = T\{x(t)\}$, then $y(t - t_0) = T\{x(t - t_0)\}$.

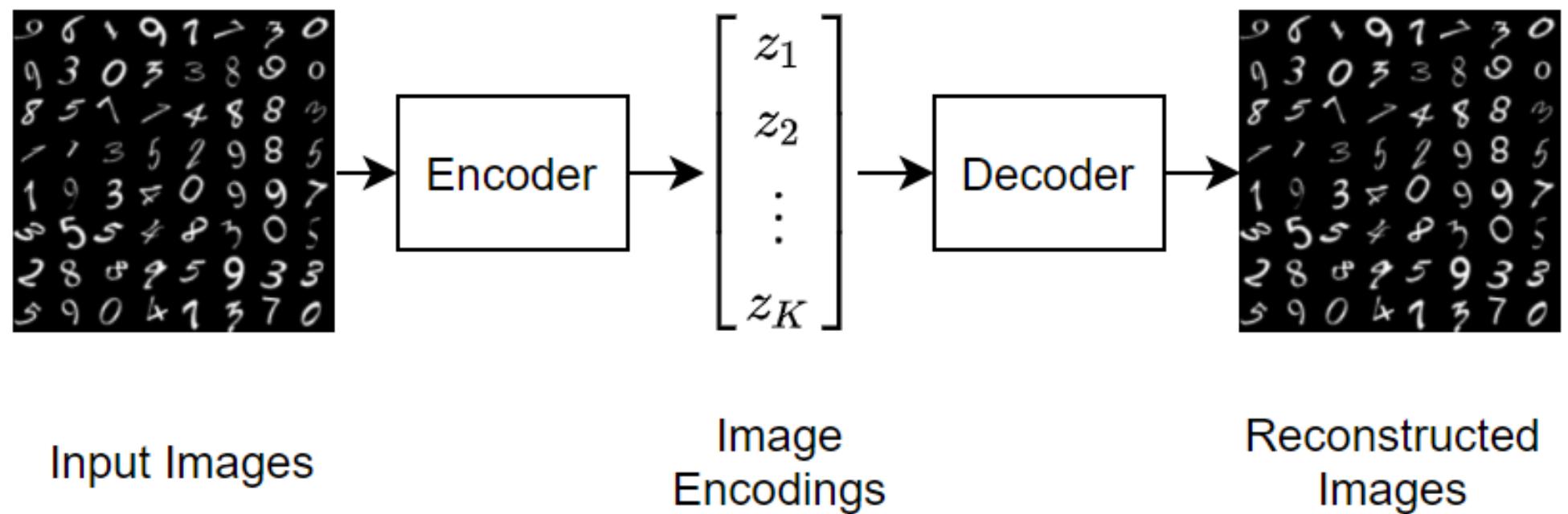
Summary



Eigenfaces

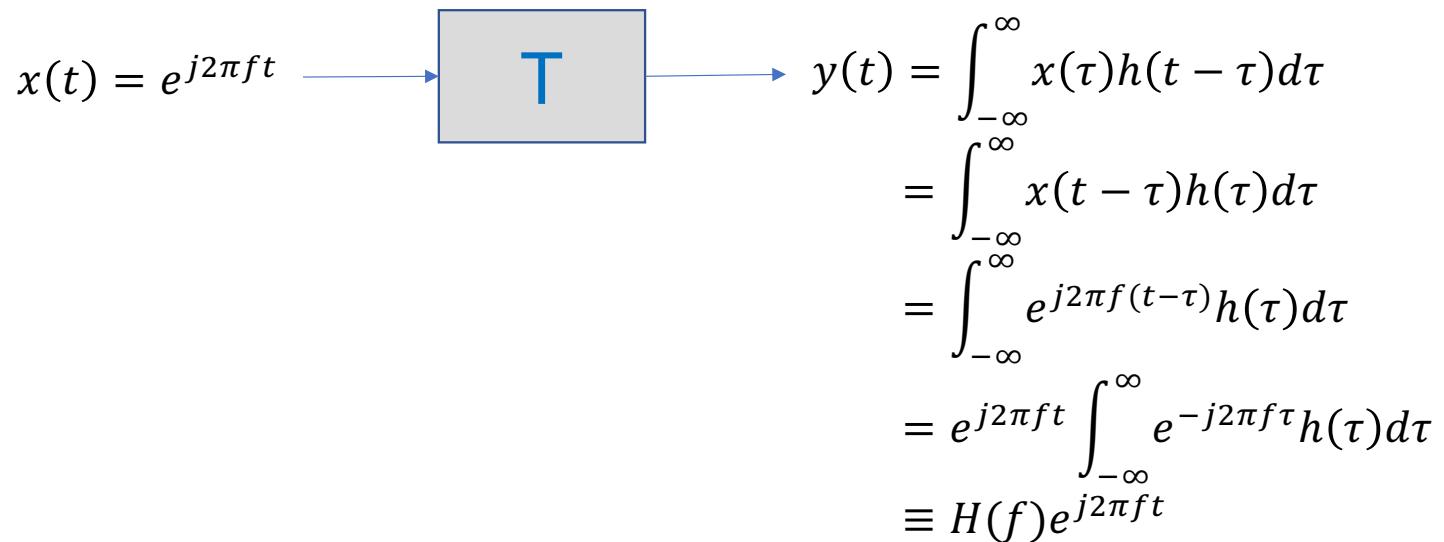


Variational Autoencoder (VAE)



Frequency-Domain Analysis (1/3)

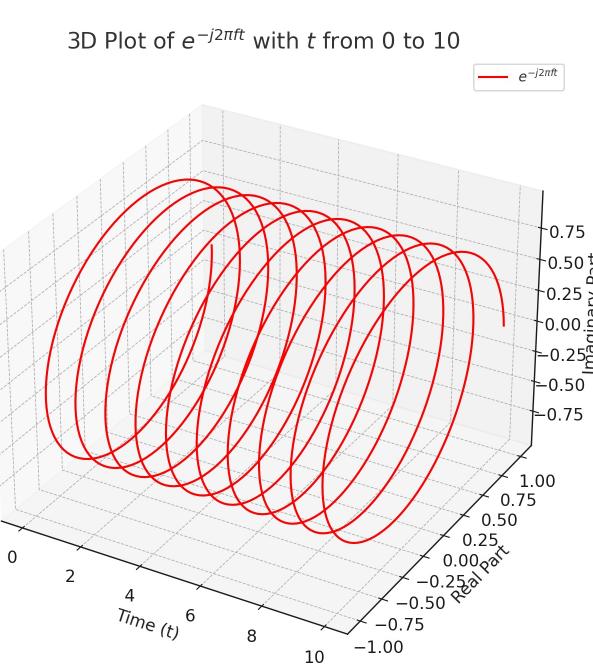
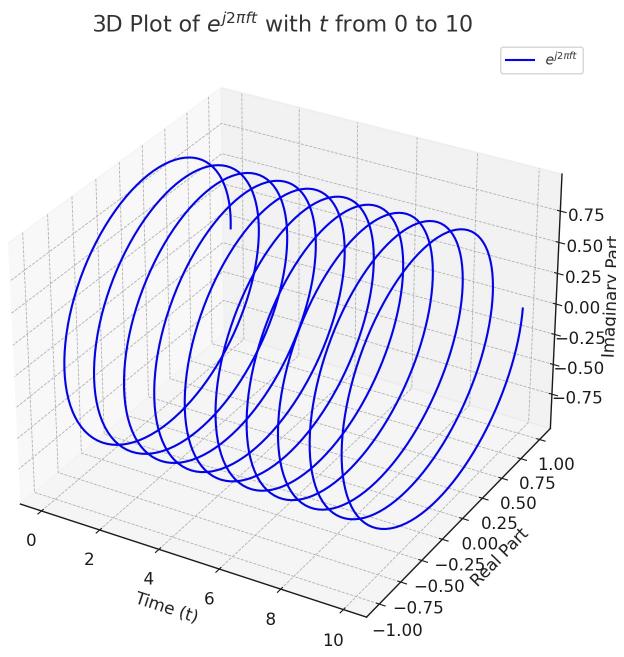
$$e^{j2\pi ft} = \cos(2\pi ft) + j\sin(2\pi ft)$$



Frequency-Domain Analysis (1/3)

$$e^{j2\pi ft} = \cos(2\pi ft) + j\sin(2\pi ft)$$

$$e^{-j2\pi ft} = \cos(2\pi ft) - j\sin(2\pi ft)$$



Frequency-Domain Analysis (1/3)

$$\begin{aligned}y(t) &= \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau \\&\equiv H(f)e^{j2\pi ft}\end{aligned}$$

H(f): Frequency Response

This is a complex-valued function that characterizes how the system responds to a sinusoid of frequency f . It typically includes both a gain (magnitude) and a phase shift.

$$y(t) = |H(f)|e^{j(2\pi ft + \arg(H(f)))}$$

Can we use this characteristic to decompose our signal?

