

# 802.11 with Multiple Antennas for Dummies



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# Wireless Channels

- Limited Capacity

- The max performance is determined by Shannon's Limit
  - **C**: Capacity (bit/sec/Hz)
  - **B**: Bandwidth (20 Mhz in 802.11a/g)
  - **S/N**: Signal to Noise Ratio in power
- Doubling of signal power yields in + 1 bit/sec/Hz
- The key for speed had lied in **SNR**

$$C = B \log_2 \left( 1 + \frac{S}{N} \right)$$

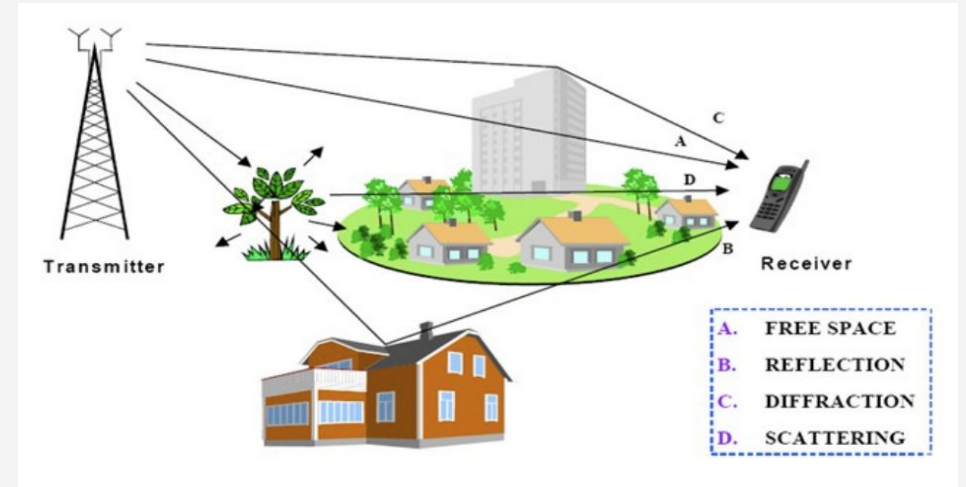
- Attenuation (Weakening)

- Slow fading
  - Gradual fluctuations of received power over long periods of time
  - *ex) path loss, shadowing*
- Fast fading
  - Large swings in received power over relatively short periods
  - *ex) multi-path, doppler effect*

# Attenuation

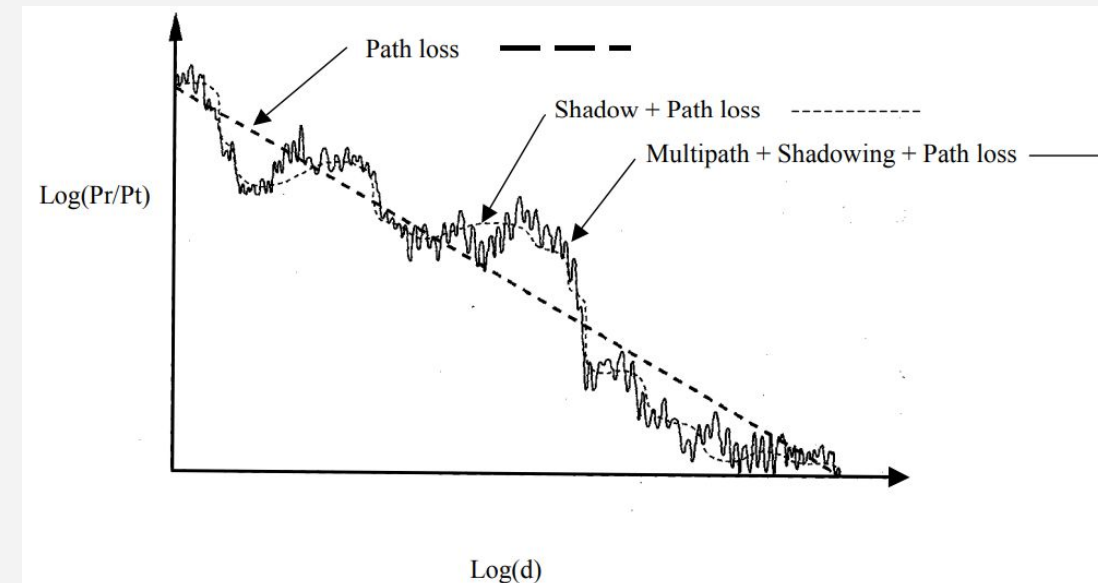
- Multi-Path

- Multi-path effects depend on the phases of signals
  - Some unlucky frequencies in 20 MHz may be wiped out  
→ Frequency Selective
- Independent for locations separated by half a wavelength
  - Half a wavelength = 180 degrees of shift = 6cm for 2.4Ghz
  - Ref. Clarke's fading model



