## FPGA Development for Radar, Radio-Astronomy and Communications MASTERS COURSE





Dept. Electrical Engineering, University of Cape Town Private Bag, Rondebosch, 7701, South Africa http://www.rrsg.uct.ac.za



Presented by John-Philip Taylor Convened by Dr Stephen Paine

Day 5 - 3 May 2022

Outline 1 of 23

Injection Testing

**Advanced Simulation** 

**IIR Filters** 

Practical – Spectrum Analyser

**Energy Counter** 





### **Outline**

### Injection Testing

Advanced Simulation

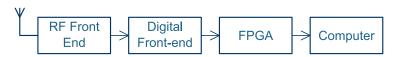
**IIR Filters** 

Practical – Spectrum Analyser

**Energy Counter** 



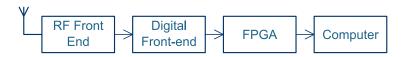




- ► The hardware (RF front-end and DFE) is not available at the time of FPGA firmware development
- There is no real-world data available
- Lab-generated signals are not suitable for DSP-chain testing
- Field tests are short-term experiments and expensive to run
- ► The DSP algorithms are experimental and uncertain



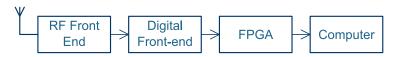




- ► The hardware (RF front-end and DFE) is not available at the time of FPGA firmware development
- ► There is no real-world data available
- Lab-generated signals are not suitable for DSP-chain testing
- Field tests are short-term experiments and expensive to run
- ► The DSP algorithms are experimental and uncertain



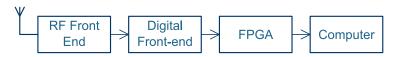




- ► The hardware (RF front-end and DFE) is not available at the time of FPGA firmware development
- ► There is no real-world data available
- Lab-generated signals are not suitable for DSP-chain testing
- Field tests are short-term experiments and expensive to run
- ► The DSP algorithms are experimental and uncertain



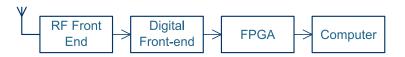




- ► The hardware (RF front-end and DFE) is not available at the time of FPGA firmware development
- ► There is no real-world data available
- Lab-generated signals are not suitable for DSP-chain testing
- Field tests are short-term experiments and expensive to run
- ► The DSP algorithms are experimental and uncertain







- ► The hardware (RF front-end and DFE) is not available at the time of FPGA firmware development
- ► There is no real-world data available
- Lab-generated signals are not suitable for DSP-chain testing
- ► Field tests are short-term experiments and expensive to run
- ► The DSP algorithms are experimental and uncertain







- ► The system is simulated
- The simulated ADC data is injected into the FPGA
- The DSP-chain is developed on this simulated data
- At first, the computer-based DSP does most, if not all, of the processing (FPGA firmware consist only of supporting infrastructure)
- As the algorithms mature, the functionality is moved into the FPGA







- ► The system is simulated
- ► The simulated ADC data is injected into the FPGA
- The DSP-chain is developed on this simulated data
- At first, the computer-based DSP does most, if not all, of the processing (FPGA firmware consist only of supporting infrastructure)
- As the algorithms mature, the functionality is moved into the FPGA







- ► The system is simulated
- ► The simulated ADC data is injected into the FPGA
- ► The DSP-chain is developed on this simulated data
- At first, the computer-based DSP does most, if not all, of the processing (FPGA firmware consist only of supporting infrastructure)
- As the algorithms mature, the functionality is moved into the FPGA







- ► The system is simulated
- The simulated ADC data is injected into the FPGA
- ► The DSP-chain is developed on this simulated data
- At first, the computer-based DSP does most, if not all, of the processing (FPGA firmware consist only of supporting infrastructure)
- As the algorithms mature, the functionality is moved into the FPGA







- ► The system is simulated
- The simulated ADC data is injected into the FPGA
- The DSP-chain is developed on this simulated data
- At first, the computer-based DSP does most, if not all, of the processing (FPGA firmware consist only of supporting infrastructure)
- As the algorithms mature, the functionality is moved into the FPGA







- ► The front-end hardware is built, but must be tested
- An FPGA-based logger is implemented that saves raw ADC data from a field trial
- This recorded data is analysed in order to test the hardware and injected into the FPGA to test the DSP-chain
- ▶ This same recorded data can be used in simulation...







