

AERO3600 — Embedded Control Systems

Practice problems: Continuous-time state-feedback control design¹

Important



This document proposes several practice problem on state-feedback control desing for continuous time systems.

Problem 1

Consider the system

$$\begin{aligned}\dot{x}_1 &= -3x_1 + x_2 + u \\ \dot{x}_2 &= -2x_1 + 1.5x_2 + 4u.\end{aligned}$$

- a) Is the system completely controllable? Justify your answer.

Check your solutions with MATLAB.

Problem 2

Consider four different LTI system written in the form

$$\dot{x} = A_i x + B_i u$$

where $i = \{1, 2, 3, 4\}$ indicate the index of the system i . The corresponding controllability matrices of the systems are as follows

$$Co_1 = \begin{bmatrix} 2 & 0 \\ 0 & -3 \end{bmatrix}, \quad Co_2 = \begin{bmatrix} -2 & 2 \\ 1 & -1 \end{bmatrix}, \quad Co_3 = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, \quad Co_4 = \begin{bmatrix} 1 & 0 & -2 & 0 \\ 0 & 1 & 0 & -2 \end{bmatrix}.$$

- a) Determine the dimension of the matrices A_i and B_i that corresponde to system i .
b) Analyse the controllability of each system. Justify your answer.

Check your solutions with MATLAB.

Problem 3

Consider the system

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 20.6 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

- a) Is the system internally stable?
b) Design a zero-state regulator controller with desired closed-loop eigenvalues at $\lambda_{1,2} = -1.8 \pm j2.4$.
c) Draw the block-diagram of the control system.

Check your solutions with MATLAB.

¹Updated: 5 May 2021.

Problem 4

Consider the system

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u.$$

- a) Is the system controllable? Justify your answer
- b) Find the controller that results in closed-loop eigenvalues $\lambda_1 = \lambda_2 = -1$.

Check your solutions with MATLAB.

Problem 5

Consider the system

$$\begin{aligned} \dot{x} &= \begin{bmatrix} 1 & 1 \\ -4 & 3 \end{bmatrix} x + \begin{bmatrix} 0 \\ 2 \end{bmatrix} u, \\ y &= \begin{bmatrix} 1 & 0 \end{bmatrix} x. \end{aligned}$$

- a) Find the state transformation T that takes the system into the companion form.
- b) Find the controller that results in closed-loop eigenvalues $\lambda_1 = -1$ and $\lambda_2 = -2$.
- c) Find the set-point regulator that takes the output y to a desired constant output \bar{y} , that is

$$\lim_{t \rightarrow \infty} y = \bar{y}$$

Check your solutions with MATLAB.

Problem 6

For each of the following systems, use linearisation to design a state-feedback controller to stabilise the origin. Select the desired closed-loop eigenvalues to achieve a particular dynamic response, for example certain overshoot and time response.

1.

$$\begin{aligned} \dot{x}_1 &= x_1 + x_2 \\ \dot{x}_2 &= 3x_1^2 x_2 + x_1 + u \\ y &= -x_1^3 + x_2 \end{aligned}$$

2.

$$\begin{aligned} \dot{x}_1 &= x_1 + x_2 \\ \dot{x}_2 &= x_1 x_2^2 - x_1 + x_3 \\ \dot{x}_3 &= u \\ y &= -x_1^3 + x_2 \end{aligned}$$

3.

$$\begin{aligned} \dot{x}_1 &= -x_1 + x_2 \\ \dot{x}_2 &= x_1 - x_2 - x_1 x_3 + u \\ \dot{x}_3 &= x_1 + x_1 x_2 - 2x_3 \\ y &= x_1 \end{aligned}$$

Check your solutions with MATLAB.

Problem 7

A simplified model of a low-frequency motion of a ship can be written as follows

$$\begin{aligned}\dot{x}_1 &= x_2 \\ \dot{x}_2 &= -10x_2 + 0.1u \\ y &= x_1\end{aligned}$$

where x_1 is the heading angle (in radians), x_2 is the heading angular rate and u is the rudder angle.

1. Find the equilibrium point and input that correspond to $\bar{y} = \frac{\pi}{4}$.
2. Use linearisation to design a state feedback controller to ensure that the output y tracks the reference values \bar{y} and such that the closed loop eigenvalues are -1 and -1.
3. Design an integral action that ensure that the output tracks the reference \bar{y} . The desired eigenvalues of the closed loop are -1 -1 -15.