Frequency-Domain Analysis (2/3)

Fourier Transform Pair

spectrum
$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$$

where $e^{-j2\pi ft} = \cos(2\pi ft) - j\sin(2\pi ft)$

Signal
Decomposition
$$x(t) = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$$

where $e^{j2\pi ft} = \cos(2\pi ft) + j\sin(2\pi ft)$

$$X(f) = Re\{X(f)\} + jIm\{X(f)\} = |X(f)|e^{j\angle X(f)}$$

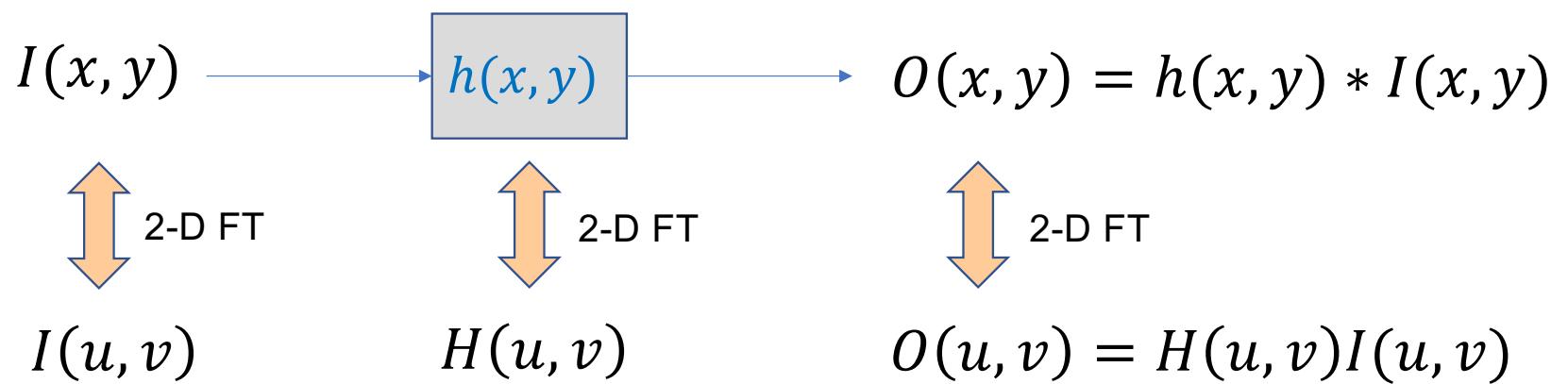
$|X(f)|$ Magnitude

$\angle X(f)$ Phase

Frequency-Domain Analysis (3/3)

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi f t} df \rightarrow \boxed{T} \rightarrow y(t) = \int_{-\infty}^{\infty} X(f) T\{e^{j2\pi f t}\} df$$
$$= \int_{-\infty}^{\infty} X(f) H(f) e^{j2\pi f t} df$$
$$\Rightarrow Y(f) = X(f)H(f)$$

2-D Linear Shift-Invariant System



h(x,y): Point Spread Function (PSF)
H(u,v): optical transfer function (OTF)

2-D Fourier Transform

$$F(u, v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi(ux+vy)} dx dy$$

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

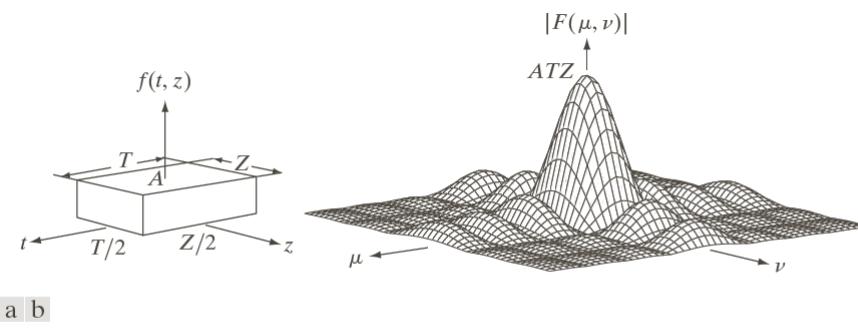
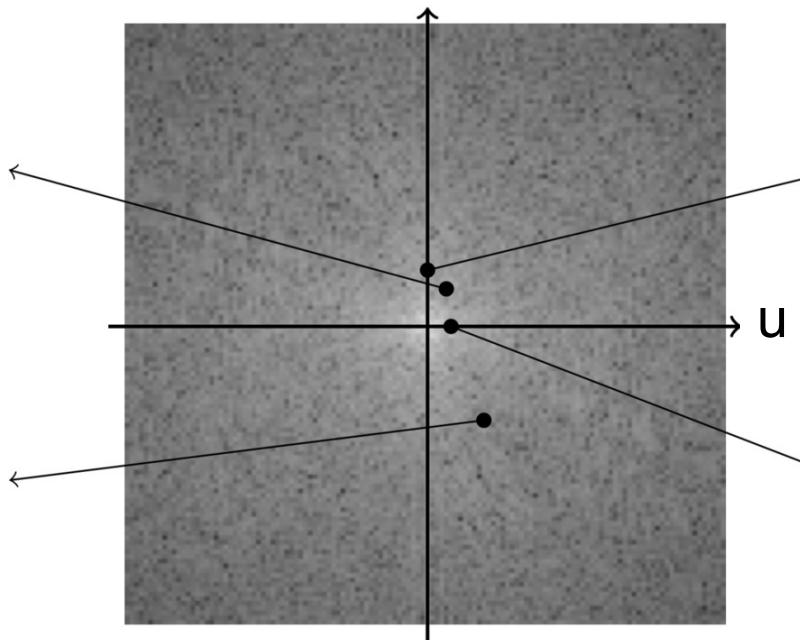
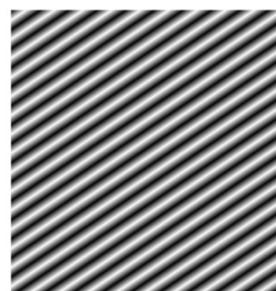
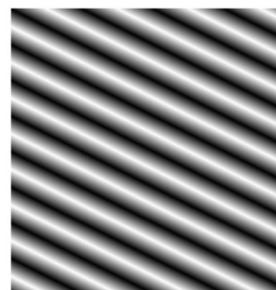


FIGURE 4.13 (a) A 2-D function, and (b) a section of its spectrum (not to scale). The block is longer along the t -axis, so the spectrum is more “contracted” along the μ -axis. Compare with Fig. 4.4.

