# **GP-based Model Predictive Control\***

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Abstract—Describe topic and relevance in a few sentences so that the reader is motivated to read the whole paper.

## I. INTRODUCTION

Model predictive control (MPC) is a popular control strategy which uses a dynamic plant model to obtain the control input that optimizes future reactions of the plant [4]. The performance of MPC depends highly on how well the model captures the dynamics of the plant [3]. But the identification of such an a priori model can be challenging and the dynamics of the plant could also change during the application [3], [7]. Therefore, a simple and fixed nominal model of the plant can be used in combination with a learned disturbance model. The disturbance model represents the error between the observed behaviour of the plant and the behaviour of the nominal model [7]. It can be modelled as a Gaussian Process (GP) regression which is a probabilistic, non-parametric model [4]. GPs have the advantage of characterizing the prediction uncertainities [4]. The mean estiamte of a GP can also be used to model the full dynamics of the plant and not only the model error. This approach was applied to a cart pole swing-up environment and an autonomous racing task in [8]. To reduce high computational costs they choose sparce spectrum GPs. Kocijan et al. [4] use an offline-identified GP model instead. Another alternative is the generation of local GPs (LGPs) where for each subspace of the GP input space different GPs are identified. While Nguyen-Tuong et al. [6] and Meier et al. [5] identify many LGPs, Ostafew et al. [7] compute one single LGP based on data within a sliding window. Other applications of GPs in a MPC framework are .

In this report we present the results of our project within the course "Statistical Learning and Stochastic Control". First, our literature research on GP-based MPC is summarized. Then a short introduction to the theory of MPC and GPs is given. Later, two examples which's implementation was part of the project are introduced in Section III and the results are discussed in Section IV.

We start by defining the used notation below.

Bold lowercase letters are used for vectors  $x \in \mathbb{R}^n$  and bold uppercase letters for matrices  $X \in \mathbb{R}^{n \times m}$ . The j-th column

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of a matrix X is denoted by  $[X]_{.,j}$  and the element in the i-th row and j-th column is  $[X]_{ij}$ . A diagonal matrix X with the diagonal elements  $x_{11}, x_{22}, \cdots, x_{nn}$  is represented by  $\operatorname{diag}(x_{11}, x_{22}, \cdots, x_{nn})$ . If  $w \sim \mathcal{N}(\mu, \Sigma)$  w is normal distributed with mean  $\mu$  and variance  $\Sigma$ .

Introduce topic and describe motivation and relevance of problem/topic.

In this paper we give an introduction to the results presented in paper(s) [2]. We present the main results, discuss ideas and illustrate the results with simulations.

Notation. Define notation.

## II. BACKGROUND

In this section a revision of GPs for our application and necessary background information on MPC are presented. The revision of GPs follows Kabzan et al. [3] as well as Rasmussen and Williams [9].

As mentioned in the Introduction, a GP is used to identify the disturbance  $d_{true}$ , which describes the error between the nominal plant dynamics  $f_{nom}$  and the true dynamics  $f_{true}$ . Thus, the true system equations are given by

$$x_{k+1} = f_{true}(x_k, u_k)$$

$$= f_{nom}(x_k, u_k) + B_d(d_{true}(z_k) + w),$$
(1)

with  $z_k = B_z x_k$  ( $B_z \in \mathcal{R}^{n_z \times n}$ ) and  $w \sim \mathcal{N}(\mathbf{0}, \Sigma_w)$  the gaussian measurement noise, where  $\Sigma_w = diag(\sigma_1^2, \cdots, \sigma_{n_d}^2)$ . The disturbance  $d = d_{true}(z_k) + w_k$  is identified using input and output data pairs  $(z_k, y_k = d_{true}(z_k) + w_k)$  which are saved in a dictionary  $\mathcal{D}$  of length m

$$\mathcal{D} = \{ \boldsymbol{Z} = [\boldsymbol{z_1}, \cdots, \boldsymbol{z_m}] \in \mathbb{R}^{n_z \times m},$$

$$\boldsymbol{Y} = [\boldsymbol{y_1}, \cdots, \boldsymbol{y_m}] \in \mathbb{R}^{n_d \times m} \}.$$
(2)

If  $m>m_{max}$ , the dictionary is updated by removing the oldest data pair.

