FPGA Development for Radar, Radio-Astronomy and Communications MASTERS COURSE





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Presented by John-Philip Taylor Convened by Dr Stephen Paine

Day 6 - 4 May 2022

Outline 1 of 40

Advanced Timing Constraints

FIR Filters

Practical - FIR Filter

Projects

Tips and Tricks

Conclusion





Outline

Advanced Timing Constraints

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Conclusion





- ► Synopsis Design Constraints start with the ideal case:
 - ► There are no PCB trace delays
 - ► The external setup requirement is zero
 - ► The external hold requirement is zero
 - ► The external propagation delay is zero
- Increase the maximum output delay for external setup timing
- Decrease the minimum output delay for external hold timing
- ► Specify the minimum and maximum input delays according to the external propagation delay parameters
- ► Worsen the situation with PCB trace delays and uncertainties (clock jitter, manufacturing tolerances, etc.)





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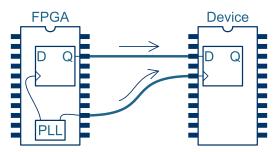




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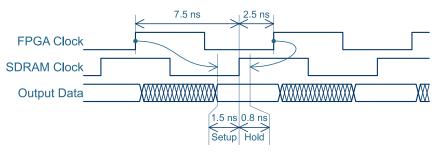




- ► FPGA internal delays and PCB trace delays cancel
- The important parameters are:
 - Setup and hold times of the external device
 - Clock jitter and other uncertainties
- ► Shift the external clock to ease the hold margin



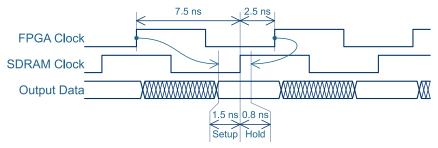




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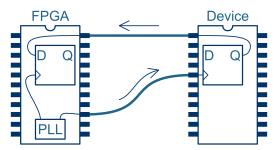


```
# Suppose 100 ps uncertainty
set_output_delay -max -clock DRAM_CLK 1.6 [get_ports opDRAM*]
set_output_delay -min -clock DRAM_CLK -0.9 [get_ports opDRAM*]
set_output_delay -max -clock DRAM_CLK 1.6 [get_ports bpDRAM*]
set_output_delay -min -clock DRAM_CLK -0.9 [get_ports bpDRAM*]
```





External Timing – Input

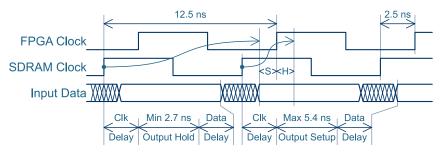


- Large FPGA internal delays and PCB trace delays
- ► Shift the external clock to ease the setup margin
- Multi-cycle requirement on the setup path (otherwise the compiler uses the 2.5 ns path) – the minimum propagation delay of 2.7 ns makes the 2.5 ns clock shift safe





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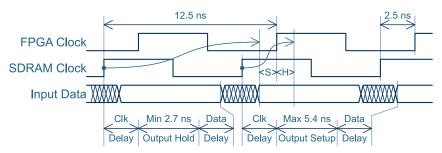


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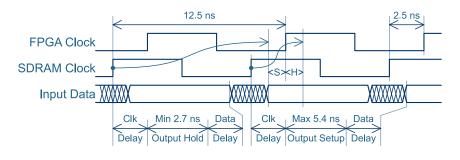
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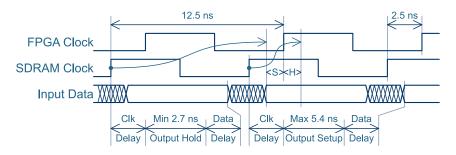




```
set_multicycle_path
  -from [get_clocks {DRAM_CLK}] \
  -to [get_clocks {*altpll_0*clk[0]}] \
  -setup 2
```



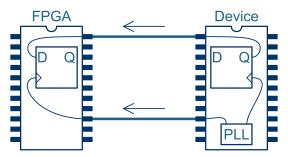




Suppose 100 ps uncertainty and 200 ps PCB delay (each way)
set_input_delay -max -clock DRAM_CLK 5.9 [get_ports bpDRAM*]
set_input_delay -min -clock DRAM_CLK 3.0 [get_ports bpDRAM*]