- ► The front-end hardware is built, but must be tested
- An FPGA-based logger is implemented that saves raw ADC data from a field trial
- This recorded data is analysed in order to test the hardware and injected into the FPGA to test the DSP-chain
- ▶ This same recorded data can be used in simulation...







- ► The front-end hardware is built, but must be tested
- An FPGA-based logger is implemented that saves raw ADC data from a field trial
- ► This recorded data is analysed in order to test the hardware and injected into the FPGA to test the DSP-chain
- ▶ This same recorded data can be used in simulation...







- ► The front-end hardware is built, but must be tested
- An FPGA-based logger is implemented that saves raw ADC data from a field trial
- ➤ This recorded data is analysed in order to test the hardware and injected into the FPGA to test the DSP-chain
- ▶ This same recorded data can be used in simulation...



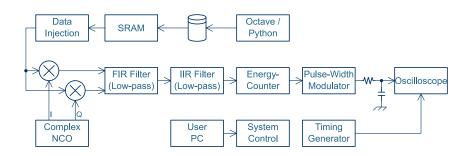




- ► The front-end hardware is built, but must be tested
- An FPGA-based logger is implemented that saves raw ADC data from a field trial
- This recorded data is analysed in order to test the hardware and injected into the FPGA to test the DSP-chain
- ► This same recorded data can be used in simulation...

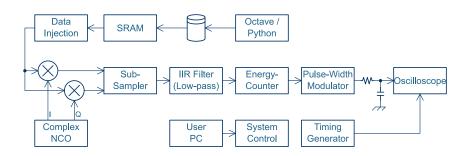






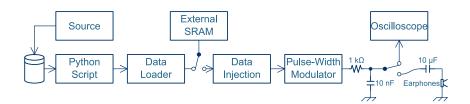
















### **Outline**

Injection Testing

#### **Advanced Simulation**

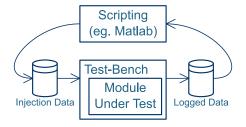
**IIR** Filters

Practical – Spectrum Analyser

**Energy Counter** 











```
`timescale 1ns/1ns
module File_IO_TB;
integer Input_File;
integer Output_File;
initial begin
  Input_File = $fopen("Test_File.txt", "rb");
  Output_File = $fopen("Output.txt" , "wb");
  $fwrite(
    Output_File,
    "Time [ns], Data [Hex], Data [Binary]\n"
  );
end
reg ipClk = 0; always #5 ipClk <= !ipClk; // 100 MHz
```





```
reg [7:0]Byte; reg [15:0]Data; integer Result;
always @(posedge ipClk) begin
  Result = $fread(Byte, Input_File); Data[ 7:0] = Byte;
  Result = $fread(Byte, Input_File); Data[15:8] = Byte;
  if(Result < 1) begin</pre>
    $fclose(Input_File); $fclose(Output_File); $stop;
  end
  $fwrite(Output_File, "%d, %04X, %s\n", $time, Data, Data);
end
endmodule
```





- ► Section 17.2.4: "Reading data from a file"
- Section 17.2.8: "Loading memory data from a file"
- \$fread reads any size register, but in big-endian
- \$fscanf reads formatted data, including raw binary
- ▶ \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change





- ► Section 17.2.4: "Reading data from a file"
- Section 17.2.8: "Loading memory data from a file"
- ▶ \$fread reads any size register, but in big-endian
- \$fscanf reads formatted data, including raw binary
- ▶ \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change





- ► Section 17.2.4: "Reading data from a file"
- ► Section 17.2.8: "Loading memory data from a file"
- \$fread reads any size register, but in big-endian
- \$fscanf reads formatted data, including raw binary
- ▶ \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change





- ► Section 17.2.4: "Reading data from a file"
- ► Section 17.2.8: "Loading memory data from a file"
- ▶ \$fread reads any size register, but in big-endian
- \$fscanf reads formatted data, including raw binary
- ▶ \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change





