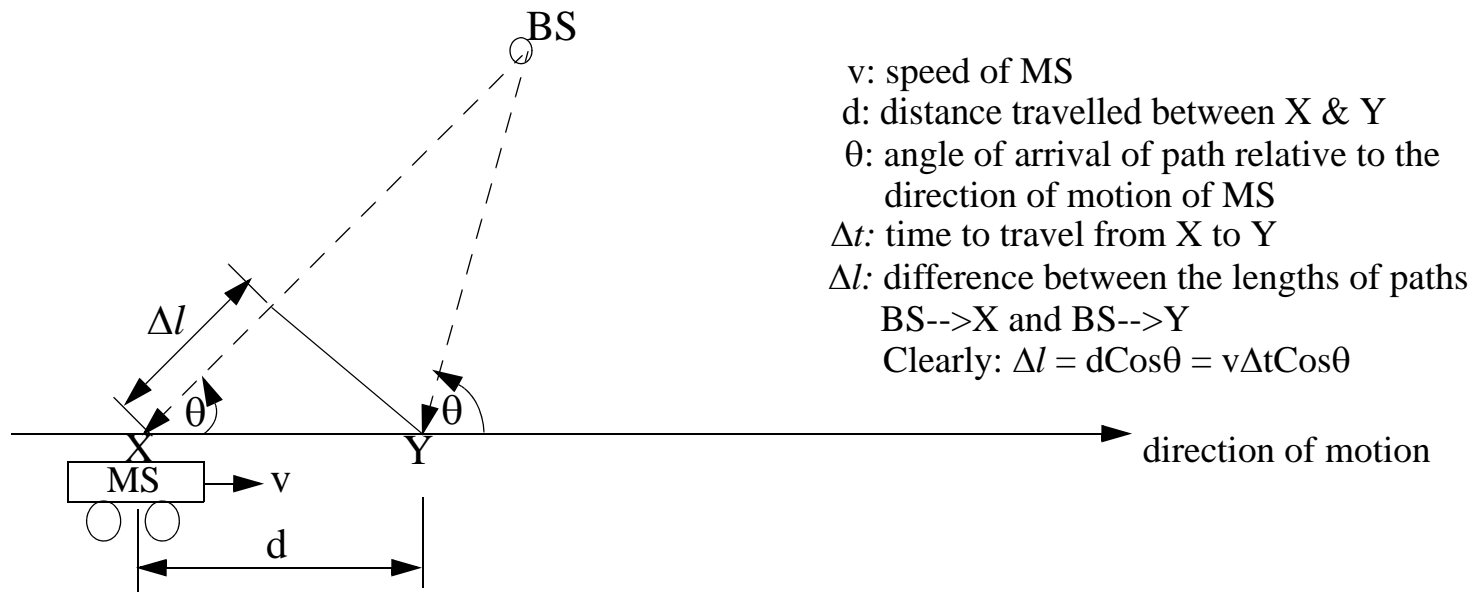


Frequency Dimension Characterization of Wireless Channels

Doppler Shift:

- shift in the operating (carrier) frequency due to relative motion between the transmitter and the receiver

Case 1: Doppler shift caused by motion in free space (i.e. only one path exists)



Assumption: BS is very far away from X and Y

Case 1: Analysis of Doppler Shift in Free Space

Assumption: BS is very far away from the MS, so that θ is approximately the same at points X and Y
(on Page 51)

Let

$\Delta\phi$: phase change in the received signal due to path length change Δl

$$\text{Now, } \Delta\phi = \left(\frac{2\pi}{\lambda}\right)\Delta l = \frac{2\pi v\Delta t}{\lambda} \cos\theta$$

Hence, the apparent change in frequency of the received signal (or Doppler Shift), f_d , is given by:

$$f_d = \frac{1}{2\pi} \frac{\Delta\phi}{\Delta t} = \frac{v}{\lambda} \cos\theta = v \frac{f_o}{c} \cos\theta = f_{d_{\max}} \cos\theta$$

where f_o is the operating frequency and $f_{d_{\max}} = v \frac{f_o}{c} = \frac{v}{\lambda}$ = maximum Doppler shift

f_{in} , the instantaneous frequency of the received signal is then given by:

$$f_{in} = f_o + f_d = f_o + f_{d_{\max}} \cos\theta \quad \text{or} \quad (f_o - f_{d_{\max}}) \leq f_{in} \leq (f_o + f_{d_{\max}})$$