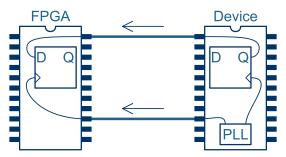




- ► The clock is sourced by the external device
- The PCB trace delays cancel
- The data is centre-aligned
- ▶ Use the dedicated serial LVDS receivers (on most FPGAs)



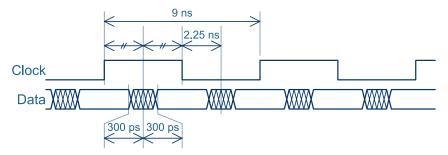




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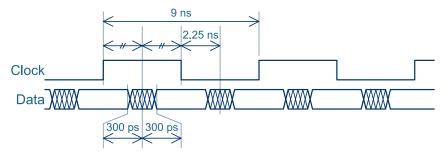




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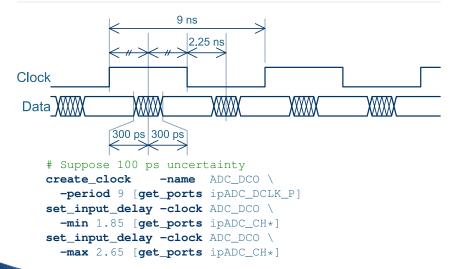


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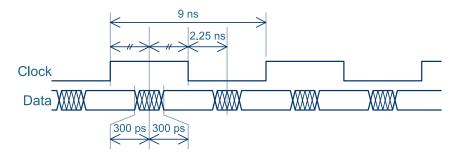


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```
set_input_delay -clock ADC_DCO -clock_fall \
    -min 1.85 [get_ports ipADC_CH*] -add_delay
set_input_delay -clock ADC_DCO -clock_fall \
    -max 2.65 [get_ports ipADC_CH*] -add_delay
```





- ▶ Use the correct I/O standard
- Set the correct output current
- Set the pin capacitance

```
IOBUF PORT "opDRAM*" IO_TYPE=LVCMOS33;
IOBUF PORT "bpDRAM*" IO_TYPE=LVCMOS33;
IOBUF PORT "clkDRAM" IO_TYPE=LVCMOS33;
```





- ▶ Use the correct I/O standard
- ► Set the correct output current
- ► Set the pin capacitance

```
IOBUF PORT "opDRAM*" IO_TYPE=LVCMOS33 DRIVE=8;
IOBUF PORT "bpDRAM*" IO_TYPE=LVCMOS33 DRIVE=8;
IOBUF PORT "clkDRAM" IO_TYPE=LVCMOS33 DRIVE=8;
```





- ▶ Use the correct I/O standard
- ► Set the correct output current
- ► Set the pin capacitance