- ► Section 17.2.4: "Reading data from a file"
- ► Section 17.2.8: "Loading memory data from a file"
- ▶ \$fread reads any size register, but in big-endian
- ▶ \$fscanf reads formatted data, including raw binary
- \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change





- ► Section 17.2.4: "Reading data from a file"
- ► Section 17.2.8: "Loading memory data from a file"
- ▶ \$fread reads any size register, but in big-endian
- ▶ \$fscanf reads formatted data, including raw binary
- ► \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change

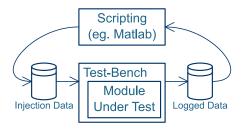




- ► Section 17.2.4: "Reading data from a file"
- ► Section 17.2.8: "Loading memory data from a file"
- ▶ \$fread reads any size register, but in big-endian
- ▶ \$fscanf reads formatted data, including raw binary
- ► \$fwrite works similar to fprintf from C, except that %u writes raw binary data
- ▶ \$fmonitor logs to file whenever the signals change



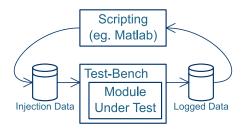




► Create a test file "ADXL345\_TB.dat": 0123456789ABCDEFGHIJKLMNOPQRSTUVWXYZ







► Create a test file "ADXL345\_TB.dat": 0123456789ABCDEFGHIJKLMNOPQRSTUVWXYZ



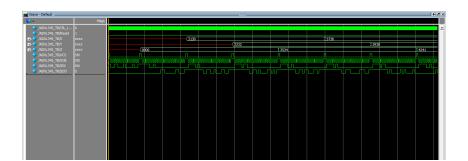


### Data Injection (ADXL345\_TB)

```
integer File, n;
reg [15:0] Data;
initial begin
  File = $fopen("../Peripherals/ADXL345_TB.dat", "rb");
  for (n = 0; n < 16; n = n + 1) @ (negedge SClk);
  forever begin
    for (n = 0; n < 8; n = n + 1) @ (negedge SClk);
    $fread(Data, File);
    for (n = 0; n < 16; n = n + 1) begin
      @ (negedge SClk);
      #40 {SDO, Data[15:1]} <= Data;
    end
  end
end
```

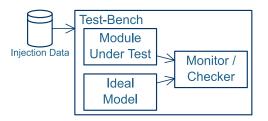










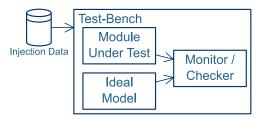


- Very useful, but time-consuming to write..
- ► The "Ideal Model" can be read from a file, or calculated in the test-bench
- ► The "Monitor" issues errors (with \$error), warnings (with \$warning) or information (with \$info) as appropriate





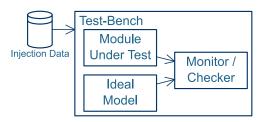
#### **Automated Test-benches**



- Very useful, but time-consuming to write...
- ► The "Monitor" issues errors (with \$error), warnings



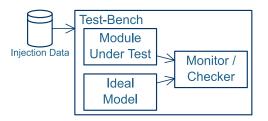




- Very useful, but time-consuming to write...
- ► The "Ideal Model" can be read from a file, or calculated in the test-bench
- ► The "Monitor" issues errors (with \$error), warnings (with \$warning) or information (with \$info) as appropriate







- ► Very useful, but time-consuming to write...
- ► The "Ideal Model" can be read from a file, or calculated in the test-bench
- ► The "Monitor" issues errors (with \$error), warnings (with \$warning) or information (with \$info) as appropriate