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

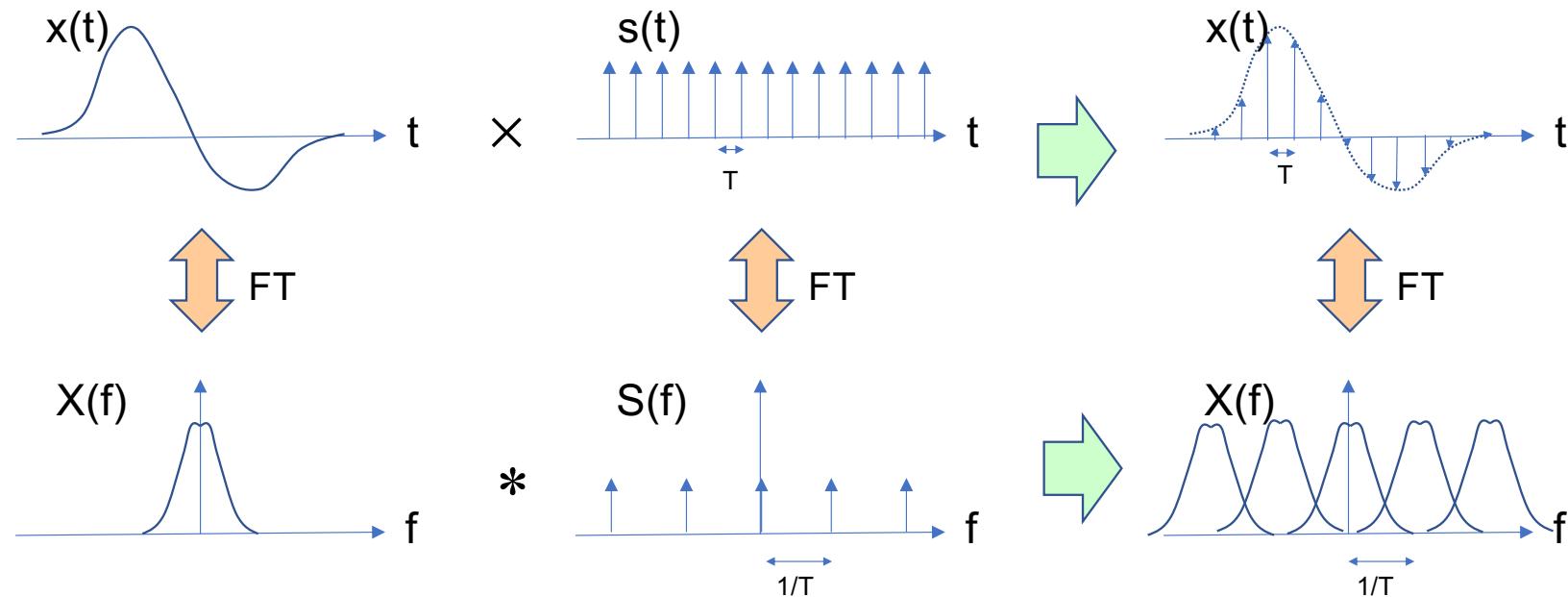
$F(u, v)$



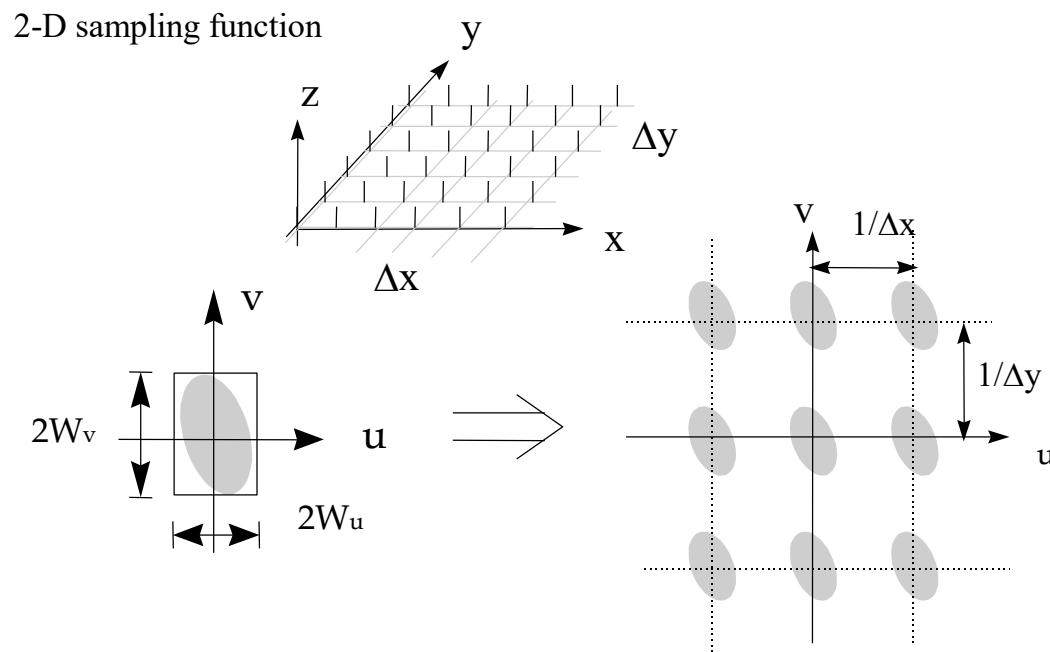
https://commons.wikimedia.org/wiki/File:2D_Fourier_Transform_and_Base_Images.png

Sampling

Whittaker-Shannon Sampling Theorem



2-D Sampling



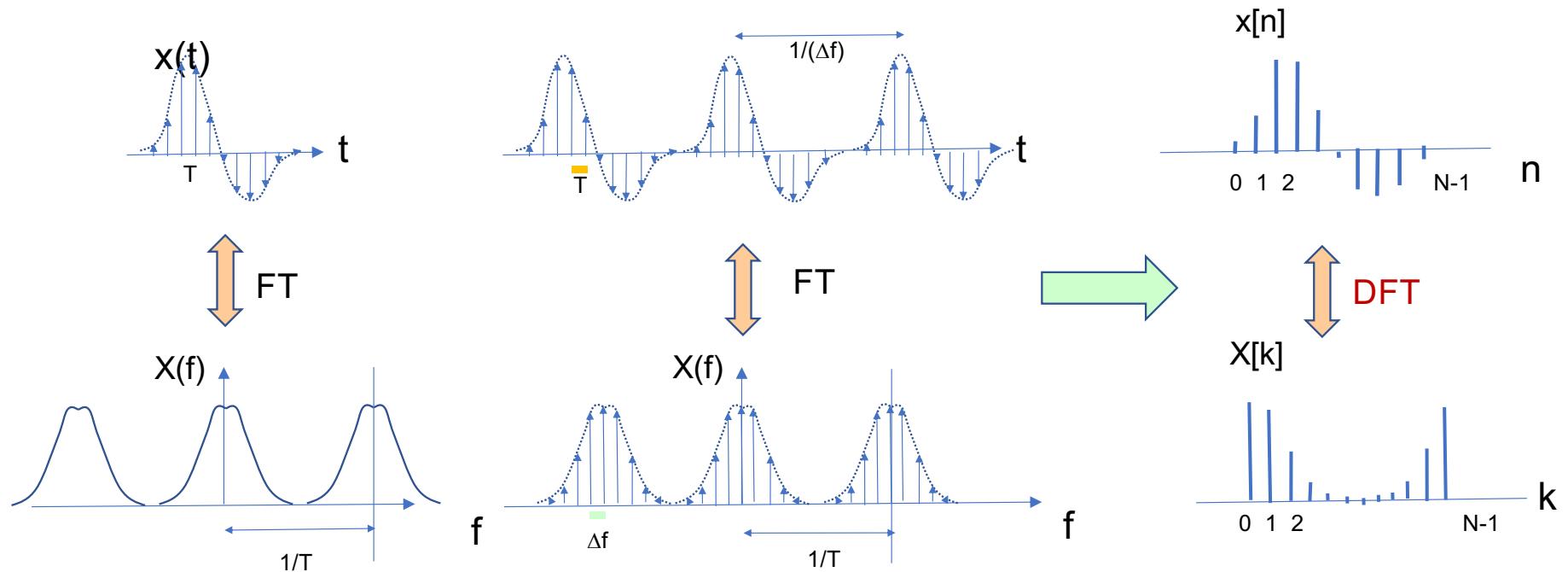
Aliasing



a b c

FIGURE 4.17 Illustration of aliasing on resampled images. (a) A digital image with negligible visual aliasing. (b) Result of resizing the image to 50% of its original size by pixel deletion. Aliasing is clearly visible. (c) Result of blurring the image in (a) with a 3×3 averaging filter prior to resizing. The image is slightly more blurred than (b), but aliasing is not longer objectionable. (Original image courtesy of the Signal Compression Laboratory, University of California, Santa Barbara.)