```
OUTPUT PORT "opDRAM*" LOAD 3.8 pF;
OUTPUT PORT "bpDRAM_DQ*" LOAD 6.0 pF;
OUTPUT PORT "CLK DRAM" LOAD 3.5 pF;
```





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- ► Convolve the impulse-response with the signal:
 - 1. Time-reverse the impulse-response
 - Within the FIR filter window, multiply the impulse-response sample by the signal sample
 - 3. Sum the products and output the result
 - 4. Move the impulse-response by one sample
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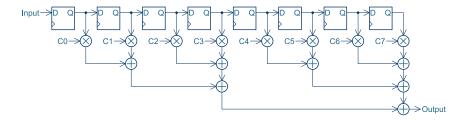




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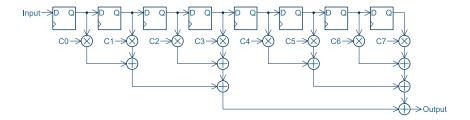




- ► Move the signal instead of the impulse-response
- To check the direction, inject an impulse (which should produce the impulse-response at the output)
- ► Pipeline the adder tree to increase the maximum clock frequency



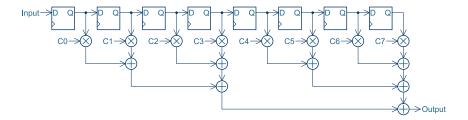




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- ► Always take the design criteria into account when designing an architecture
- ► What happens when you add decimation (i.e. you don't need an output sample every clock cycle)?
- ▶ What if the FPGA clock is much faster than the sample rate?
- What about a combination of the above?
- ► Always consider the scenario: sample rate, clock speed, power requirements, throughput requirements, decimation (if any), available resources, etc...





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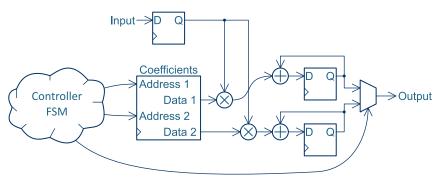




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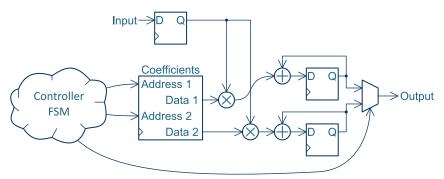




- ► This example decimates by *N*/2: a 512-point filter will decimate by 256
- ▶ The coefficient addresses are run N/2 out of phase



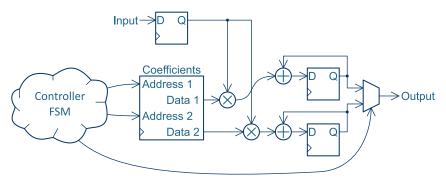




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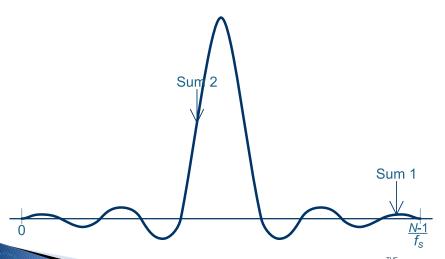




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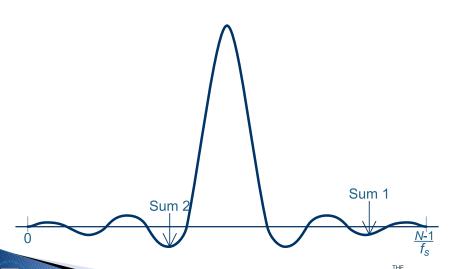






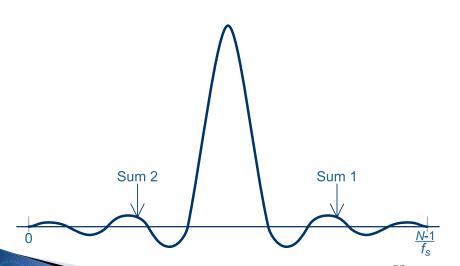






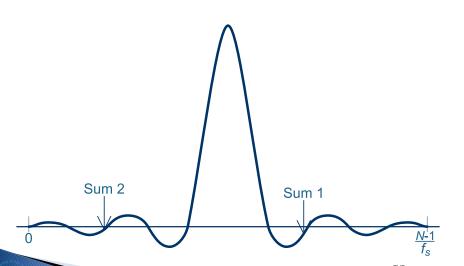






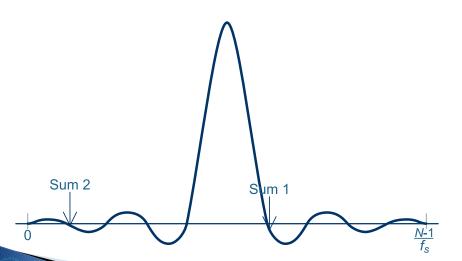






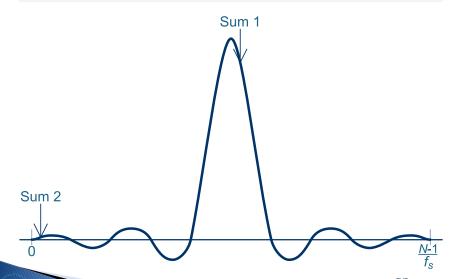






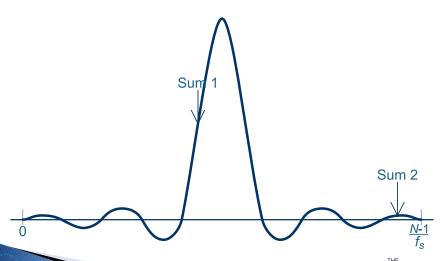








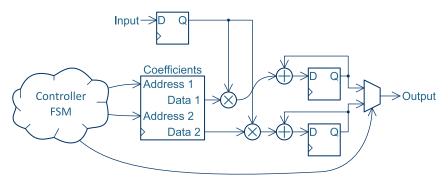








Faster System Clock

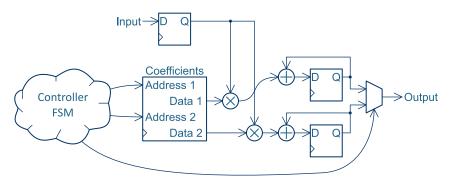


- ► If the sample clock is lower than the system clock, this same architecture can decimate by less:
- ► Keep more than 2 sums sometimes more efficient to keep these in BRAM as well





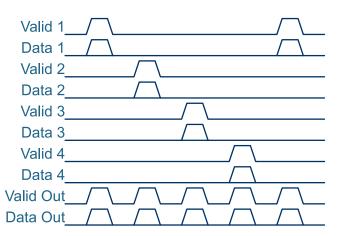
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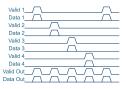








FIR Filter 12 of 40

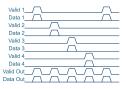


- Use multiple instances of a filter that decimates more than desired
- Reset the counters of the units out of phase
- ► Combine the outputs with a simple AND-OR circuit





FIR Filter 12 of 40

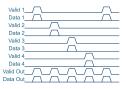


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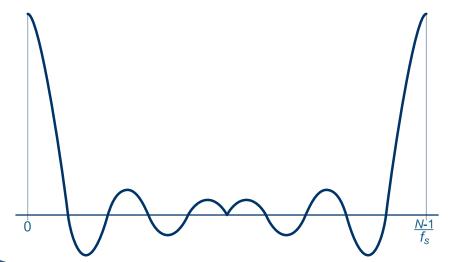






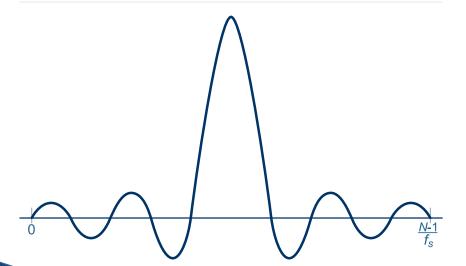








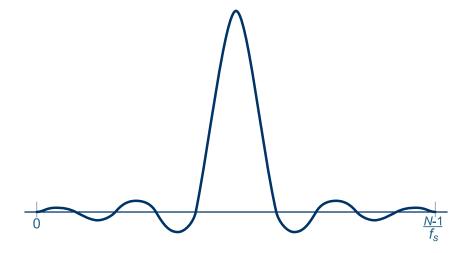






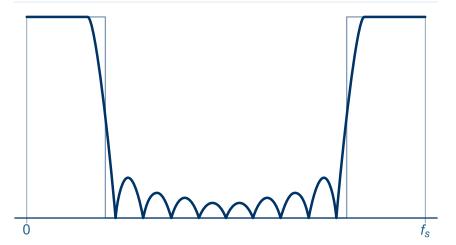












