Assignments for Chapter 1

- 1.20
- 1.21 (c) (f)
- 1.24 (a)
- **1.26**
- 1.27 (a) (f)
- 1.41

Periodicity of DT Complex Exponentials

Important difference between $e^{j\omega_0 n}$ and $e^{j\omega_0 t}$:

• $e^{j\omega_0 n}$ is a periodic signal only when $\frac{\omega_0}{2\pi}$ is a rational number

$$e^{j\omega_0 n}=e^{j\omega_0(n+N)}$$
 \longrightarrow $e^{j\omega_0 N}=1$ \longrightarrow $\omega_0 N=2\pi m$ Hence, $\frac{\omega_0}{2\pi}=\frac{m}{N}$

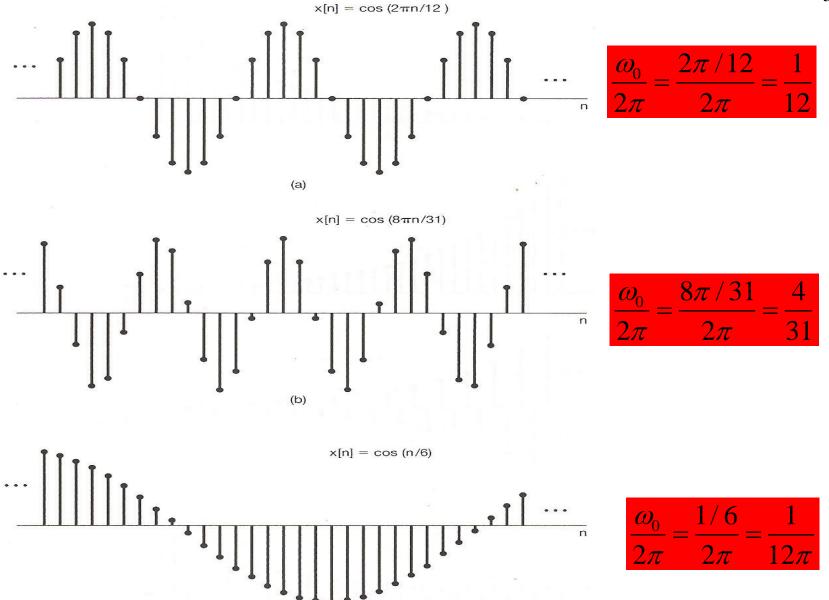


Figure 1.25 Discrete-time sinusoidal signals.

(c)

How to determine the fundamental period of $e^{j\omega_0 n}$?

Solution:

Let N be the fundamental period, then

$$e^{j\omega_0(n+N)} = e^{j\omega_0n} \rightarrow e^{j\omega_0N} = 1.$$

- \exists integer m, $\omega_0 N = 2\pi m$.
- Therefore,

$$N=\frac{2\pi}{\omega_0}m.$$

• Hence, N is the minimum positive integer in the set $\{\frac{2\pi}{\omega_0}m|\forall\ integer\ m\}$.

Example

• What is the fundamental period of $e^{jrac{6}{5}\pi n}$?

$$\left\{ \frac{2\pi}{\omega_0} m \middle| \forall integer m \right\} = \left\{ \frac{5}{3} m \middle| \forall integer m \right\}$$
$$= \left\{ \dots, 0, \frac{5}{3}, \frac{10}{3}, 5, \frac{20}{3}, \dots \right\}$$

Hence, the fundamental period is 5 and fundamental frequency is $\frac{2\pi}{5}$.

DT Harmonically Related Set

Harmonically related discrete-time signal sets

$$\{\phi_k[n] = e^{jk(\frac{2\pi}{N})n}, \quad \mathbf{k} = 0, \pm 1, \pm 2, \dots \}$$

all with common period N

There are only N elements in the above set.