Discrete Fourier Transform (1/2)



Discrete Fourier Transform (2/2)

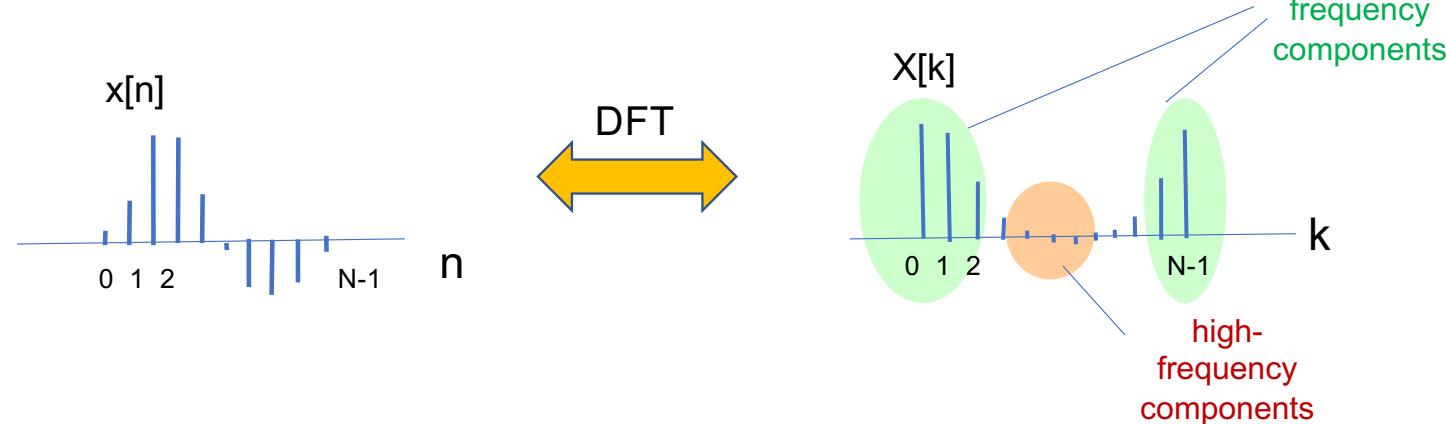
Fourier Transform Pair

$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$$

$$x(t) = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$$

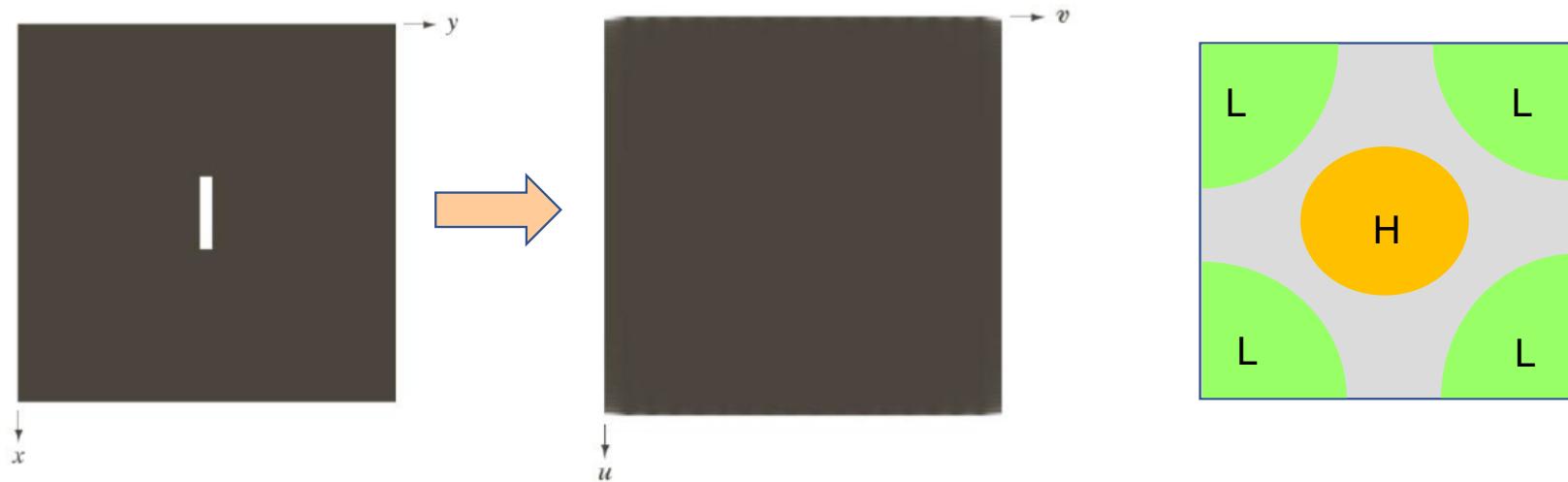
$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n]e^{-j\frac{2\pi kn}{N}} \quad k = 0, 1, 2, \dots, N-1$$

$$x[n] = \sum_{k=0}^{N-1} X[k]e^{j\frac{2\pi kn}{N}} \quad n = 0, 1, 2, \dots, N-1$$



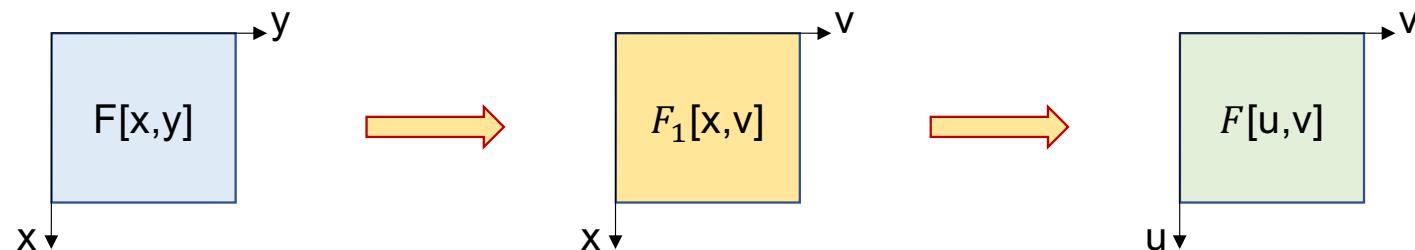
2-D DFT

$$F[u, v] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m, n] e^{-j2\pi(\frac{um}{M} + \frac{vn}{N})}$$
$$f[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F[u, v] e^{j2\pi(\frac{um}{M} + \frac{vn}{N})}$$



Properties of DFT - Separability

$$\begin{aligned} F(u, v) &= \frac{1}{N^2} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x, y) e^{-j2\pi \frac{ux+vy}{N}} \\ &= \frac{1}{N} \sum_{x=0}^{N-1} e^{-j2\pi \frac{ux}{N}} \left\{ \sum_{y=0}^{N-1} \frac{1}{N} f(x, y) e^{-j2\pi \frac{vy}{N}} \right\} \end{aligned}$$



$$F[u, v] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m, n] e^{-j2\pi(\frac{um}{M} + \frac{vn}{N})}$$

$$f[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F[u, v] e^{j2\pi(\frac{um}{M} + \frac{vn}{N})}$$

- **Average Value** $\bar{f}[x, y] = \frac{1}{N^2} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f[x, y] = F[0, 0]$
- **Periodicity** $f[x, y] = f[x + kN, y + lN]$
 $F[u, v] = F[u + kN, v + lN]$ $k, l = 0, 1, 2, \dots$
- **Translation**