Case 2: Analysis of Doppler Shift in a Multipath Channel

In a multipath channel, the transmitted signal arrives at the receiving antenna via

- a random number N paths, each with random amplitude a_i and random phase ϕ_i
- each path i arrives at a random angle θ_i , uniformly distributed between 0 and 2π

With relative motion, each path i experiences its own Doppler shift $f_{d,i} = f_{d_{max}} \cos(\theta_i)$ (from Page 52)

Let $r(t) = \sum_{i=1}^N a_i \exp[j(\phi_i + 2\pi f_{d,i} t)]$ = low-pass complex equivalent of received signal envelope

Let $R(\Delta t)$ = autocorrelation function of $r(t)$ and $S_d(f)$ = power spectrum

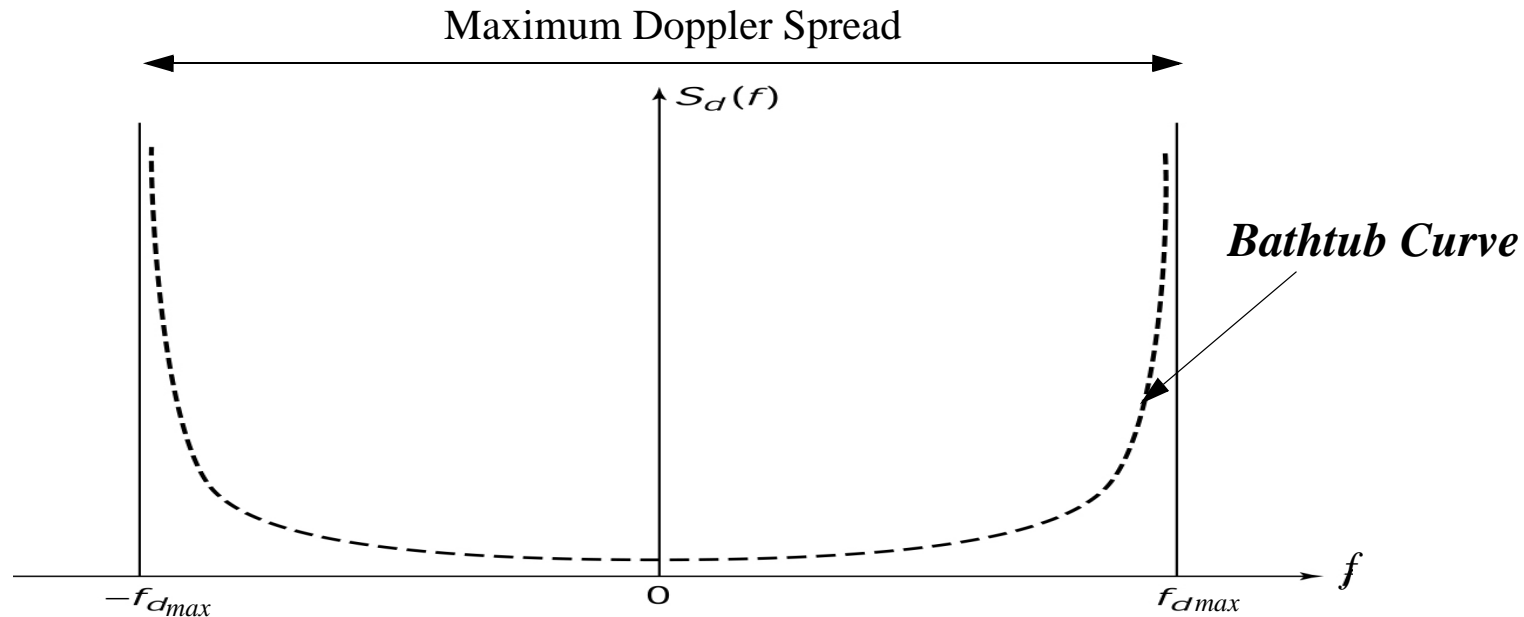
By definition: $R(\Delta t) = E[r(t)r^*(t + \Delta t)]$ and $S_d(f) = F\{R(\Delta t)\}$ = Doppler spectrum

where $r^*(t)$ is the complex conjugate of $r(t)$, $E[.]$ is the expectation operator, and $F\{\}$ is the Fourier transform

