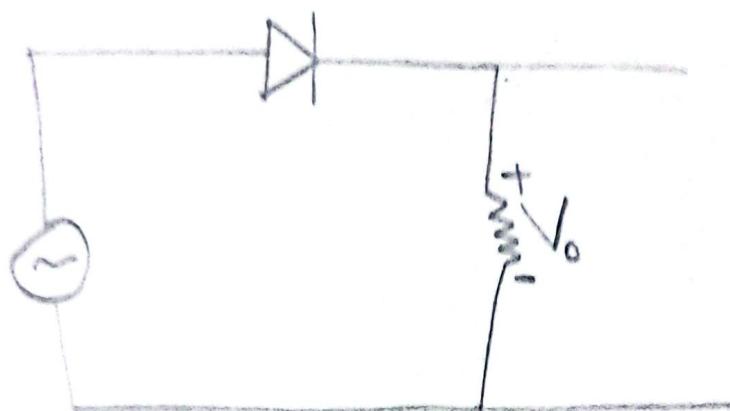
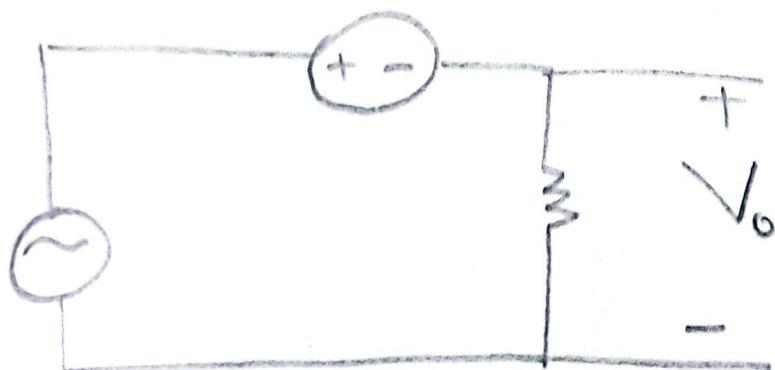


Rectifier

When the diode is on,



$$V_o = V_{in} - V_{D0}$$

$$I = \frac{V_{in} - V_{D0}}{R}$$

For the diode to be on,

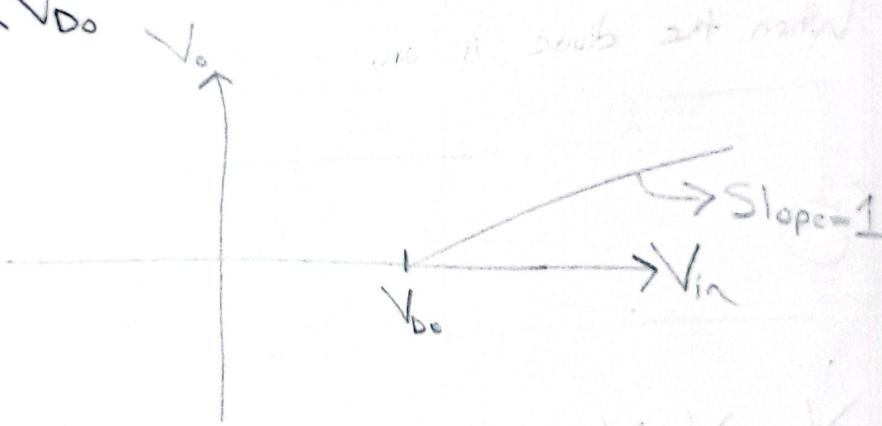
$$V_{in} > V_{D0}$$

When the diode is off.



