

Differential Equations and Linear Algebra

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All errors, typographical and substantive, and other offenses, are entirely my own.

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## 2 - Second Order Equations

### 2.1 - Second Derivatives in Science and Engineering

**Question: 2.1.1**

Find a cosine and a sine that solve  $d^2y/dt^2 = -9y$ . This is a second order equation so we expect *two constants*  $C$  and  $D$  (from integrating twice):

$$\text{Simple harmonic motion } y(t) = C \cos(\omega t) + D \sin(\omega t)$$

What is  $\omega$ ? If the system starts from rest (this means  $dy/dt = 0$  at  $t = 0$ ), which constant  $C$  or  $D$  will be zero?

Differentiating  $y(t)$  twice, we get a  $\omega^2$  term. Making the necessary substitutions, we can see that  $\omega^2 = 9$ , implying  $\omega = 3$ . Thus we have  $y = \sin(3t)$  and  $y = \cos(3t)$ . The constants  $C$  and  $D$  are determined by the initial conditions. Assuming the system starts at rest, we must have

$$\frac{dy}{dt} = -3C \sin(3t) + 3D \cos(3t) = 0$$

which implies that  $dy/dt_{t=0} = 3D = 0$ , thus  $D = 0$ .

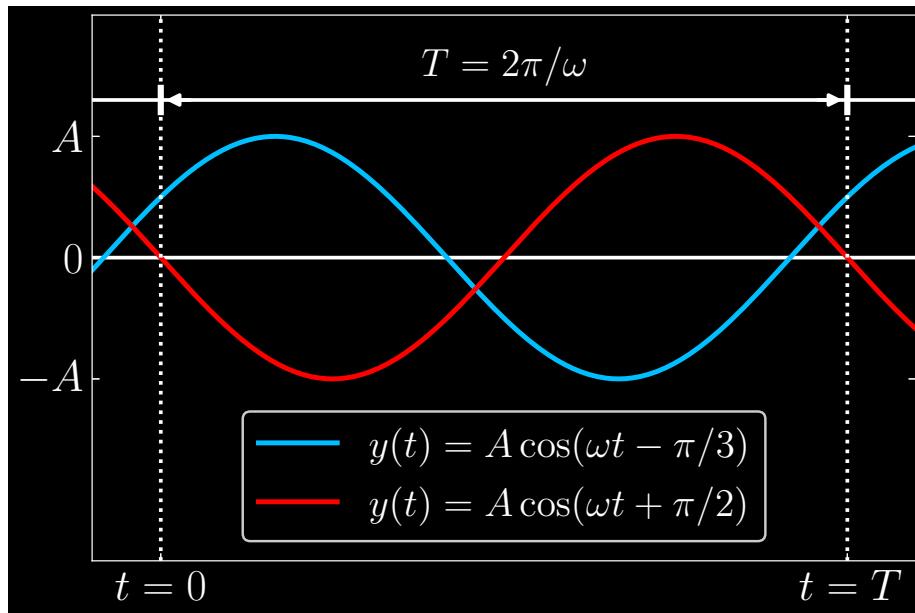
**Question: 2.1.2**

In Problem 1, which  $C$  and  $D$  will give the starting values  $y(0) = 0$  and  $y'(0) = 1$ ?

We have  $y(0) = C = 0$  and  $y'(0) = 3D = 1$ , or  $D = 1/3$ .

**Question: 2.1.3**

Draw Figure 2.3 to show simple harmonic motion  $y = A \cos(\omega t - \alpha)$  with phases  $\alpha = \pi/3$  and  $\alpha = -\pi/2$ .

**Question: 2.1.4**

Suppose the circle in Figure 2.4 has radius 3 and circular frequency  $f = 60$  Hertz. If the moving point starts at the angle  $-45^\circ$ , find its  $x$ -coordinate  $A \cos(\omega t - \alpha)$ . The phase lag is  $\alpha = 45^\circ$ . When does the point first hit the  $x$ -axis?

The circular motion of the point is expressed by the sinusoidal

$$3 \cos\left(120\pi t - \frac{\pi}{4}\right)$$

Note that since the frequency is  $f = 60$  Hertz, the angular frequency is  $2\pi \cdot 60 = 120\pi$  radians  $s^{-1}$ . The point hits the  $x$ -axis when the argument of the cosine is zero, namely at

$$120\pi t - \frac{\pi}{4} = 0 \implies t = \frac{1}{480}s$$

**Question: 2.1.5**

If you drive at 60 miles per hour on a circular track with radius  $R = 3$  miles, what is the time  $T$  for one complete circuit? Your circular frequency is  $f = \underline{\hspace{2cm}}$  and your angular frequency is  $\omega = \underline{\hspace{2cm}}$  (with what units?). The period is  $T$ .