Check your solutions with MATLAB.

Problem 8

Consider the system

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 1 & -3 \end{bmatrix} x + \begin{bmatrix} 0 \\ 2 \end{bmatrix} u.$$

- a) Determine if the system is controllable. Justify your answer.
- b) Find the gain of the controller $u = -[k_1 \ k_2]x$, such that the closed loop eigenvalues are $\lambda_1 = -1$ and $\lambda_2 = -2$.
- c) Determine if it possible to arbitrarily assign the closed loop eigenvalues, that is, can we assign the values λ_1 and λ_2 arbitrarily?
- d) Consider now the controller $u = -[k_3 \ 3k_3]x$. Determine the gain k_3 such that one of the closed loop eigenvalues $\lambda_1 = -1$. Determine if it is possible to arbitrarily assign the second closed loop eigenvalue λ_2 .
- e) Consider the controller in d). Determine if the closed loop is stable. Justify your answer.

Check your solutions with MATLAB.

Problem 9

Consider the dynamics of the magnetic levitation system²

$$\begin{aligned}35\ddot{\phi} + 4\ddot{\theta} - 100\sin(\phi) &= -\tau \\ 4\ddot{\phi} + \ddot{\theta} - 4\sin(\phi)\dot{\phi}^2 &= \tau \\ y_1 &= \phi \\ y_2 &= \theta\end{aligned}$$

where ϕ is the absolute angle of the body and θ is angle of the ball.

1. Write the system in state space form with $x_1 = \phi$, $x_2 = \theta$, $x_3 = \dot{\phi}$ and $x_4 = \dot{\theta}$.
2. Find the equilibrium point and input that correspond to $\bar{x}_1 = 0$ and $\bar{x}_2 = 0$.
3. Use linearisation to design a state feedback controller to ensure that the output y tracks the reference values \bar{y} and such that the closed loop eigenvalues are -10, -10, -15 and -15.
4. Design an integral action that ensures that the outputs track the reference values $\bar{y}_1 = 0$ and $\bar{y}_2 = 4$. If possible, the desired eigenvalues of the closed loop should be -10, -10, -15, -15, -5 and -5.

Solve numerically using MATLAB and simulate the closed loop.

²See for example the article *Balancing and Transferring Control of a Ball Segway Using a Double-Loop Approach* in IEEE Control Systems Magazine, vol. 38, no. 2, pp. 15-37, 2018, D. B. Pham, H. Kim, J. Kim and S. Lee.

Problem 10

Consider the dynamics of the robot known as Ballbot system³

$$\begin{aligned}35\ddot{\phi} + 4\ddot{\theta} - 100\sin(\phi) &= -\tau \\4\ddot{\phi} + \ddot{\theta} - 4\sin(\phi)\dot{\phi}^2 &= \tau \\y_1 &= \phi \\y_2 &= \theta\end{aligned}$$

where ϕ is the absolute angle of the body and θ is angle of the ball.

1. Write the system in state space form with $x_1 = \phi$, $x_2 = \theta$, $x_3 = \dot{\phi}$ and $x_4 = \dot{\theta}$.
2. Find the equilibrium point and input that correspond to $\bar{x}_1 = 0$ and $\bar{x}_2 = 0$.
3. Use linearisation to design a state feedback controller to ensure that the output y tracks the reference values \bar{y} and such that the closed loop eigenvalues are -10, -10, -15 and -15.
4. Design an integral action that ensures that the outputs track the reference values $\bar{y}_1 = 0$ and $\bar{y}_2 = 4$. If possible, the desired eigenvalues of the closed loop should be -10, -10, -15, -15, -5 and -5.

Solve numerically using MATLAB and simulate the closed loop.

Additional exercises

You can find additional exercises in the chapter 11—the Design of State variable feedback systems—of the book *Modern Control Systems* by R. C. Dorf and R. H. Bishop. The book is available at UON Library.

³See for example the article *Balancing and Transferring Control of a Ball Segway Using a Double-Loop Approach* in IEEE Control Systems Magazine, vol. 38, no. 2, pp. 15-37, 2018, D. B. Pham, H. Kim, J. Kim and S. Lee.