Each output dimension  $a \in \{1, \dots, n_d\}$  is treated as a different GP with the kernel  $k^a$ , which results in the posterior distribution mean  $\mu^a(z)$  and variance  $\Sigma^a(z)$ 

$$\mu^{a}(z) = k_{zz}^{a} (K_{zz}^{a} + I\sigma_{a})^{-1} [Y]_{.,i},$$
 (3)

$$\Sigma^{a}(z) = k_{zz}^{a} - k_{zz}^{a} (K_{ZZ}^{a} + I\sigma_{a})^{-1} k_{Zz}^{a}$$
(4)

in dimension a for a test point z. The expressions  $k_{zz}^a$ ,  $k_{zz}^a$  and  $K_{Zz}^a$  are compact notations for  $k^a(z,z) \in \mathbb{R}$ ,  $[k_{zz}^a]_j = k^a(z,z_j)$ ,  $[k_{Zz}^a]_j = k^a(z_j,z)$  and  $[K_{zz}^a]_{ij} = k^a(z_i,z_j)$ , respectively. It holds  $k_{zz}^a = (k_{zz}^a)^T \in \mathbb{R}^m$ .

For each output dimension a we make use of the squared exponentional kernel given by

$$k^{a}(\boldsymbol{z}, \bar{\boldsymbol{z}}) = \boldsymbol{\sigma_{f,a}}^{2} \exp(-0.5(\boldsymbol{z} - \bar{\boldsymbol{z}})^{T} \boldsymbol{M}^{-1}(\boldsymbol{z} - \bar{\boldsymbol{z}})), \quad (5)$$

where  $\sigma_{f,a}^2$  and M are the squared output variance and the positive diagonal length scale matrix, respectively. The practical implementation of Gaussian process regression from page 19 of Rasmussen and Williams [9] which uses the Cholesky factorization to address the matrix inversion is applied. This results in the following algorithm

$$L^{a} = \operatorname{cholesky}(K(Z, Z) + \sigma_{n}(a, a)^{2}I)$$
 (6)

$$\alpha^{a} = L^{T} \setminus (L \setminus Y(:, a)), \tag{7}$$

where  $[K(Z, Z)]_{ij} = k(z_i, z_j)$  with  $z_i, z_j \in Z$ . Thus, the predictive mean and variance are given by

$$\mu_y^a = K(z, Z)^T \alpha^a \tag{8}$$

$$\Sigma_{u}^{a} = k(\boldsymbol{z}, \boldsymbol{z}) - \boldsymbol{v}^{aT} \boldsymbol{v}^{a}, \tag{9}$$

with  $v^a = L^a/K(z,Z)$ .

Combining the GPs of each dimension results in the multivariate GP approximation

$$d(z) \sim \mathcal{N}(\mu^d(z), \Sigma^d(z)),$$
 (10)

where the mean  $\mu^d(z)\in \mathcal{R}^{n_d}$  and the variance  $\mathbf{\Sigma}^d(z)\in \mathcal{R}^{n_d imes n_d}$  are given by

$$\mu^{d}(z) = [\mu^{1}(z), \cdots, \mu^{n_{d}}(z)]^{T},$$
 (11)

$$\Sigma^{d}(z) = \operatorname{diag}([\Sigma^{1}(z), \cdots, \Sigma^{n_{d}}(z)]). \tag{12}$$

The model considered for control including the identified disturbance d is given by

$$x_{k+1} = f_{nom}(x_k, u_k) + B_d d(z_k). \tag{13}$$

Model predictive control (MPC), receding horizon control or moving horizon control are all names for a control strategy which predicts the future dynamic behaviour within a finite prediction horizion and chooses the control input such that a perfomance functional is minimized [1]. Since the predicted behaviour is not equal to the system behaviour due to disturbances and model-plant mismatch, only the first input of the computed control input sequence is applied [1]. Using the new measurement one sampling time later, the procedure is repeated to find a new control sequence within the receding horizon.

Evaluating the GP model  $d(x_k, u_k)$  results in a stochastic distribution, which leads to a stochastic distribution of the state x. The distribution at each prediction step is assumed to be Gaussian with  $x_k \sim \mathcal{N}(\mu_k^x, \Sigma_k^x)$ . To evaluate the uncertainty over the prediction horizon N the mean and variance of x are propagated forward which results in

$$\boldsymbol{\mu_{k+1}^x} = f_{nom}(\boldsymbol{\mu_k^x}, \boldsymbol{u_k}) + \boldsymbol{B_d} \boldsymbol{\mu^d}(\boldsymbol{\mu_k^x}, \boldsymbol{u_k}) \qquad (14)$$

$$\Sigma_{k+1}^x = ????. \tag{15}$$

For more information on the propagation see ???? Appendix. Thus, the MPC problem is given by

$$\min_{\tilde{u}} \left( \sum_{k=1}^{N-1} f_o(t_k, \boldsymbol{\mu_k^x}, \boldsymbol{\Sigma_k^x}, \tilde{\boldsymbol{u}_k}, r(t)) \right) + f_{end}(t_N, \boldsymbol{\mu_N^x}, \boldsymbol{\Sigma_N^x}, r(t_N))$$
(16)