Using dimensional analysis, the period is

$$T = \frac{1 \text{ hr}}{60 \text{ mi}} (3 \text{ mi}) = \frac{1}{20} \text{ hr}$$

The circular frequency is

$$f = \frac{60 \text{ mi}}{\text{hr}} \frac{1 \text{ hr}}{3600 \text{ s}} \frac{1 \text{ cycle}}{3 \text{ mi}} = \frac{1}{180} \text{ s}^{-1}$$

with angular frequency

$$\omega = 2\pi f = \frac{\pi}{90} \text{ rad s}^{-1}$$

### Question: 2.1.6

The total energy  $E$  in the oscillating spring-mass system is

$$\begin{aligned} E &= \text{kinetic energy in mass} + \text{potential energy in spring} \\ &= \frac{m}{2} \left( \frac{dy}{dt} \right)^2 + \frac{k}{2} y^2 \end{aligned}$$

Compute  $E$  when  $y = C \cos(\omega t) + D \sin(\omega t)$ . The energy is constant!

Given  $y(t)$ , we have first time-derivative

$$\frac{dy}{dt} = -\omega C \sin(\omega t) + \omega D \cos(\omega t)$$

squaring gives us

$$\left( \frac{dy}{dt} \right)^2 = \omega^2 (C^2 \sin^2(\omega t) - 2CD \sin(\omega t) \cos(\omega t) + D^2 \cos^2(\omega t))$$

Lastly, the  $y^2$  term is

$$y^2 = C^2 \cos^2(\omega t) + 2CD \sin(\omega t) \cos(\omega t) + D^2 \sin^2(\omega t)$$

Combining our ingredients, with  $\omega = \sqrt{k/m}$ , observe that energy  $E$  reduces to a constant:

$$\begin{aligned} E &= \frac{m}{2} \left( \frac{dy}{dt} \right)^2 + \frac{k}{2} y^2 \\ &= \frac{m}{2} \left( \frac{k}{m} \right) (C^2 \sin^2(\omega t) - 2CD \sin(\omega t) \cos(\omega t) + D^2 \cos^2(\omega t)) \\ &\quad + \frac{k}{2} (C^2 \cos^2(\omega t) + 2CD \sin(\omega t) \cos(\omega t) + D^2 \sin^2(\omega t)) \\ &= C^2 + D^2 \end{aligned}$$

**Question: 2.1.7**

Another way to show that the total energy  $E$  is constant:

Multiply  $my'' + ky = 0$  by  $y'$ . Then integrate  $my'y''$  and  $kyy'$ .

Take the first term and integrate:

$$m \int y' y'' dt$$

By integration by parts, let

$$u = \frac{dy}{dt}, \quad \frac{du}{dt} = \frac{d^2y}{dt^2} \quad \Rightarrow \quad du = \frac{d^2y}{dt^2} dt$$

Then we can rewrite the first term as

$$m \int u du = \frac{m}{2} u^2 + C = \frac{m}{2} (y')^2 + C$$

as for the second term:

$$k \int yy' dt = k \int y \frac{dy}{dt} dt = k \int y dy = \frac{k}{2} y^2 + C$$

Summing the two pieces (sans the constants) restores the original energy function:

$$E = \frac{m}{2} \left( \frac{dy}{dt} \right)^2 + \frac{k}{2} y^2$$

and since the derivative of a constant is zero, it must be the case that  $E$  is a constant.

**Question: 2.1.8**

A **forced oscillation** has another term in the equation and  $A \cos(\omega t)$  in the solution:

$$\frac{d^2y}{dt^2} + 4y = F \cos(\omega t) \quad \text{has} \quad y = C \cos(2t) + D \sin(2t) + A \cos(\omega t)$$

- (a) Substitute  $y$  into the equation to see how  $C$  and  $D$  disappear (they give  $y_n$ ). Find the forced amplitude  $A$  in the particular solution  $y_p = A \cos(\omega t)$ .
- (b) In case  $\omega = 2$  (forcing frequency = natural frequency), what answer does your formula give for  $A$ ? The solution formula for  $y$  breaks down in this case.

The second time-derivative is

$$\frac{d^2y}{dt^2} = -4C \cos(2t) - 4D \sin(2t) - \omega^2 A \cos(\omega t)$$

(a) We have

$$\frac{d^2y}{dt^2} + 4y = (4 - \omega^2)A \cos(\omega t) = F \cos(\omega t)$$

implying  $A = \frac{F}{4 - \omega^2}$ .

(b) When  $\omega = 2$ ,  $A$  is undefined.