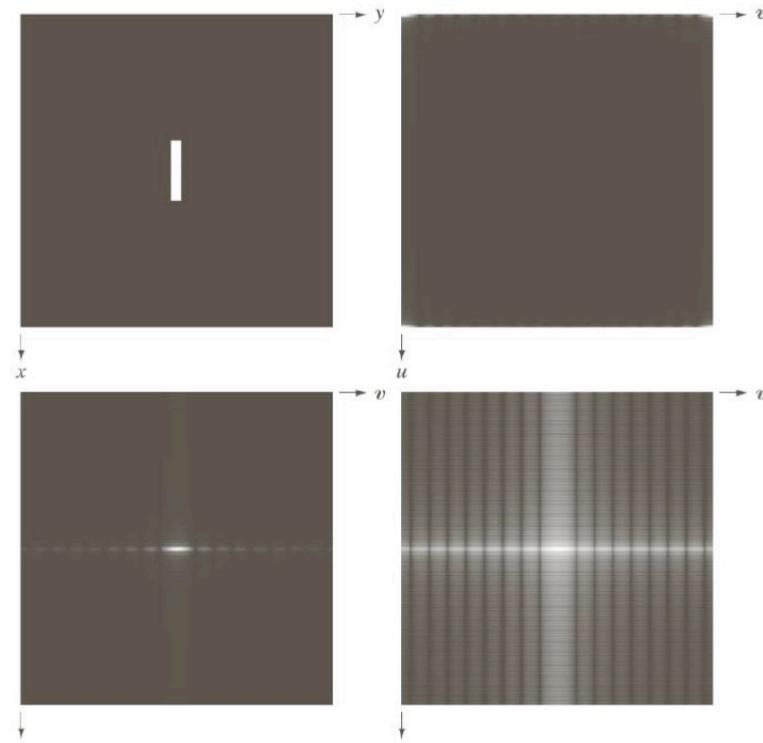
$f[x - x_0, y - y_0]$		$F[u, v] e^{-j2\pi(\frac{ux_0 + vy_0}{N})}$
$f[x, y] e^{j2\pi(\frac{u_0x + v_0y}{N})}$		$F[u - u_0, v - v_0]$

If $u_0 = v_0 = N/2$

$$f[x, y](-1)^{(x+y)}$$

↓ DFT

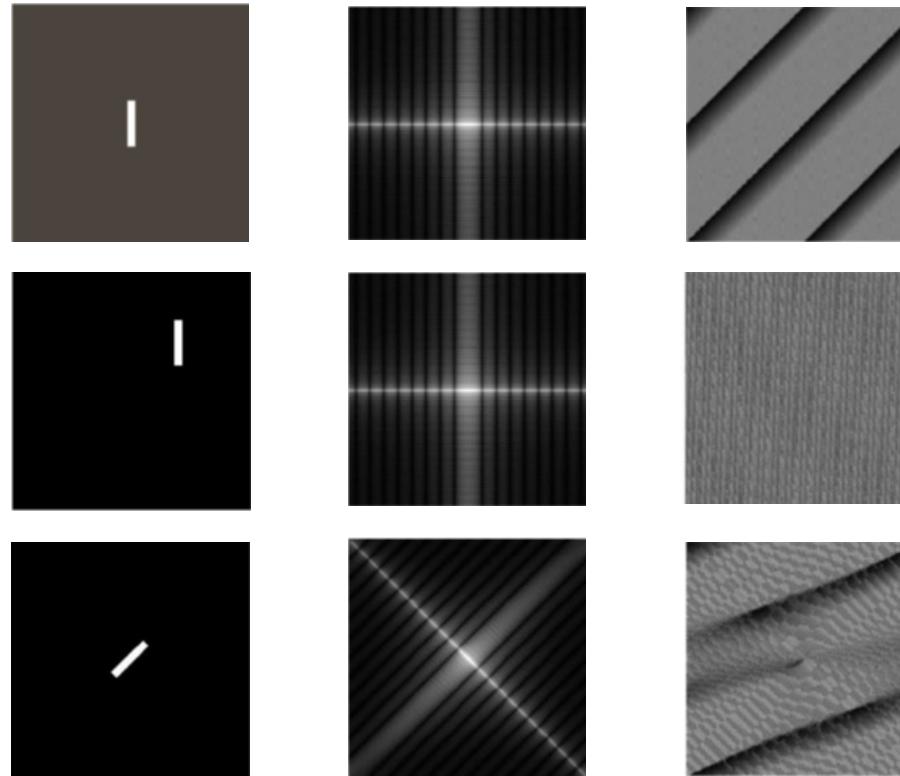
$$F[u - \frac{N}{2}, v - \frac{N}{2}]$$



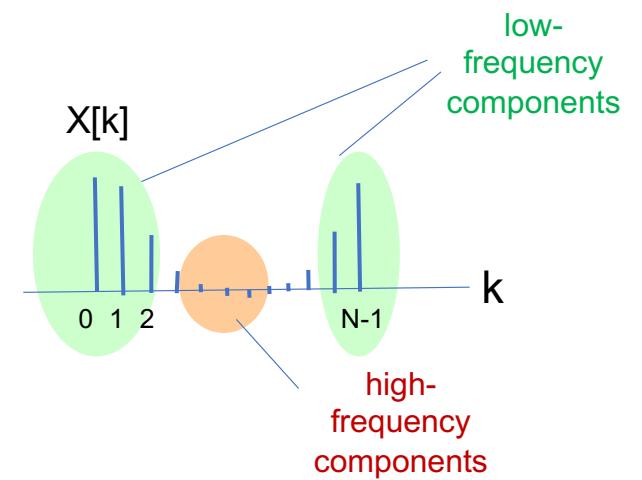
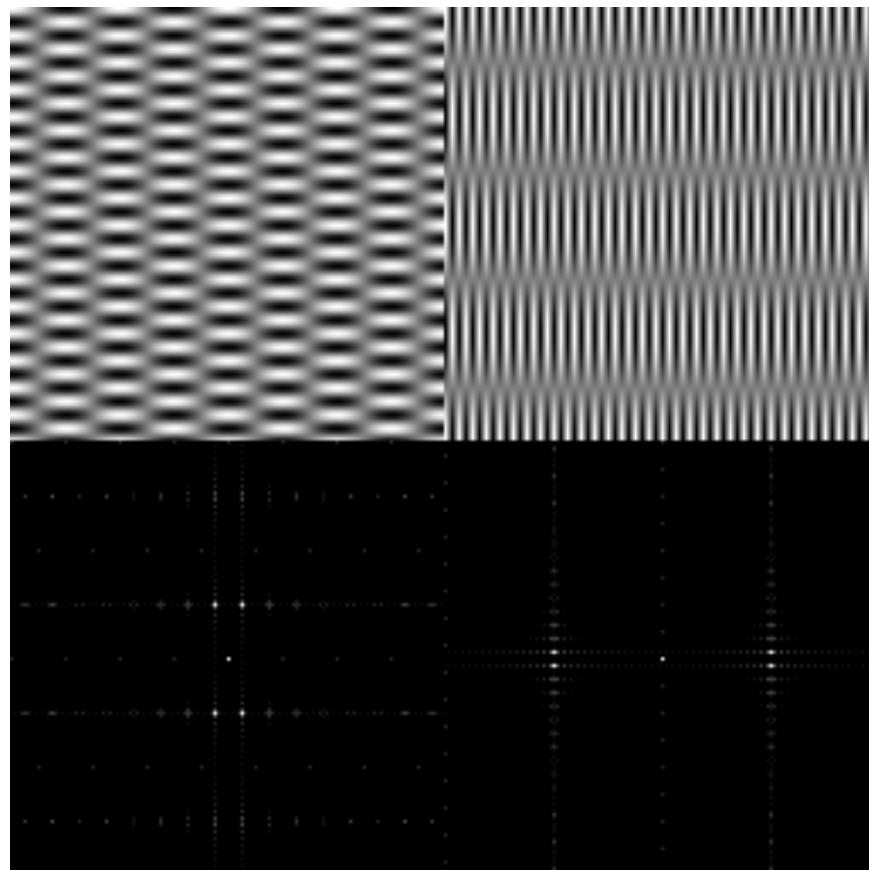
Log function

