

Stationary and non-stationary sinusoidal model synthesis with phase vocoder and FFT⁻¹

Clément Cazorla, Vincent Chrun, Bastien Fundaro, Clément Maliet
ENSEEIH - 3EN TSI 2016-2017

March 7, 2017

Abstract

The present document is to serve as both a technical report and a documentation for the code produced during the long project. As such the first part will explain the theoretical framework and the state-of-the-art of the field. The second part will give some insight about the code structure and the conventions that were adopted and last but not least, the third part will serve as an actual documentation and details every class, methods and attributes.

Contents

I	Theoretical framework	4
1	Sound synthesis	4
1.1	Frequency domain	4
1.2	Sinusoidal model	4
1.3	Phase Vocoder	4
2	Theoretical synthesis with phase vocoder and FFT^{-1}	4
2.1	Stationary case	4
2.2	Non-stationary case	4
II	Code structure and conventions	5
3	Conventions	5
3.1	Naming conventions	5
3.2	Spectrum and sinusoids parameters classes	6
3.2.1	Spectrum	6
3.2.2	Parameters	7
3.2.3	NonStationaryParameters	7
4	Code structure	8
4.1	General structure	8
4.2	Class structure	8
III	Documentation	9
5	Core module	9
5.1	Synthesizer	9
5.2	StationarySynthesizer	9
5.3	NonStationarySynthesizer	9
6	Spectrum generation module	9
6.1	SpectrumGenerator	9
6.2	StationarySpectrumGenerator	9
6.3	NonStationarySpectrumGenerator	9
6.4	StationnaryLobe	9
6.5	NonStationaryLobeGenerator	9
7	Phase Vocoder module	9
7.1	PhaseVocoder	9
7.2	StationaryPhaseVocoder	9
7.3	NonStationaryPhaseVocoder	9

A	Phase advance and propagation along the signal	10
A.1	On the first attempt at Phase Vocoder use	10
A.1.1	"Pure" synthesis	10
A.1.2	"Parametered" synthesis	12
A.2	Theoretical phases advance	12

Part I

Theoretical framework

1 Sound synthesis

1.1 Frequency domain

1.2 Sinusoidal model

1.3 Phase Vocoder

2 Theoretical synthesis with phase vocoder and FFT^{-1}

2.1 Stationary case

2.2 Non-stationary case

Part II

Code structure and conventions

3 Conventions

In this section we remind the reader of a few coding convention necessary to ensure a seamless work flow and a bug free program as much as possible. Files should contains an entire module (as described in 4.1) not just a single class to limit the number of files and ease the bug tracking. Imports in files should be kept to a minimum and left in namespaces (do not use the **from** module **import** * syntax). It is preferable to import a whole module if more than three elements from the module are needed in the file, otherwise consider the **from** module **import** element1 , element2 syntax to avoid unnecessary memory flooding. If conflicts exists, notwithstanding the number of elements needed, the whole modules are to be loaded with a namespace. Namespaces may be abbreviated to the programmer's convenience however some abbreviation are to be universally respected :

- (i) `numpy` should always be imported as `np`
- (ii) `matplotlib.pyplot` should always be imported as `plt`

Finally math functions should always come from the `numpy` module and not python's `math` module to guarantee a universal behaviour across the program.

3.1 Naming conventions

Naming conventions are freely adapted from Python recommended conventions defined in PEP8 [1], as such :

- (i) *Class* should be named in **CapitalizedWord**
- (ii) *Methods* and *functions* should be named in **lower_case_with_underscores**
- (iii) *Attributes* and *variables* should be names in **lower_case_with_underscores**
- (iv) *Instantiation* following the fact that everything is an object in python should be named as *variables*.

Moreover during class declaration, the following principles should be adopted :

- (i) Non-public methods and attributes should use one leading underscore.
- (ii) Elements that conflicts with python reserved name should use one trailing underscore rather that simplification or a misspelling.
- (iii) Accessors or mutators using one leading underscore should be interpreted as properties of their associated attribute. As such it should be guaranteed that they induce a low computational cost.