**Question: 2.1.9**

Following Problem 8, write down the complete solution  $y_n + y_p$  to the equation

$$m \frac{d^2y}{dt^2} + ky = F \cos(\omega t) \quad \text{with } \omega \neq \omega_n = \sqrt{k/m} \quad (\text{no resonance})$$

The answer  $y$  has free constants  $C$  and  $D$  to match  $y(0)$  and  $y'(0)$  ( $A$  is fixed by  $F$ ).

Per Problem 8, we have solution

$$y = \underbrace{C \cos\left(\sqrt{\frac{k}{m}}t\right)}_{y_n} + \underbrace{D \sin\left(\sqrt{\frac{k}{m}}t\right)}_{y_p} + \underbrace{\frac{F}{k - m\omega^2} \cos(\omega t)}$$

All this involves is dividing the equation through by  $m$ , understanding that the angular frequency of the sinusoids in the null solution is the square root of  $k/m$ , and making the changes to our formula for  $A$  accordingly.

**Question: 2.1.10**

Suppose Newton's Law  $F = ma$  has the force  $F$  in the *same* direction as  $a$ :

$$my'' = +ky \quad \text{including } y'' = 4y$$

Find two possible choices of  $s$  in the exponential solutions  $y = e^{st}$ . The solution is not sinusoidal and  $s$  is real and the oscillations are gone. Now  $y$  is unstable.

Substituting in  $y$ , we find

$$ms^2 e^{st} = ke^{st}$$

This forces

$$s = \pm \sqrt{\frac{k}{m}}$$

**Question: 2.1.11**

Here is a *fourth* order equation:  $d^4y/dt^4 = 16y$ . Find *four* values of  $s$  that give exponential solutions  $y = e^{st}$ . You could expect four initial conditions on  $y$ :  $y(0)$  is given along with what three other conditions?

Equivalently, we find the four complex roots of 16:

$$s^4 = 16$$

which are  $s = \pm 2, \pm 2i$ .

**Question: 2.1.12**

To find a particular solution to  $y'' + 9y = e^{ct}$ , I would look for a multiple  $y_p(t) = Ye^{ct}$  of the forcing function. What is that number  $Y$ ? When does your formula give  $Y = \infty$ ? (Resonance needs a new formula for  $Y$ .)

Let  $y_p = Ye^{ct}$ . Substituting, we find

$$Yc^2e^{ct} + 9Ye^{ct} = e^{ct}$$

Solving for  $Y$  yields  $Y = \frac{1}{c^2 + 9}$ . When  $c \rightarrow \pm 3$ , we have resonance, and  $Y \rightarrow \infty$ .

**Question: 2.1.13**

In a particular solution  $y = Ae^{i\omega t}$  to  $y'' + 9y = e^{i\omega t}$ , what is the amplitude  $A$ ? The formula blows up when the forcing frequency  $\omega =$  what natural frequency?

Substituting, we derive

$$-\omega^2 Ae^{i\omega t} + 9Ae^{i\omega t} = e^{i\omega t}$$

which gives us  $A = \frac{1}{9 - \omega^2}$ . Resonance is when  $\omega = 3$ .

**Question: 2.1.14**

Equation (10) says that the tangent of the phase angle is  $\tan(\alpha) = y'(0)/\omega y(0)$ . First, check that  $\tan(\alpha)$  is dimensionless when  $y$  is in meters and time is in seconds. Next, if that ratio is  $\tan(\alpha) = 1$ , should you choose  $\alpha = \pi/4$  or  $\alpha = 5\pi/4$ ? Answer:

Separately you want  $R \cos(\alpha) = y(0)$  and  $R \sin(\alpha) = y'(0)/\omega$

If those right hand sides are positive, choose the angle  $\alpha$  between 0 and  $\pi/2$ .

If those right hand sides are negative, add  $\pi$  and choose  $\alpha = 5\pi/4$ .

*Question:* If  $y(0) > 0$  and  $y'(0) < 0$ , does  $\alpha$  fall between  $\pi/2$  and  $\pi$  or between  $3\pi/2$  and  $2\pi$ ? If you plot the vector from  $(0, 0)$  to  $(y(0), y'(0)/\omega)$ , its angle is  $\alpha$ .

As  $y(0) > 0$  and  $y'(0) < 0$  requires positive cosine and negative sine,  $\alpha$  falls between  $3\pi/2$  and  $2\pi$ .

**Question: 2.1.15**

Find a point on the sine curve in Figure 2.1 where  $y > 0$  but  $v = y' < 0$  and also  $a = y'' < 0$ . The curve is sloping down and bending down.

Find a point where  $y < 0$  but  $y' > 0$  and  $y'' > 0$ . The point is below the  $x$ -axis but the curve is sloping \_\_\_\_ and bending \_\_\_\_.

One area corresponding to the first set of conditions is  $\pi/2 < t < \pi$ . As for the second, we have  $3\pi/2 < t < 2\pi$ .