- Attenuated Signal

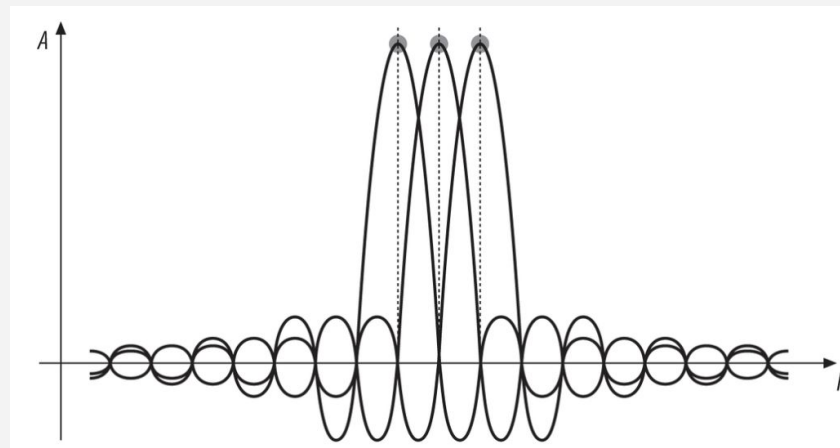
- Slow fading + fast fading
- Signal fluctuates in effect of multipath and doppler



# OFDM

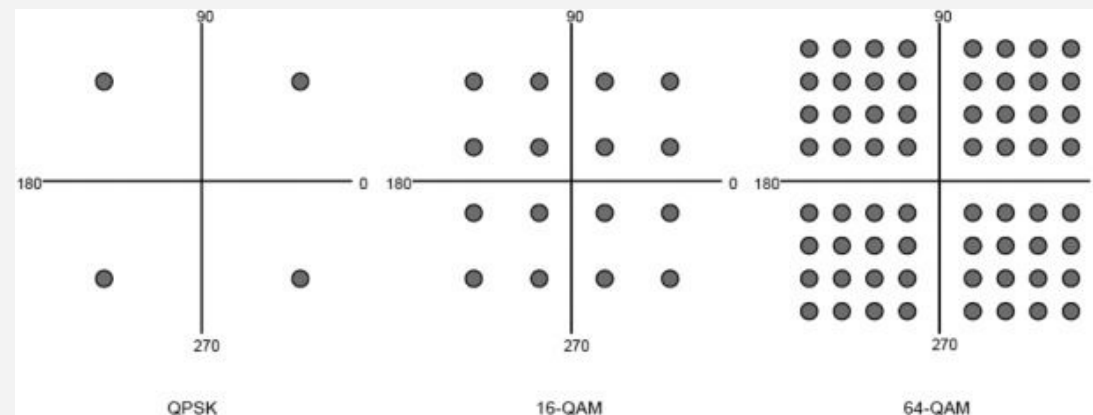
- Orthogonal Frequency Division Modulation

- 20 Mhz into 64 subcarriers
  - 48 subcarriers for data transfer
  - 4 subcarriers for pilot tones
  - 12 subcarriers for guard (unused)
- Subcarriers are orthogonally divided in frequency domain
  - All other subcarriers do not contribute to the carrier's waveform (amplitude)
- OFDM benefits from frequency, time and spatial *diversity*
  - *Diversity*: spreading of information with redundancy



