Long Project with Audiogaming

Additive Synthesis with Inverse Fourier Transform for Non-Stationary Signals

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March 03, 2017



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The company

Audio Gaming NATURAL BORN INTERACTIVE

- Location: Toulouse, Paris
- Activity: Audio plug-in (VSTs and RTAS)
- Main customers: Film and Video Game Industry (Sony, Ubisoft)
- 10 employees

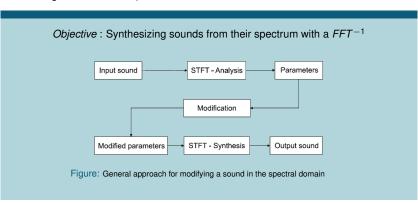


Figure: Audiofire: audio plug-in that recreates fire sound



Objective

 We are continuing the Audiogaming long project from 2015 (Emilie Abia, Lili Zheng, Quentin Biache)



■ We have to implement a new method of additive synthesis ⇒ computationally very fast



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Introduction

Context of the Project

■ 6 weeks only ⇒ Focus on the synthesis method only.

Given codes in Python and Matlab from the 2015 project :

- Python: Analysis estimator of sinus parameters and sinus generation with those parameters (only stationary)
- Matlab : Some reasearch on the Non-stationary synthesis with the LUT of lobes
- We made our own Object Oriented Programmation tree structure in Python
- We remade all the codes to be coherent with the OOP tree structure



Introduction Work Environment









Figure: PvCharm as Python IDE. Slack to communicate. GitHub to stock the codes and have a versionning. Freedcamp to plan the project events



Project Management: Gantt Chart (expected event)





Project Management: Gantt Chart now





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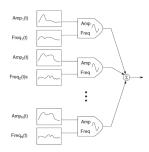
Additive Synthesis

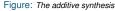
Time Domain

The sound signal is represented as a sum of N sinusoids:

$$x(t) = \sum_{n=1}^{N} a_n sin(2\pi f_n t + \phi_n)$$

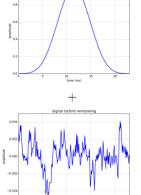
- Very costly to implement
- Impossible to compute in real-time







Method Overview : Windowing Analysis



Windowing step:

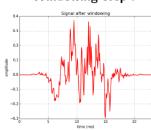


Figure: Windowing step



Method Overview : Peak detection in Frequency Domain Analysis

Peak detection and extraction of parameters by STPT (particular Short Time Fourier Transform):

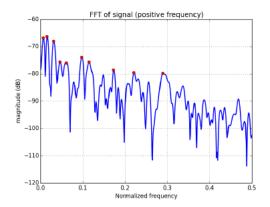


Figure: Peak detection



Method Overview : Result (FFT⁻¹)

Synthesis

Introduction

Additive synthesis with FFT⁻¹ according to the parameters from the analysis:

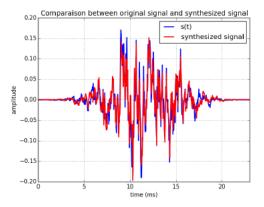


Figure: Synthesized frame vs Original frame



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Stationary sinusoidal model

Mathematical model:

$$s(t) = a_0 \exp[j(2\pi f_0 t + \phi_0)] \tag{1}$$

- 3 parameters: a_0 (amplitude), f_0 (frequency) and ϕ_0 (phase).
- Simplest model but useful for certain kinds of signals.
- Each spectral bin represents a stationary sinusoid.
- ⇒ generate a synthetic spectrum with the desired parameters
- \Rightarrow generate a main lobe derived from the Fourier transform of the normalized window w supposedly 1 used during analysis
- \Rightarrow place it at the right position on the spectrum.



¹Because no actual analysis happened

Lobe generation

We generate the sinusoids in frequency domain:

- Window the signal to maximize the energy in the main lobe
- We only keep the main lobe for each sine (11 points)
- We assume that the parameters (amplitude, frequency, phase) are already given by the analysis
- We interpolate the relevant bins value if by any chance the wanted frequency \hat{f} is not exactly on a bin, that is to say if $\hat{f} \notin \{\frac{2k\pi}{N}\}_{k=0...N-1}$

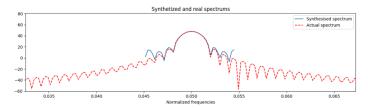


Figure: Windowed sine lobe



Frames separation

The sound signal is a frame-by-frame signal:

The analysis hop size will be called R_a and the synthesis hop size R_s (moving step of the frame)

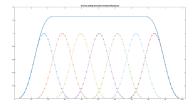


Figure: Sum of small size Hanning windows

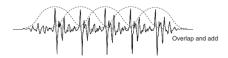


Figure: Overlap and add



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Phase Coherence

Phase coherence

The Phase coherence is not a problem in the Stationary case:

- We don't know the window effect on the phase : $f_{w,\hat{f}}(\phi): \phi \mapsto \tilde{\phi}$ ⇒ We calculate its influence on the first frame and assume the same influence on the other frame.
- We then multiply the generated lobe by $\frac{A}{2}$ and set the lobe phase to $\tilde{\phi} + 2\pi \hat{f} R_a$
- In the purely stationary case, the expected phase shift is the theoretical phase shift:

$$\begin{cases} \tilde{\phi}_i = \tilde{\phi}_{i-1} + 2\pi \hat{f} R_a \\ \tilde{\phi}_0 = \tilde{\phi} \end{cases} \tag{2}$$



Quasi-Stationary Case

What is changing

Introduction

- In a Quasi-stationary case, the sine wave can change a little bit in frequency. Main problem ⇒ Phase coherence!
- We need to implement a a method to correct the phase : Phase Vocoder!