#### **Outline**

Injection Testing

**Advanced Simulation** 

**IIR Filters** 

Practical – Spectrum Analyser

**Energy Counter** 





$$H(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}, \quad \omega_0 = 2\pi f_0, \quad \zeta = \frac{1}{\sqrt{2}}$$
$$s = (1 - z^{-1})f_s, \quad f_s = 781.25 \text{ kHz}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\omega_0^2}{a - bz^{-1} + cz^{-2}}, \begin{cases} a = f_s^2 + 2\zeta\omega_0 f_s + \omega_0^2 \\ b = 2f_s^2 + 2\zeta\omega_0 f_s \\ c = f_s^2 \end{cases}$$

$$Y(z) \left( a - bz^{-1} + cz^{-2} \right) = X(z) \left( \omega_0^2 \right)$$

$$(2^N) \left( \omega_0^2 x_n + by_{n-1} - cy_{n-2} \right)$$





$$H(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}, \quad \omega_0 = 2\pi f_0, \quad \zeta = \frac{1}{\sqrt{2}}$$
$$s = (1 - z^{-1})f_s, \quad f_s = 781.25 \text{ kHz}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\omega_0^2}{a - bz^{-1} + cz^{-2}}, \begin{cases} a = f_s^2 + 2\zeta\omega_0 f_s + \omega_0^2 \\ b = 2f_s^2 + 2\zeta\omega_0 f_s \\ c = f_s^2 \end{cases}$$

$$Y(z) \left( a - bz^{-1} + cz^{-2} \right) = X(z) \left( \omega_0^2 \right)$$

$$(2^N) \left( \omega_0^2 x_n + by_{n-1} - cy_{n-2} \right)$$





$$H(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}, \quad \omega_0 = 2\pi f_0, \quad \zeta = \frac{1}{\sqrt{2}}$$
  $s = (1 - z^{-1})f_s, \quad f_s = 781.25 \text{ kHz}$ 

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\omega_0^2}{a - bz^{-1} + cz^{-2}}, \begin{cases} a = f_s^2 + 2\zeta\omega_0 f_s + \omega_0^2 \\ b = 2f_s^2 + 2\zeta\omega_0 f_s \\ c = f_s^2 \end{cases}$$

$$Y(z)\left(a-bz^{-1}+cz^{-2}\right)=X(z)\left(\omega_0^2\right)$$

$$y_n = \left(\frac{2^N}{a}\right) \left(\frac{\omega_0^2 x_n + b y_{n-1} - c y_{n-2}}{2^N}\right)$$





$$H(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}, \quad \omega_0 = 2\pi f_0, \quad \zeta = \frac{1}{\sqrt{2}}$$
$$s = (1 - z^{-1})f_s, \quad f_s = 781.25 \text{ kHz}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\omega_0^2}{a - bz^{-1} + cz^{-2}}, \begin{cases} a = f_s^2 + 2\zeta\omega_0 f_s + \omega_0^2 \\ b = 2f_s^2 + 2\zeta\omega_0 f_s \\ c = f_s^2 \end{cases}$$
$$Y(z) \left( a - bz^{-1} + cz^{-2} \right) = X(z) \left( \omega_0^2 \right)$$





$$H(s) = \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}, \quad \omega_0 = 2\pi f_0, \quad \zeta = \frac{1}{\sqrt{2}}$$
 $s = (1 - z^{-1})f_s, \quad f_s = 781.25 \text{ kHz}$ 

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\omega_0^2}{a - bz^{-1} + cz^{-2}}, \quad \begin{cases} a = f_s^2 + 2\zeta\omega_0 f_s + \omega_0^2 \\ b = 2f_s^2 + 2\zeta\omega_0 f_s \\ c = f_s^2 \end{cases}$$
$$Y(z) \left( a - bz^{-1} + cz^{-2} \right) = X(z) \left( \omega_0^2 \right)$$
$$y_n = \left( \frac{2^N}{a} \right) \left( \frac{\omega_0^2 x_n + by_{n-1} - cy_{n-2}}{2^N} \right)$$





$$y_n = \frac{Ax_n + By_{n-1} - Cy_{n-2}}{2^N}, \quad \left\{ \begin{array}{l} A = (2^N/a) \cdot \omega_0^2 \\ B = (2^N/a) \cdot b \\ C = (2^N/a) \cdot c \\ N = 32 \end{array} \right.$$

$f_0$	Α	В	С
10	28	8 589 446 092	4 294 478 824
100	2 775	8 585 049 594	4 290 085 073
1 000	274 663	8 541 087 701	4 246 395 068
10 000	24 799 428	8 104 255 079	3 834 087 211
100 000	997 792 625	4 839 800 553	1 542 625 882