Conclusion: Relative motion causes an impulse transmitted at a single frequency to spread out in frequency at the receiver and, in the presence of multipath, each path experiences a different Doppler shift. This phenomenon is referred to as Doppler Spread or Frequency Dispersion

Example Doppler Spectrum: Jake's Spectrum

$$S_d(f) = \frac{S_o}{\pi f_{d_{max}} \sqrt{1 - \left(\frac{f}{f_{d_{max}}}\right)^2}}, \quad |f| \leq f_{d_{max}} \quad S_o = \text{a constant}$$



- Notes: 1. Doppler spectrum $S_d(f)$ = power spectrum of the received signal at different frequencies
2. Maximum Doppler Spread = $2f_{d_{max}}$, where $f_{d_{max}}$ = maximum Doppler shift

Frequency-Dimension Metric: Doppler Spread or Frequency Dispersion

Doppler Spread:

- defined as the range of frequency over which the channel transfer function $H(f)$ is spread due to relative motion between the transmitter and the receiver
- occurs in the frequency domain
- Mathematically, Doppler Spread $\zeta = 2f_d$

Statistic: maximum Doppler Spread $\zeta_{max} = 2f_{d_{max}}$

Effect of Doppler Spread: Frequency distortion (a.k.a “random” FM)

Physical Manifestation of Doppler Spread in the Time Domain: Coherence Time

Coherence Time, T_c : characterizes frequency-dispersiveness of the channel in the *time domain*

- T_c is defined as the period of time over which the impulse response $h(t)$ of the channel is approximately the same

$$h(t) \approx h(t + \Delta t), \quad |\Delta t| \leq T_c$$

$$h(t) \neq h(t + \Delta t), \quad |\Delta t| > T_c$$

- Alternatively, T_c is the period of time over which the channel amplitudes are correlated
- T_c is a derived quantity from the max Doppler spread (T_c is inversely proportional to ζ_{max})

NOTE: $T_c \neq \frac{1}{B_c}$

Example Relationships between T_c and ζ_{max} :

$$T_c \approx \frac{9}{16\pi\zeta_{max}} = \text{time interval for which the envelope correlation function} > 50\%$$

$$T_c \approx \left(\frac{9}{16\pi\zeta_{max}^2} \right)^{\frac{1}{2}} = \frac{0.423}{\zeta_{max}} = \text{geometric mean of } \{T_c = 1/\zeta_{max}\} \text{ and } \{T_c = 9/(16\pi\zeta_{max})\}$$

Physical Manifestation of Doppler Spread in the Time Domain: Time Selective vs. Time Flat Channel

Time Selectivity of the Channel: occurs at non-zero Doppler spread (i.e. $0 < \zeta_{max} < \infty$) $\implies T_c < \infty$

- different components of the impulse response experience different attenuation (i.e. channel is time-varying)
- the magnitude of the impulse response at different times are uncorrelated

Time Flatness of the Channel: occurs at zero Doppler spread (i.e. $\zeta_{max} = 0$) $\implies T_c = \infty$

- different components of the impulse response experience the same attenuation (i.e. channel is time-invariant or time-flat)
- the magnitude of the impulse response at different times are correlated

Notes:

- 1) zero Doppler spread means the absence of relative motion (e.g. MS is stationary)
- 2) For a given value of $T_c < \infty$, time-selective channels can be classified into slow and fast channels, depending on the relative value of symbol duration T_s .
- 3) A time-selective channel is slow when the symbol duration T_s is much less than T_c the coherence time. I.e. $T_s \ll T_c$. Conversely, the channel is fast when $T_s \gg T_c$.

