

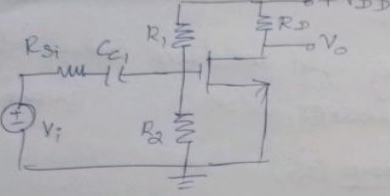
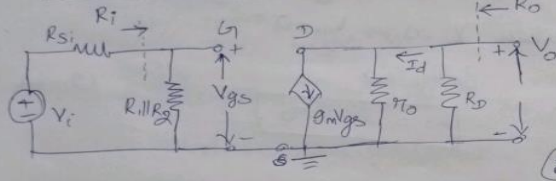
Test: CLAT- 2
Course Code & Title: 18ECC201J – Analog Electronic Circuits
Year & Sem: II / IV
Date: 24-05-2022
Duration: 2 periods
Max. Marks: 50
Course Articulation Matrix :

18ECC201J - Analog Electronic Circuits		Program Outcomes (POs)														
		Graduate Attributes												PSO		
COs	Course Outcomes (COs)	1	2	3	4	5	6	7	8	9	10	11	12	1	2	3
CO-1	Analyze bipolar amplifier circuits and their frequency response.	1	2	3	-	-	-	-	-	-	-	-	-	-	-	-
CO-2	Develop MOSFET amplifier circuits and their frequency response.	1	2	3	-	-	-	-	-	-	-	-	-	-	-	-
CO-3	Compile various negative feedback amplifier and oscillator circuits.	1	-	3	-	-	-	-	-	-	-	-	-	-	-	-
CO-4	Demonstrate the different classes of power amplifiers according to their performance characteristics.	1	2	3	-	-	-	-	-	-	-	-	-	-	-	-
CO-5	Construct the basic circuit building blocks that are used in the design of IC amplifiers, namely current mirrors and sources.	1	2	3	-	-	-	-	-	-	-	-	-	-	-	-
CO-6	Organize analog electronic circuits using discrete components to measure various analog circuits' performance.	-	-	3	-	-	-	-	-	2	-	-	-	3	1	-

Part - A
(10 x 1 = 10 Marks)
Instructions: Answer ALL the Questions

Q. No	Question	Marks	BL	CO	PO
1.	at negative peak	1	1	2	1
2.	It blocks the noise & for ac signal it acts a short circuit resulting in grounding source terminal.	1	2	2	1
3.	common gate	1	2	2	1
4.	45MHz	1	2	2	2
5.	mid-frequency	1	2	2	1
6.	Shunt-Series	1	2	3	2
7.	Series-Series	1	2	3	1
8.	Shunt-Shunt	1	2	3	1
9.	one frequency	1	2	3	1
10	capacitors	1	2	3	1

Part - B
(4 x 10 = 40 Marks)
SECTION B1
Instructions: Answer ANY 2 Questions

11	$V_{GSQ} = 12 \times \frac{450 \times 10^3}{(170 + 450) \times 10^3} = 8.70 \text{ V} \quad \text{--- (1 mark)}$ $I_{DQ} = K_n (V_{GSQ} - V_{TN})^2 \quad \text{--- (1 mark)}$ $= 4 \times 10^{-3} (8.7 - 2.5)^2$ $= 153 \text{ mA}$ $g_m = 2K_n (V_{GSQ} - V_{TN})$ $= 2 \times 4 (8.70 - 2.5)$ $= 49.6 \text{ mA/V} \quad \text{--- (2 mark)}$ $r_o = \left(\frac{1}{\lambda I_{DQ}} \right)^{-1} = (0.01)(153)^{-1}$ $= 0.65 \text{ k}\Omega = 650 \Omega \quad \text{--- (1 mark)}$ $A_v = \frac{g_m (R_S \parallel r_o)}{1 + g_m (R_S \parallel r_o)} \frac{R_1 \parallel R_2}{R_S + (R_1 \parallel R_2)}$ $R_1 \parallel R_2 = 123 \text{ k}$ $A_v = \frac{49.6 (575.22)}{1 + 49.6 (575.22)} \frac{123 \times 10^3}{5 \text{ k} + 123 \times 10^3}$ $= 0.96 \quad \text{--- (5 mark)}$	10	3	2	2
12	 <p>Equivalent circuit</p>  <p>Voltage gain</p> $A_v = V_o / V_i$ $V_o = -g_m V_{gs} (r_o \parallel R_D)$ $V_{gs} = V_i \times \frac{R_1 \parallel R_2}{R_{Si} + (R_1 \parallel R_2)} \quad \text{--- (5 mark)}$ <p>Subs V_{gs} in V_o</p> $\therefore V_o = -g_m (r_o \parallel R_D) V_i \times \frac{R_1 \parallel R_2}{R_{Si} + (R_1 \parallel R_2)}$ $A_v = \frac{V_o}{V_i} = -g_m (r_o \parallel R_D) \frac{(R_1 \parallel R_2)}{R_{Si} + (R_1 \parallel R_2)}$	10	3	2	2