**Question: 2.1.16**

- (a) Solve  $y'' + 100y = 0$  starting from  $y(0) = 1$  and  $y'(0) = 10$ .  
**(This is  $y_n$ .)**
- (b) Solve  $y'' + 100y = \cos(\omega t)$  with  $y(0) = 0$  and  $y'(0) = 0$ .  
**(This can be  $y_p$ .)**

- (a) Let  $y = c_1 \cos(10t) + c_2 \sin(10t)$ . Then  $y(0) = c_1 = 1$  and  $y'(0) = 10c_2 = 10$  implies  $c_2 = 1$ . Then the null solution is

$$y_n = \cos(10t) + \sin(10t)$$

- (b) Let  $y_p = R \cos(\omega t)$ . Substitute to find

$$-\omega^2 R \cos(\omega t) + 100R \cos(\omega t) = \cos(\omega t)$$

Isolate  $R$  to derive

$$R = \frac{1}{100 - \omega^2}$$

The solution to this set of initial conditions requires  $y(0) = 0$  and  $y'(0) = 0$ . Begin with

$$y(t) = c_1 \cos(10t) + c_2 \sin(10t) + \frac{1}{100 - \omega^2} \cos(\omega t)$$

From the first condition, we have  $c_1 = -\frac{1}{100 - \omega^2}$ . From the second, we have  $c_2 = 0$ . The full solution is

$$y(t) = \frac{1}{100 - \omega^2} (\cos(\omega t) - \cos(10t))$$

**Question: 2.1.17**

Find a particular solution  $y_p = R \cos(\omega t - \alpha)$  to  $y'' + 100y = \cos(\omega t) - \sin(\omega t)$ .

Substituting  $y_p$  gives us

$$\begin{aligned} & -\omega^2 R \cos(\omega t - \alpha) + 100R \cos(\omega t - \alpha) \\ &= (100R - \omega^2 R)[\cos(\omega t) \cos(\alpha) + \sin(\omega t) \sin(\alpha)] \end{aligned}$$

This implies that

$$\begin{aligned} R \cos(\alpha)(100 - \omega^2) &= 1 \\ R \sin(\alpha)(100 - \omega^2) &= -1 \end{aligned}$$

enabling us to conclude that  $\alpha = 7\pi/4$ , and amplitude

$$R = \frac{\sqrt{2}}{100 - \omega^2}$$

Ergo, the particular solution is

$$y_p = \frac{\sqrt{2}}{100 - \omega^2} \cos\left(\omega t - \frac{7\pi}{4}\right)$$

**Question: 2.1.18**

Simple harmonic motion also comes from a linear pendulum (like a grandfather clock). At time  $t$ , the height is  $A \cos(\omega t)$ . What is the frequency  $\omega$  if the pendulum comes back to the start after 1 second? The period does not depend on the amplitude (a large clock or a small metronome or the movement in a watch can all have  $T = 1$ ).

The angular frequency is  $2\pi \cdot f = 2\pi$  radians s<sup>-1</sup>.

**Question: 2.1.19**

If the phase lag is  $\alpha$ , what is the time lag in graphing  $\cos(\omega t - \alpha)$ ?

Put differently, we want to find the value of  $t'$  such that we are able to restore  $\omega t$  as the cosine's argument. If we have

$$t' = t + \alpha/\omega$$

then we get

$$\cos(\omega t' - \alpha) = \cos(\omega(t + \alpha/\omega) - \alpha) = \cos(\omega t)$$

Thus the time lag term is  $\alpha/\omega$ .

**Question: 2.1.20**

What is the response  $y(t)$  to a delayed impulse if  $my'' + ky = \delta(t - T)$ ?

The full solution will be a factor of the step function, given by:

$$y(t) = \int_0^{t-T} \frac{\sin(\omega_n(t - T - s))}{m\omega_n} \delta(s) ds = \frac{\sin(\omega_n(t - T))}{m\omega_n} H(t - T)$$

Intuitively, when  $t \leq T$ , the right-hand side vanishes. No impulse is imparted, and thus there is no response. But when we are at time  $t \geq T$  – after the threshold – the response kicks in.

**Question: 2.1.21**

(Good challenge) Show that  $y = \int_0^t g(t-s) f(s) ds$  has  $my'' + ky = f(t)$ .

1. Why is  $y' = \int_0^t g'(t-s) f(s) ds + g(0) f(t)$ ? Notice the two  $t$ 's in  $y$ .
2. Using  $g(0) = 0$ , explain why  $y'' = \int_0^t g''(t-s) f(s) ds + g'(0) f(t)$ .
3. Now use  $g'(0) = 1/m$  and  $mg'' + kg = 0$  to confirm  $my'' + ky = f(t)$ .