 $\boldsymbol{\mu_0^x} = \boldsymbol{x}(t) \tag{17}$ 

$$\mu_{k\perp 1}^x = f_{nom}(\mu_k^x, \tilde{u}_k) + B_d \mu_k^d \tag{18}$$

$$\Sigma_{k+1}^{x} = \nabla^{T} f_{nom}(\mu_{k}^{x}, \tilde{u}_{k}), \Sigma_{xdw} \nabla f_{nom}(\mu_{k}^{x}, \tilde{u}_{k}),$$
(19)

$$\mu_k^x \in \mathcal{X},$$
 (20)

$$u_k \in \mathcal{U},$$
 (21)

where  $\tilde{\boldsymbol{u}}(\dot):[t,t+N]\to\mathcal{U}$  and  $\tilde{u}_k$  refers to the k-th element of  $\tilde{\boldsymbol{u}}$ . The reference trajectory is given by r(t). The input and state constraint sets  $\mathcal{U}$  and  $\mathcal{X}$  can be defined as box constraints according to Allgöwer et al. [1]

$$\mathcal{U} = \{ u \in \mathcal{R}^p | \boldsymbol{u_{min}} \le \boldsymbol{u} \le \boldsymbol{u_{max}} \}, \tag{22}$$

$$\mathcal{X} = \{ u \in \mathcal{R}^n | \boldsymbol{x_{min}} \le \boldsymbol{x} \le x_{max} \}. \tag{23}$$

According to the control problem, a cost function  $f_o$  and the final cost  $f_{end}$  are defined. The cost functions, which were used during this project, are further described for each example in Section III. Solving the above optimal control problem at each sampling time  $t=j\delta$  with  $j=0,1,\cdots$  results in an optimal control sequence  $\tilde{u}^*$  for each sample time, where only value is applied  $u^*=\tilde{u}^*1$ .

# III. MAIN RESULTS

Two examples were considered during this project. An inverted pendulum and autonomous driving based on a single track model. Both examples are introduced in this chapter.

# A. Inverted Pendulum

The motion of the 2-dimensional inverted pendulum is described by

$$(M_c + M_p)\ddot{x} + b\dot{x} + \frac{1}{2}M_pl\ddot{\theta}\cos\theta - \frac{1}{2}M_pl\dot{\theta}^2\sin\theta = F$$
(24)

$$(I + M_p \left(\frac{l}{2}\right)^2)\ddot{\theta} - \frac{1}{2}M_p gl\sin\theta + M_p l\ddot{x}\cos\theta = 0.$$
(25)

A schematic drawing of the inverted pendulum including the parameters can be seen in Figure 1. The standard gravity on the surface of the earth g is set to be g=9.8. The objective of this system is to achieve an upright pendulum position by applying force to the cart. The nonlinear state space model with the states  $[x, \dot{x}, \theta, \dot{\theta}]$  is given by

$$\begin{bmatrix} \dot{x} \\ \ddot{x} \\ \dot{\theta} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} \dot{x} \\ \dot{\theta} \end{bmatrix} \tag{26}$$

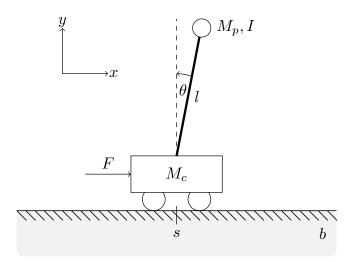


Fig. 1. A schematic drawing of the inverted pendulum on a carriage. The mass of the carriage and the pole are given by  $M_c$  and  $M_p$ , respectively. The pole is defined by its length l and its moment of inertia I. b is the friction coefficient between the carriage and the floor. The states of the state-space model are the carriage position s, its derivative  $\dot{s}$ , the pole angle with the vertical  $\theta$  and its derivative  $\dot{\theta}$ . F is the applied force.

The above state space model defines the nominal dynamics  $f_{nom}$  of the inverted pendulum. A true disturbance  $d_{true}$  is added to the nominal model to form the true dynamics  $f_{true}$ . Representing a defect in the joint connecting the cart and the pole the disturbance is given by

$$d_{true} = 0.1\theta - 0.01\dot{\theta} + 0.0524$$
  
= 0.1 $\mathbf{z}(\mathbf{1}) - 0.01\mathbf{z}(\mathbf{2}) + 0.0524$ , (27)

with  $oldsymbol{z} = oldsymbol{B_z} oldsymbol{x}$  and

$$\boldsymbol{B_z} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}. \tag{28}$$

The disturbance only affects  $\theta$  which is implemented by setting  $\boldsymbol{B_d} = [0,0,1,0]$ . In our example the parameters of the inverted pendulum were set to  $M_c = 5$ ,  $M_p = 2$ , I = 0.6, l = 3 and b = 0.1. The implemented MPC has a prediction horizon N of N = 10 and the number of iterations to optimize the cost function at each time step is limited to ten. The control formulation aims to stabilize the pole at  $\theta = \pi$ . Therefore, the cost function  $f_c$  is defined to

$$f_c = (C_k \mu_x - r(t))^T Q (C_k \mu_x - r(t)) + Ru^2, \quad (29)$$

where Q and R are weight matrices/values penalizing the deviation between a selection of mean states  $C_k \mu_x$  and the reference signal as well as the amplitude of the input force F, respectively. The reference r is defined to  $\dot{x}=0$ ,  $\theta=0$  and  $\dot{\theta}=0$ .