Answer to a some practice problems

Page 1

(state-space feedback control design)

Problem 1

$$A = \begin{bmatrix} -3 & 1 \\ -2 & 1.5 \end{bmatrix} ; B = \begin{bmatrix} 1 \\ 4 \end{bmatrix} ; C_{AB} = \begin{bmatrix} 1 & 1 \\ 4 & 4 \end{bmatrix}$$

$\text{rank}(C_{AB}) = 1 \rightarrow$ The matrix C_{AB} is square (2-by-2) and its determinant is zero $\Rightarrow \text{rank}(C_{AB}) = 1$

Also notice that the columns of C_{AB} are linearly dependent. (see the note on mathematical review for the definition of rank of a matrix)

Since $\text{rank}(C_{AB})$ is smaller than the number of the states \Rightarrow the system is not completely controllable

Problem 2

First notice that for a system $\dot{x} = Ax + Bu$, with $x \in \mathbb{R}^n$, $A \in \mathbb{R}^{n \times n}$ and $u \in \mathbb{R}^m$, where n is the number of states and m is number of inputs, the controllability matrix is.

$$C_0 = \begin{bmatrix} \underbrace{B}_{n \times m} & \underbrace{AB}_{n \times m} & \underbrace{A^2B}_{n \times m} & \dots & \underbrace{A^{n-1}B}_{n \times m} \end{bmatrix} \Rightarrow C_0 \in \mathbb{R}^{n \times nm}, \text{ that is the matrix } C_0 \text{ is a } n\text{-by-}nm \text{ matrix}$$

then for.

- System 1: $C_{01} \in \mathbb{R}^{2 \times 2} \Rightarrow A_1 \in \mathbb{R}^{2 \times 2}; B_1 \in \mathbb{R}^{2 \times 1}; \left. \begin{matrix} \text{rank}(C_{01}) = 2 \\ \text{n}^\circ \text{ of states} = 2 \end{matrix} \right\} \Rightarrow$ System 1 is completely controllable.

- System 2: $C_{02} \in \mathbb{R}^{2 \times 2} \Rightarrow A_2 \in \mathbb{R}^{2 \times 2}; B_2 \in \mathbb{R}^{2 \times 1}; \left. \begin{matrix} \text{rank}(C_{02}) = 1 \\ \text{n}^\circ \text{ of states} = 2 \end{matrix} \right\} \Rightarrow$ System 2 is not completely controllable

- System 3: $C_3 \in \mathbb{R}^{3 \times 3} \Rightarrow A_3 \in \mathbb{R}^{3 \times 3}; B_3 \in \mathbb{R}^{3 \times 1}; \text{rank}(C_3) = 3$
 $\left. \begin{array}{l} \text{rank}(C_3) = 3 \\ \text{n}^\circ \text{ of states} = 3 \end{array} \right\} \Rightarrow \text{System 3 is completely Controllable.}$

- System 4: $C_4 \in \mathbb{R}^{2 \times 4} \Rightarrow A_4 \in \mathbb{R}^{2 \times 2}; B_4 \in \mathbb{R}^{2 \times 2}; \text{rank}(C_4) = 2$
 $\left. \begin{array}{l} \text{rank}(C_4) = 2 \\ \text{n}^\circ \text{ of states} = 2 \end{array} \right\} \Rightarrow \text{System 4 is completely Controllable.}$

Problem 3

$$A = \begin{bmatrix} 0 & 1 \\ 20.6 & 0 \end{bmatrix}; B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

a) Eigenvalues of A $\rightarrow \det[\lambda I - A] = \det \begin{bmatrix} \lambda & -1 \\ -20.6 & \lambda \end{bmatrix} = \lambda^2 - 20.6 \rightarrow \lambda_{1,2} = \pm \sqrt{20.6}$

the system is internally unstable because A has a positive eigenvalue.