Input resistance (R_i)

$$R_i = R_1 \parallel R_2$$

Low frequency input resistance looking into the gate of MOSFET is ∞ . (1 mark)

output resistance (R_o)

R_o is found by setting $V_i = 0$

When $V_i = 0 \Rightarrow V_{gs} = 0$

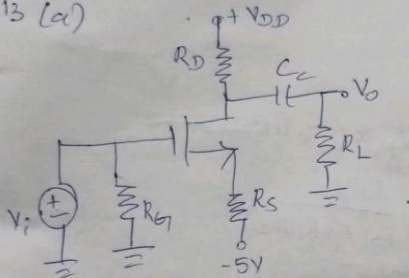
$$g_m V_{gs} = 0$$

$$\therefore R_o = R_D \parallel r_o$$

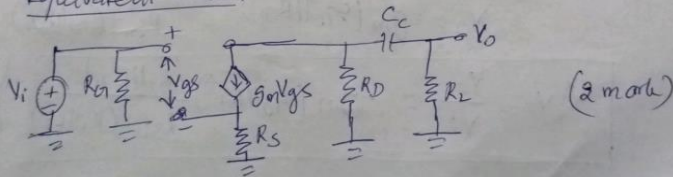
(2 mark)

13

13 (a)



Equivalent circuit : r_o - assumed to be ∞



(2 mark)

The maximum output voltage assuming C_c is a short circuit.

$$|V_o|_{\max} = g_m V_{gs} (R_D \parallel R_L)$$

$$V_i = V_{gs} + g_m V_{gs} R_S$$

$$V_{gs} = \frac{V_i}{1 + g_m R_S}$$

$$V_o = g_m \frac{V_i}{1 + g_m R_S} (R_D \parallel R_L)$$

$$A_{v \max} = V_o / V_i = \frac{g_m (R_D \parallel R_L)}{1 + g_m R_S}$$

The time constant is a function of the effective resistance seen by capacitor C_c which is determined by setting independent sources equal to zero.

$V_i = 0$, $g_m V_{gs} = 0$ then

$$\tau_s = C_c (R_D + R_L)$$

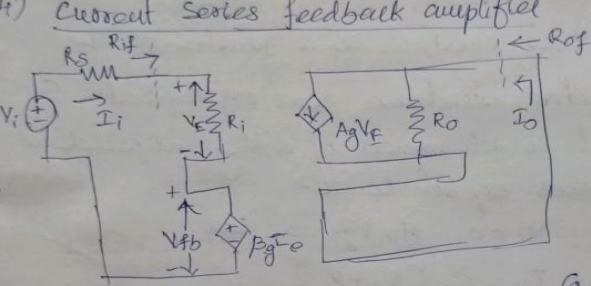
$$f_L = \frac{1}{2\pi\tau_s}$$

(3 mark)

	<p>b.</p> $R_i = 80\text{ k}\Omega \parallel 30\text{ k}\Omega = 21.8\text{ k}\Omega \quad (2\text{ mark})$ $R_o = R_D \parallel r_o$ $r_o = (\lambda I_{DQ})^{-1}$ $I_{DQ} = k_n (V_{GS} - V_{TN})^2$ $= 0.5 (2.72 - 1.5)^2 = 0.613\text{ mA}$ $r_o = (\lambda I_{DQ})^{-1} = 163\text{ k}\Omega \quad (3\text{ mark})$ $R_o = R_D \parallel r_o = 5.78\text{ k}\Omega$	5	3	2	3
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SECTION B2