Use the Leibniz integral rule:

$$\frac{d}{dt} \left( \int_0^t g(t-s) f(s) ds \right) = g(0) f(t) + \int_0^t \frac{\partial}{\partial t} g(t-s) f(s) ds$$

- (1) By above, applying the partial derivative inside the second term yields the desired result.
- (2) One more application of the Leibniz rule gives us

$$y'' = \int_0^t g''(t-s) f(s) ds + g(0) f'(t) + g'(0) f(t)$$

With the premise  $g(0) = 0$ , the second term vanishes and gives us the expected result.

(3) Derive

$$my'' + ky = \int_0^t [mg''(t-s) + kg(t-s)] f(s) ds + f(t) = f(t)$$

where the last equality follows by appealing to the nullity of  $g(t)$ .

**Question: 2.1.22**

With  $f = 1$  (direct current has  $\omega = 0$ ) verify that  $my'' + ky = 1$  for this  $y$ :

**Step response**

$$y(t) = \int_0^t \frac{\sin(\omega_n(t-s))}{m\omega_n} \cdot 1 ds = y_p + y_n = \frac{1}{k} - \frac{1}{k} \cos(\omega_n t)$$

We have second derivative

$$y''(t) = \frac{\omega_n^2}{k} \cos(\omega_n t)$$

Since  $\omega_n = \sqrt{k/m}$ ,  $my''(t) = \cos(\omega_n t)$ . Ergo, we have

$$my'' + ky = \cos(\omega_n t) + k \left[ \frac{1}{k} - \frac{1}{k} \cos(\omega_n t) \right] = 1$$

**Question: 2.1.23**

(Recommended) For the equation  $d^2y/dt^2 = 0$  find the null solution. Then for  $d^2g/dt^2 = \delta(t)$  find the fundamental solution (start the null solution with  $g(0) = 0$  and  $g'(0) = 1$ ). For  $y'' = f(t)$  find the particular solution using formula (16).

Integrating twice, the null solution is

$$y(t) = C_1 t + C_2$$

To find the fundamental solution  $g(t)$ , imposing the initial conditions  $g(0) = 0$  and  $g'(0) = 1$  forces  $C_1 = 1$  and  $C_2 = 0$ , thus

$$g(t) = t$$

Lastly, if  $y'' = f(t)$ , then

$$y_p(t) = \int_0^t (t-s) f(s) ds$$

**Question: 2.1.24**

For the equation  $d^2y/dt^2 = e^{i\omega t}$  find a particular solution  $y = Y(\omega) e^{i\omega t}$ . Then  $Y(\omega)$  is the frequency response. Note the “resonance” when  $\omega = 0$  with the null solution  $y_n = 1$ .

One particular solution is

$$y(t) = -\frac{e^{i\omega t}}{\omega^2}$$

meaning that  $Y(\omega) = -1/\omega^2$ . When  $\omega = 0$ , we have resonance, and this particular solution breaks down.

**Question: 2.1.25**

Find a particular solution  $Y e^{i\omega t}$  to  $my'' - ky = e^{i\omega t}$ . The equation has  $-ky$  instead of  $ky$ . What is the frequency response  $Y(\omega)$ ? For which  $\omega$  is  $Y$  infinite?

We have

$$\frac{d^2}{dt^2}(Y e^{i\omega t}) = -Y \omega^2 e^{i\omega t}$$

Substituting and canceling the  $e^{i\omega t}$  terms, we get

$$-mY\omega^2 - kY = 1$$

Implying

$$Y(\omega) = -\frac{1}{m\omega^2 + k}$$

If we have

$$\omega = i\sqrt{\frac{k}{m}}$$

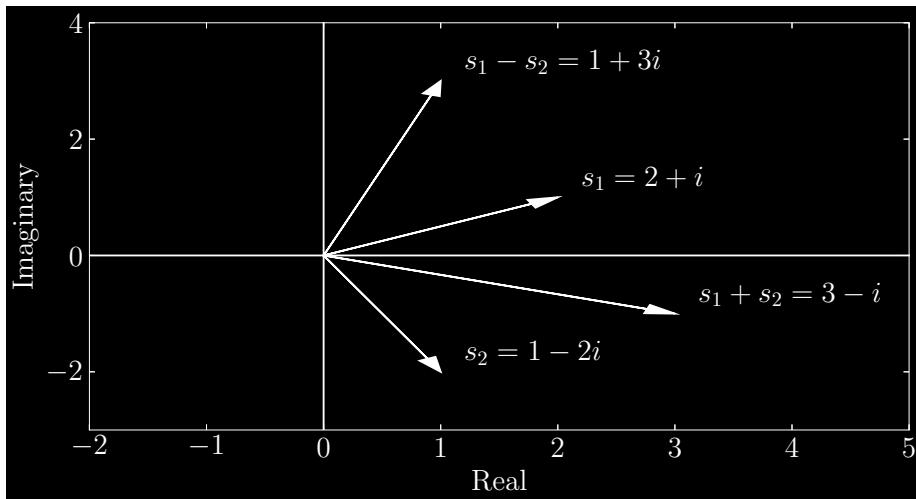
then  $Y(\omega)$  diverges to infinity – meaning all real frequencies will not lead to this!