$$V_o = 0$$

$$V_{in} < V_{D0}$$

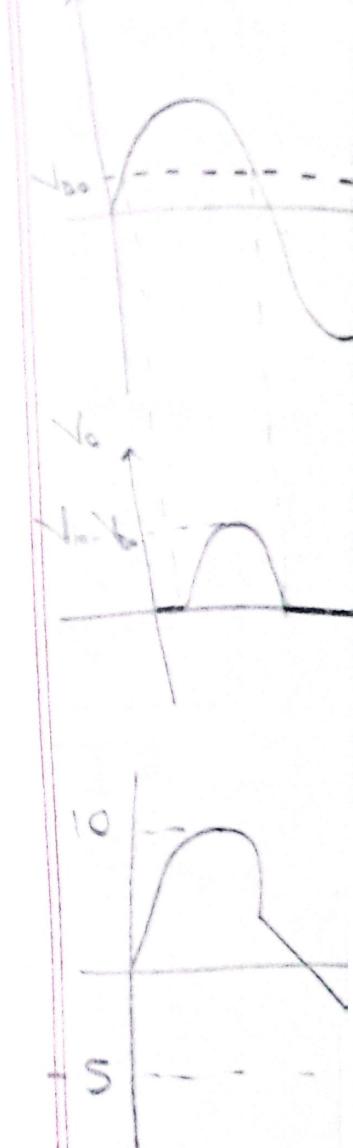


• VTC graph of Rectifier if Diode is on

no diode acts as a short

if $V_o = 0$

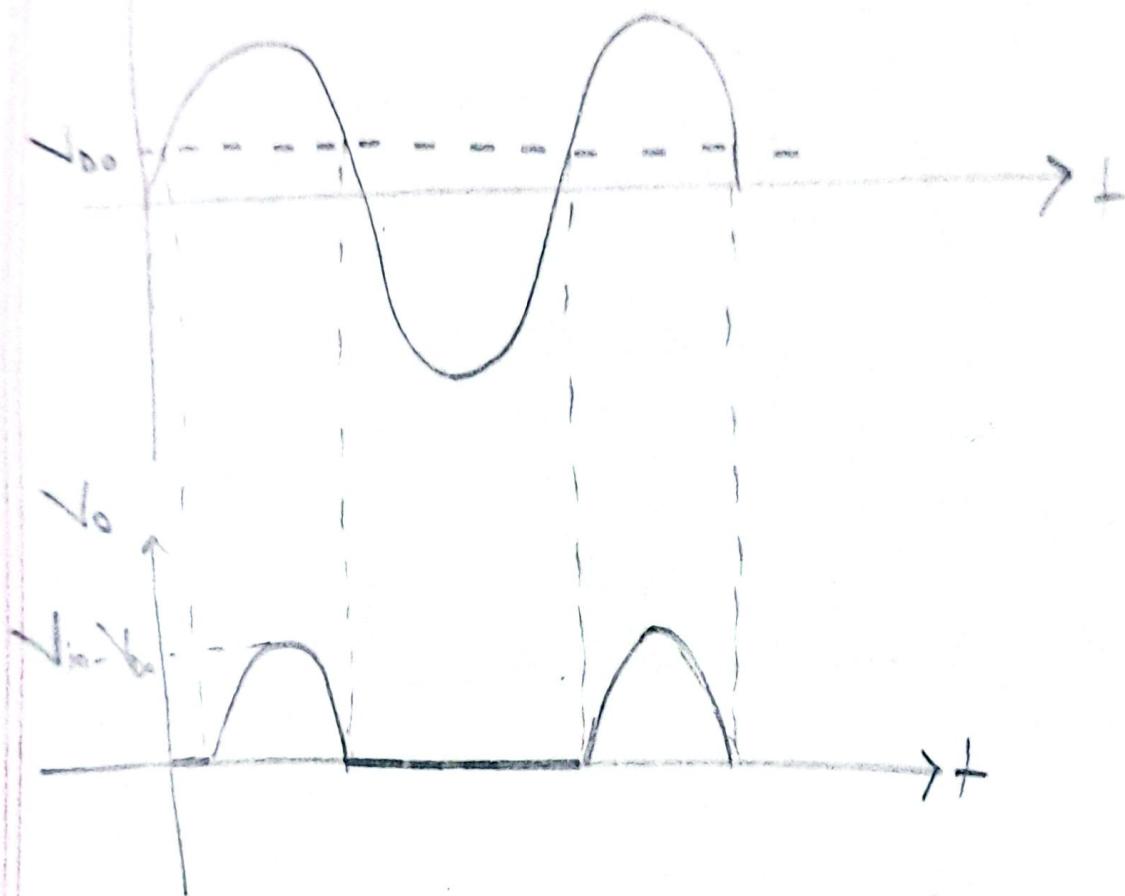
② Draw the V_o 's



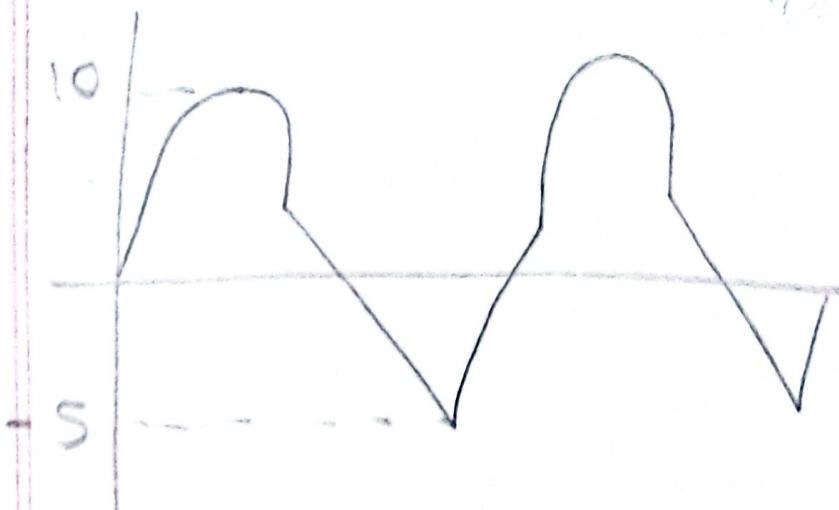
Coverage of
 V_{out}

V_m = Input

2) Draw the V_o vs t graph from V_{in} vs t graph



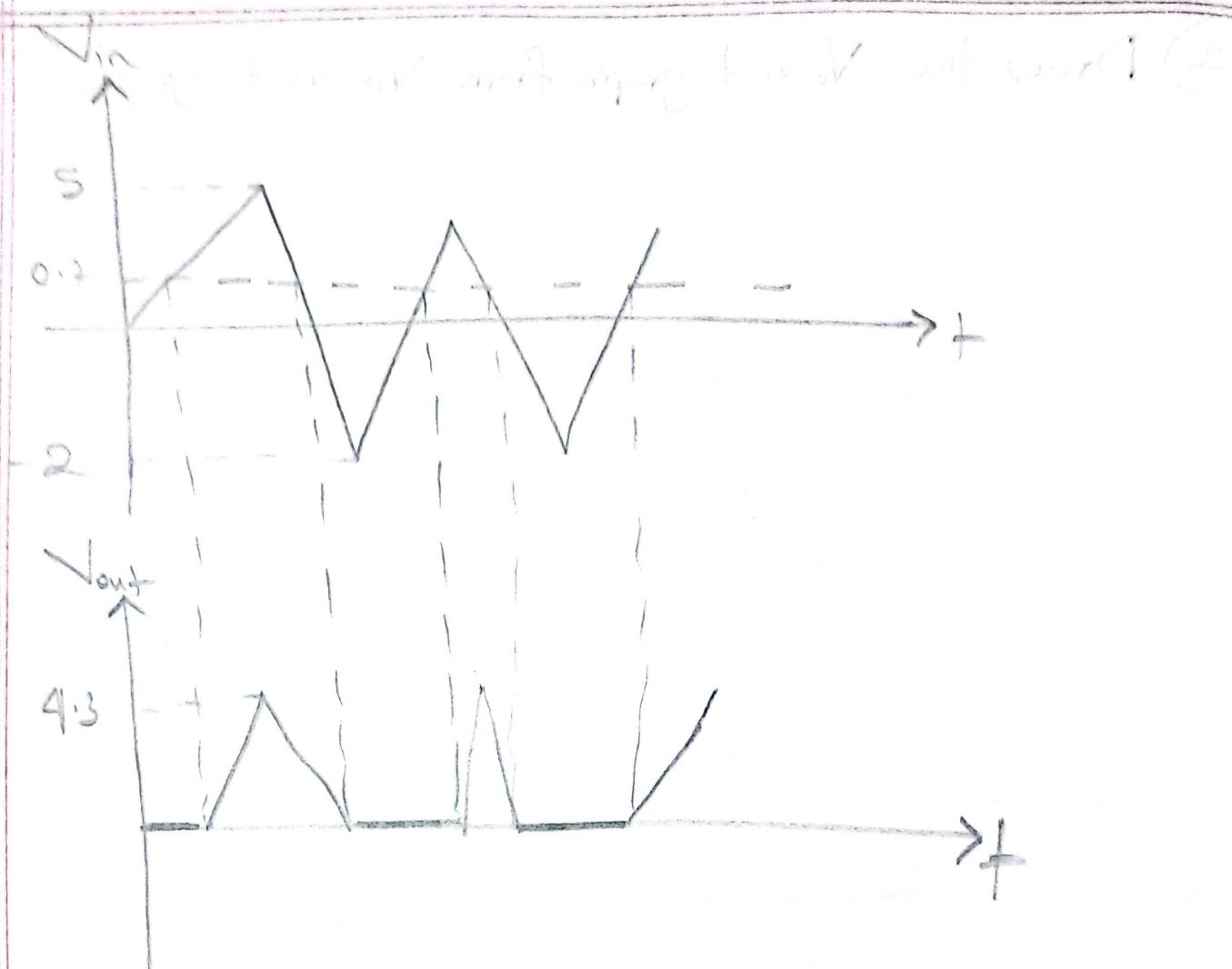
original example



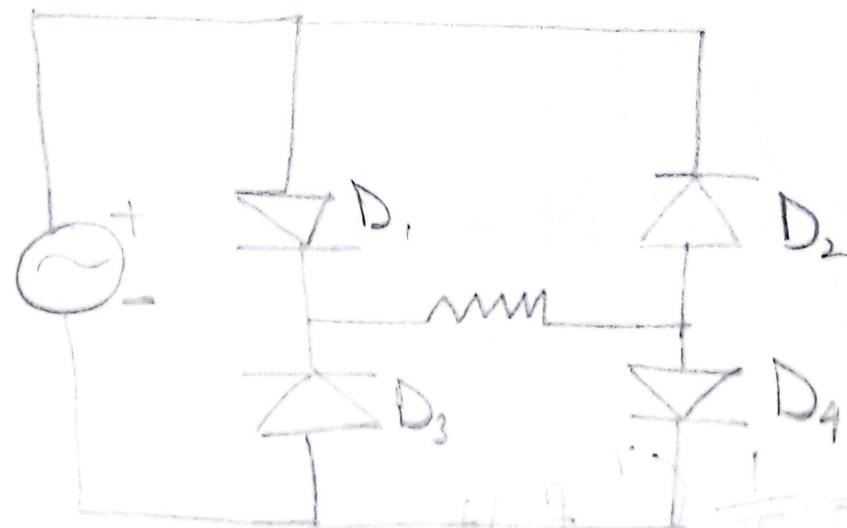
$$\text{average of } f = \frac{1}{T} \int_0^T f dt$$

$$V(\text{avg}) = \frac{V_m}{\pi} - \frac{V_{00}}{2}$$

$$V_{00} = \text{long time avg } V_{00} =$$



Full-wave Rectifier



$$\begin{aligned}
 \frac{V}{T} &= \frac{V_0}{T} = 7.70 \text{ Vpp} \\
 \frac{V}{T} - \frac{V_0}{T} &= (6.65) \text{ V} \\
 \frac{V}{T} &= \sqrt{2} \cdot \text{Half-Wave} = \text{aV}
 \end{aligned}$$