Instructions: Answer ANY 2 Questions

14	<p>a.</p> <p>i) Current series feedback amplifier</p>  <p>Transfer gain (A_{gf})</p> $A_{gf} = I_o / V_i$ $I_o = A_g V_E \Rightarrow V_E = I_o / A_g$ $V_i = V_E + V_{fb} = \frac{I_o}{A_g} + \beta_g I_o = \left(1 + \frac{\beta_g}{A_g}\right) I_o$ $A_{gf} = \frac{I_o}{V_i} = \frac{A_g}{1 + \beta_g A_g} \quad (2\text{ mark})$ <p>Input resistance (R_{if})</p> $R_{if} = \frac{V_i}{I_i}, \quad V_i = V_E + V_{fb}, \quad V_{fb} = \beta_g I_o$ $V_{fb} = \beta_g A_g V_E$ <p>Subs in V_i, $V_i = V_E + \beta_g A_g V_E = V_E (1 + \beta_g A_g)$</p> $V_E = I_i R_i, \quad V_i = I_i R_i (1 + \beta_g A_g)$ $R_{if} = \frac{V_i}{I_i} = R_i (1 + \beta_g A_g) \quad (2\text{ mark})$ <p>Output resistance (R_{of})</p> <p>Setting $V_i = 0$, applying a test current I_x in output circuit.</p> $R_{of} = \frac{V_x}{I_x}, \quad I_x = \frac{V_x}{R_o} + A_g V_E$ <p>As $V_i = 0$, $V_E + V_{fb} = 0 \Rightarrow V_E = -V_{fb} = -\beta_g I_x$</p> <p>Subs V_{fb} in I_x, $I_x = \frac{V_x}{R_o} + A_g (-\beta_g I_x)$</p> $I_x (1 + \beta_g A_g) = V_x / R_o$ $R_{of} = \frac{V_x}{I_x} = R_o (1 + \beta_g A_g) \quad (2\text{ mark})$	8	3	3	2
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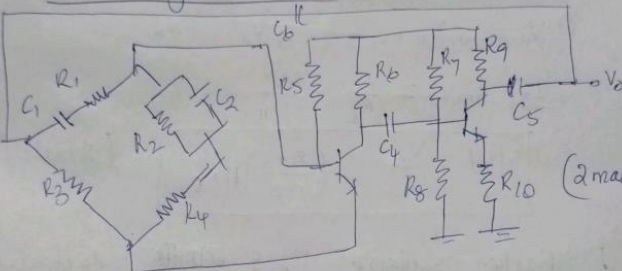
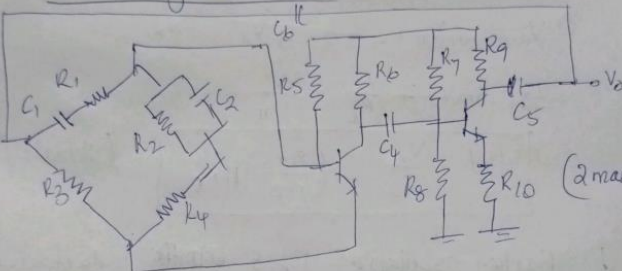
1. The total phase shift around the closed loop should be 0° (or) 360° . (1 mark)
$$\angle AP = 360^\circ / 0^\circ$$
2. The product of open loop gain of the amplifier and the feedback network should be unity. i.e. $|AP| = 1$ (1 mark)