## 2.2 - Key Facts About Complex Numbers

### Question: 2.2.1

Mark the numbers  $s_1 = 2 + i$  and  $s_2 = 1 - 2i$  as points in the complex plane. (The plane has a real axis and an imaginary axis.) Then mark the sum  $s_1 + s_2$  and the difference  $s_1 - s_2$ .

Arithmetic involving complex numbers simply is vector superposition:



### Question: 2.2.2

Multiply  $s_1 = 2 + i$  times  $s_2 = 1 - 2i$ . Check absolute values:  $|s_1||s_2| = |s_1s_2|$ .

We have

$$(2 + i)(1 - 2i) = 2 - 4i + i + 2 = 4 - 3i$$

with  $|s_1| = |s_2| = \sqrt{5}$ ,  $|s_1s_2| = \sqrt{16 + 9} = 5$ , ascertaining  $|s_1||s_2| = |s_1s_2|$ .

### Question: 2.2.3

Find the real and imaginary parts of  $1/(2 + i)$ .

Multiply by  $(2 - i)/(2 - i)$ :

$$\frac{1}{2+i} \frac{2-i}{2-i} = \frac{2-i}{|2+i|^2} = ?$$

The denominator is 5, so we have

$$\underbrace{\frac{2}{5}}_{\text{real}} - \underbrace{\frac{i}{5}}_{\text{imaginary}}$$

**Question: 2.2.4**

*Triple angles* Multiply equation (10) by another  $e^{i\theta} = \cos(\theta) + i \sin(\theta)$  to find formulas for  $\cos(3\theta)$  and  $\sin(3\theta)$ .

Derive

$$\begin{aligned} (\cos(\theta) + i \sin(\theta))^3 &= (\cos^2(\theta) - \sin^2(\theta) + 2i \cos(\theta) \sin(\theta))(\cos(\theta) + i \sin(\theta)) \\ &= \cos^3(\theta) - \sin^2(\theta) \cos(\theta) + 2i \cos^2(\theta) \sin(\theta) \\ &\quad + i \cos^2(\theta) \sin(\theta) - i \sin^3(\theta) - 2 \cos(\theta) \sin^2(\theta) \\ &= \underbrace{\cos^3(\theta) - 3 \sin^2(\theta) \cos(\theta)}_{\text{real}} + i \underbrace[3 \cos^2(\theta) \sin(\theta) - \sin^3(\theta)]_{\text{imaginary}} \end{aligned}$$

**Question: 2.2.5**

*Addition formulas* Multiply  $e^{i\theta} = \cos(\theta) + i \sin(\theta)$  times  $e^{i\phi} = \cos(\phi) + i \sin(\phi)$  to get  $e^{i(\theta+\phi)}$ . Its real part is  $\cos(\theta + \phi) = \cos(\theta) \cos(\phi) - \sin(\theta) \sin(\phi)$ . What is its imaginary part  $\sin(\theta + \phi)$ ?

Derive

$$\begin{aligned} [\cos(\theta) + i \sin(\theta)][\cos(\phi) + i \sin(\phi)] &= \underbrace{\cos(\theta) \cos(\phi) - \sin(\theta) \sin(\phi)}_{\text{real}} \\ &\quad + i \underbrace{[\cos(\theta) \sin(\phi) + \sin(\theta) \cos(\phi)]}_{\text{imaginary}} \end{aligned}$$

**Question: 2.2.6**

Find the real part and the imaginary part of each cube root of 1. Show directly that the three roots add to zero, as equation (11) predicts.

The cube roots, in polar form, are given by

$$\begin{aligned} e^{i2\pi/3} &= -\frac{1}{2} + i \frac{\sqrt{3}}{2} \\ e^{i4\pi/3} &= -\frac{1}{2} - i \frac{\sqrt{3}}{2} \\ 1 & \end{aligned}$$

which vanish when summed.

**Question: 2.2.7**

The three cube roots of 1 are  $z$  and  $z^2$  and 1, when  $z = e^{2\pi i/3}$ . What are the three cube roots of 8 and the three cube roots of  $i$ ? (The angle for  $i$  is  $90^\circ$  or  $\pi/2$ , so the angle for one of its cube roots will be \_\_\_\_\_. The roots are spaced by  $120^\circ$ .

In polar form, we can express 8 as  $8e^{2\pi i}$ . Let  $z = e^{2\pi i/3}$ . Its cube roots are then  $2z$ ,  $2z^2$ , and 2.