Positive Half Cycle

- D_1 and D_4 will be on for positive half-cycle

$$V_o = V_{in} - 2V_{D0}$$

$$V_{in} > 2V_{D0}$$

For V_o to be positive

Negative Half Cycle

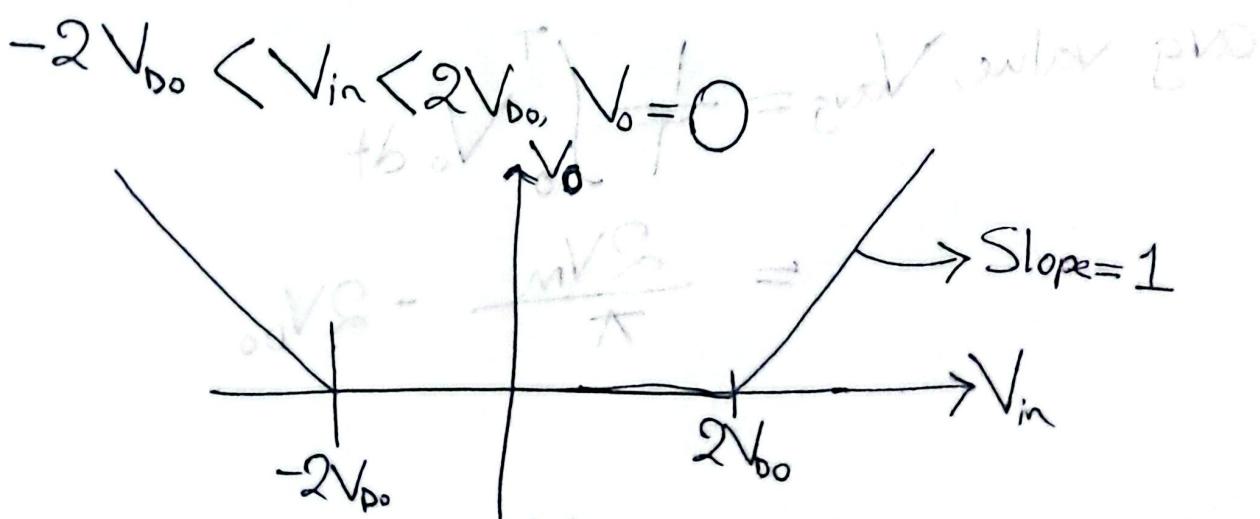
$$V_o = -V_{in} - 2V_{D0}$$

For,

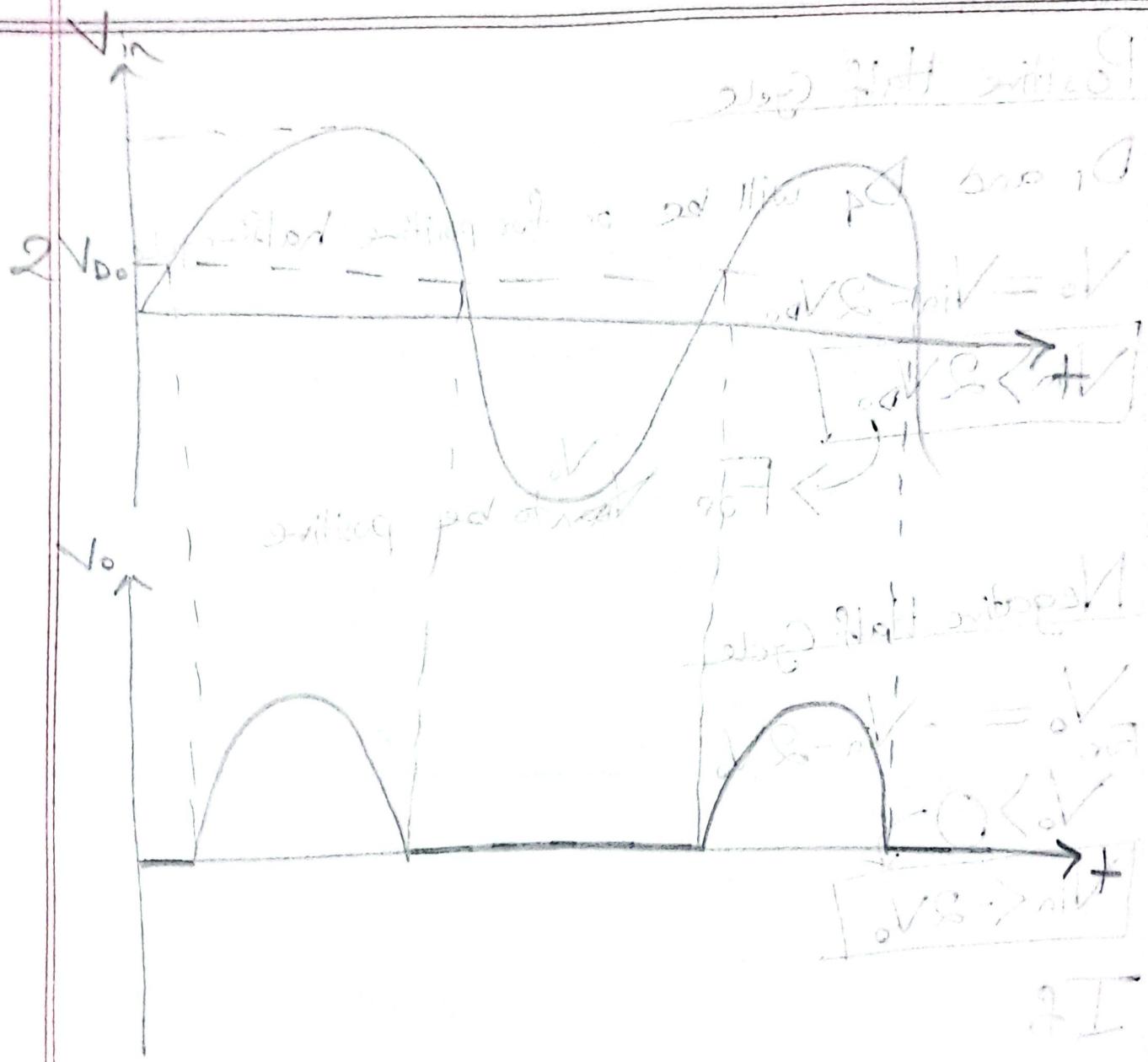
$$V_o > 0$$

$$V_{in} < -2V_{D0}$$

If,

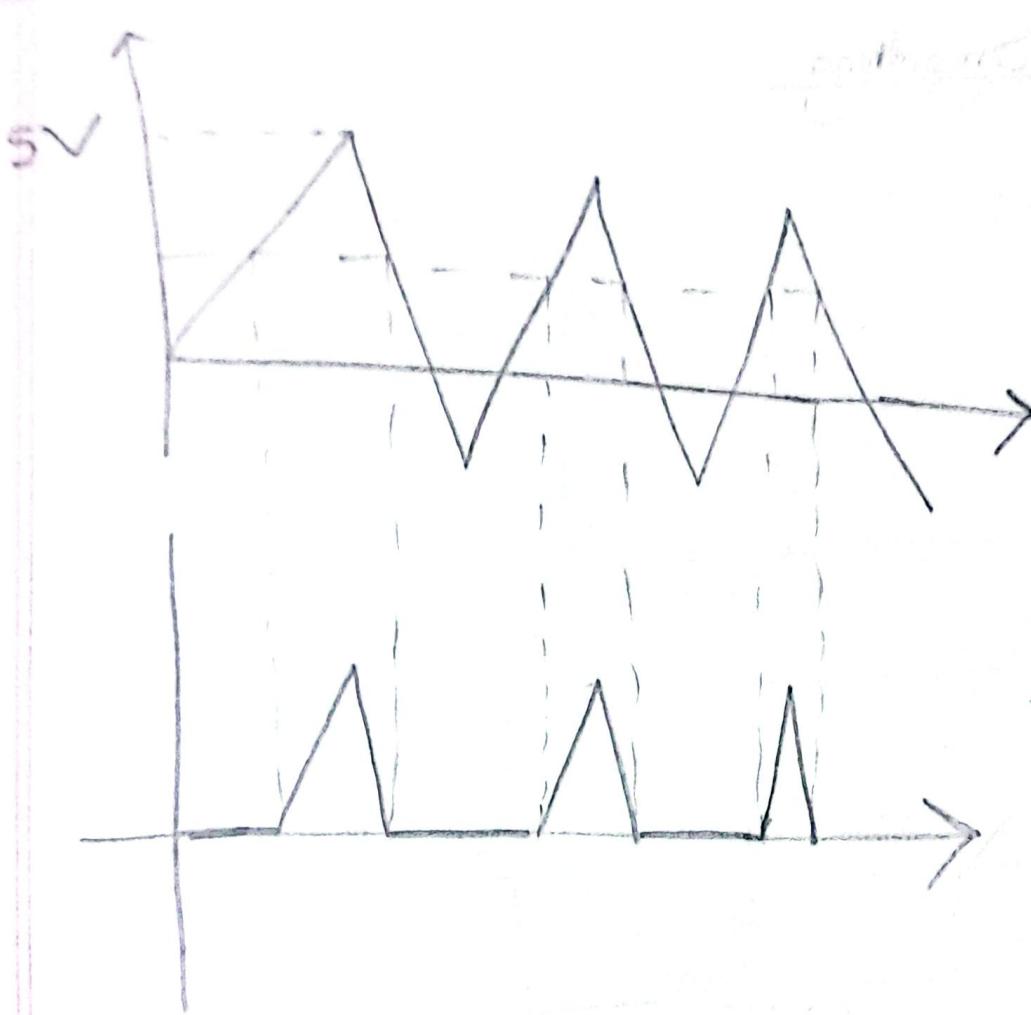


VTC graph for full wave Rectifier.



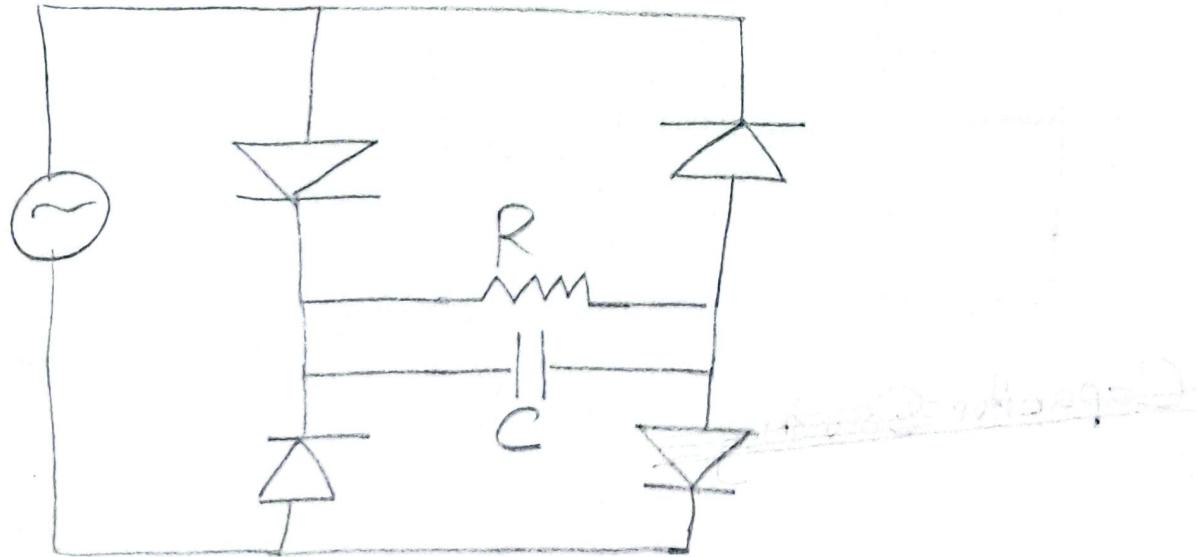
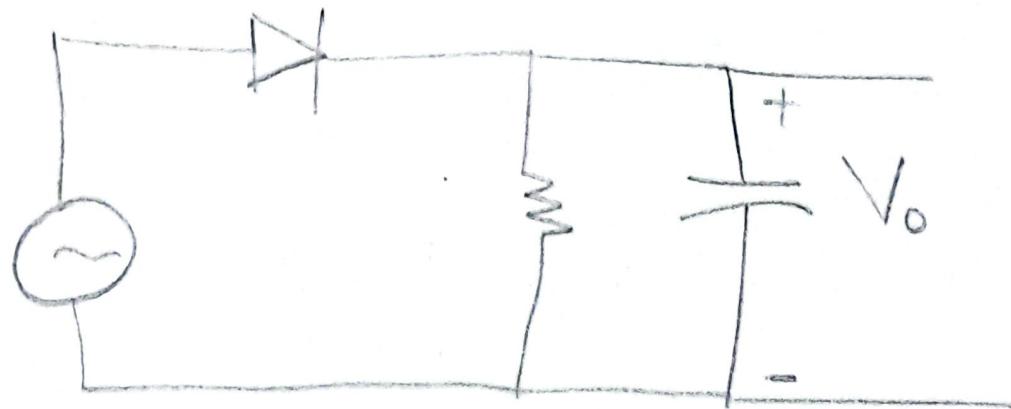
avg value, $V_{avg} = \frac{1}{T} \int_{0}^{T} V_o dt$