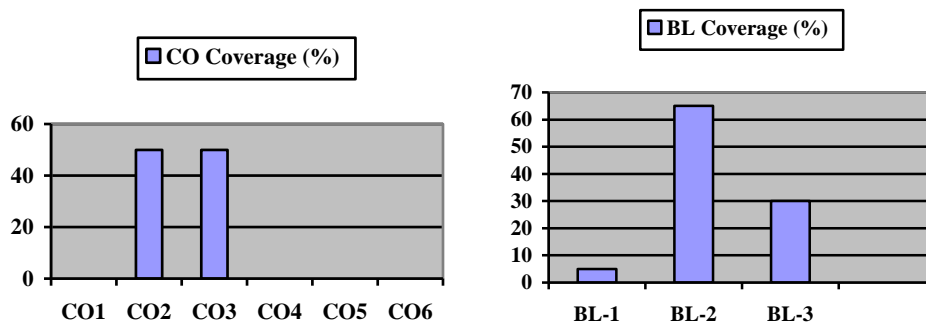
2

a.

$$R_{of} = \frac{V_x}{I_o} = \left(\frac{g_m}{1+\beta} \right) \parallel R_E \quad (2 \text{ mark})$$

3

	<p>b.</p> <p>b) Hartley oscillator – uses tapped inductance Colpitts oscillator – uses tapped capacitance</p> <p>(ii) In Hartley oscillator, the feedback for the active device is taken from a voltage divider made of two inductors in series across the capacitor whereas in Colpitts oscillator, the voltage divider is made of two capacitors in series across the inductor</p>	2	2	3	1
16	<p>a.</p> <p>16(a) $f_0 = \frac{1}{2\pi\sqrt{LC}}$ (or) $C = \frac{1}{(2\pi f_0)^2 L}$</p> <p>Given $f_0 = (540-1650) \text{ kHz}$, $L = 1 \text{ mH}$</p> <p>$f_0 = 540 \text{ kHz}$</p> $C_{\text{max}} = \frac{1}{4\pi^2 \times (540 \times 10^3)^2 \times 1 \times 10^{-3}}$ <p>$f_0 = 1650 \text{ kHz}$ $\Rightarrow 86.86 \text{ pF}$</p> $C_{\text{min}} = \frac{1}{4\pi^2 \times (1650 \times 10^3)^2 \times 1 \times 10^{-3}}$ <p>$= 9.3 \text{ pF}$</p> <p>The range of $C \Rightarrow 9.3 \text{ pF}$ to 86.86 pF.</p> <p>b.</p> <p>(b) Weinbridge oscillator</p>  <p>$f_0 = \frac{1}{2\pi\sqrt{R_1 C_1 R_2 C_2}}$ if $R_1 = R_2$, $C_1 = C_2$ (1 mark)</p> <p>$f_0 = \frac{1}{2\pi RC}$ (1 mark)</p> <p>Condition for oscillation</p> <p>$R_3/R_4 = 2$</p> <p>Explanation – (2 marks)</p>	4	3	3	3
	<p>(b) Weinbridge oscillator</p>  <p>$f_0 = \frac{1}{2\pi\sqrt{R_1 C_1 R_2 C_2}}$ if $R_1 = R_2$, $C_1 = C_2$ (1 mark)</p> <p>$f_0 = \frac{1}{2\pi RC}$ (1 mark)</p> <p>Condition for oscillation</p> <p>$R_3/R_4 = 2$</p> <p>Explanation – (2 marks)</p>	6	2	3	2



Evaluation Sheet

Name of the Student:

Register No.:

Part- A (10 x 1= 10 Marks)					
Q. No	CO	PO	Maximum Marks	Marks Obtained	Total
1	CO2	1	1		
2	CO2	1	1		
3	CO2	1	1		
4	CO2	2	1		
5	CO2	1	1		
6	CO3	2	1		
7	CO3	1	1		
8	CO3	1	1		
9	CO3	1	1		
10	CO3	1	1		
Part- B (4 x 10= 40 Marks)					
11	CO2	3	10		
12	CO2	2	10		
13.a	CO2	2	5		
13.b	CO2	3	5		
14.a	CO3	2	8		
14.b	CO3	2	2		
15.a	CO3	3	8		
15.b	CO3	1	2		
16.a	CO3	3	4		
16.b	CO3	2	6		

Consolidated Marks:

CO	Maximum Marks	Marks Obtained
2		
3		
Total		

PO	Maximum Marks	Marks Obtained
1		
2		
3		
Total		

Signature of Course Teacher