- (iv) Non-public elements that should not be inherited or may cause conflicts during inheritance should use two leading underscore and make use of Python name-mangling.

To seamlessly manipulate both *stationary* and *non stationary* models, class that are inherited in two versions are preceded with either **Stationary** or **NonStationary** respectively.

3.2 Spectrum and sinusoids parameters classes

Because many spectra, main lobes and sinusoidal model parameters have to be traded between modules we created two classes, respectively **Spectrum** and **Parameters**. They mainly serve as containers, holding the data and returning them in a point wise fashion. This way we can prevent conflicts and errors that would come from a non-uniform data sharing protocol and as well ensure that every operation performed on either spectra or parameters are made following the same principles and algorithms.

3.2.1 Spectrum

Spectrum
<pre> _amplitude : np.array _phase : np.array _nfft : int ----- __init__(self, amplitude, phase) __add__(self, other) __iadd__(self, other) __mul__(self, other) __imul__(self, other) from_complex_spectrum(cls, complex_spectrum) @classmethod void_spectrum(cls) @classmethod set_spectrum(self, amplitude, phase, start_bin=None, stop_bin=None) set_complex_spectrum(self, complex_spectrum, start_bin=None, stop_bin=None) get_amplitude(self, k) get_phase(self, k) get_nfft(self) </pre>

The Spectrum class stores a spectrum in amplitude and phase, however it may be created or changed from a complex **np.array** respectively with the class method **from_complex_spectrum** and the method **set_complex_spectrum**. Those two methods may take optional parameters **start_bin** and **stop_bin** if one need to update only a part of the spectrum, for example a single lobe. The class checks that the given data are consistent upon instantiation. The **+** operation as well as the **+** = operation have been defined between two **Spectrum** objects and between a **Spectrum** object and an array of complex numbers.

The \times operation as well as the $\times =$ operation have been defined between a **Spectrum** object and an array of complex numbers. Addition and multiplication attempts between other data type will result in a **NotImplementedError** exception.

3.2.2 Parameters

Parameters
<pre> _amplitudes : np.array _frequencies : np.array _phases : np.array _number_sinuses : int ----- __init__(self, amplitudes, frequencies, phases) get_amplitude(self, k) get_frequency(self, k) get_phase(self, k) get_number_sinuses(self) </pre>

The **Parameters** class is more of a structure than a class and only contains the stationary sinusoidal model parameters and their respective accessors. It also stores the number of sinuses and checks that the given data are consistent upon instantiation.

In the stationary sinusoidal model the signal $s(t)$ is defined as follow¹ :

$$s(n) = \sum_{i=1}^{N_{sinus}} \alpha_i \sin(2\pi \tilde{f}_i n + \phi_i)$$

with $\tilde{f}_i = \frac{f_i}{f_s}$ the normalised frequency.

We then store the parameters as follows :

_amplitudes stores the α_i

_frequencies stores the \tilde{f}_i

_phases stores the ϕ_i

3.2.3 NonStationaryParameters

NonStationaryParameters(Parameters)
<pre> _acrs : np.array _fcrs : np.array ----- __init__(self, amplitudes, frequencies, phases, acrs, fcrs) get_acr(self, k) get_fcr(self, k) </pre>

The stationary sinusoidal model can be extended to the first order development

to better model fast amplitude and frequency change over time. The signal $s(t)$ can then be expressed as a sum of linearly varying chirps¹ :

$$s(n) = \sum_{i=1}^{N_{sinus}} (\alpha_i + \mu_i \cdot nT_s) \sin(2\pi\tilde{f}_i n + \frac{\psi_i}{2}(nT_s)^2 + \phi_i)$$

Where we define the **Amplitude Change Rate** μ and the **Frequency Change Rate** ψ .