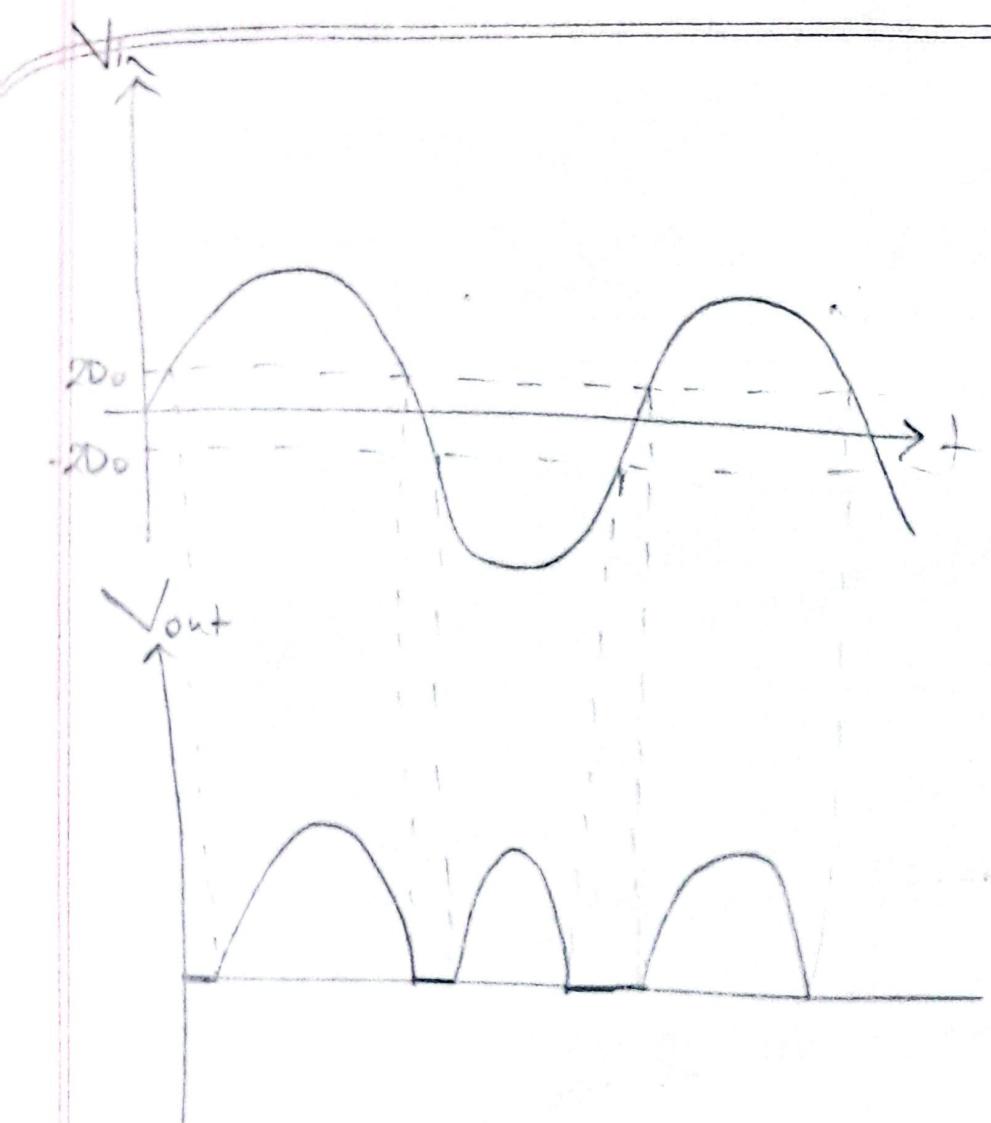
$= \frac{2V_m}{\pi} - 2V_0$

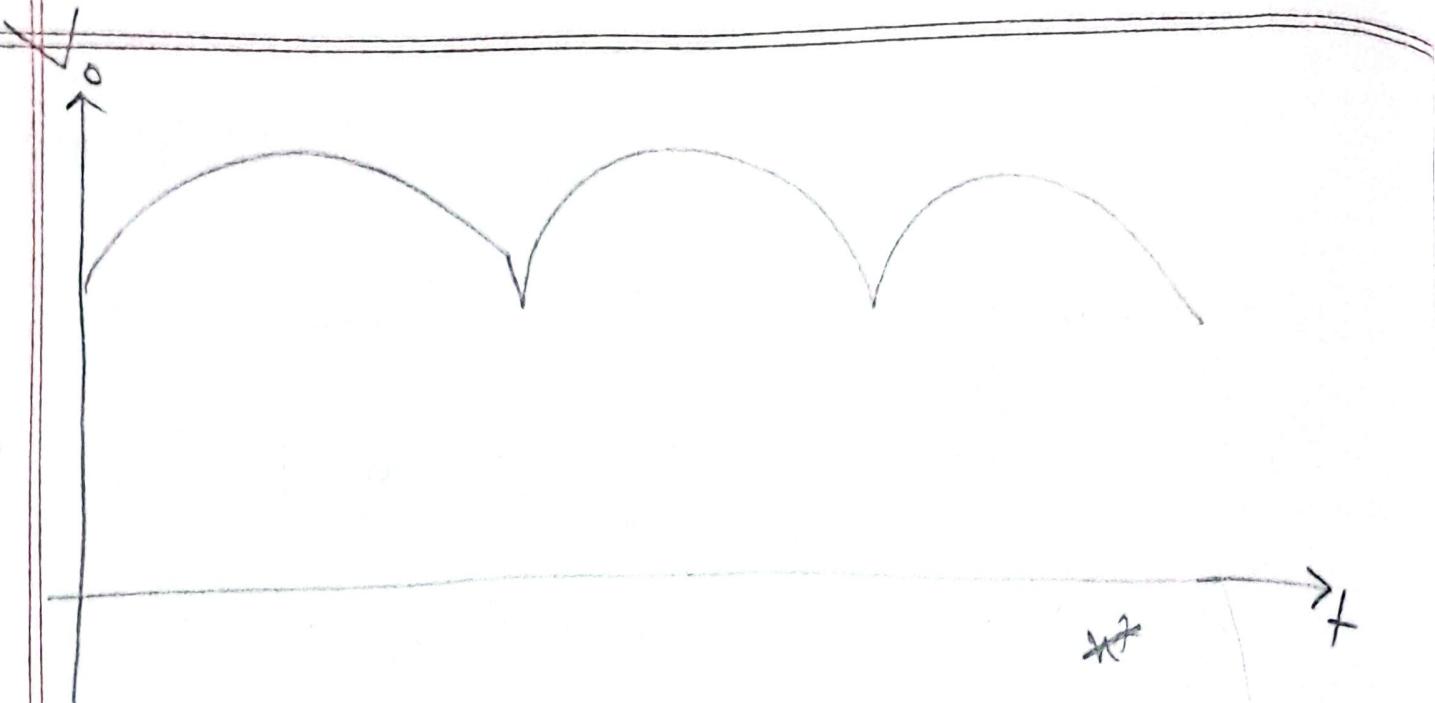


Capacitor Smoothing

Capacitor Smoothing







With capacitor

	Half-Wave	Full Wave
Output Peak (V _p)	$V_m - V_{D0}$	$V_m - 2V_{D0}$
Ripple Frequency (f ₂ (f ₁))	f_1	$2f_1$ (This is output frequency)
Ripple (peak-to-peak)	$\frac{V_p}{f_2 \cdot R \cdot C}$	$\frac{V_p}{2f_1 \cdot R \cdot C}$
Ripple, rms	$\frac{V_p(p-p)}{2\sqrt{3}}$	$\frac{V_p(p-p)}{2\sqrt{3}}$
V _{oc}	$V_p - \frac{V_{D0}(p-p)}{2}$	$V_p - \frac{V_{D0}(p-p)}{2}$

100 200

100 200

Without Capacitor

	Half-Wave	Full-Wave
Peak Value V_p	$V_m - V_{D0}$	$V_m - 2V_{D0}$
Average Value V_{avg}/N_{dc}	$\frac{V_m}{\pi} - \frac{V_{D0}}{2}$	$\frac{2V_m}{\pi} - \frac{2V_{D0}}{2}$

V_m = input peak

Past Midterm questionsFall-2022

4) $V_i = 8 \sin(100\pi t)$ $V_m = 8$

$$R = 10 \times 10^3$$

$$V_{D0} = 0.8V$$

$$f_0 = 100 \text{ Hz}$$

a) ^{Full} Half-wave rectifier is used

b) SKIP

c) Draw the two graph

3) $V_{DC} = \frac{2V_m}{\pi} - \frac{V_{D0}}{2} - 2V_{os}$

$$= \frac{8}{\pi} - \frac{0.8}{2}$$

$$= 2.16 - 0.4V = 1.76V$$

$$V_p = V_m - V_{D0} = 8 - 0.8 = 7.2V$$

$$V_{r(p-p)} = \frac{5}{100} \times 7.2$$

$$= 0.36V \text{ or } 0.32V$$

$$e) V_{r(p-p)} = \frac{2\sqrt{p}}{f_R \cdot R \cdot C}$$

$$0.36 = \frac{7.2 \times 2 \times 6.4}{100 \times 10,000 \times C}$$

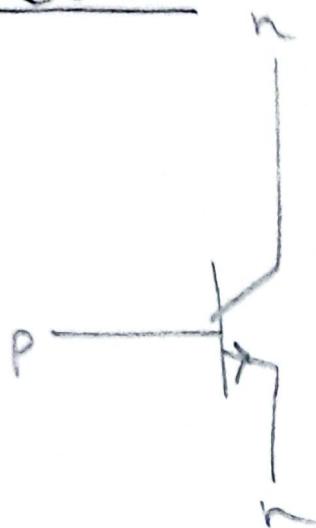
$$C = 2 \times 10^{-5} F$$

$$= 20 \mu F$$

$$f) V_{dc} = V_p - \frac{V_{r(p-p)}}{2}$$

$$= 7.2 - \frac{0.362}{2}$$

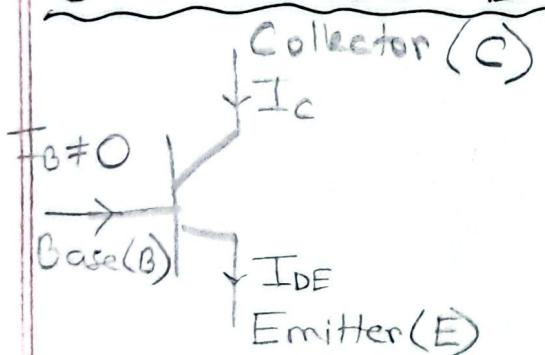
$$= 7.02624V$$

Lecture-15BJT

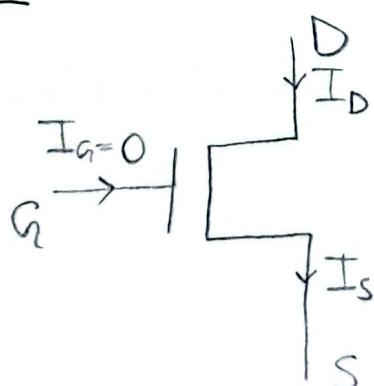
npn-BJT



pnp-BJT

BJT vs MOSFET

$$I_B + I_C = I_E$$



Since $I_G = 0$,

$$I_D = I_S = I_{DS}$$

- MOSFET is voltage controlled and BJT is current controlled.