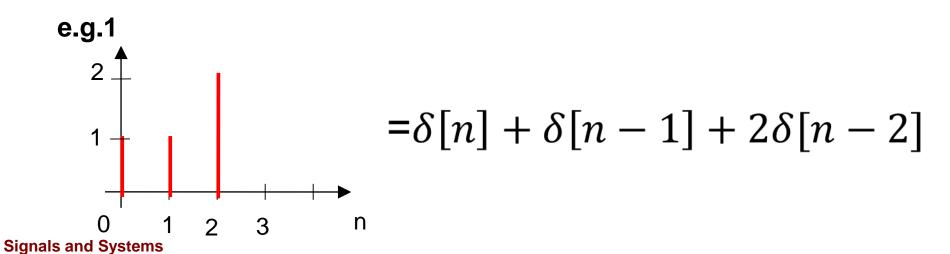
Proof:
$$\phi_{k+N}[n] = e^{j(k+N)(\frac{2\pi}{N})n} = e^{jk(\frac{2\pi}{N})n} \cdot e^{j2\pi n} = e^{jk(\frac{2\pi}{N})n} = \phi_k[n]$$

This is different from continuous case. Only Ndistinct signals in this set.

DT Unit Impulse Function

Discrete-time $\delta[n] = \begin{cases} 0, & n \neq 0 \\ 1, & n = 0 \end{cases}$

 As a basic building function, we can use unit impulse function to represent other different signals.

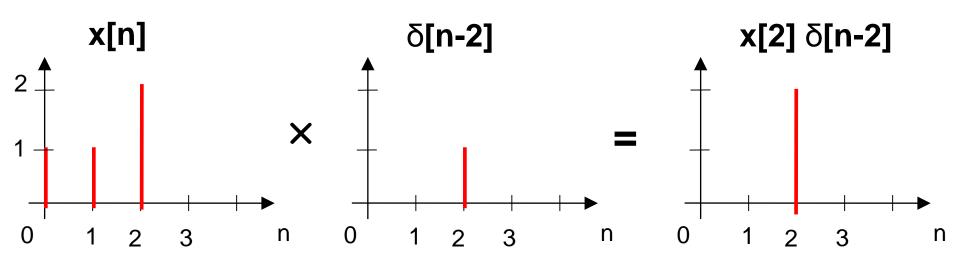


DT Unit Impulse Function (cont.)