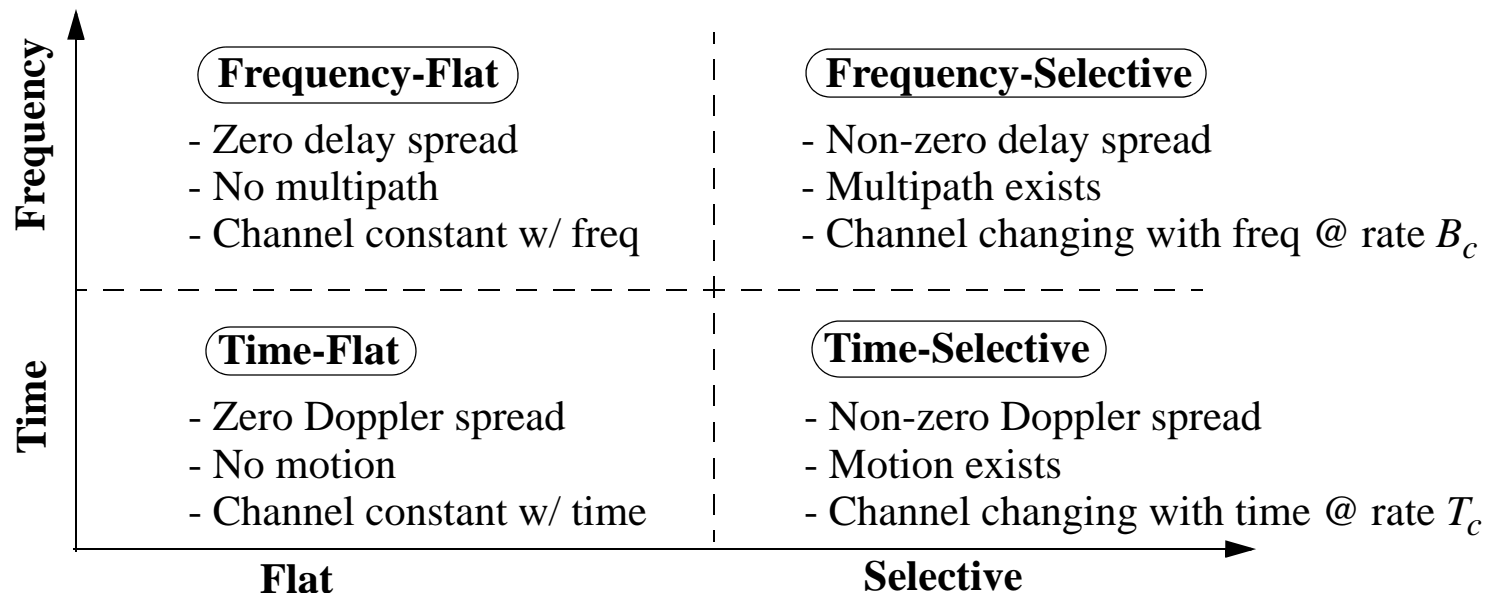
Summary of Channel Types in both Time and Frequency Dimensions

Time Dimension Model: Zero delay spread vs. non-zero delay spread (i.e. no multipath vs. multipath)

Channel types in freq domain: Frequency-flat vs. Frequency-selective channels

Freq. Dimension Model: zero Doppler spread vs. non-zero Doppler spread (i.e. no motion vs. motion)

Channel types in time domain: Time-flat vs. Time-selective channels



Summary of Channel Models in Distance, Time and Frequency Dimensions & User Impacts

Dimension	Channel Model	User Impact
Distance	Path loss	Quality
	Lognormal shadowing	Quality
	Multipath (Rayleigh & Rician)	Quality
Time	Zero delay spread vs. non-zero delay spread (Channel types: Frequency-flat vs. Frequency-selective)	Data rate & quality
Frequency	Zero Doppler spread vs. Non-zero Doppler spread (Channel types: Time-flat vs. Time-selective)	Data rate & quality

Class Example

Problem Statement:

Consider an antenna transmitting at 900 MHz. The receiver, an MS, is travelling at a speed of 30 km/hr and is receiving/transmitting data at 200 kbps. Examine whether the channel fading is slow or fast. Assume that the coherence time, $T_c = 9/(16\pi\zeta_{max})$

Simulation of Rayleigh Fading Signal Case 1: MS is stationary

Inputs:

- Operating frequency, f_o
- Average Power of the received signal, $E[P] = 2\sigma^2$ (from Page 40)

Implementation:

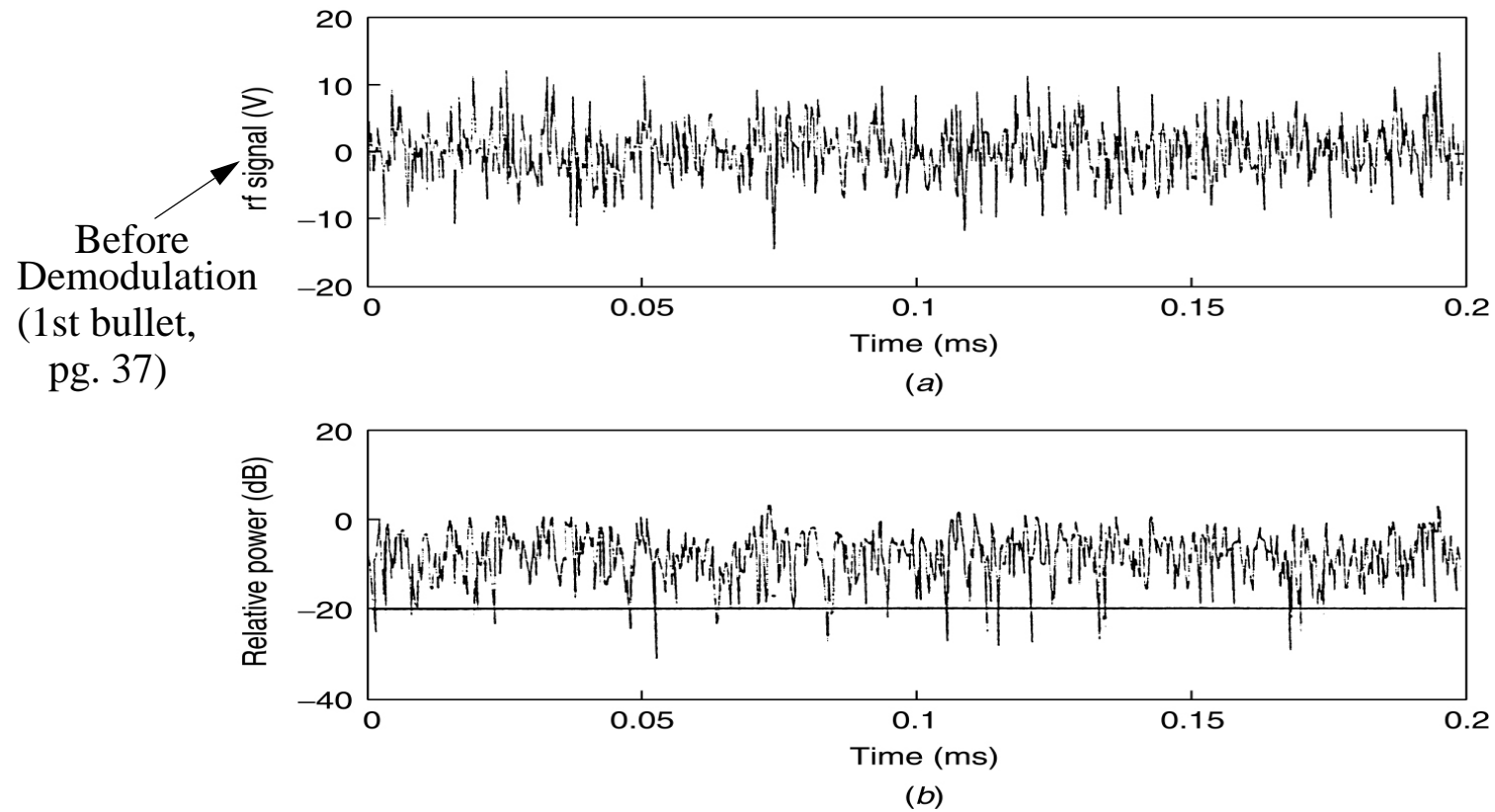
Step 1: Calculate the sampling rate, $f_s = M * f_o$, (typically $M = 4$). Hence, sampling interval, $\Delta t = 1/f_s$

Step 2: At each time instant, generate two Gaussian random variables X and Y , where $X \sim \text{Gaussian}(0, \sigma^2)$ and $Y \sim \text{Gaussian}(0, \sigma^2)$ (Step 2 follows from 3rd bullet on page 37)