For  $i$ , we have polar form  $e^{\pi i/2}$ . Then its cube roots are  $e^{\pi i/6}$ ,  $e^{5\pi i/6}$ , and  $e^{3\pi i/2}$ .

Argue by vector superposition that the cube roots in all cases sum to zero.

**Question: 2.2.8**

- (a) The number  $i$  is equal to  $e^{\pi i/2}$ . Then its  $i^{\text{th}}$  power  $i^i$  comes out equal to a real number, using the fact that  $(e^s)^t = e^{st}$ . What is that real number  $i^i$ ?
- (b)  $e^{i\pi/2}$  is also equal to  $e^{5\pi i/2}$ . Increasing the angle by  $2\pi$  does not change  $e^{i\theta}$  - it comes around a full circle and back to  $i$ . Then  $i^i$  has another real value  $(e^{5\pi i/2})^i = e^{-5\pi/2}$ . What are all the possible values of  $i^i$ ?

(a) We have  $(e^{\pi i/2})^i = e^{-\pi/2}$ .

(b) All possible values of  $i^i$  are  $e^{(-\pi \pm 4n\pi)/2}$ .

**Question: 2.2.9**

The numbers  $s = 3 + i$  and  $\bar{s} = 3 - i$  are complex conjugates. Find their sum  $s + \bar{s} = -B$  and their product  $(s)(\bar{s}) = C$ . Then show that  $s^2 + Bs + C = 0$  and also  $\bar{s}^2 + B\bar{s} + C = 0$ . Those numbers  $s$  and  $\bar{s}$  are the two roots of the quadratic equation  $x^2 + Bx + C = 0$ .

The sum is  $s + \bar{s} = 6$ , so  $B = -6$ . Their product is  $s\bar{s} = 10 = C$ . Then we have

$$s^2 + Bs + C = s^2 - (s + \bar{s})s + s\bar{s} = 0$$

$$\bar{s}^2 + B\bar{s} + C = \bar{s}^2 - (s + \bar{s})\bar{s} + s\bar{s} = 0$$

**Question: 2.2.10**

The numbers  $s = a + i\omega$  and  $\bar{s} = a - i\omega$  are complex conjugates. Find their sum  $s + \bar{s} = -B$  and their product  $(s)(\bar{s}) = C$ . Then show that  $s^2 + Bs + C = 0$ . The two solutions of  $x^2 + Bx + C = 0$  are  $s$  and  $\bar{s}$ .

The sum is  $s + \bar{s} = 2a$ , so  $B = -2a$ . Their product is  $s\bar{s} = a^2 + \omega^2$ . By the same argument as the previous problem, the solutions of the given polynomial are  $s$  and  $\bar{s}$ .

**Question: 2.2.11**

- (a) Find the numbers  $(1+i)^4$  and  $(1+i)^8$ .
- (b) Find the polar form  $re^{i\theta}$  of  $(1+i\sqrt{3}) / (\sqrt{3}+i)$ .

(a) We have  $1+i = \sqrt{2}e^{i\pi/4}$ . Then

$$(\sqrt{2}e^{i\pi/4})^4 = 4e^{i\pi} = -4$$

Raising to the eighth power, we simply square our previous answer to get  $(1+i)^8 = 16$ .

(b) The numerator is  $2e^{i\pi/3}$  and the denominator  $2e^{i\pi/6}$ . The quotient is  $e^{i\pi/6}$ .

**Question: 2.2.12**

The number  $z = e^{2\pi i/n}$  solves  $z^n = 1$ . The number  $Z = e^{2\pi i/2n}$  solves  $Z^{2n} = 1$ . How is  $z$  related to  $Z$ ? (This plays a big part in the Fast Fourier Transform.)

They are related by  $z = Z^2$ .

**Question: 2.2.13**

- (a) If you know  $e^{i\theta}$  and  $e^{-i\theta}$ , how can you find  $\sin(\theta)$ ?
- (b) Find all angles  $\theta$  with  $e^{i\theta} = -1$  and all angles  $\phi$  with  $e^{i\phi} = -i$ .

(a) One way to write  $\sin(\theta)$  is

$$\frac{e^{i\theta} - e^{-i\theta}}{2} = \sin(\theta)$$

(b) We have  $\theta = \pi \pm 2n\pi$  (which eliminates the imaginary term), and  $\phi = -\frac{\pi}{2} \pm 2n\pi$ .