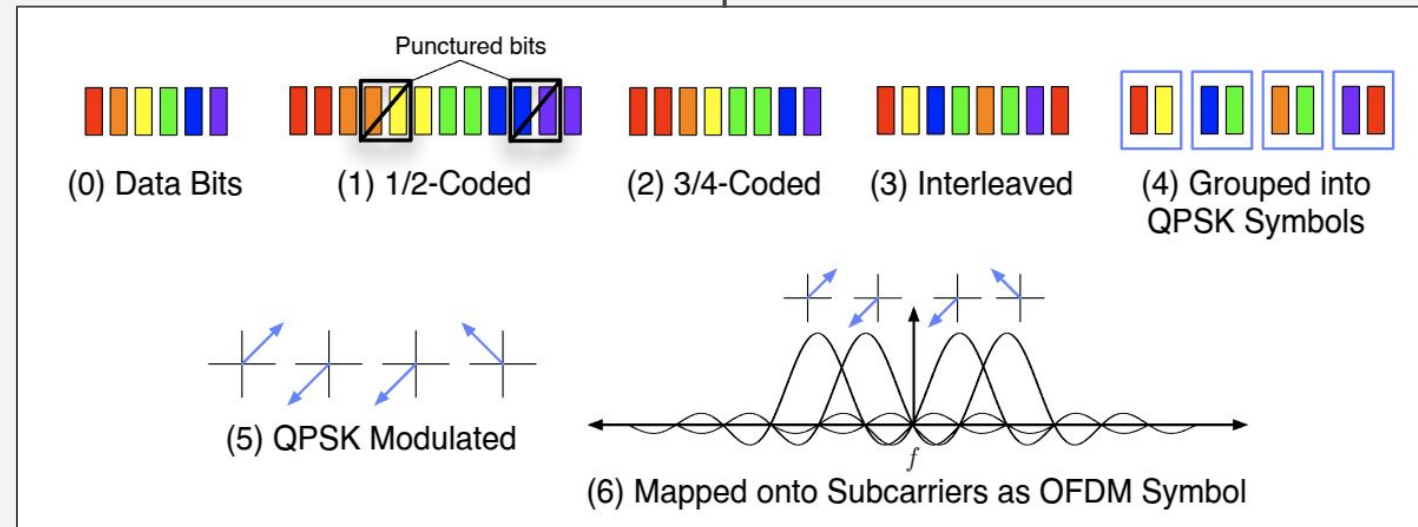
- Quadrature Amplitude Modulation (QAM)

- M-ary bit modulation scheme in IQ\* plane
  - ex) 16-QAM(4 bits), 32-QAM(5 bits), ...
- Each symbol is carried to a subcarrier



# Diversity Benefits of OFDM

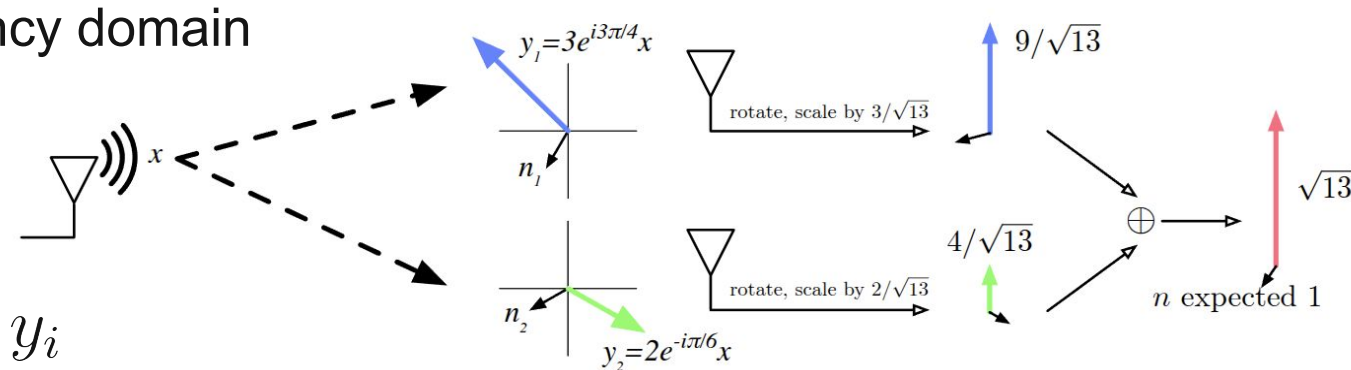
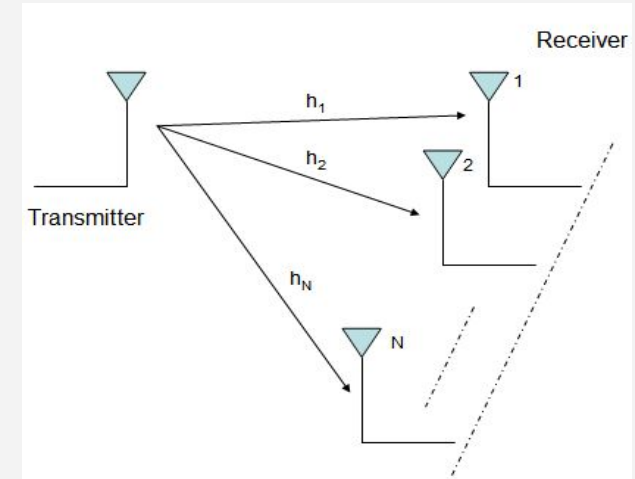
- Frequency diversity
  - Different frequencies, different fading  
→ *Frequency selective*
  - Spread data bits into multiple subcarriers
    - **Coding scheme:** adds redundant bits by coding rate R
    - **Interleaving:** spread the redundancy across subcarriers
  - Less likely dropping a frame out of deep fade
- Time diversity
  - Slower symbols in parallel
    - 312.5 kHz per symbol
      - $T_{SYM} = \frac{1}{312.5k} = 3.2\mu s$
  - Instead of fast symbols on a wideband  
→ Less likely to be distorted by quick change