(Zero-pad to see the side-lobes)





- ► Implement a N-point FIR filter that decimates by N (i.e. one sample output for every N samples input)
- ▶ Use an appropriate cut-off frequency and a Hann window $w(n) = \sin^2\left(\frac{\pi n}{N-1}\right)$
- ► Use Matlab / Octave / Python to generate the FIR filter constants and memory initialisation file
- ► Verify through simulation
- ▶ Verify on FPGA
- ► Combine eight filter units to drop the sub-sampling rate to N/8
- ► Verify on FPGA





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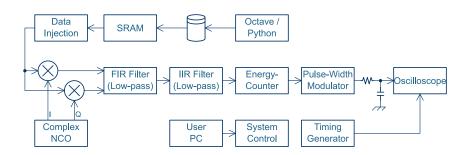
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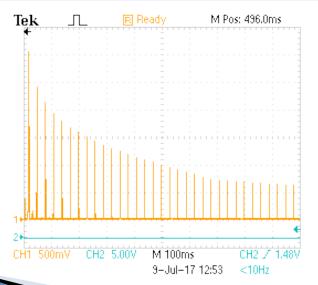








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- Everybody must do a different project, but the projects are interlinked. Too make it more fun, make sure the system parameters are compatible across projects.
- ► You can propose a project: preferably in line with your current MSc research
- ➤ You need to design the DSP chain and choose appropriate system parameters
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 - ▶ Demonstrate a working FPGA-based DSP chain
 - Give a 15-minute presentation on the design and performance results
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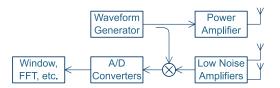








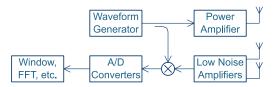




- Sparse-array FMCW RADAR (see Project 4)
- Choose system parameters appropriate for a practical radar – typical parameters include:
 - Sweep time of about 1 ms (sweep faster for fast-moving targets, and sweep slower for more range)
 - RF bandwidth of 500 MH;
 - 256 samples per sweep
 - 256 sweeps per burst (this is used for Doppler processing later in the chain)



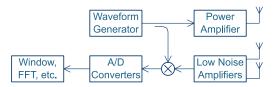




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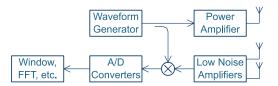




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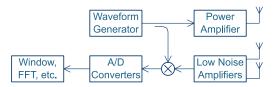




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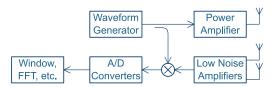




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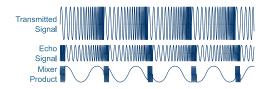




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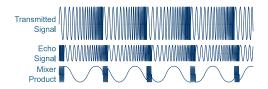




- You can assume that the FPGA generates the transmit sweep triangle
- ► Take the FFT of each sweep (range FFT), organise them into bursts and store the results in SRAM
- ▶ Do the FFT in-place instead of streaming, which saves on multipliers and RAM blocks
- ► This is the end of this project another project could potentially take this output data and processes it further



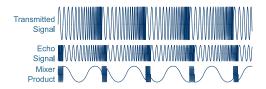




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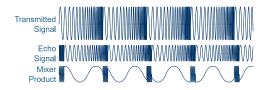




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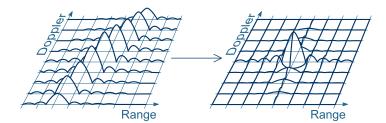




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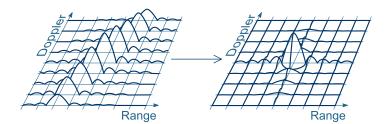




- Simulate the data output of Project 1 and inject the anticipated result into the SRAM
- ▶ Play back the corner-turned data and take the Doppler FFTs
- Store the resulting range-Doppler maps in SRAM for the next step



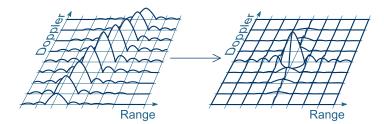