Thus we inherit the **Parameters** class to add the two additional parameters as follow :

`_acrs` stores the μ_i

`_fcrs` stores the ψ_i

4 Code structure

4.1 General structure

4.2 Class structure

¹Please look up section 1.2 page 4 for more details

Part III

Documentation

5 Core module

5.1 Synthesizer

5.2 StationarySynthesizer

5.3 NonStationarySynthesizer

6 Spectrum generation module

6.1 SpectrumGenerator

6.2 StationarySpectrumGenerator

6.3 NonStationarySpectrumGenerator

6.4 StationnaryLobe

6.5 NonStationaryLobeGenerator

7 Phase Vocoder module

7.1 PhaseVocoder

7.2 StationaryPhaseVocoder

7.3 NonStationaryPhaseVocoder

Conclusion

A Phase advance and propagation along the signal

A.1 On the first attempt at Phase Vocoder use

In the stationary case, the Phase Vocoder is expressed in 3 equations as follow :

$$\Delta\Phi_k^u = \angle X(t_a^u, \Omega_k) - \angle X(t_a^{u-1}, \Omega_k) - R_a\Omega_k \quad (1)$$

$$\hat{\omega}_k(t_a^u) = \Omega_k + \frac{\Delta_p\Phi_k^u}{R_a} \quad (2)$$

$$\angle Y(t_s^u, \Omega_k) = \angle Y(t_s^{u-1}, \Omega_k) + R_s\hat{\omega}_k(t_a^u) \quad (3)$$

What (1) means is that we look for the true phase shift (the analysis phase shift) during the frame u and $u-1$ which is $\angle X(t_a^u, \Omega_k) - \angle X(t_a^{u-1}, \Omega_k)$ and we compute the error in phase shift, that is to say the difference between the *true* phase shift and the *expected* phase shift.

In (2) we use the error in phase shift to compute the deviation in frequency $\frac{\Delta_p\Phi_k^u}{R_a}$ from the expected frequency Ω_k and thus compute the *true* frequency at which the bin was excited between t_a^{u-1} and t_a^u .

Finally (3) assume the correct synthesis phase shift will be $R_s\hat{\omega}_k(t_a^u)$ that is the true frequency times the synthesis hop.

The issue with such an approach in our case is that we dropped the analysis. We want to use the Phase Vocoder to *create* phase shifts in our spectra, but the Phase Vocoder is in fact nothing but a fancy way to copy existing phase shifts while taking into account a different hop during analysis and synthesis. I will try and break down the issues I have into two cases.

A.1.1 "Pure" synthesis

We wish to synthesize a stationary or a sum of stationary sinusoid from scratch. For simplicity's sake and without loss of generality we will take the one sinusoid case. That is to say that we want, without prior knowledge to generate $s(n)$ such as :

$$s(n) = \alpha \cos(2\pi\tilde{f}n + \phi) \quad (4)$$

knowing only α , \tilde{f} and ϕ .

We will also, to ease the process, assume that we know the application¹ $f_{w,\tilde{f}}(\phi) : \phi \mapsto \tilde{\phi}$ which takes into account the effect of windowing on the phase

¹And this this is a very strong hypothesis in the sense that it will never be true, but this is not the core issue at stake here.

of the frame spectrum's phase².

The first step is then to generate a synthetic spectrum with the desired parameters. To do this we only generate a main lobe derived from the Fourier transform of the normalized window w supposedly³ used during analysis, and place it at the right position on the spectrum. This involves to interpolate the relevant bins value if by any chance the wanted frequency \tilde{f} is not exactly on a bin, that is to say if $\tilde{f} \notin \{\frac{2k\pi}{N}\}_{k=0\dots N-1}$. We then multiply the generated lobe by $\frac{A}{2}$ and set the lobe phase to $\tilde{\phi} + 2\pi\tilde{f}R_a$ ⁴. We then wished to use the phase vocoder to advance the phase (compute the needed phase shift). To get the temporal frame, we theoretically only have to compute the inverse Fourier transform of the generated spectrum.