b) Desired characteristic polynomial

$$\begin{aligned} (s + 1.8 + j2.4)(s + 1.8 - j2.4) &= s^2 + 1.8s - j2.4s + 1.8s + 1.8^2 - j2.4 \cdot 1.8 + j2.4s \\ &\quad + j2.4 \cdot 1.8 + 2.4^2 \\ &= s^2 + 3.6s + 9 \quad (*) \end{aligned}$$

$$K = [k_0 \ k_1] \Rightarrow (A - BK) = \begin{bmatrix} 0 & 1 \\ -(-20.6 + k_0) & -k_1 \end{bmatrix}$$

$$\text{Char}(A - BK) = s^2 + k_1 s + (-20.6 + k_0) \quad (**)$$

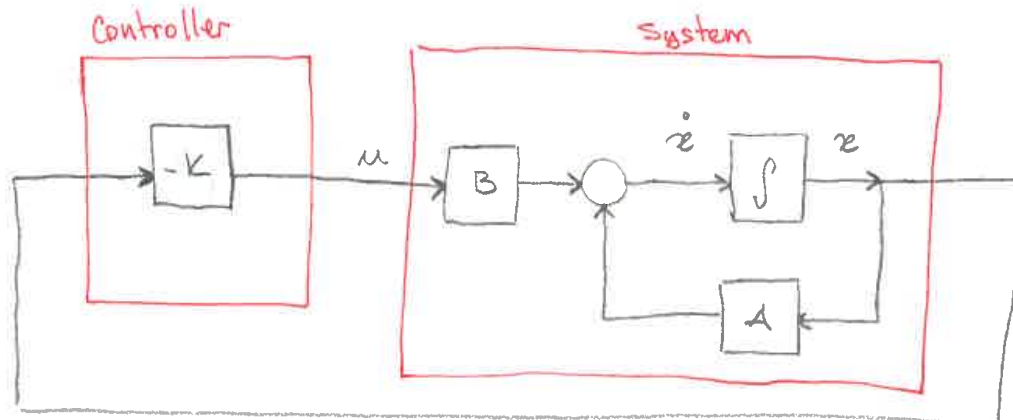
From (*) and (**) $\Rightarrow k_1 = 3.6$

$$k_0 = 29.6$$

$$K = [29.6 \quad 3.6]$$

$$u = -K \cdot x$$

c)



Problem 4

$$a) \quad A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} ; \quad B = \begin{bmatrix} 0 \\ 1 \end{bmatrix} ; \quad C_{AB} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} ; \quad \left. \begin{array}{l} \text{rank}(C_{AB}) = 2 \\ \text{n}^\circ \text{ of states} = 2 \end{array} \right\} \Rightarrow \text{the system is completely Controllable}$$

b). Desired characteristic polynomial

$$(s+1)(s+1) = s^2 + 2s + 1$$

the system is in companion form, then with $K = [k_0 \ k_1]$

$$(A - BK) = \begin{bmatrix} 0 & 1 \\ -k_0 & -k_1 \end{bmatrix} \rightarrow \text{char}(A - BK) = s^2 + k_1 s + k_0$$

Matching the desired char. polynomial and $\text{char}(A - BK) \Rightarrow k_0 = 1 ; k_1 = 2$

$$u = -[1 \ 2] x$$

Problem 5

$$a) \quad A = \begin{bmatrix} 1 & 1 \\ -4 & 3 \end{bmatrix} ; \quad B = \begin{bmatrix} 0 \\ 2 \end{bmatrix} ; \quad C_{AB} = \begin{bmatrix} 0 & 2 \\ 2 & 6 \end{bmatrix} ; \quad \left. \begin{array}{l} \text{rank}(C_{AB}) = 2 \\ \text{n}^\circ \text{ of states} = 2 \end{array} \right\} \Rightarrow \text{the system is completely Controllable}$$

$$\text{char}(A) = \det(sI - A) = \det \begin{bmatrix} s-1 & -1 \\ 4 & s-3 \end{bmatrix} = s^2 - 4s + 7$$

then, the companion form is

$$A_c = \begin{bmatrix} 0 & 1 \\ -7 & 4 \end{bmatrix}; \quad B_c = \begin{bmatrix} 0 \\ 1 \end{bmatrix}; \quad C_{AcBc} = \begin{bmatrix} 0 & 1 \\ 1 & 4 \end{bmatrix}$$

$$\exists T \mid z = T^{-1}x \rightarrow \dot{z} = A_c z + B_c u, \text{ with } A_c = T^{-1}AT$$