MOSFET vs BJT (I-V graph)

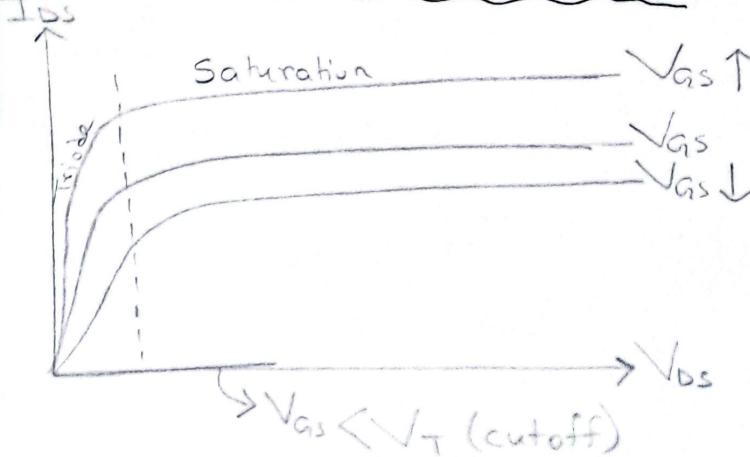
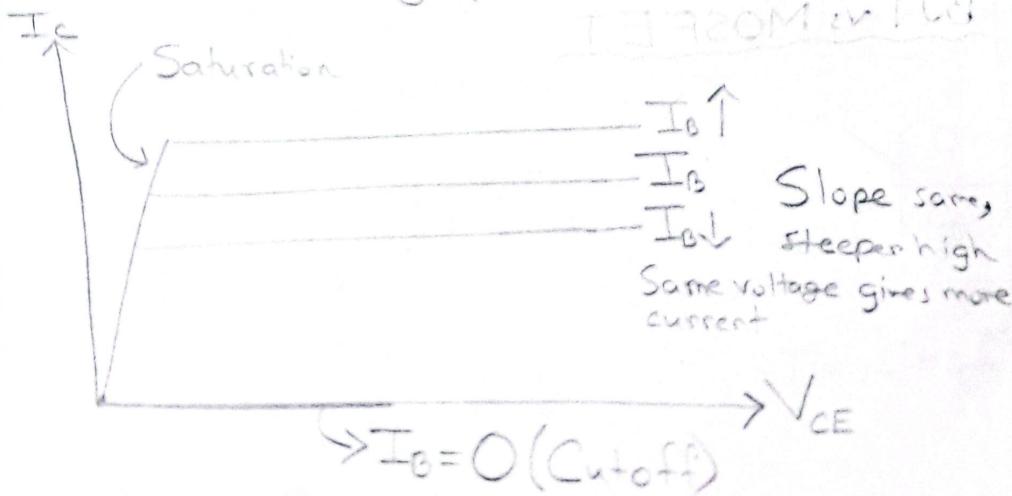
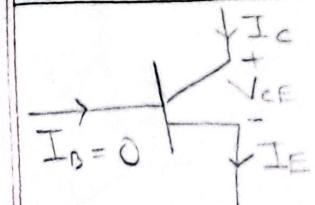


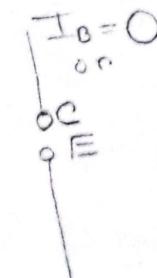
Fig: MOSFET graph



BJT-S Model



When $I_C < I_{Th}$

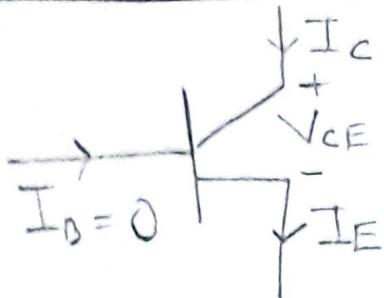


$I_C = 0$

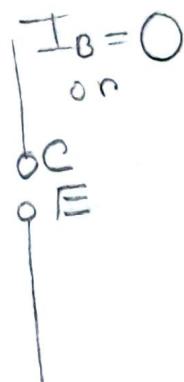
When $I_C > I_{Th}$



BJT-S Model

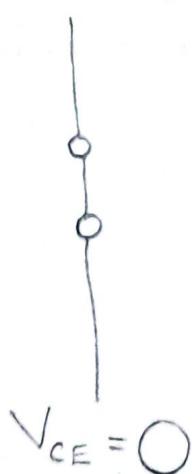


When $I_B < I_{Th}$



$$I_C = 0$$

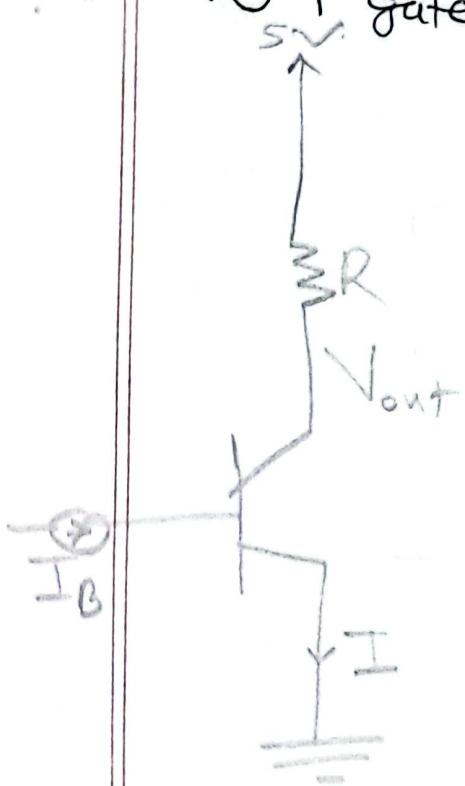
When $I_B > I_{Th}$ (0.5mA)