- ► Be careful with overflows perform a limit operation
- ► Be careful with the signedness of the operation
- ► To help you keep track of what bit represents what in fixed-point representations, use negative indices:

```
// Constants are in the range [0, 2), but
// Verilog still thinks they're unsigned integers
wire [0:-32]B = 33'd_8_589_446_092;
// Internal registers are in the range [-1, 1)
reg [0:-39]y_1;
// Products are in the range [-2, 2)
wire [1:-71]B y 1;
```





- ► Be careful with overflows perform a limit operation
- ► Be careful with the signedness of the operation
- ► To help you keep track of what bit represents what in fixed-point representations, use negative indices:

```
// Constants are in the range [0, 2), but
// Verilog still thinks they're unsigned integers
wire [0:-32]B = 33'd_8_589_446_092;
// Internal registers are in the range [-1, 1)
reg [0:-39]y_1;
// Products are in the range [-2, 2)
wire [1:-71]B v 1:
```





- ► Be careful with overflows perform a limit operation
- ► Be careful with the signedness of the operation
- ► To help you keep track of what bit represents what in fixed-point representations, use negative indices:

```
// Constants are in the range [0, 2), but
// Verilog still thinks they're unsigned integers
wire [0:-32]B = 33'd_8_589_446_092;
// Internal registers are in the range [-1, 1)
reg [0:-39]y_1;
// Products are in the range [-2, 2)
wire [1:-71]B v 1:
```



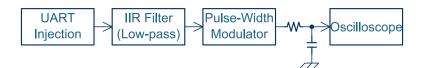


- ► Be careful with overflows perform a limit operation
- ► Be careful with the signedness of the operation
- ► To help you keep track of what bit represents what in fixed-point representations, use negative indices:

```
// Constants are in the range [0, 2), but
// Verilog still thinks they're unsigned integers
wire [0:-32]B = 33'd_8_589_446_092;
// Internal registers are in the range [-1, 1)
reg [0:-39]y_1;
// Products are in the range [-2, 2)
wire [1:-71]B_y_1;
```

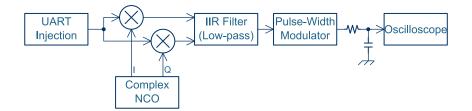






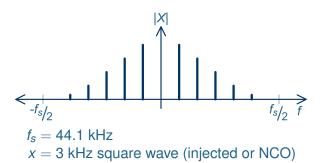








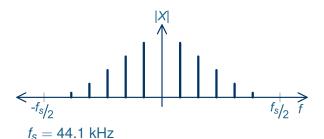




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz





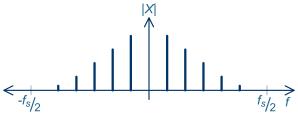


x = 3 kHz square wave (injected or NCO)

- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz







- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz



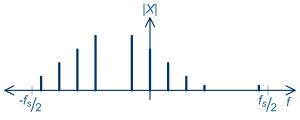




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz



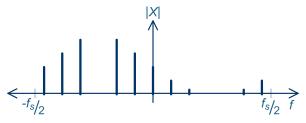




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz



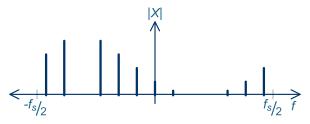




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz



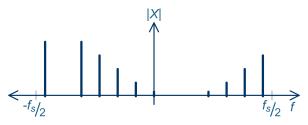




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz



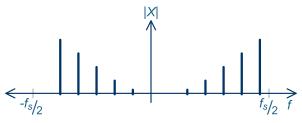




- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz







- ► Make the resolution bandwidth 200 Hz (100 Hz low-pass filter)
- ► Sweep the complex NCO from DC to 20 kHz