$$B_c = T^{-1}B$$

$$C_{AcBc} = [B_c \quad A_c B_c] = [T^{-1}B \quad T^{-1}AT T^{-1}B] = T^{-1}[B \quad AB] = T^{-1}C_{AB} \Rightarrow$$

$$\Rightarrow T = C_{AB} C_{AcBc}^{-1} \quad \text{and} \quad T^{-1} = C_{AcBc} \cdot C_{AB}^{-1}$$

$$C_{AcBc}^{-1} = \begin{bmatrix} -4 & 1 \\ 1 & 0 \end{bmatrix} \Rightarrow T = \begin{bmatrix} 0 & 2 \\ 2 & 6 \end{bmatrix} \begin{bmatrix} -4 & 1 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 2 & 0 \\ -2 & 2 \end{bmatrix}$$

b)

Desired characteristic polynomial $\lambda_1 = -1$; $\lambda_2 = -2$

$$(s+1)(s+2) = s^2 + 3s + 2$$

Design using the companion form for the state space model $\dot{z} = A_c z + B_c u \Rightarrow$

$$u = -k_z z; \text{ with } k_z = [k_{z0} \quad k_{z1}].$$

$$(A_c - B_c k_z) = \begin{bmatrix} 0 & 1 \\ -(7+k_{z0}) & -(4+k_{z1}) \end{bmatrix}; \quad \text{Char}(A_c - B_c k_z) = s^2 + (-4+k_{z1})s + (7+k_{z0})$$

$$\Rightarrow \begin{cases} 7+k_{z0} = 2 \\ -4+k_{z1} = 3 \end{cases} \Rightarrow \begin{cases} k_{z0} = -5 \\ k_{z1} = 7 \end{cases} \Rightarrow k_z = [-5 \quad 7] \rightarrow u = -k_z z$$

However, the controller has to be expressed in the original states

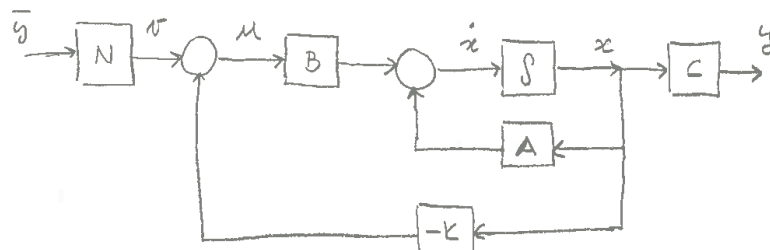
$$u = -k_z T^{-1}x = -[-5 \quad 7] \begin{bmatrix} \frac{1}{2} & 0 \\ \frac{1}{2} & \frac{1}{2} \end{bmatrix} x = -[1 \quad \frac{7}{2}] x$$

$$k = [1 \quad \frac{7}{2}]$$

c)

Page 5

Alternative solution



$$\dot{x} = (A - BK)x + BN\bar{y}$$

$$y = Cx$$

$$\rightarrow \text{new states } z \triangleq x + (A - BK)^{-1}BN\bar{y}$$

$$y \triangleq C[z - (A - BK)^{-1}BN\bar{y}]$$

$$\dot{z} = \dot{x} = (A - BK)x + BN\bar{y} = (A - BK)z$$

The system in states z is

$$\begin{cases} \dot{z} = (A - BK)z \\ y = Cz - C(A - BK)^{-1}BN\bar{y} \end{cases}$$

Since $\text{eig}(A - BK) \begin{matrix} \nearrow \\ \searrow \end{matrix} \begin{matrix} -1 \\ -2 \end{matrix}$, then the system is stable and thus $\lim_{t \rightarrow \infty} z = 0$, which implies that

$$\lim_{t \rightarrow \infty} y = -C(A - BK)^{-1}BN\bar{y}. \text{ On the other side, it}$$

is required that $\lim_{t \rightarrow \infty} y = \bar{y}$.

therefore

$$C(A - BK)^{-1}BN = -1 \Rightarrow [1 \ 0] \begin{bmatrix} 1 & 1 \\ -6 & -4 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 2 \end{bmatrix} N = -1 \Rightarrow$$

$$\Rightarrow [1 \ 0] \begin{bmatrix} -2 & -\frac{1}{2} \\ 3 & \frac{1}{2} \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix} N = -1 \Rightarrow -1N = -1 \Rightarrow \boxed{N = 1}$$