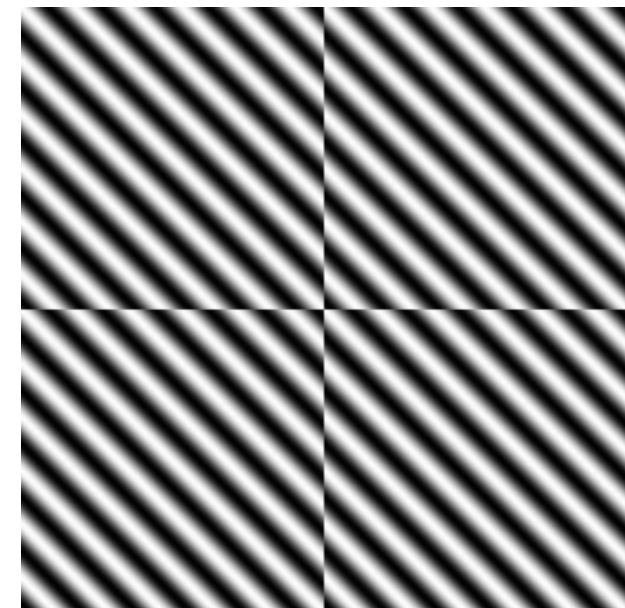
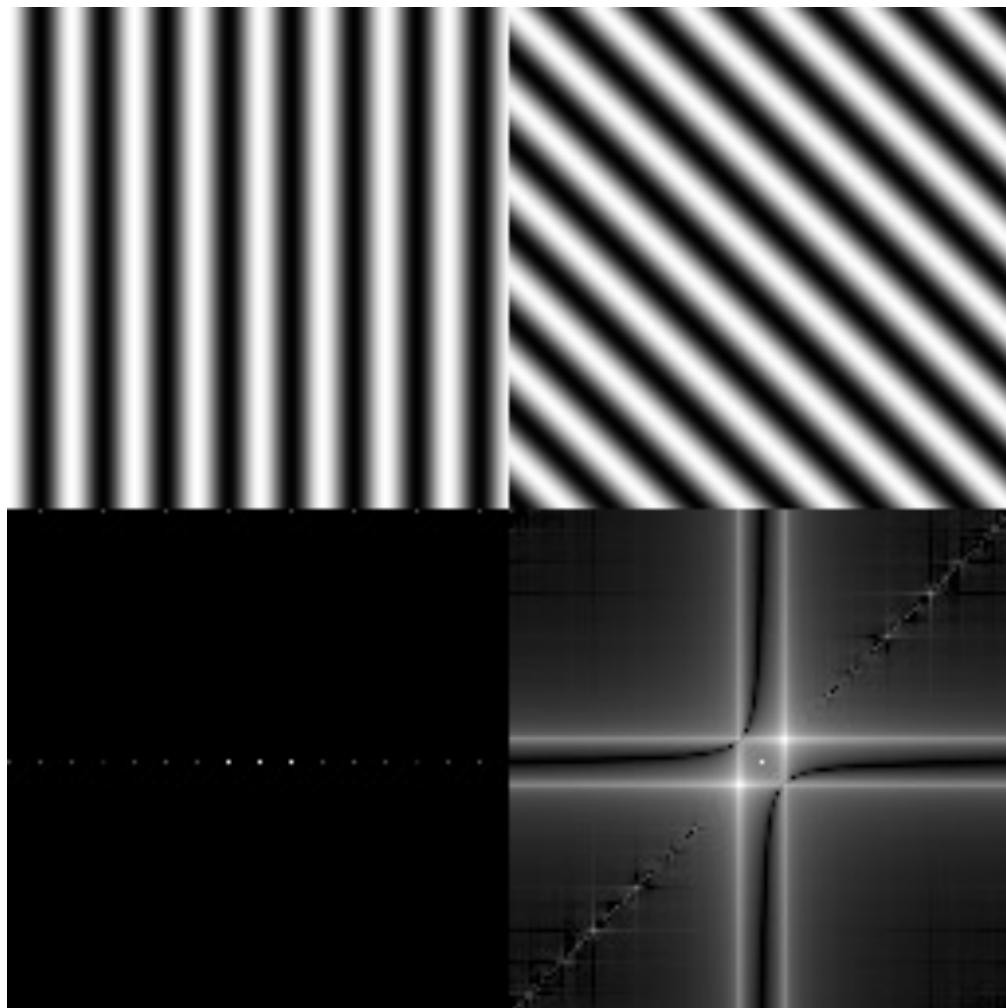
Rotation

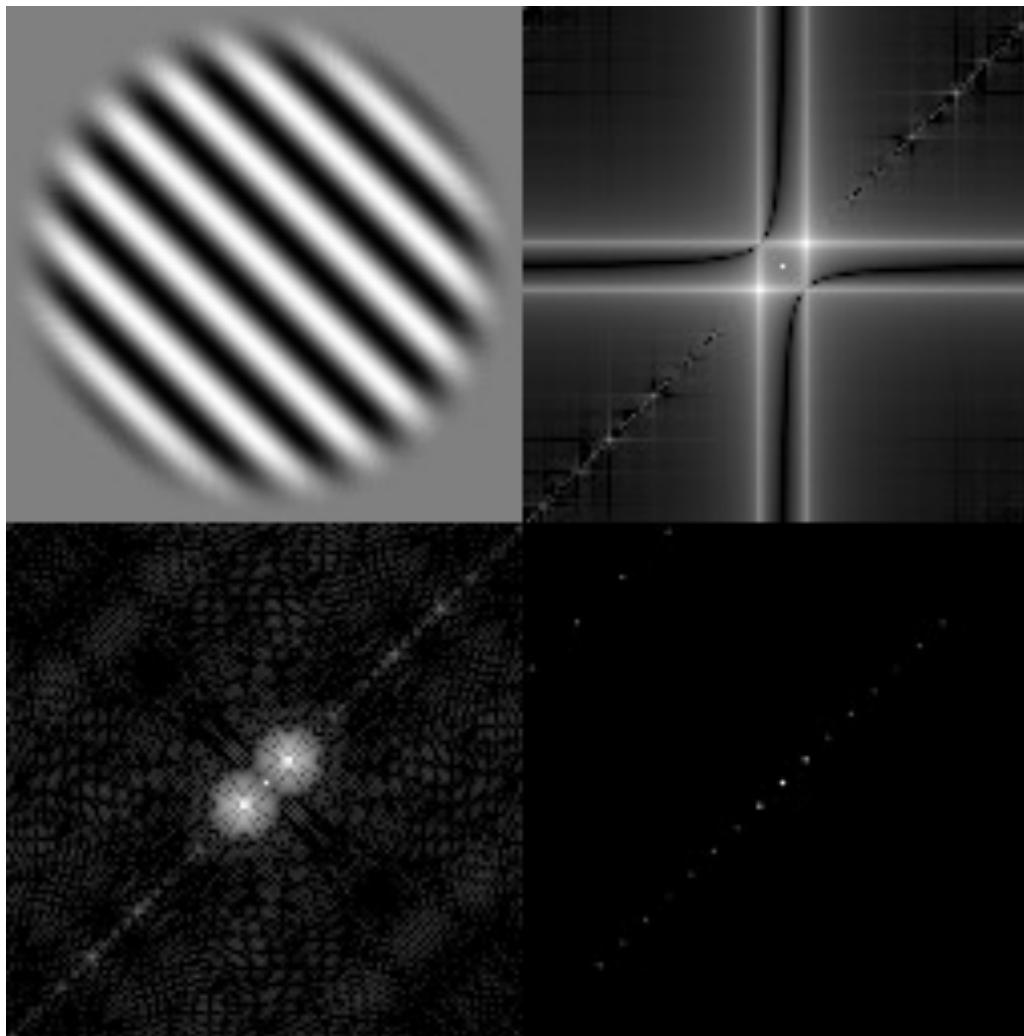
In principle, $f[r, \theta + \theta_0] \leftrightarrow F[w, \phi + \theta_0]$