In order to use the Phase Vocoder we assumed the generated spectrum to be equivalent to the analysis spectrum $X(t_a^u)$ and the antecedently phase corrected spectrum to be equivalent to the past synthesis spectrum $Y(t_a^{u-1})$.

At the first iteration :

- $X(t_a^{u-1})$ is void because by hypothesis, nothing happened before.
- $X(t_a^u)$ is the freshly generated spectrum
- $Y(t_a^{u-1})$ is also void for the same reasons

Equation 1 gives, for $k \in 1\dots N-1$ s.t $X(t_a^u, \Omega_k)$ is a bin of the lobe a phase shift error of $\tilde{\phi}$.

Then after 2 and 3 we obtain :

$$\begin{aligned}\angle Y(t_s^u, \Omega_k) &= \angle Y(t_s^{u-1}, \Omega_k) + R_s \hat{\omega}_k(t_a^u) \\ &= \angle Y(t_s^{u-1}, \Omega_k) + \frac{R_s}{R_a} \tilde{\phi} + R_s \Omega_k\end{aligned}$$

If $R_s = R_a$ we have a perfect reconstruction of the time synthesized overlap-add test signal. However, in that case, the Phase Vocoder is perfectly irrelevant to the synthesis, indeed, since we have no *actual* analysis phase, we only need to modify R_a to change the length of the final signal.

At the following iteration, we have to update the phase of the generated spectrum to $\tilde{\phi} + 2 \times 2\pi\tilde{f}R_a$ instead of $\tilde{\phi} + 2\pi\tilde{f}R_a$ (as computed in ??), recursively, we can define the phase of the lobe the i th generated spectrum as :

$$\begin{cases} \tilde{\phi}_i = \tilde{\phi}_{i-1} + 2\pi\tilde{f}R_a \\ \tilde{\phi}_0 = \tilde{\phi} \end{cases} \quad (5)$$

²For more details on the theory, please read part I page 4

³Because no actual analysis happened

⁴This is because we wish to generate frame spaced by R_a so we have to compensate the *expected* phase shift by hand. In fact, in the purely stationary case, the expected phase shift is the theoretical phase shift.

A.1.2 "Parametered" synthesis

In this case, we will not synthesize a truly stationary signal but we assume that the signal is quasi-stationary, which is to say that given a small enough analysis window, it can be considered stationary within that frame.

$$s(n) \cdot \mathbb{1}_{[t_a^{u-1}, t_a^u]} \simeq \alpha \cos(2\pi \tilde{f}n + \phi) \quad (6)$$

Note that the initial phase ϕ is constant from one frame to the other *by definition*, indeed, a change of phase is equivalent to a sweep in frequency. Also this is more of a constrain of stability on frequency than it is on amplitude because of the nature of the overlap-add process.

We can assume that under the right conditions the method developed in subsection A.1.1 above still holds, if we update the parameters \tilde{f} and α in the spectrum generation.

A.2 Theoretical phases advance

For stationary signals expressed as in (4) the phase advance from one frame to the other is elementary to compute :

$$\begin{aligned} \Delta\Phi_f^{t_a} &= \Phi_f^{t_a+H_a} - \Phi_f^{t_a} \\ &= 2\pi f(t_a + H_a) - 2\pi f t_a \\ &= 2\pi f H_a = 2\pi \tilde{f} R_a \end{aligned} \quad (7)$$

For non stationary signal expressed as :

$$s(n) = (\alpha + \mu \cdot t_a) \sin(2\pi f t_a + \frac{\psi}{2} t_a^2 + \phi) \quad (8)$$

References

- [1] N. C. Guido van Rossum, Barry Warsaw, “Style guide for python code,” Aug. 2013.