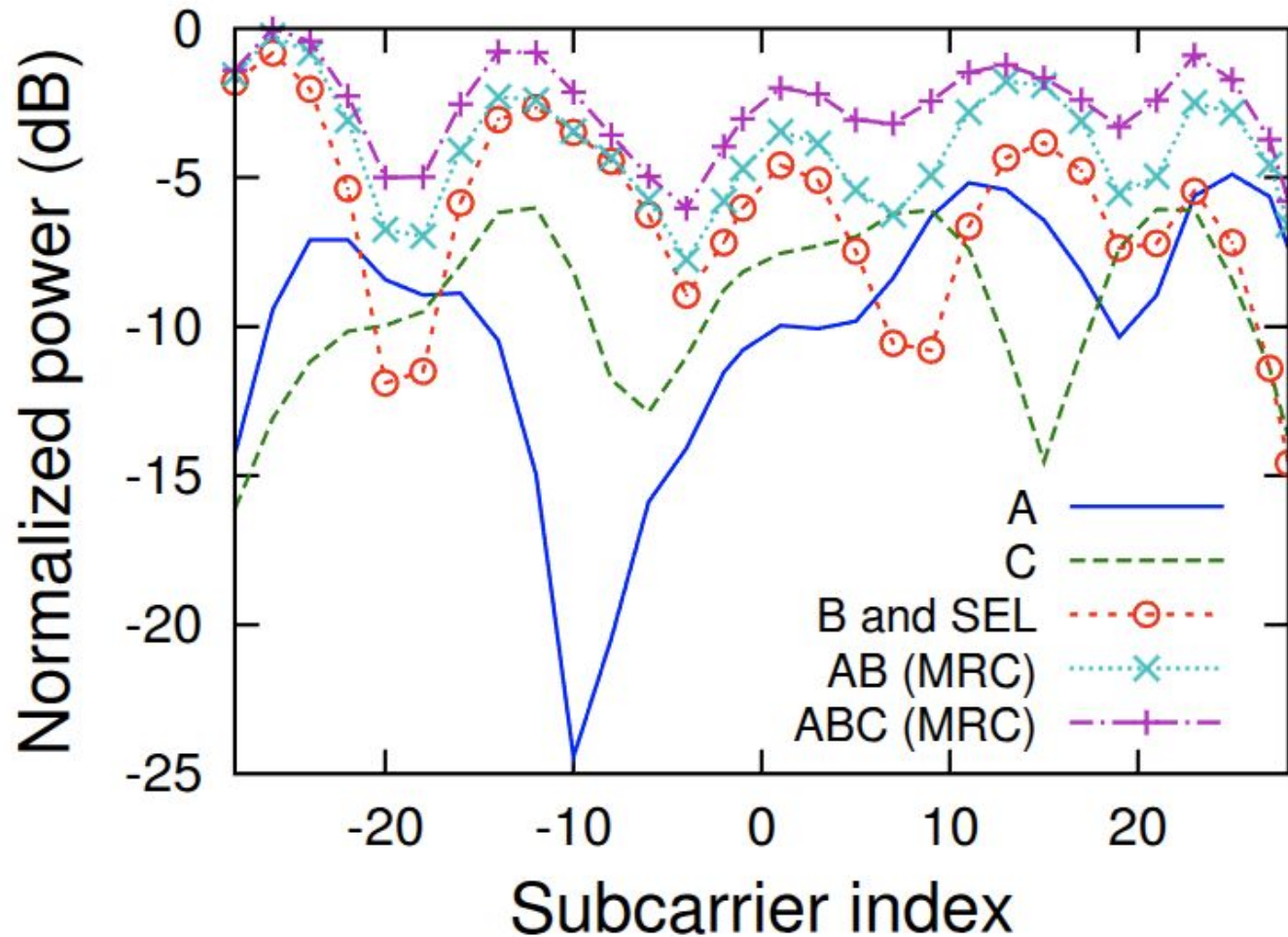
# Spatial Diversity - Receiver

- Selection Combining (SEL)
  - Select the antenna with the strongest SNR\*
    - all the other antennas are ignored
  - Standard method of 802.11a/g APs
- Maximum Ratio Combining (MRC)
  - 1. Estimate the channel gains  $H$  in frequency domain
    - By computing from preamble's training field
  - 2. Rotate and scale the received signals  $Y$ 
    - By applying the complex conjugate of  $h$
  - 3. Sum up the aligned channel responses  $y_i$