Example







Frequency-domain Filtering

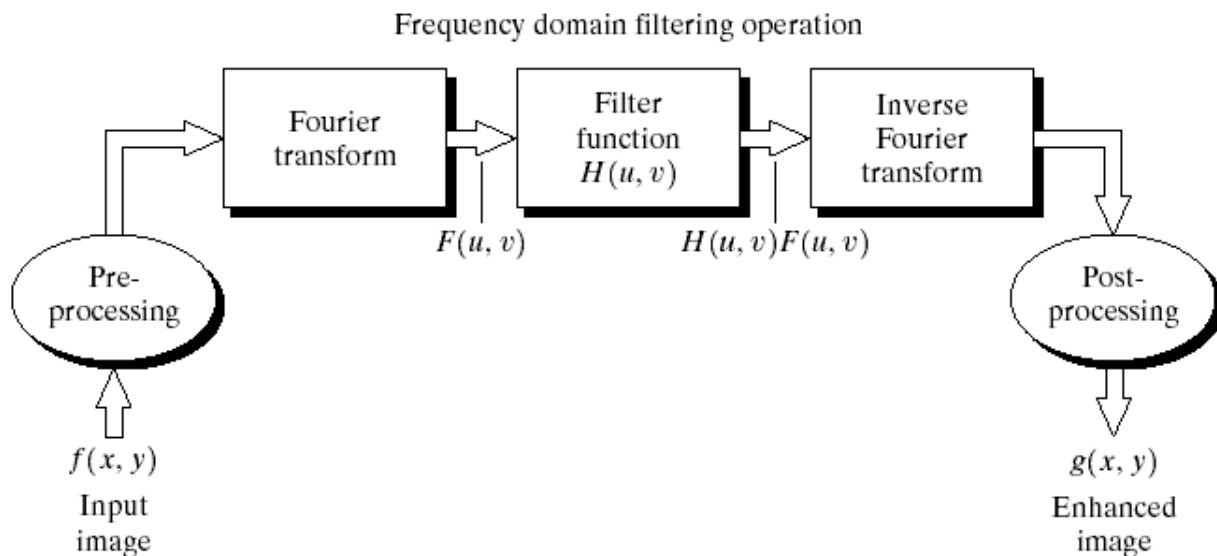
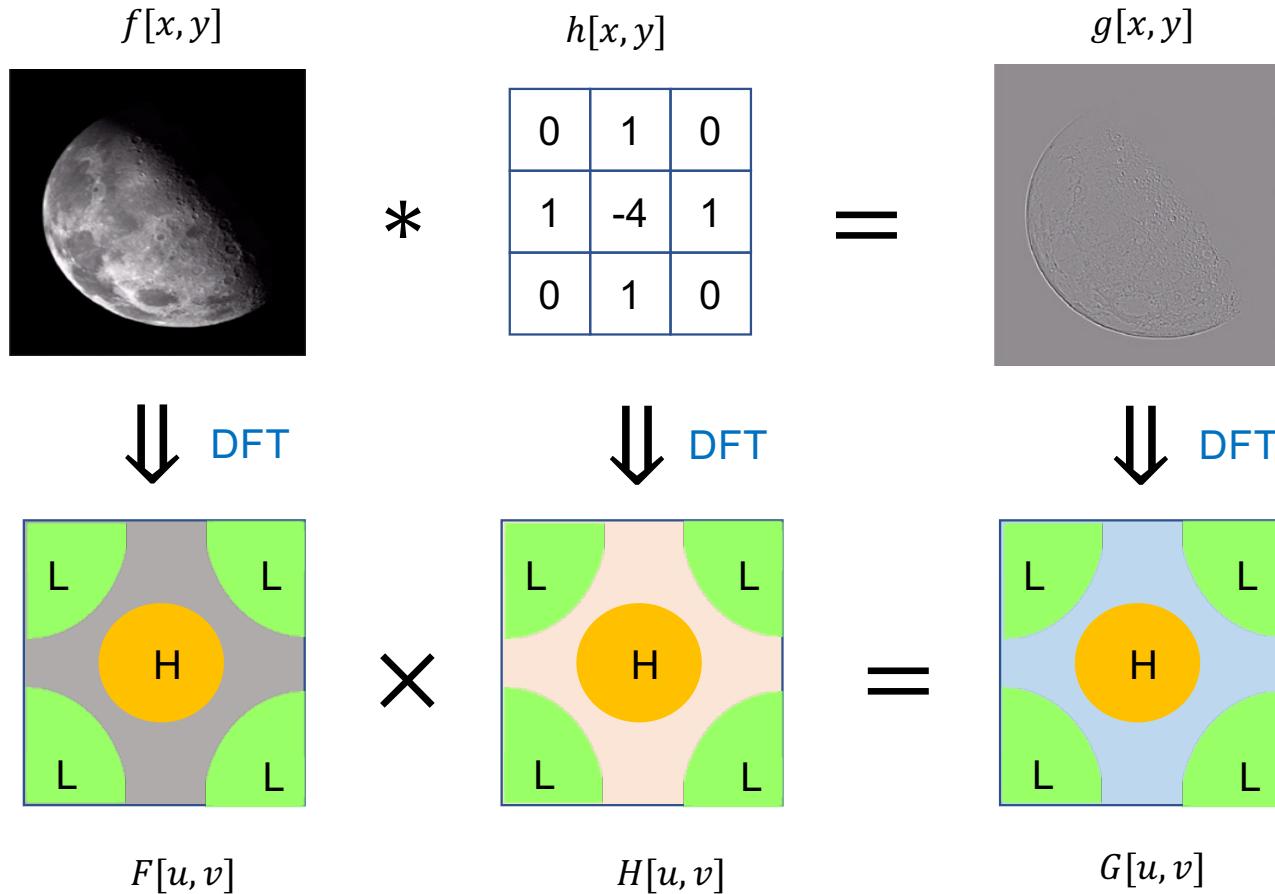
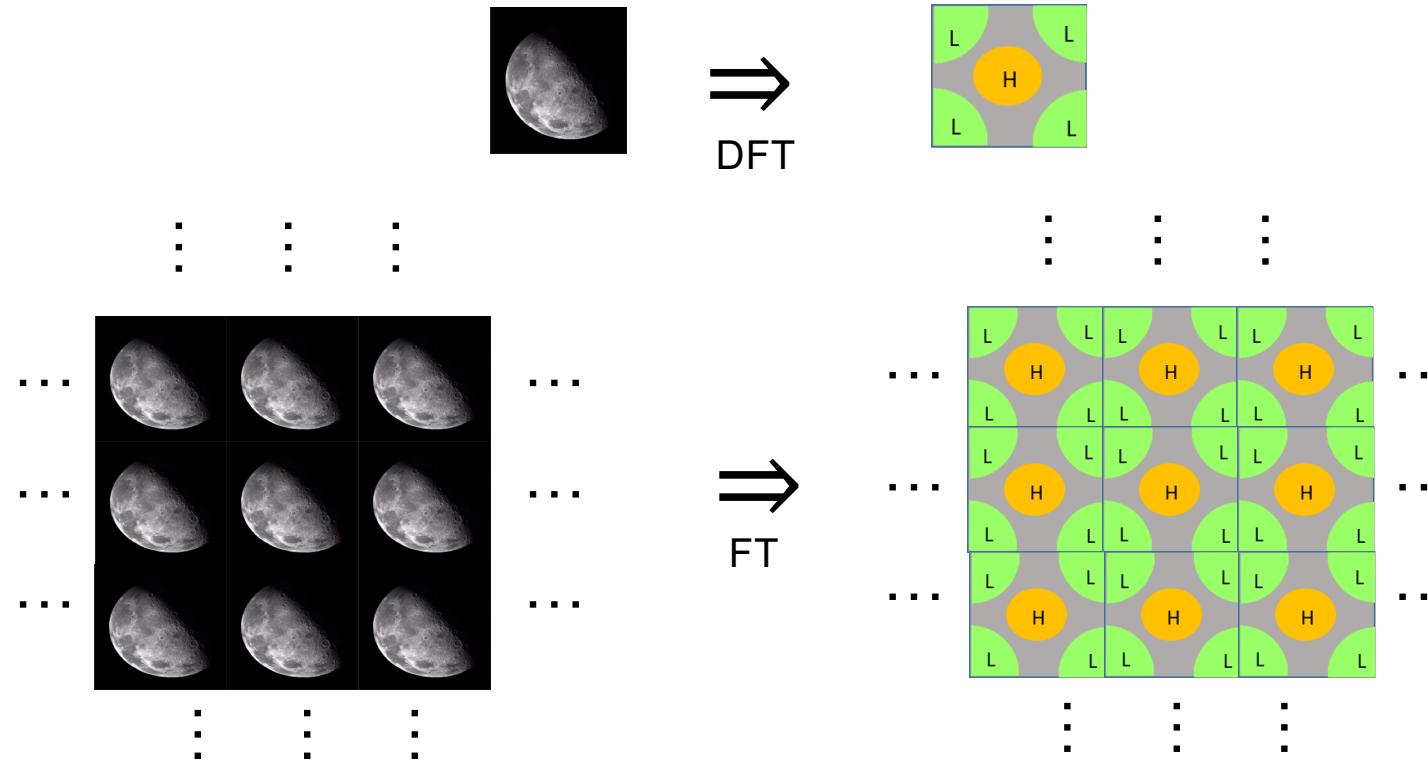