The final cost is defined by

$$f_{end} = (C_k \mu_x - r(t))^T Q (C_k \mu_x - r(t)). \tag{30}$$

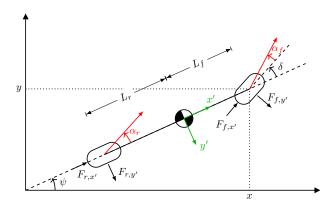


Fig. 2. A schematic drawing of the single-track model. The vehicle mass and the vehicle moment of inertia (yaw axis) are given by M and I, respectively. The distance from the front wheel as well as from the rear wheel to the center of mass are defined by  $L_f$  and  $L_f$ , respectively.  $\psi$  is the yaw angle. The forces which act on the rear and front wheel in longitudinal and lateral direction are named  $F_{r,x'}$ ,  $F_{r,y'}$ ,  $F_{f,x'}$  and  $F_{f,y'}$ . The rear-slip angle is  $\alpha_r$  and the front-slip angle is  $\alpha_f$ 

#### B. Single-Track System

The single-track model is a simplified version of the more complex twin-track model, where the front and rear axis only carry one wheel each instead of two. A schematic drawing of the single-track model including the relevant parameters can be seen in Figure 2.

The state-space representation of the single-track model is given by

$$\dot{I}_x = V_{x'}\cos(\psi) - V_{y'}\sin(\psi) \tag{31}$$

$$\dot{I}_y = V_{x'}\sin(\psi) + V_{y'}\cos(\psi) \tag{32}$$

$$\psi = \psi \tag{33}$$

$$\dot{V}_{x'} = \frac{1}{M} (F_{r,x'} + F_{f,x'} * \cos(\delta) - F_{f,y'} \sin(\delta) + V_{y'} \dot{\psi})$$
(34)

$$\dot{V}_{y'} = \frac{1}{M} (F_{r,y'} + F_{f,x'} \sin(\delta) + F_{f,y'} \cos(\delta) - V_{y'} \dot{\psi})$$
(35)

$$\ddot{\psi} = \frac{1}{I} (F_{f,y'} L_f \cos(\delta) + F_{f,x'} L_f \sin(\delta) - F_{r,y'} L_r)$$
(36)

$$\dot{d_{track}} = v_{track},\tag{37}$$

where the steering angle  $\delta$ , the wheel torque gain T and the track center line velocity  $v_{track}$  are inputs. The wheel torque gain is bounded by -1 (maximum deceleration) and 1 (maximum acceleration). Additionally, the steering angle is bounded between -30 degree and 30 degree. The wheel forces in vehicle coordinates are defined to

$$F_{r,x'} = \zeta F_W \tag{38}$$

$$F_{r,u'} = c_r \alpha_r \tag{39}$$

$$F_{f,x'} = (1 - \zeta)F_W$$
 (40)

$$F_{f,y'} = c_f \alpha_f, \tag{41}$$

where  $\zeta=0.5$  is longitudinal wheel torque distribution and  $c_f=c_r=14000$  front and rear cornering stiffness. The is total wheel force  $F_W$  given by

$$F_W = T((T \ge 0)4000 + (T \le 0)8000 sign(V_{x'})), \quad (42)$$

where the factor 4000 allows 1g (1 gM) acceleration and the factor 8000 2g (2 gM) breaking. The wheel slip angles  $\alpha_r$  and  $\alpha_f$  are defined by

$$\alpha_r = \operatorname{atan2}(V_{y'} - L_r \dot{\psi}, V_{x'}) \tag{43}$$

$$\alpha_f = \operatorname{atan2}(V_{y'} + L_f \dot{\psi}, V_{x'}) - \delta. \tag{44}$$

Ideas, theorems, proofs and discussions ..... Ideas, theorems, proofs and discussions .....

#### IV. EXAMPLES

Show and discuss simulation examples etc....

# V. CONCLUSIONS

Summarize the main points (with more details than in the preceding introduction). The paper should not be between 4 and 8 pages.

# **APPENDIX**

Appendixes should appear before the acknowledgment.

# **ACKNOWLEDGMENT**

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