→ MRC is known to be optimal to maximize SIMO capacity



# Comparison: SEL and MRC

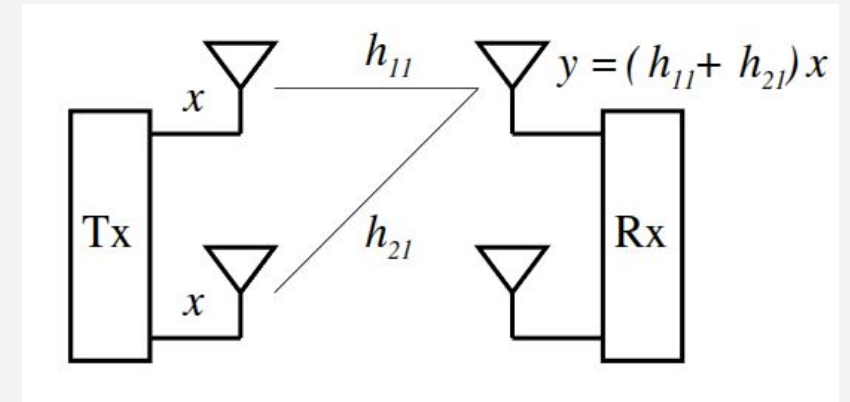


- Antenna A has the worst fluctuations
  - The signal variation of 20 dB
- Antenna B is chosen in SEL
  - The signal variation of 15 dB
- Weighted sum of signals in MRC
  - The signal variation of 5 dB



# Spatial Diversity - Transmitter

- Selection Combining (SEL)
  - Select the antenna with the strongest SNR\*
- Maximum Ratio Combining (MRC)
  - Channel state feedback
    - Transmitter must know channel's state beforehand
      - Alternatively, transmitter can learn the channel gains when it receives a packet
  - Precoding
    - Phase adjustment (based on phase shift of channel)
    - Amplitude weighting (based on SNR)
  - Beamforming
    - The signals combine constructively at the receiver's antenna



# Direct-Mapped MIMO

- Direct-Mapped
  - Transmitter is blind to its channel  $\rightarrow$  directly sends data stream without precoding
- Zero Forcing (ZF)
  - Eliminate interference  $H$  to zero
  - $H$  is not always invertible in practice
    - $|H| \rightarrow 0$ , then  $H^{-1} \rightarrow \infty$   
 $\rightarrow$  **Noise amplification** problem
- Minimum Mean Squared Error (MMSE)
  - Find  $H$  that minimizes MSE between  $\vec{x}$  and  $W\vec{y}$ 
    - Minimizing the total error of interference and noise together
  - MMSE suppresses noise amplification problem

$$\vec{y} = H\vec{x} + \vec{n} \rightarrow H^{-1}\vec{y} = \vec{x} + \underbrace{H^{-1}\vec{n}}_{\text{noise}}$$

$$\vec{y} = H\vec{x} + \vec{n} \rightarrow \vec{x} \approx W\vec{y}$$

$$\operatorname{argmin}_W \text{MSE} = \mathbb{E} [\|\mathbf{x} - \mathbf{W}\mathbf{y}\|^2]$$



$$W = (H^H H + \sigma_n^2 \mathbf{I})^{-1} H^H$$

# Precoded MIMO

- Challenge of Direct-mapped
  - Paths are assumed to be uncorrelated → not often
    - LoS(Line of Sight) makes  $H$  correlated
      - Noise amplification problem
- Solution
  - Use Singular Value Decomposition for precoding
    - $U$  : unitary matrix (receiver-side rotation)
    - $S$  : diagonal matrix (scaling factor)
    - $V$  : unitary matrix (transmitter-side rotation)
      - Use  $V$  for precoding

$$H = USV^H$$

# Precoded MIMO (2)

- Operation

Transmitter:

$$\hat{x} = Vx$$



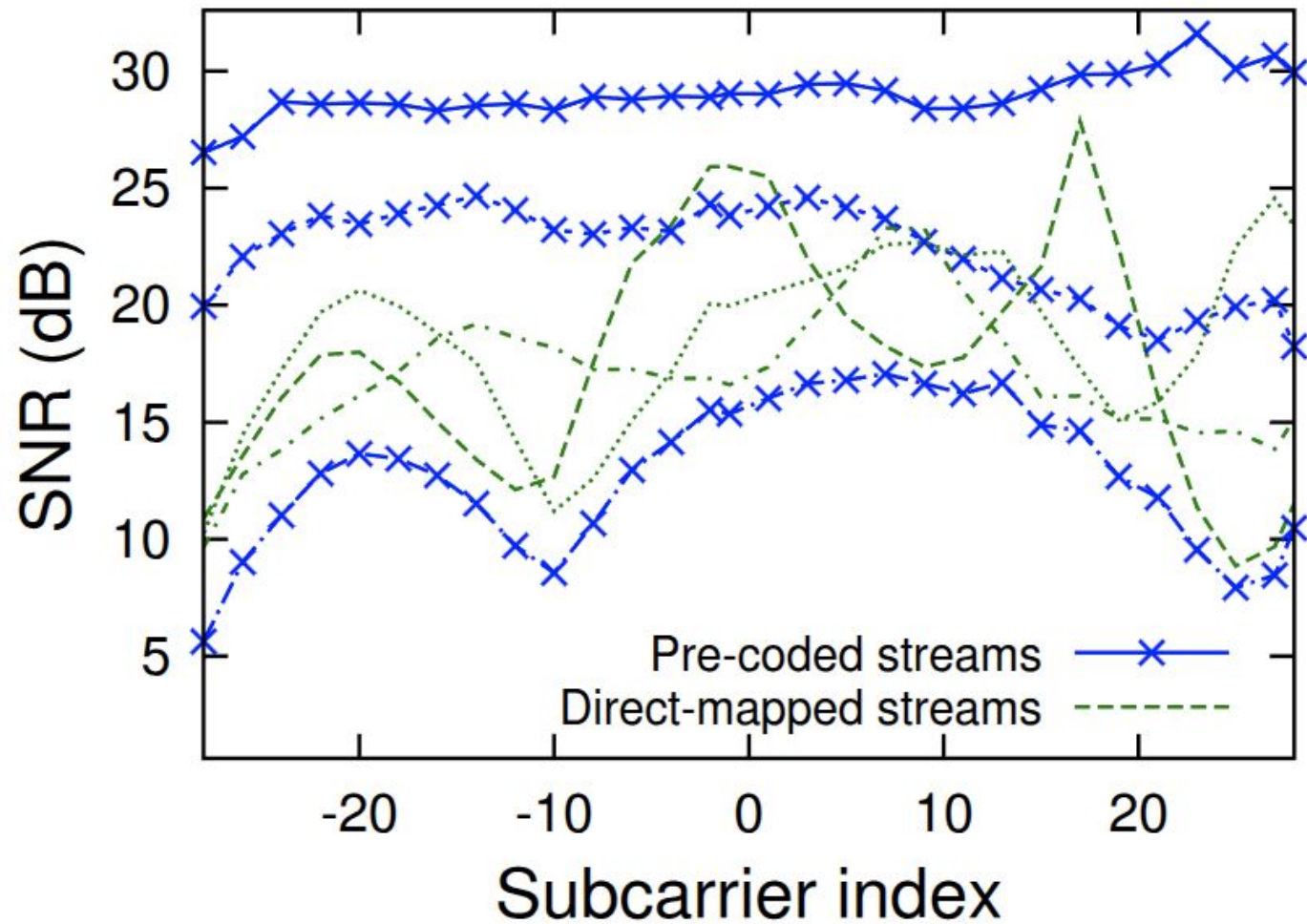
Channel:  $y = H\hat{x} + n = (US)x + n$



Receiver:  $\hat{y} = U^H y = Sx + U^H n$

- No noise amplification
  - $U$  is unitary matrix  $\rightarrow$  only rotates the noise

# Comparison: Direct-mapped and Precoded



# Doppler Effect

- Frequency/Phase Shift

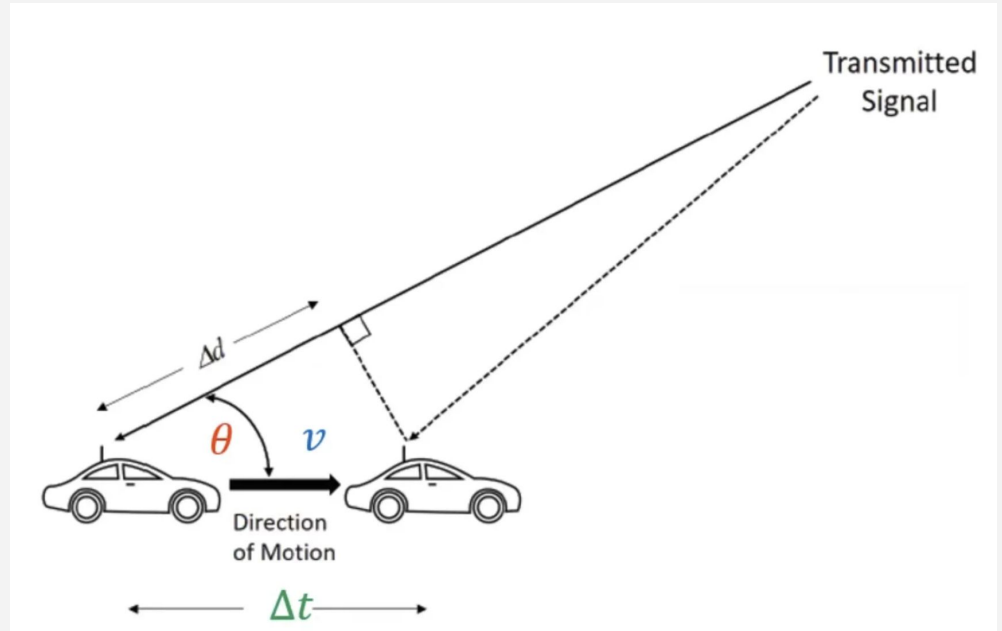
- $v$  : speed of the receiver (m/s)
- $\lambda$  : wavelength of the signal (m)
- $\theta$  : angle to direction of motion