FIGURE 4.5 Basic steps for filtering in the frequency domain.

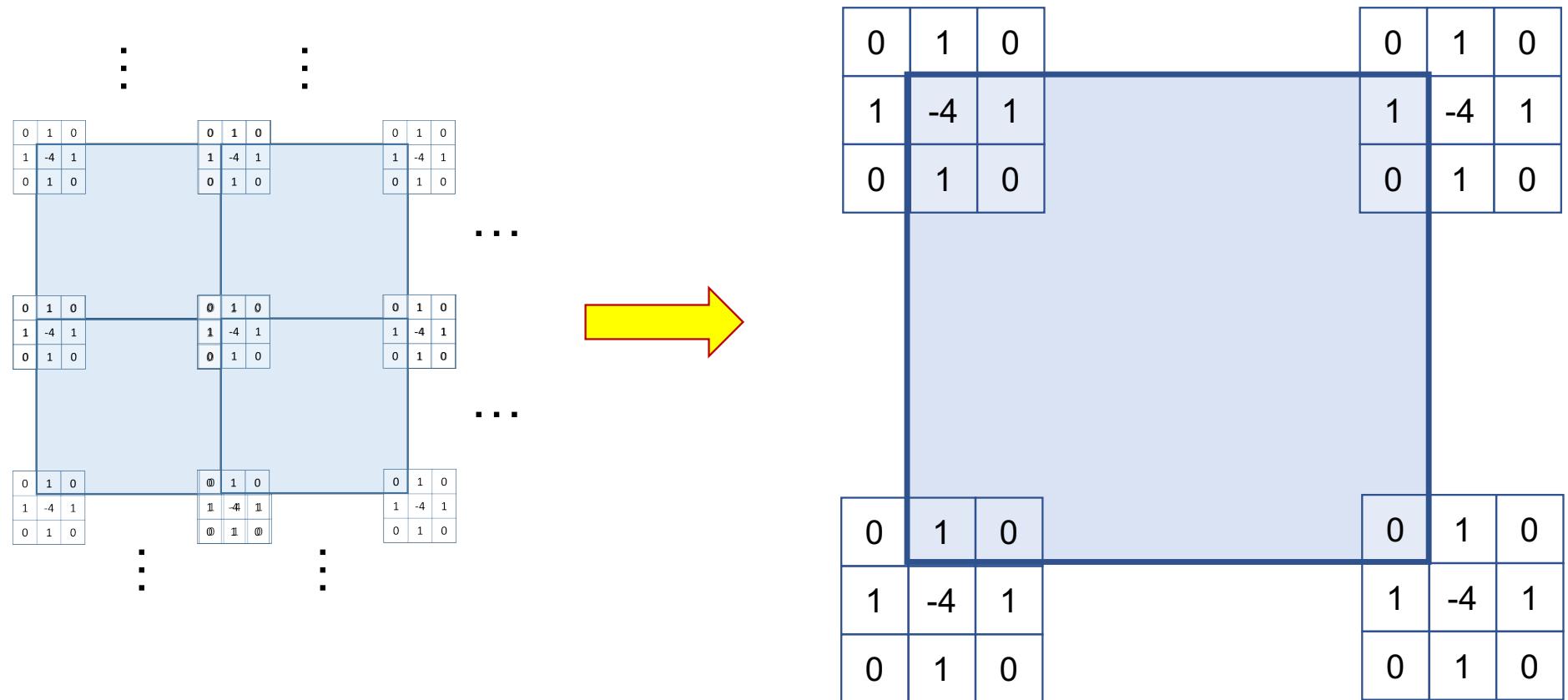
Linear Filtering in Frequency-domain (1/4)



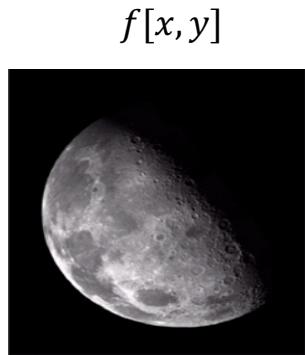
Linear Filtering in Frequency-domain (2/4)



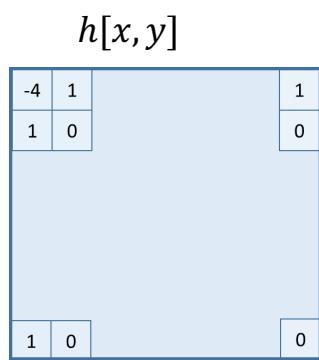
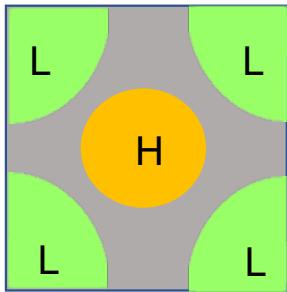
Linear Filtering in Frequency-domain (3/4)



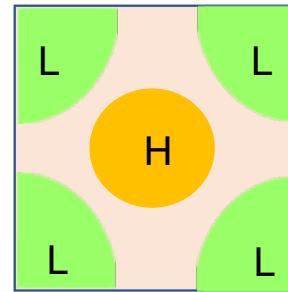
Linear Filtering in Frequency-domain (4/4)



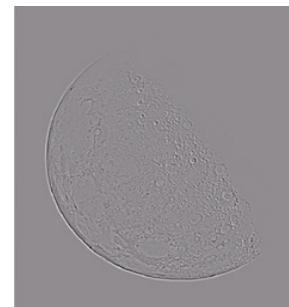
\Downarrow DFT



\Downarrow DFT

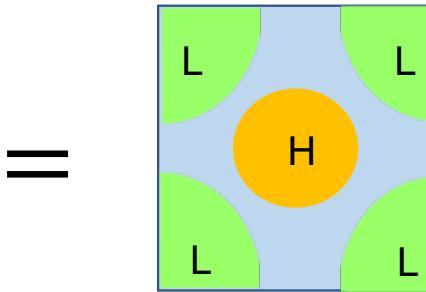


$g[x, y]$



\Downarrow DFT

$$F[u, v] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m, n] e^{-j2\pi(\frac{um}{M} + \frac{vn}{N})}$$



Next: Image Enhancement via Spatial-domain and Frequency-domain Operators