Logic Function with BJT

Lab M2 - FG8

1) NOT gate

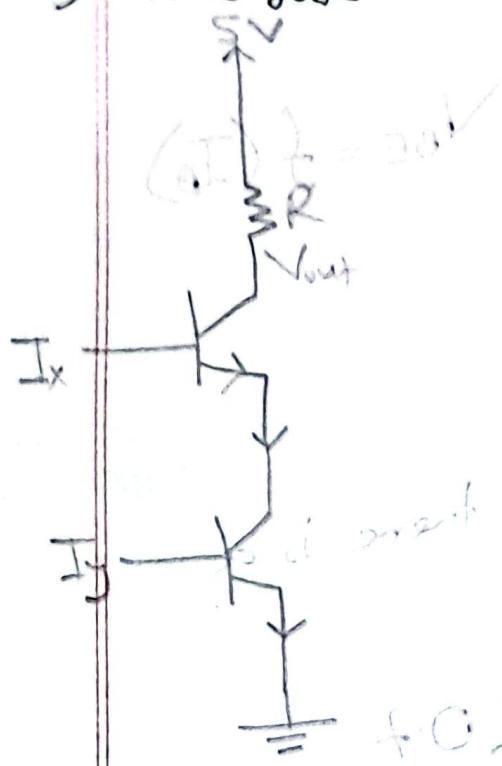


$$I_C > \alpha I_B \text{ and } I_C$$

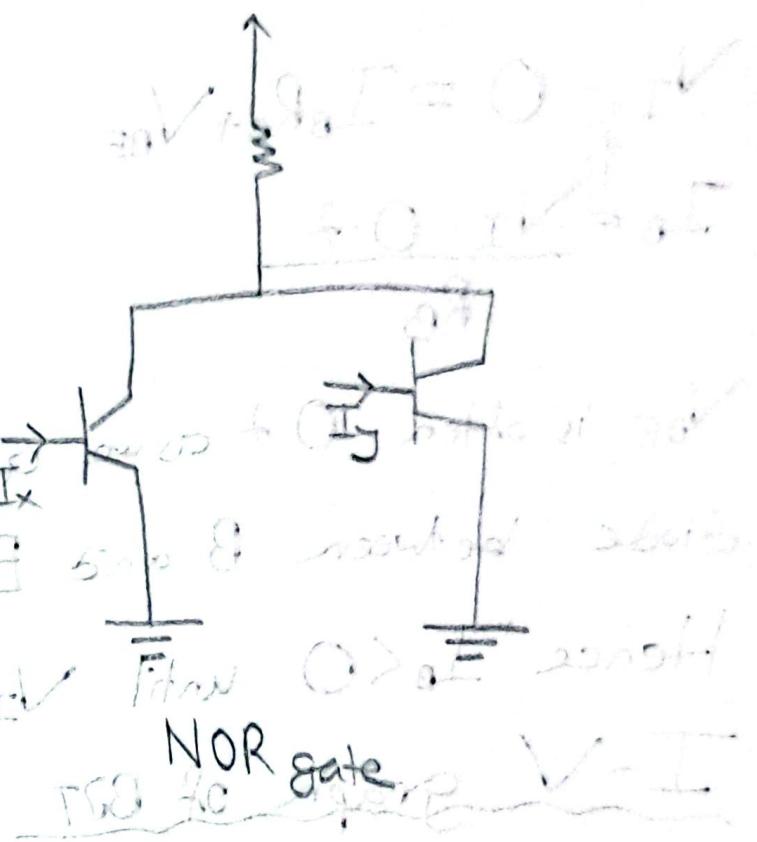
NOT gate with BJT/inverter

(and 0) $I_C < I_B$ when

2) NAND gate



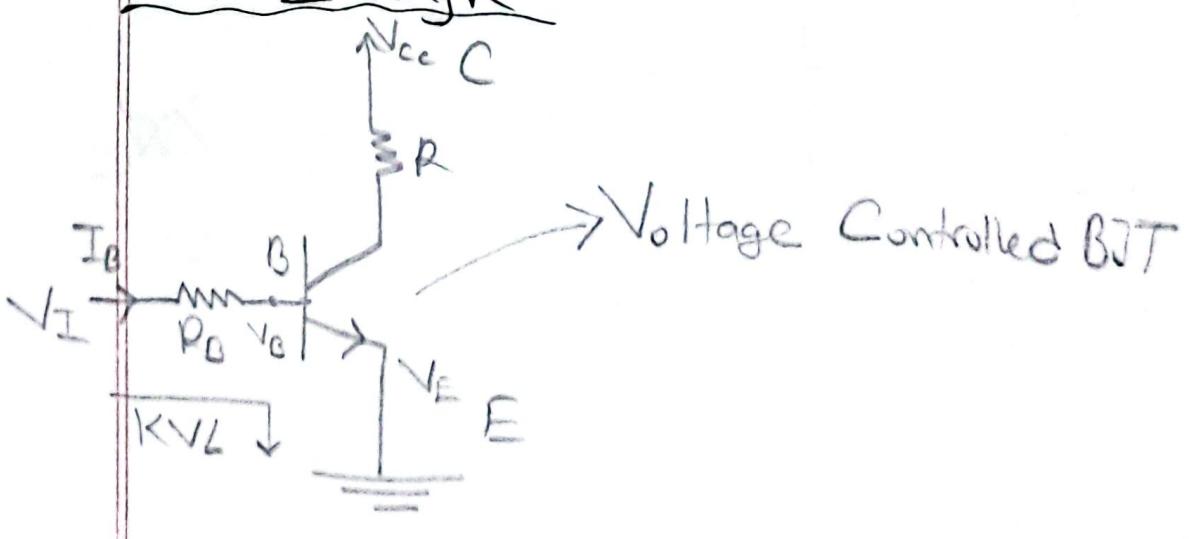
3) NOR gate



Drawback of BJT

- 1) We cannot cascade BJT directly
- 2) No current source exist to drive BJT

V to I using R



KVL:

~~using ACME~~

using ACME

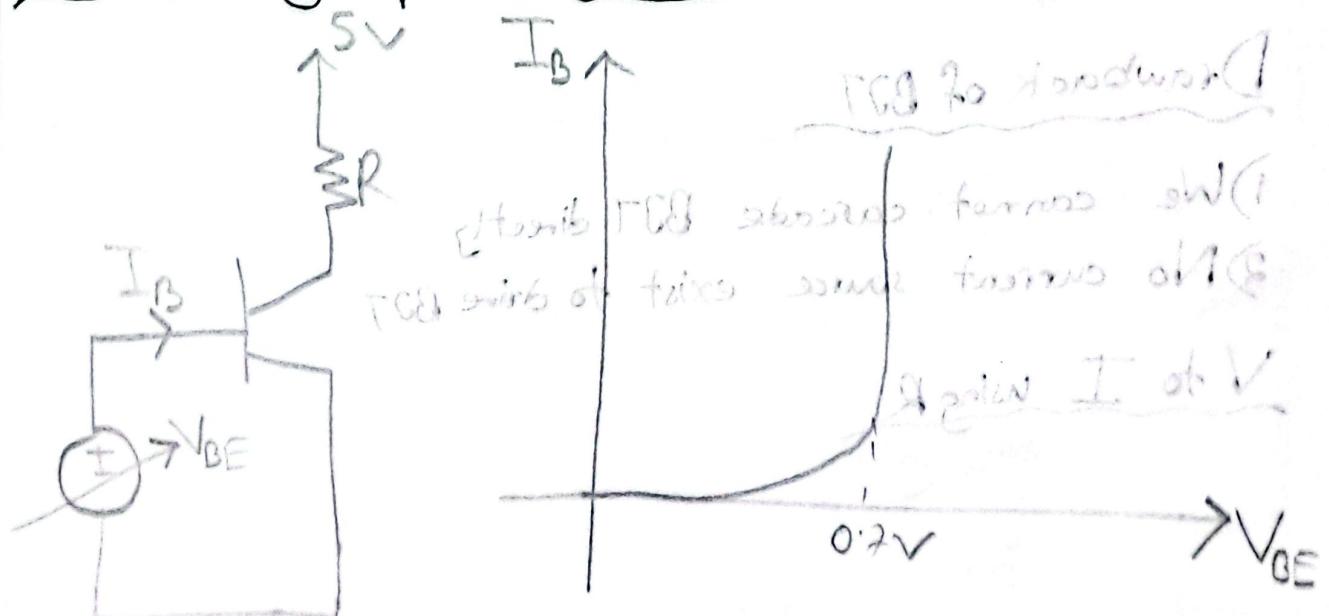
$$V_I - 0 = I_B R_B + V_{BE}$$

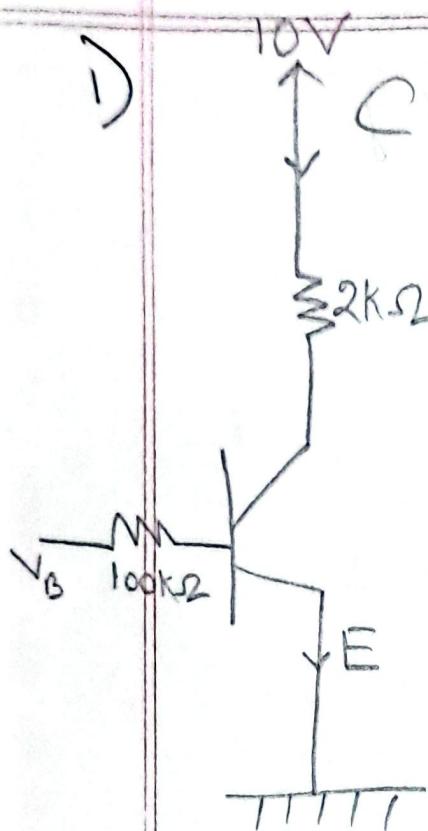
$$V_{BE} = f(I_B)$$

$$I_B = \frac{V_I - 0.7}{R_B}$$

- V_{BE} is often 0.7 as we assume there is a diode between B and E.
- Hence $I_B < 0$ until $V_I \geq 0.7$

I-V graph of BJT





(3m)

Given,

$$I_{Th} = 0.2\text{mA}$$

Q2) Find V_{In} for $I_B \geq I_{Th}$

If BJT is ON,

$$V_{BE} = 0.7$$

$$V_B = I_B R_B + 0.7$$

$$\begin{aligned}
 V_B &= (0.2 \times 100) + 0.7 \\
 &= 20 + 0.7 \\
 &= 20.7
 \end{aligned}$$

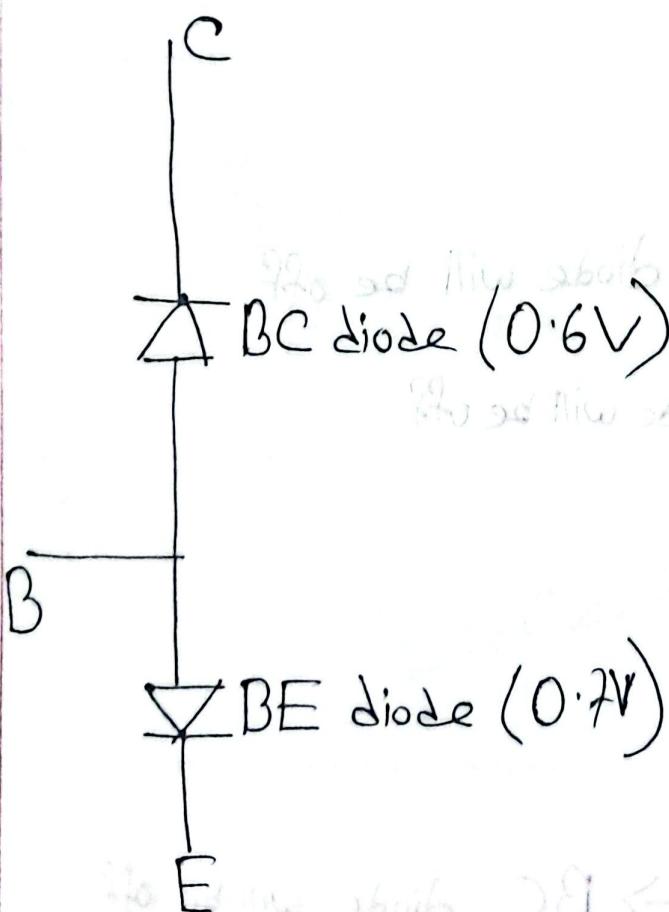
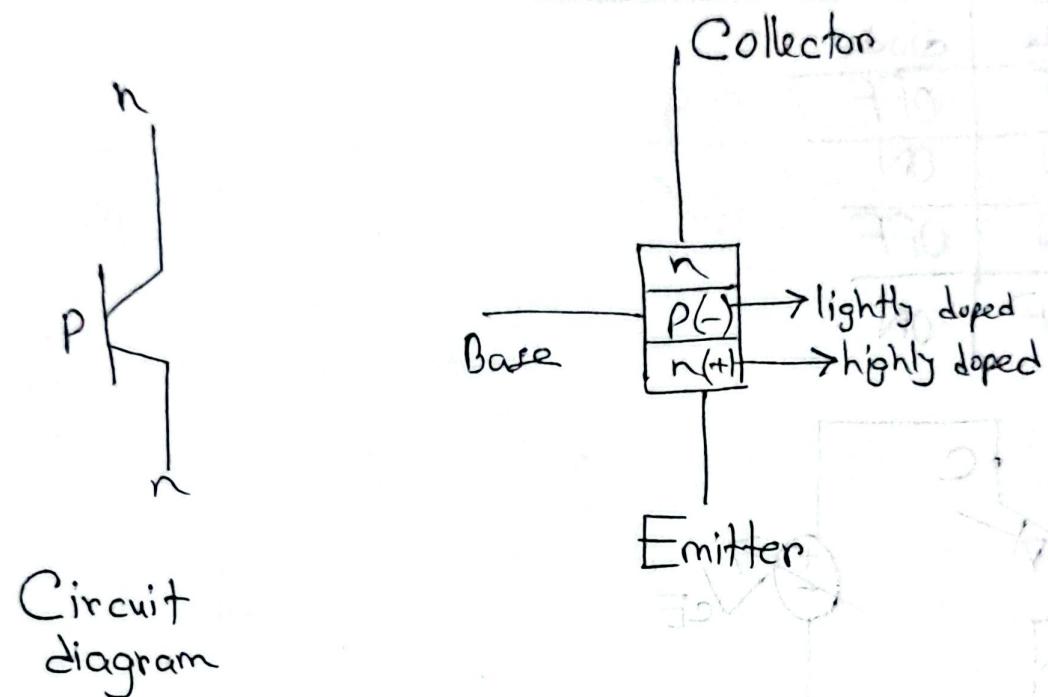
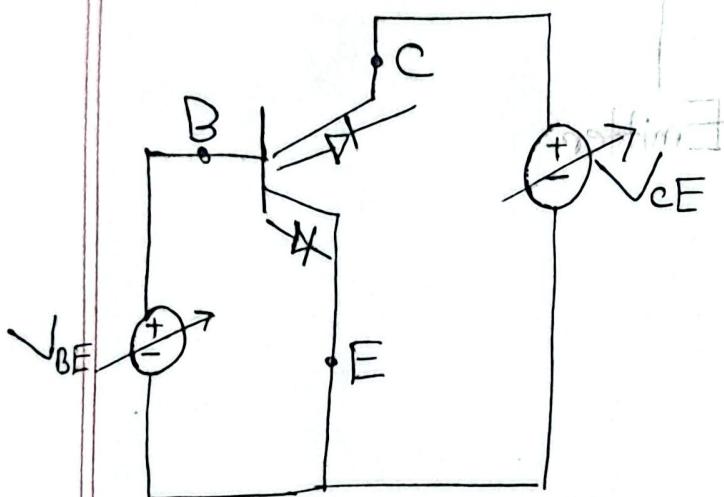
lecture-16