Sampling property

$$x[n] \delta[n] = x[0] \delta[n]$$

 $x[n] \delta[n-n_0] = x[n_0] \delta[n-n_0]$



DT Unit Step Function

Discrete-time

$$u[n] = \begin{cases} 1, & n \ge 0 \\ 0, & n < 0 \end{cases}$$

$$-3-2-1 & 0 & 1 & 2 & 3 & n \end{cases}$$

Relation between unit impulse and unit step functions

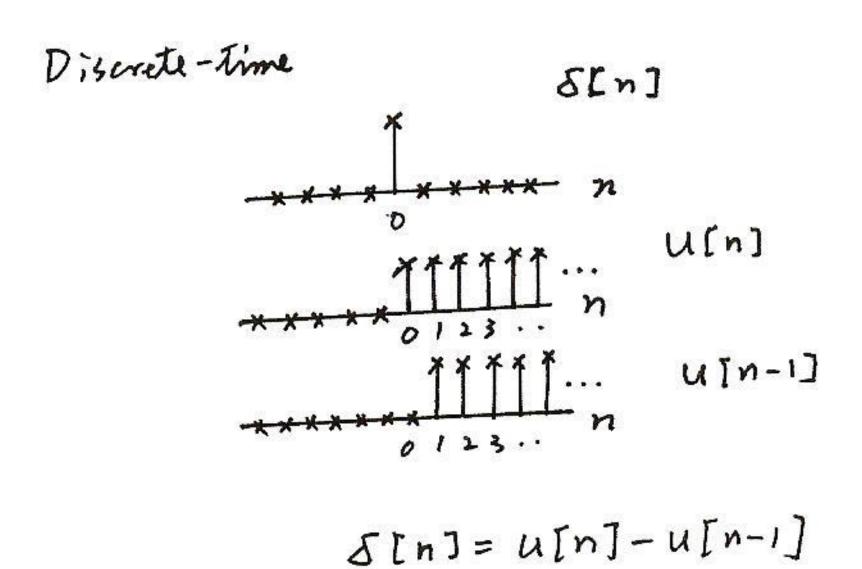
First difference

$$\delta[n] = u[n] - u[n-1]$$

- Running Sum
$$u[n] = \sum_{m=-\infty}^{n} \delta[m] = 0, n < 0$$

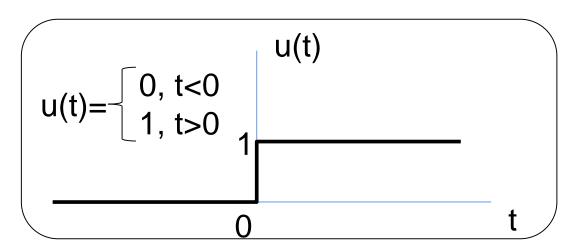
$$u[n] = \sum_{m=-\infty}^{\infty} \delta[n-k]$$

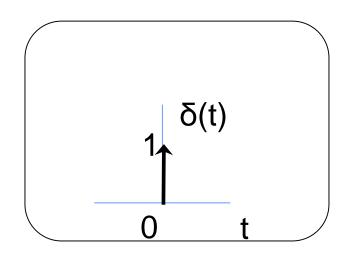
DT Unit Step Function: First Difference



CT Unit Impulse and Unit Step Functions

Continuous-time



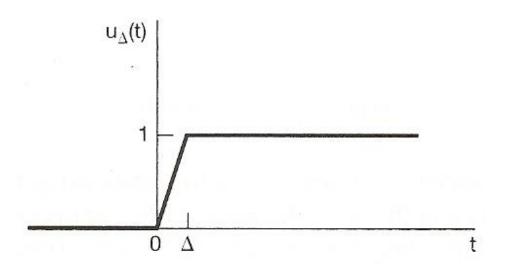


Relation between unit impulse and unit step functions

$$\delta(t) = \frac{du(t)}{dt}$$

$$u(t) = \int_{-\infty}^{t} \delta(\tau) d\tau$$

CT Unit Impulse and Unit Step Functions: Asymptotic View



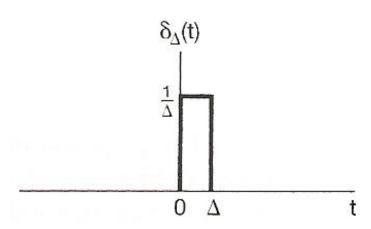


Figure 1.33 Continuous approximation to the unit step, $u_{\Delta}(t)$.

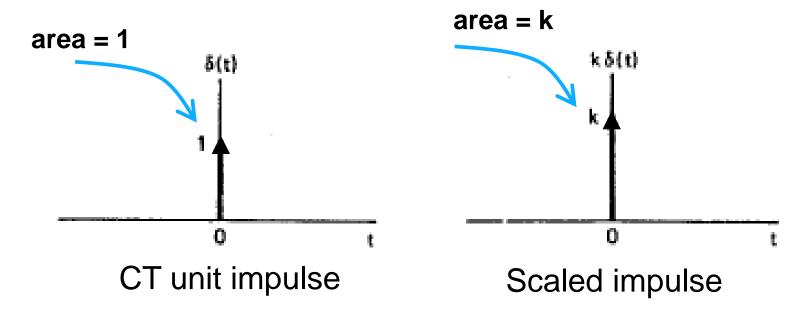
Figure 1.34 Derivative of $u_{\Delta}(t)$.

$$u(t) = \lim_{\Delta \to 0} u_{\Delta}(t)$$

$$\delta(t) = \lim_{\Delta \to 0} \delta_{\Delta}(t)$$

More on CT unit impulse function:

δ(t) has in effect no duration, but unit area.