- Example on OFDM

- 64 QAM at 2.4 GHz band
  - $\lambda = 0.125(m)$
  - $v = 100km/h \approx 30m/s$
  - $T_{SYM} = 4\mu s$ 
    - $\rightarrow f_D = 240Hz$
    - $\Delta\phi = 2\pi f_D T_{sym} \approx 0.006 \text{ rad } (\approx 0.34^\circ)$
- if 100 OFDM symbols, then 34 degrees of shift

$$f_D = \frac{v}{\lambda} \cos\theta$$

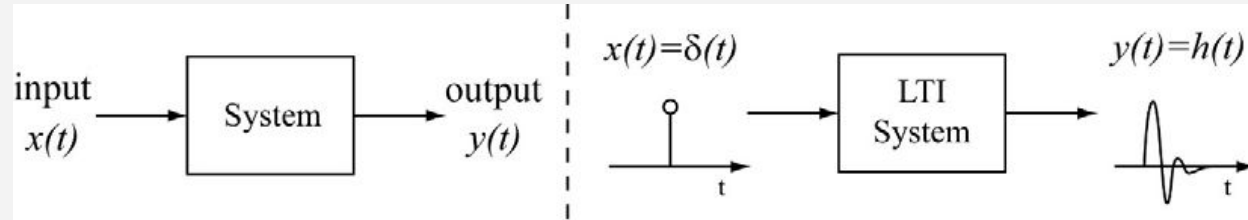
$$\left( \because \frac{\Delta\phi}{\Delta t} = w = 2\pi f_D, \quad \Delta\phi = \frac{2\pi}{\lambda} \Delta d \right)$$



# LTI System

## • Linear Time Invariant (LTI)

- Linear relationship between input and output
  - Additivity and homogeneity (scaling)
- The output is shifted as well as the input
  - $x(t - t_0) \xrightarrow{\text{system}} h(t - t_0)$
- The system corresponds to a channel in wireless communication



## • Impulse Response $h(x)$

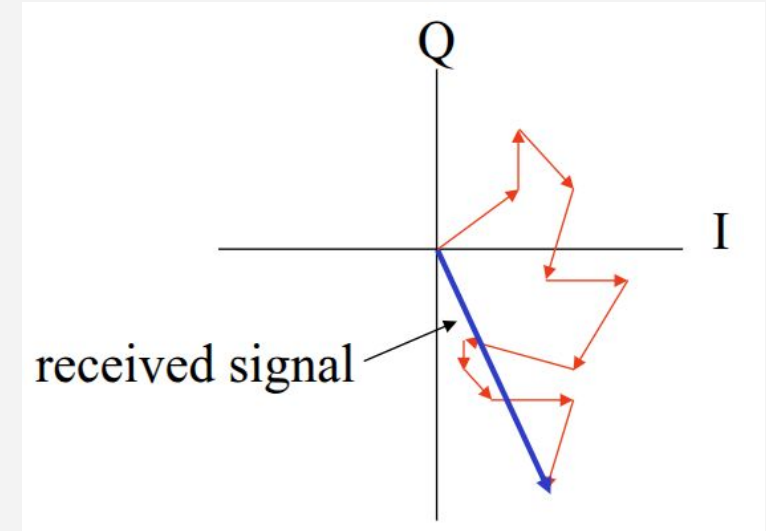
- System's response when the input is Dirac Delta  $\delta(t)$
- Impulse response completely defines LTI's behavior
  - By convolving the continuous input with  $h(x)$
- 'Channel gain' is the impulse response in frequency domain

$$\begin{aligned}
 &\delta(t - \tau) \rightarrow h(t - \tau) \\
 &\quad \downarrow \text{Homogeneity} \\
 &x_i \delta(t - t_i) \rightarrow x_i h(t - t_i) \\
 &\quad \downarrow \text{Additivity} \\
 &x(t) = \sum x_i \delta(t - t_i) \rightarrow \sum x_i h(t - t_i) \\
 &\quad \downarrow \Delta t \rightarrow 0 \\
 &x(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t - \tau) d\tau \rightarrow y(t) = x(t) * h(t)
 \end{aligned}$$