Fig: Diode Equivalent Circuit

	BE diode	BC diode
Cutoff	OFF	OFF
Saturation	ON	ON
Active	ON	OFF
Reverse- Active	OFF	ON



$V_{BE} < 0.7$ then BE diode will be off

$V_{CE} < 0.7$ then CE diode will be off

$$V_{BC} < 0.7$$

$$(V_B - V_C) < 0.6$$

$$(V_B - V_E) - (V_C - V_E)$$

If

$$V_{CE} > V_{BE} - 0.6 \rightarrow BC \text{ diode will be off}$$

1) Cutoff

BE diode off

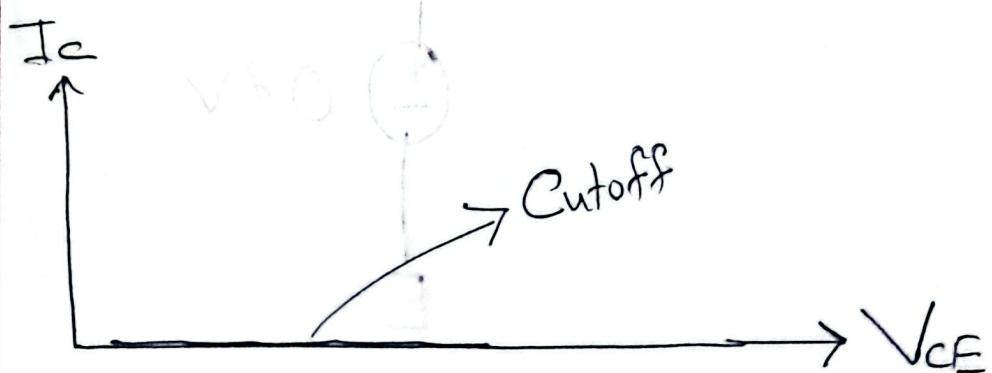
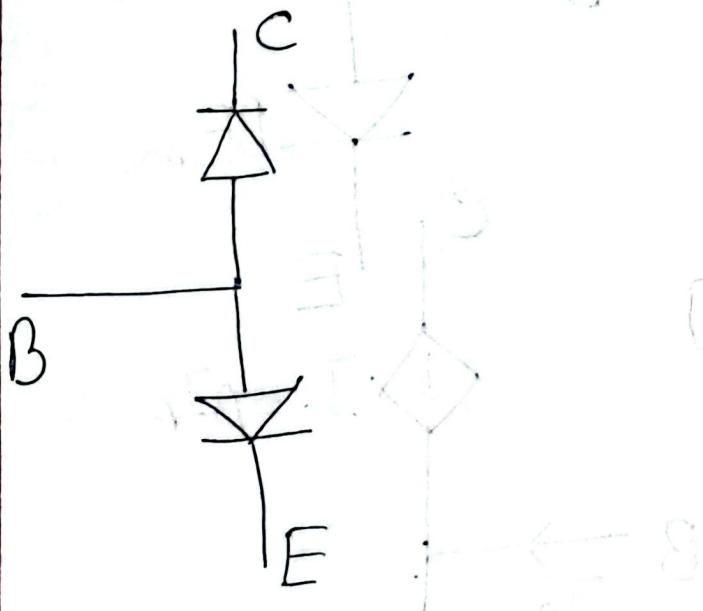
$$V_{BE} < 0.7$$