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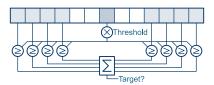


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Project 3 – Range CFAR

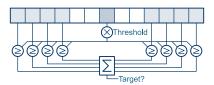


- Simulate the data output of Project 2 and inject the anticipated result into the SRAM
- Play back the corner-turned data and process the range CFAR
- ► Create a stream of detected targets (store the range, Doppler and phase of the two incoming channels
- ► In a real system, this stream would go directly to the next step, but for this project, store the stream in SRAM





Project 3 - Range CFAR

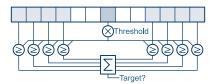


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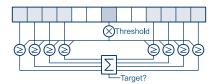


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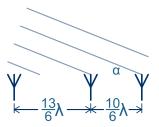
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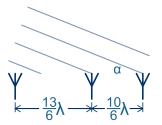




- Simulate the data output of Project 3 and inject the anticipated result into the SRAM
- Play back the target stream and perform angle extraction for the sparse array





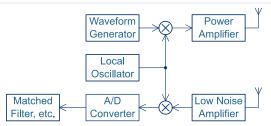


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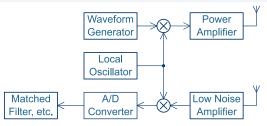
27 of 40



- Design and implement a chirped pulse RADAR front-end
- Inject raw ADC data and implement a matched filter by means of convolution (essentially a FIR filter)
- Display the PRI's on the oscilloscope





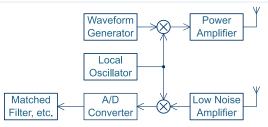


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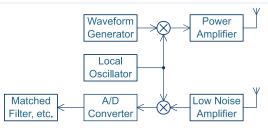


- Design and implement a chirped pulse RADAR front-end
- ► Inject raw ADC data and implement a matched filter by means of convolution (essentially a FIR filter)
- Display the PRI's on the oscilloscope (Range is about 6.7 μs/km, so it is practical to output the signal from a long-range RADAR (350 km or so) on 40 kHz bandwidth PWM (filter time-constant of 4 μs)...





27 of 40

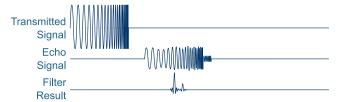


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- ► Inject raw ADC data and implement a matched filter by means of convolution (essentially a FIR filter)
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- The rest of a typical processing chain is conceptually similar to projects 2 to 4, which can be adapted to process the results from this front-end





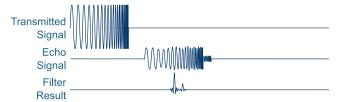
Project 5 – Pulsed Front-end



- You can assume that the FPGA generates the transmit timing control, so you can use the SRAM address to determine where you are in the current PRI
- ► After doing matched filtering, organise the results into bursts and store them in SRAM
- ➤ This is the end of this project another project could potentially take this output data and processes it further



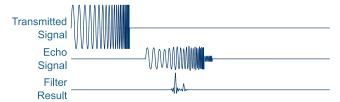




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- ▶ Design and implement a commensal RADAR front-end
- Inject raw ADC data and implement range extraction (correlate the direct path with the echo signal)



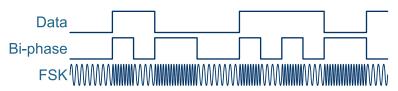




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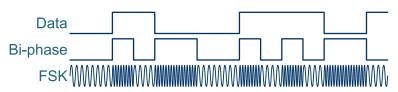