• Or the integration of CT unit impulse function is unit. $\int_{-\infty}^{\infty} \delta(t) dt = 1$

Sampling Property

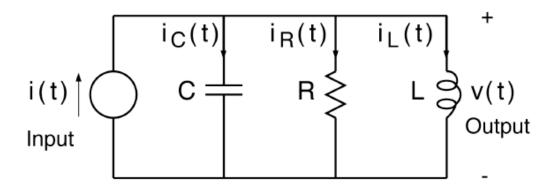
Sampling property

$$x(t) \delta(t-t_0) = x(t_0) \delta(t-t_0)$$

System Examples

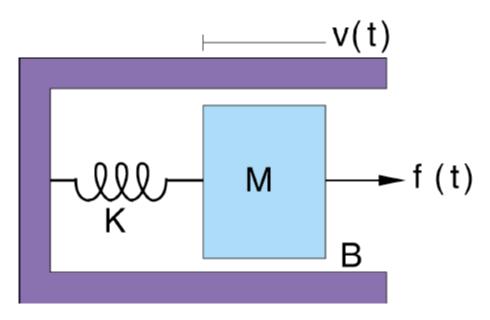
$$x(t)$$
 \longrightarrow CT System \longrightarrow $y(t)$ $x[n]$ \longrightarrow DT System \longrightarrow $y[n]$

Ex. #1 RLC circuit — an electrical system



$$i(t) = C \frac{dv(t)}{dt} + \frac{v(t)}{R} + \frac{1}{L} \int_{-\infty}^{t} v(\tau) d\tau$$
capacitance resistance

Ex. #2 A shock absorber – a mechanical system

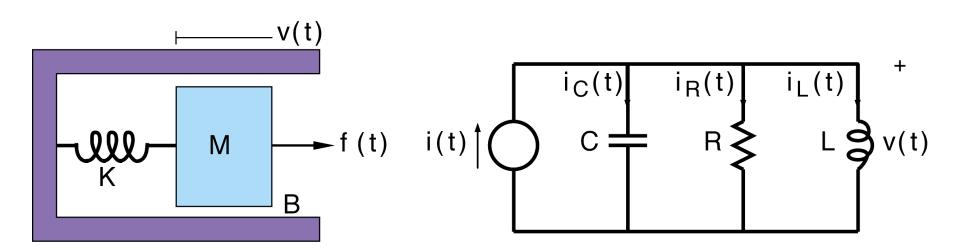


Force Balance:

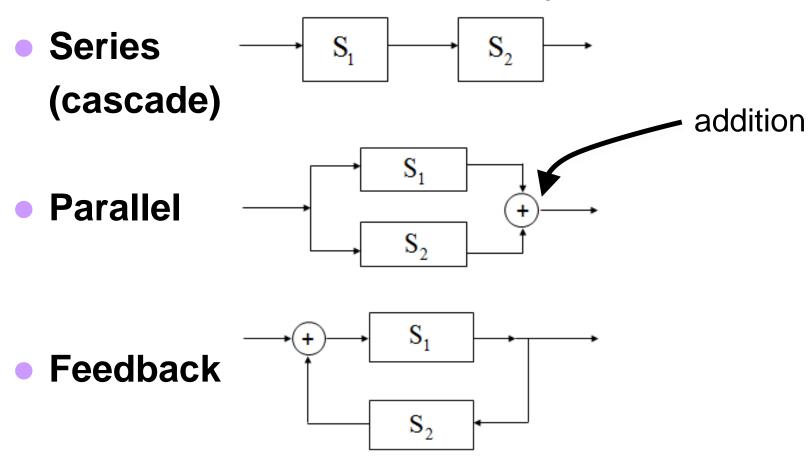
$$f(t) = M \frac{dv(t)}{dt} + \underbrace{Bv(t)}_{friction} + \underbrace{K \int_{-\infty}^{t} v(\tau) d\tau}_{spring \ force}.$$

This equation looks quite familiar, we just saw it earlier.

- Observation: different systems could be described by the same input/output relations
- In this course, we focus on the mathematical relation between input and output



Interconnection of Systems



System Properties: 1) Memoryless or With Memory

Memoryless: output at a given time depends only on the input at the same time

eg.
$$y[n] = (ax[n] - x^{2}[n])^{2}$$

With Memory

eg.
$$y[n] = \sum_{k=-\infty}^{n} x[k]$$

summer or accumulator

2) Invertability

invertible: distinct inputs lead to distinct outputs, i.e. an inverse system exits



No information loss

eg.
$$y[n] = \sum_{k=-\infty}^{n} x[k]$$

 $z[n] = y[n] - y[n-1] = x[n]$

3) Causality

- <u>Causality</u>: A system is causal if the output does not anticipate future values of the input, i.e., if the output at any time depends only on values of the input up to that time.
- All real-time physical systems are causal, because time only moves forward, and effect occurs after cause.
 (Imagine if you own a noncausal system whose output depends on tomorrow's stock price.)
 - Do not apply to spatially varying signals. (We can move both left and right, up and down.)
 - Do not apply to systems processing recorded (or nonrealtime) signals, e.g. taped sports games vs. live broadcast.