# Multipath Fading Models

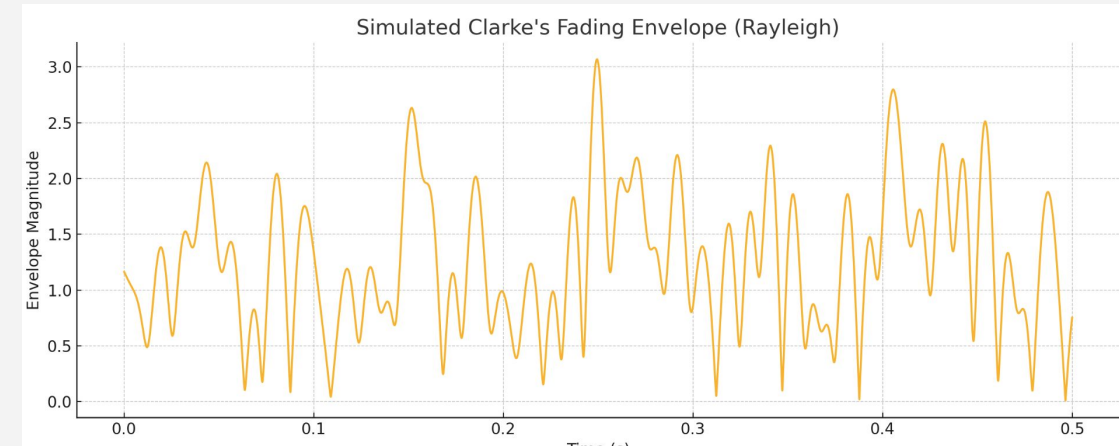
- Rayleigh's Fading

- No dominant line-of-sight path
- Many multipath components
- Amplitudes follow Rayleigh Distribution
  - $R = \sqrt{Z_1^2 + Z_2^2}$ ,  $Z \sim N(0, 1)$
- Phases follow uniform distribution (random)



- Clarke's Model

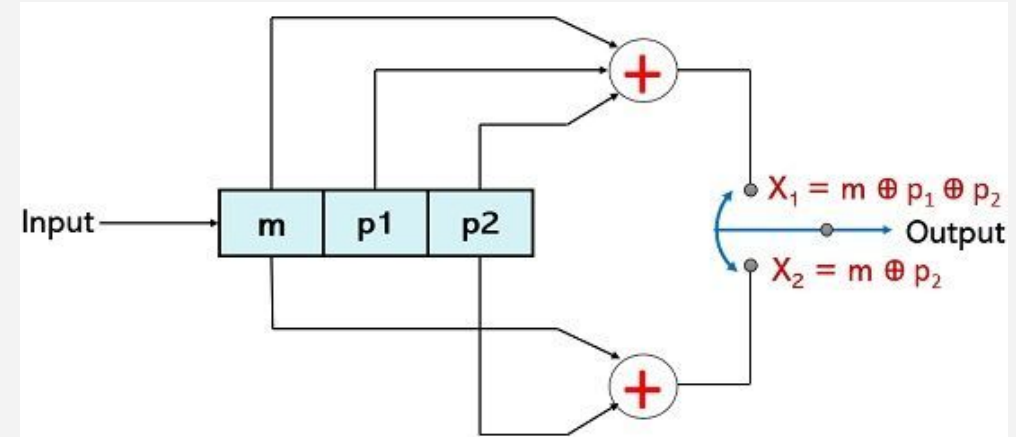
- Theoretically based on Rayleigh's fading
- The model assumes receiver is moving
  - Doppler effect is added to Rayleigh Distribution
- The model gives fading waveform over time





# Convolutional Code

- Forward Error Correction (FEC)
  - Let receiver detects and corrects error bits
    - ↔ Backward Error Correction (BEC)
      - Receiver only detects, then requests re-transmission
- XOR encoder with shifting
  - Constraint length **K**
    - The encoder remembers K bits at a time (K=3 for m, p1, p2)
  - Coding rate **R**
    - The number of output functions (R=2 for  $X_1$ ,  $X_2$ )
  - Shift, multiply and add : convolution-like



Input	Register	Output ( $X_1X_2$ )
1	100	11
0	010	10
1	101	00

ex) 101 → 00 10 11,  $K=3$ ,  $R=1/2$