- Implement an FSK-based, bi-phase-coded communication channel
- Use S/PDIF as inspiration, with synchronisation word, etc.
- Inject simulated ADC data and demodulate by whatever means is convenient (matched filter, most likely)
- Display the received bit-stream on the oscilloscope
- ► Analyse channel performance in the presence of noise



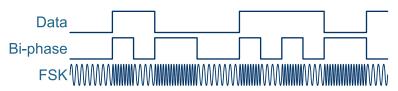




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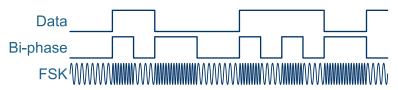




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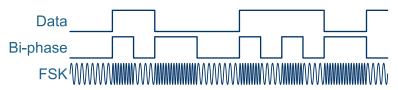




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- Design and implement a 16-QAM modulator
- Inject a data stream and produce a 16-QAM output stream
- Including a synchronisation header and run-length limit
- ► Store the resulting modulated signal in SRAM







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- Simulate the data output of Project 8 and inject the anticipated result into the SRAM
- Assume an ideal RF front-end

 (i.e. no need to perform carrier-recovery)
- ► Play back the data stream and recover synchronisation
- Demodulate the data stream to obtain the original data







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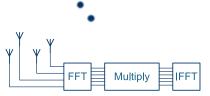




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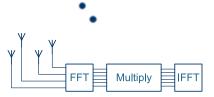




- ➤ Simulate a sky with at least two sources and inject ADC data from 4 receivers (4-bit per sample each)
- Perform correlation between all combinations of input channels (FFT, multiply, IFFT)
- Store the result in SRAM
- ► Keep things simple: the earth is flat and stationary, etc.
- ▶ If parallel FFTs don't fit, do them sequentially



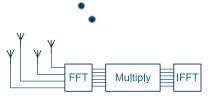




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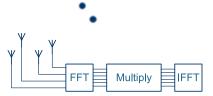




- ➤ Simulate a sky with at least two sources and inject ADC data from 4 receivers (4-bit per sample each)
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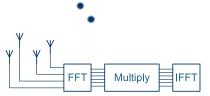




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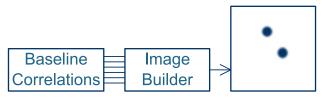




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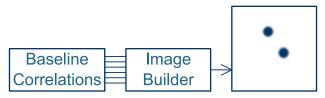




- ► Simulate the output of Project 10 and inject the anticipated result into the SRAM
- Build an image from the correlation data
- Store the resulting image in SRAM so that it can be viewed on the PC



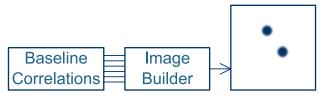




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- ▶ Use module parametrisation when appropriate
- Use standardised bus structures and interfaces, and the same interface family across all projects
- Use consistent naming conventions
- Clearly mark negative logic in the name





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Libraries 36 of 40

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Libraries 36 of 40

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 http://www.asic-world.com/
- Jean P. Nicolle
 FPGA 4 Fun
 http://www.fpga4fun.com/





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Presented by John-Philip Taylor Convened by Dr Stephen Paine

Day 6 - 4 May 2022