Causal or Non-causal?

•
$$y(t) = x^2(t-1)$$

•
$$y(t)=x(t+1)$$

•
$$y(t)=x(t) \cos(t+1)$$

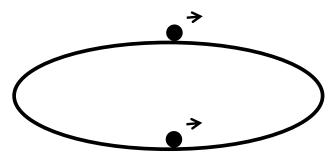
•
$$y[n]=(1/2)^{n+1} x^3[n-1]$$

4) Stability

 If the input to a stable system is bounded, the output must also be bounded.

• e.g.: S_1 : y(t) = t x(t)

 S_2 : y(t) = $e^{x(t)}$



5) Time Invariance (TI)

DT: A system x[n] → y[n] is TI if for any input x[n] and any time shift n₀

If
$$x[n] \rightarrow y[n]$$

then $x[n - n_0] \rightarrow y[n - n_0]$.

Similarly for CT time-invariant system

If
$$x(t) \rightarrow y(t)$$

then $x(t - t_0) \rightarrow y(t - t_0)$.

Time-invariant or Time-varying?

- Steps:
- 1) Calculate $y_1(t) \leftarrow x_1(t)$
- 2) Calculate $y_2(t) \leftarrow x_2(t) = x_1(t-t_0)$
- 3) Does $y_1(t-t_0)$ equal $y_2(t)$?

e.g.:
$$y[n] = \left(\frac{1}{2}\right)^{n+1} x^3[n-1]$$

$$\begin{cases} X_{1}[n] = X_{1}[n-n_{0}] \\ X_{2}[n] = Y_{1}[n-n_{0}] \\ = Y_{1}[n-n_{0}] \end{cases}$$

Now we can deduce something:

 If the input to a TI system is periodic, then the output is also periodic with the same period (Problem 1.43 (a)).

Proof: Suppose
$$x(t+T) = x(t)$$
 and $x(t) \rightarrow y(t)$

Then by TI
$$x(t+T) \rightarrow y(t+T)$$

But these are So these must be the same input! the same output, i.e., y(t) = y(t+T)

Linearity

Suppose $x_1(t) \rightarrow y_1(t)$ and $x_2(t) \rightarrow y_2(t)$, such system is linear, if

- 1) Additivity property: $x_1(t) + x_2(t) \rightarrow y_1(t) + y_2(t)$
- 2) Scaling (or homogeneity) property:

$$a \mathbf{x}_1(\mathbf{t}) \rightarrow a \mathbf{y}_1(\mathbf{t})$$

where a is a complex number

e.g.:
$$y(t) = 2 x(t)$$
 $y(t) = x^2(t)$

Linear system or not?

- Steps
- 1) Have $y_1(t)$ and $y_2(t)$ as output signals to $x_1(t)$ and $x_2(t)$
- 2) Have $y_3(t)$ as output signal to $x_3(t) = a x_1(t) + b x_2(t)$
- 3) Does $y_3(t)$ equal "a $y_1(t)$ + b $y_2(t)$ "?

More examples on textbook Read Example 1.17 ~ 1.20

Linearity (cont.)

Superposition

If
$$x_k[n] \xrightarrow{\text{System}} y_k[n] \text{ k=1,2,3,...}$$

Then
$$\sum_k a_k x_k[n] \xrightarrow{\text{System}} \sum_k a_k y_k[n]$$

 This property seems to be almost trivial now, but it is one of the most important ones

Linear Time-invariant (LTI) Systems

LTI: Linear + Time invariant

 A basic fact: If we know the response of an LTI system to some inputs, we actually know the response to many inputs.

Example: DT LTI System