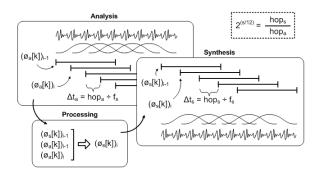


Figure: Phase Vocoder overview



Quasi-Stationary Case

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Phase Vocoder

In this case, the phase changing is different from the stationary case. We have to calculate the instantaneous frequency for the kth bin:

$$\hat{\omega}_k(t_a^u) = \Omega_k + \frac{\Delta_p \Phi_k^u}{R_a} \tag{3}$$

Where:

$$\Delta \Phi_k^u = \angle X(t_a^u, \Omega_k) - \angle X(t_a^{u-1}, \Omega_k) - R_a \Omega_k \tag{4}$$

Hence,

$$\angle Y(t_s^u, \Omega_k) = \angle Y(t_s^{u-1}, \Omega_k) + R_s \hat{\omega}_k(t_a^u)$$
(5)

We replace the output signal phase by this one.

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Non-Stationary Case

A different approach

Mathematical model:

$$s(t) = \exp[(\lambda_0 + \mu_0 t) + j(\phi_0 + 2\pi f_0 t + \frac{\psi_0}{2} t^2)]$$
 (6)

5 parameters:

 $(\lambda_0 + \mu_0 t)$ (overall amplitude)

 \hat{f}_0 (frequency)

 ϕ_0 (phase)

 μ_0 (amplitude change rate (ACR))

 ψ_0 (frequency change rate (FCR))

- The analysis part give us all those parameters
- To manage the influence of the ACR and the FCR on the lobe ⇒ Interpolation of Look-up table of already saved lobes with different (ACR,FCR).



Non-Stationary Case

Look up table

Introduction

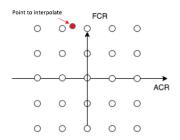


Figure: ACR/FCR grid



Figure: Look-up table

Non-Stationary Case

Phase Vocoder: Scaled-Phase Locking

- We can use the phase vocoder to correct the phase, but the main problem is sine waves that switch from a frequency bin to another bin. The phase might change a lot from one frame to another.
- ⇒ We use a refined version of the phase vocoder : Scaled-Phase Locking ⇒ It takes into account the frequency trajectory of each lobes.

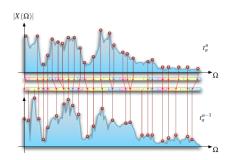


Figure: Peak coherence from a frame to another



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Non-Stationary Case

Phase Vocoder: Scaled-Phase Locking

Scaled-Phase Locking

We find each region corresponding to each peaks, and then we use the phase vocoder algorithm for each region :

 k_0 is the precedent frequency bin for the peak - k_1 is the current frame peak bin :

$$\hat{\omega}_{k_1}(t_a^u) = \Omega_k + \frac{\Delta_p \Phi_k^u}{R_a} \tag{7}$$

Where:

$$\Delta \Phi_{k_1}^u = \angle X(t_a^u, \Omega_{k_1}) - \angle X(t_a^{u-1}, \Omega_{k_0}) - R_a \Omega_{k_1}$$
 (8)

Hence.

$$\angle Y(t_s^u, \Omega_{k_1}) = \angle Y(t_s^{u-1}, \Omega_{k_0}) + R_s \hat{\omega}_{k_1}(t_a^u)$$
 (9)

Then for each bin in the region:

$$\angle Y(t_s^u, \Omega_k) = \angle Y(t_s^u, \Omega_{k_1}) + \beta [\angle X(t_a^u, \Omega_k) - \angle X(t_a^u, \Omega_{k_1})]$$
(10)

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Non-Stationary Case

Scaled-Phase Locking Problem

Moreover, we do not know for now how to manage the peaks that appear and disappear

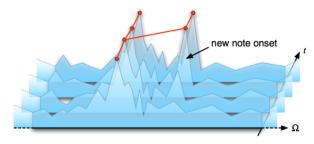


Figure: Phase-locking problem



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Protocol

- A one second triangular signal consisting in 84 frames
- We first vary the number of harmonics and compare the time of execution
- In a second time, for a 10 harmonics signal, we vary the frequency and investigate the reconstruction error.



Triangular wave synthesis

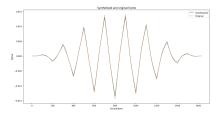


Figure: original vs. synthesized triangular wave



Time and Relative-Time

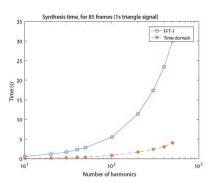


Figure: Time of execution

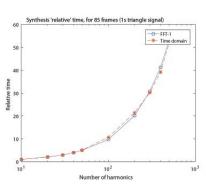


Figure: Relative Time of Execution



Stationary Case

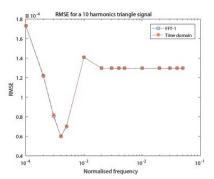


Figure: RMSE (original vs. synthesized triangular wave



Non-Stationary Case

Chirps

The idea is to try the method on some chirps signal. And then on real sounds, like instruments and voices.

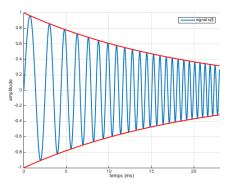


Figure: Chirp signal to test

We can measure the error of the lobe interpolation with the Look-Up Table. (Not done Neellell &

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Conclusion

- - 6 weeks only
 - Research subject ⇒ Can it really works?
 - Lots of trouble when we try to understand the phase coherence problem

Do you have any question?



References

Introduction

References

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