BC diode off

$$V_{CE} > V_{BE} - 0.6$$

Condition for cutoff,

$$I_B = I_C = I_E = 0$$



ii) Active

BE ON

$I_B > 0$

BC off

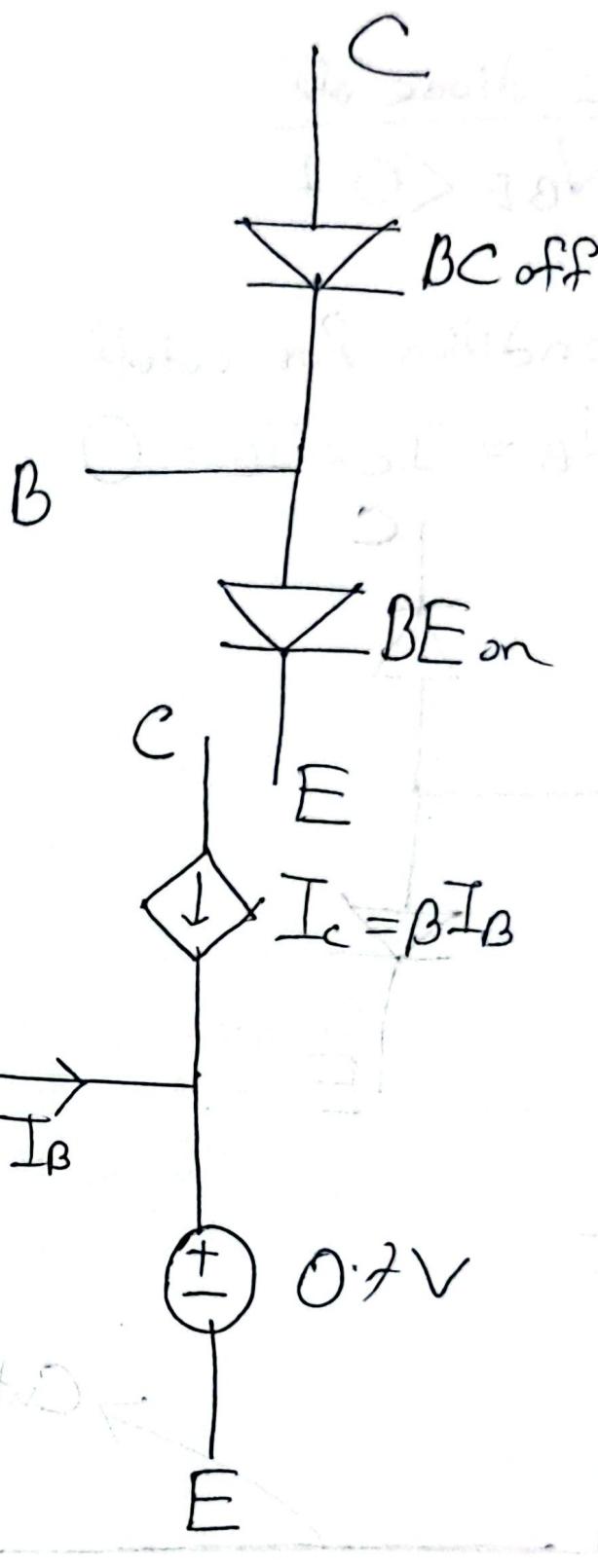
$V_{CE} > V_{BE} - 0.6$

In active mode

$$I_C = \beta I_B$$

$\beta = \text{constant } [50 \sim 200]$

$\beta \sim 80$



For all modes,

$$I_E = I_B + I_O$$

For active mode,

$$I_E = I_B + \beta I_B$$

$$I_E = I_B(1 + \beta)$$

~~$$I_E$$~~

$$I_E = \frac{I_C}{\beta} + I_C$$

$$I_E = \left(1 + \frac{1}{\beta}\right) I_C$$

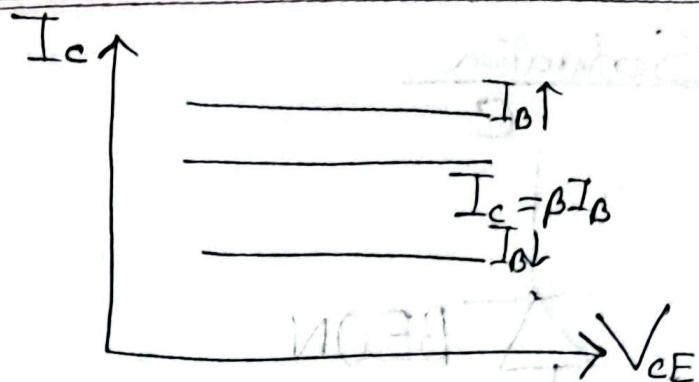
$$I_E = \frac{1 + \beta}{\beta} I_C$$

$$\boxed{\alpha = \frac{\beta}{1 + \beta}}$$
$$I_C = \alpha I_E$$

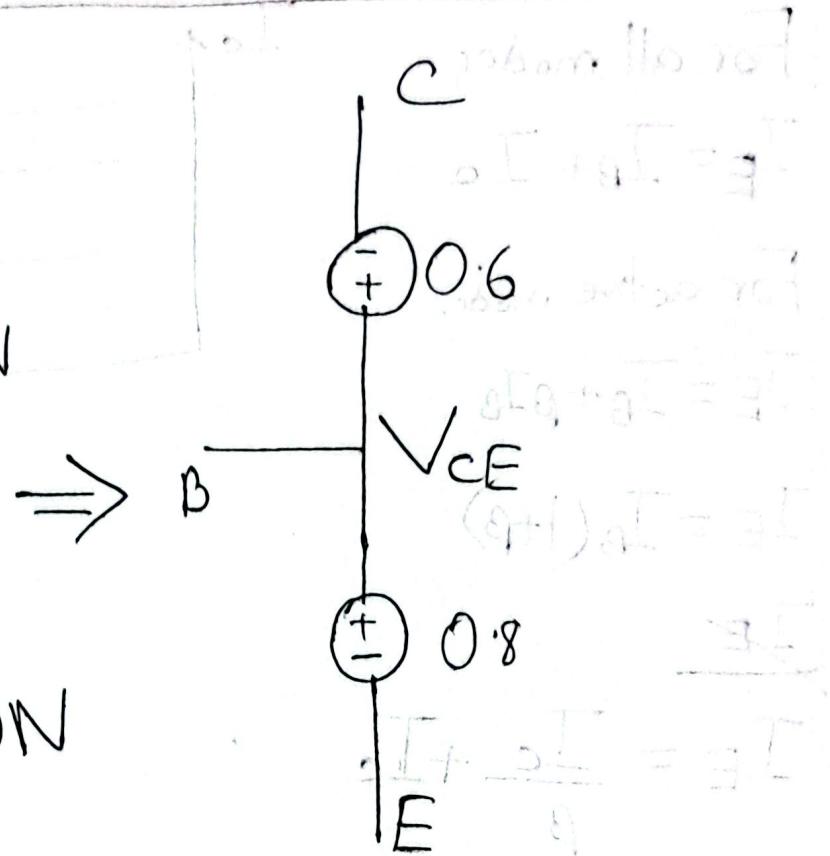
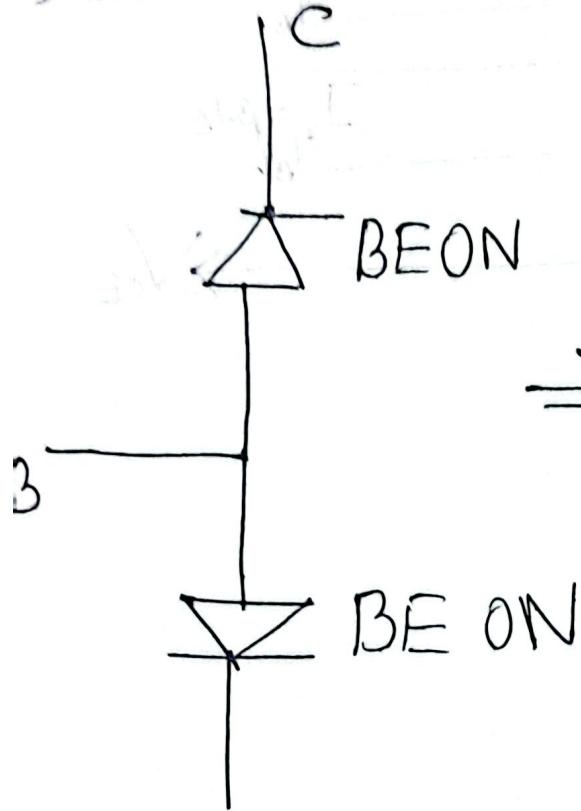
$$\text{If } \beta = 100$$

$$I_C = 100 I_B$$

$$I_E = 0.99 I_C$$



iii) Saturation



$$\begin{aligned}
 0.6 + 0.8 &= 1.4V \\
 V_{CE} &= V_C - V_E \\
 &= -0.6 + 0.8 \\
 &= 0.2V
 \end{aligned}$$

Cutoff

$$V_{BE} < V_{th}$$

Triode

$$V_{DS} < V_{ov}$$

Saturation

$$V_{DS} > V_{ov}$$

$$0.6 + 0.8 = 1.4V$$

$$0.6 + 0.8 = 1.4V$$

For Active,

$$\beta = \frac{I_c}{I_B}$$

For saturation,

$$\frac{I_c}{I_B} < \beta$$

Summary

State

Cutoff

Active

Saturation

Equation

$$I_c = I_B = I_E = 0$$

$$V_{BE} = 0.7$$

$$\frac{I_c}{I_B} = \beta$$

$$I_E = \alpha I_c$$

$$\alpha = \frac{\beta}{1+\beta}$$

$$V_{BE} = 0.8V$$

$$V_{CE} = 0.2V$$

Conditions

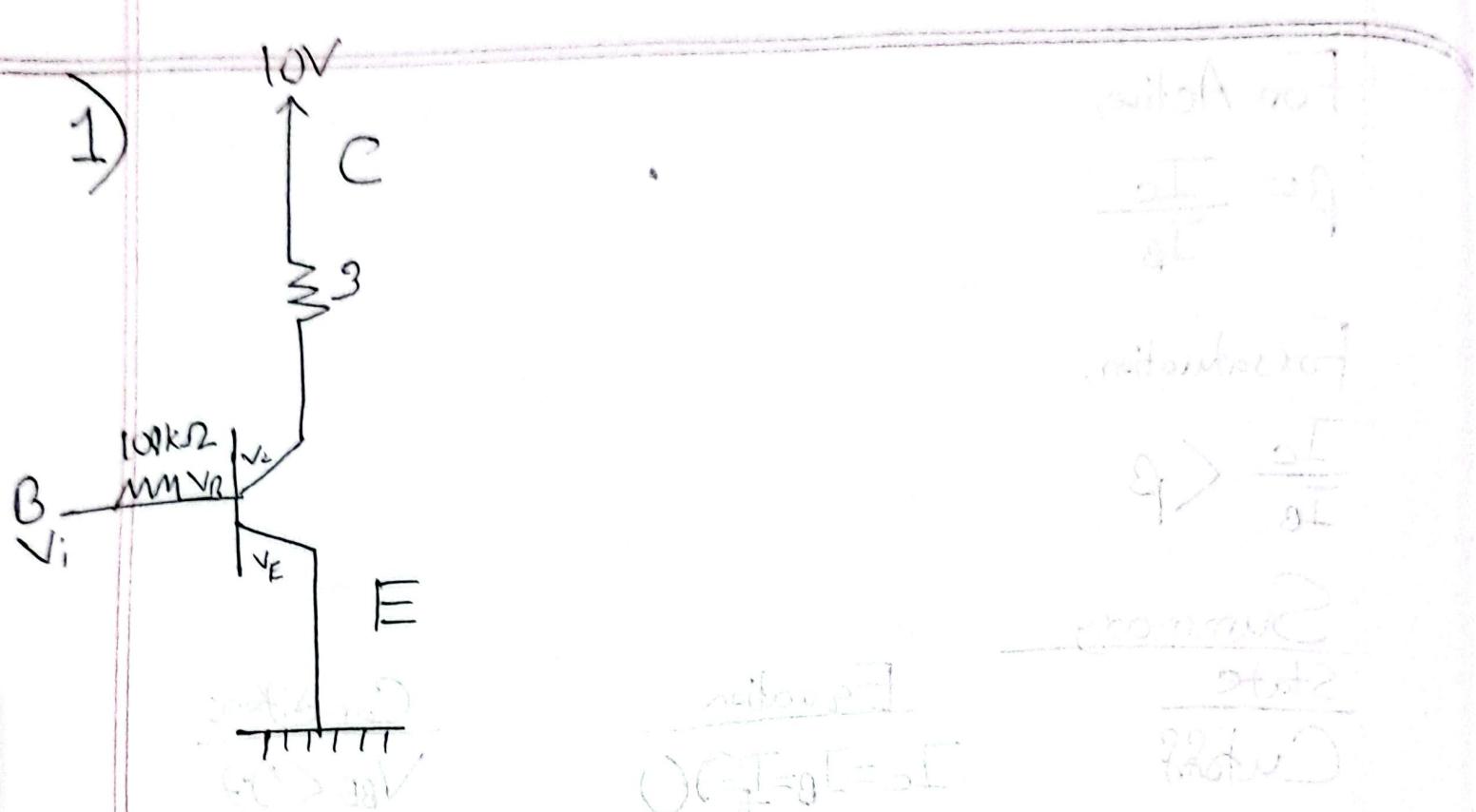
$$V_{BE} < 0.7$$

$$V_{CE} < 0.5$$

$$V_{CE} > 0.2V$$

$$\boxed{\begin{aligned} V_i &= I_B R_B + V_{BE} \\ I_c &= \beta I_B \\ I_E &= \beta I_B + I_B \end{aligned}}$$

$$\frac{I_c}{I_B} < \beta$$



$$V_{in} = 1V \quad \beta = 100$$

$$\therefore \cancel{V_B = 1V} \quad 6.0 = 30V$$

$$V_E = 0V$$

Assuming Cutoff,

$$V_B = 1V \quad (\text{Since, } i_B = 0)$$

$$V_C = 10V \quad (\text{Since } i_C = 0)$$

$$V_E = 0V \quad (\text{Since ground})$$

$$V_{BE} = V_B - V_E \\ = 1V > 0.7$$

$$V_{BE} = 30V \\ V_{CE} = 30V$$

$$V_{BC} = V_B - V_C$$

$$= 1 - 10$$

$$= -9V < 0.5$$

Since $V_{BE} > 0.7$ the BJT is not in cutoff

Assuming Active mode

$$V_{CE} > 0.7V$$

$$V_{BE} = 0.7V$$

$$V_B - V_E = 0.7$$

$$V_B - 0 = 0.7$$

$$V_B = 0.7V$$

$$I_C = \beta I_B$$

$$I_B = \frac{V_B - V_{in}}{100}$$

$$= \frac{1 - 0.7}{100}$$

$$= 0.003mA$$

$$I_C = 0.3mA$$

$$I_E = 0.99 \times 0.3$$

$$= 0.297mA$$

182 320

1.652mA

$$\frac{10 - V_C}{3} = 0.3$$

$$V_C = 9.1V$$

$$V_{CE} = 9.1V$$

Since $V_{CE} > 0.2$, BJT is in active mode