### **Outline**

Injection Testing

**Advanced Simulation** 

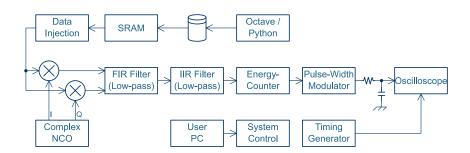
**IIR Filters** 

Practical – Spectrum Analyser

**Energy Counter** 

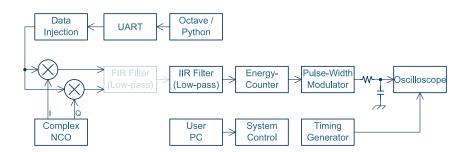
















#### **Outline**

Injection Testing

**Advanced Simulation** 

**IIR Filters** 

Practical – Spectrum Analyser

**Energy Counter** 





$$P = \frac{1}{T_p} \int_{T_p} |x(t)|^2 dt$$
$$= \frac{1}{N} \sum_{n=0}^{N-1} (x_n x_n^*)$$

- ► Use a variable *N* (time-window) use power-of-two sizes
- Use a LUT to convert the signal power to log-scale





$$P = \frac{1}{T_p} \int_{T_p} |x(t)|^2 dt$$
$$= \frac{1}{N} \sum_{n=0}^{N-1} (x_n x_n^*)$$

- ► Use a variable *N* (time-window) use power-of-two sizes
- Use a LUT to convert the signal power to log-scale



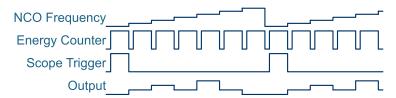


$$P = \frac{1}{T_p} \int_{T_p} |x(t)|^2 dt$$
$$= \frac{1}{N} \sum_{n=0}^{N-1} (x_n x_n^*)$$

- ► Use a variable *N* (time-window) use power-of-two sizes
- ► Use a LUT to convert the signal power to log-scale



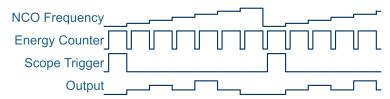




- Many parameters are involved...
- ► Set the parameters by means of registers: let the PC do all the calculations and sanity-checks
- Or you can use look-up tables to simplify the registers (eg. resolution bandwidth selection by constants 1 → 5, translated to the filter constants, energy-counter window size, counter limits, NCO step-sizes, etc.)



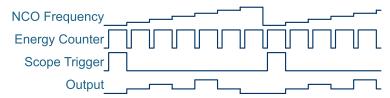




- Many parameters are involved...
- Set the parameters by means of registers: let the PC do all the calculations and sanity-checks
- Or you can use look-up tables to simplify the registers (eg. resolution bandwidth selection by constants 1 → 5, translated to the filter constants, energy-counter window size, counter limits, NCO step-sizes, etc.)



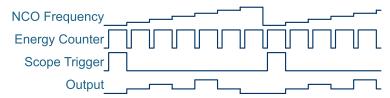




- Many parameters are involved...
- Set the parameters by means of registers: let the PC do all the calculations and sanity-checks
- Or you can use look-up tables to simplify the registers (eg. resolution bandwidth selection by constants 1 → 5, translated to the filter constants, energy-counter window size, counter limits, NCO step-sizes, etc.)







- ► Many parameters are involved...
- ► Set the parameters by means of registers: let the PC do all the calculations and sanity-checks
- Or you can use look-up tables to simplify the registers (eg. resolution bandwidth selection by constants 1 → 5, translated to the filter constants, energy-counter window size, counter limits, NCO step-sizes, etc.)





#### **Select References**

- Stephen Brown and Zvonko Vranesic Fundamentals of Digital Logic with Verilog Design, 2<sup>nd</sup> Edition ISBN 978-0-07-721164-6
- Merrill L Skolnik Introduction to RADAR Systems ISBN 978-0-07-288138-7
- Mark A. Richards and James A. Scheer Principles of Modern Radar: Basic Principles ISBN 978-1-89-112152-4
- Deepak Kumar Tala
  World of ASIC
  http://www.asic-world.com/
- Jean P. Nicolle
  FPGA 4 Fun
  http://www.fpga4fun.com/





## FPGA Development for Radar, Radio-Astronomy and Communications MASTERS COURSE





Dept. Electrical Engineering, University of Cape Town Private Bag, Rondebosch, 7701, South Africa http://www.rrsg.uct.ac.za



Presented by John-Philip Taylor Convened by Dr Stephen Paine

Day 5 - 3 May 2022