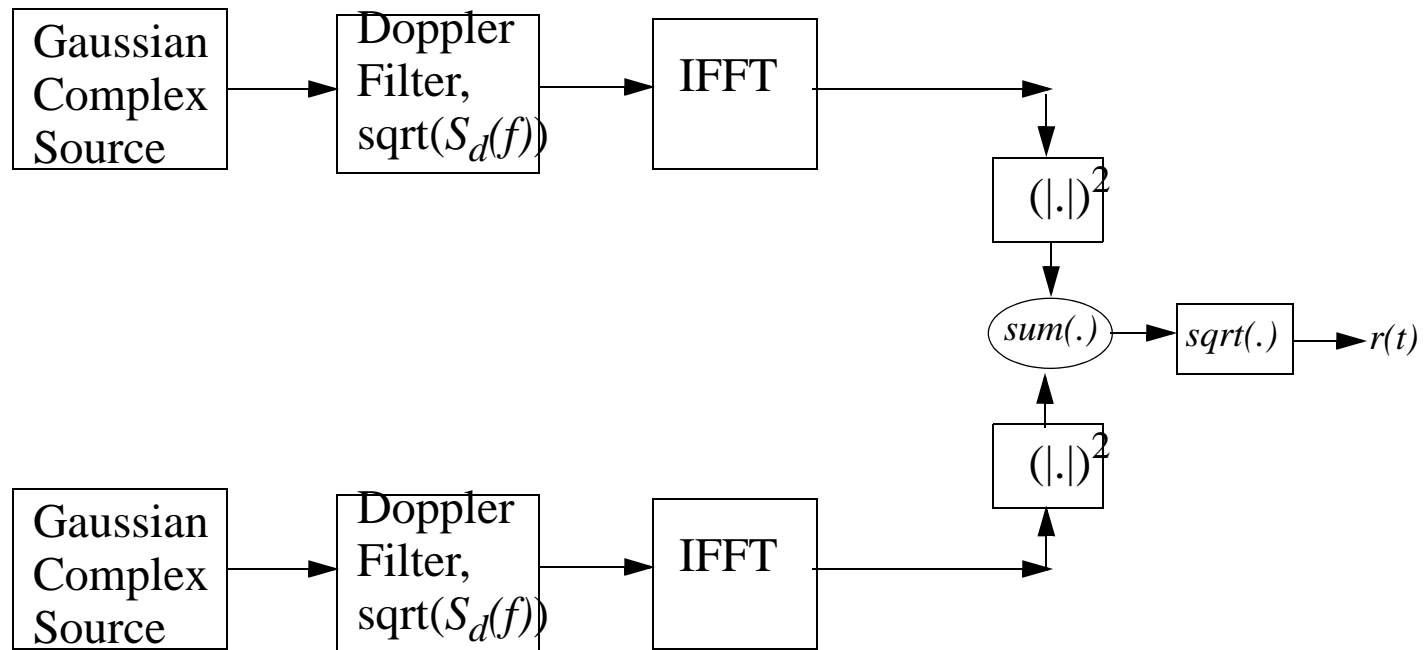
Step 3: Calculate the envelope $A = \sqrt{X^2 + Y^2}$ (Step 3 follows from 4th bullet on page 37)

Step 4: Calculate the power $P = A^2$ (Step 4 follows from the definition of power on page 40)

Illustrative Rayleigh faded Signal



Simulation of Rayleigh fading signal, Case 2: MS is in motion (i.e., with Doppler)



IFFT: Inverse Fast Fourier Transform
 $S_d(f)$: Doppler (power) spectrum

Simulation of Rayleigh fading signal: MS is in motion

Inputs:

- Operating frequency f_o and Speed of motion v . (Hence can calculate $f_{d_{max}} = (vf_o)/c$)
- Doppler Spectrum, $S_d(f)$ (see page 54 for an example $S_d(f)$)
- Number of sampling points in the frequency domain, M_f (M_f is usually a power of 2)

Implementation:

- Step 1: Generate two sets (in-phase and quadrature) of complex Gaussian random variables for each sampling interval Δf ($= f_{d_{max}} / M_f$) in the frequency domain
- Step 2: Generate shaped filter ($\sqrt{S_d(f)}$) samples at each Δf
- Step 3: Multiply the in-phase and quadrature noise sources by the fading spectrum $\sqrt{S_d(f)}$
- Step 4: Perform an Inverse Fast Fourier Transform (IFFT) on the resulting freq domain signals
- Step 5: Add the squares of each signal point in time
- Step 6: Take the square root of the signal obtained in step 5 to calculate the Rayleigh distributed envelope