**Question: 2.2.14**

Locate all these points on one complex plane:

(a)  $2 + i$

(b)  $(2 + i)^2$

(c)  $\frac{1}{2+i}$

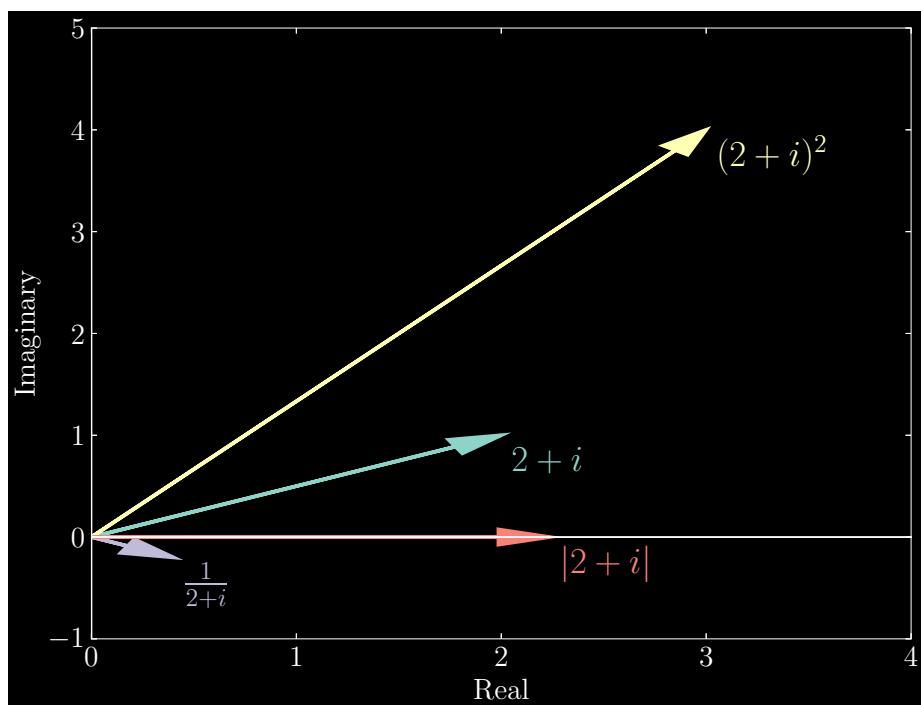
(d)  $|2 + i|$

(a)  $2 + i$

(b)  $(2 + i)^2 = 3 + 4i$

(c)  $\frac{1}{2+i} = \frac{2-i}{5}$

(d)  $|2 + i| = \sqrt{5}$



**Question: 2.2.15**

Find the absolute values  $r = |z|$  of these four numbers. If  $\theta$  is the angle for  $6 + 8i$ , what are the angles for these four numbers?

- (a)  $6 - 8i$
- (b)  $(6 - 8i)^2$
- (c)  $\frac{1}{6 - 8i}$
- (d)  $8i + 6$

- (a)  $r = 10, \phi = -\theta$ , in polar form  $z = 10e^{-i\theta}$
- (b)  $r = 100, \phi = -2\theta$ , in polar form  $z = 100e^{-2\theta}$
- (c)  $r = 1/10, \phi = \theta$ , in polar form  $z = \frac{1}{10}e^{i\theta}$
- (d)  $r = 10, \phi = \theta$ , the original number is unchanged

**Question: 2.2.16**

What are the real and imaginary parts of  $e^{a+i\pi}$  and  $e^{a+i\omega}$ ?

The real parts are  $-e^a$  and  $e^a \cos(\omega)$  and the imaginary parts are 0 and  $e^a \sin(\omega)$ .

**Question: 2.2.17**

- (a) If  $|s| = 2$  and  $|z| = 3$ , what are the absolute values of  $sz$  and  $s/z$ ?
- (b) Find upper and lower bounds in  $L \leq |s + z| \leq U$ . When does  $|s + z| = U$ ?

- (a)  $|sz| = |s||z| = 6, |s/z| = |s|/|z| = 2/3$
- (b) The lower bound is achieved when  $s$  and  $z$  are antiparallel, meaning the modulus of their vector sum is at best  $L = |s + z| = 1$ . The upper bound is reached when they are parallel, so  $U = |s + z| = 5$ .

**Question: 2.2.18**

- (a) Where is the product  $(\sin(\theta) + i \cos(\theta))(\cos(\theta) + i \sin(\theta))$  in the complex plane?
- (b) Find the absolute value  $|S|$  and the polar angle  $\phi$  for  $S = \sin(\theta) + i \cos(\theta)$ .

This is my favorite problem, because  $S$  combines  $\cos(\theta)$  and  $\sin(\theta)$  in a new way. To find  $\phi$ , you could plot  $S$  or add angles in the multiplication of part (a).

- (a) Note that  $\sin(\theta) = \cos\left(\frac{\pi}{2} - \theta\right)$  and  $\cos(\theta) = \sin\left(\frac{\pi}{2} - \theta\right)$ . Then the first term is equal to  $\cos\left(\frac{\pi}{2} - \theta\right) + i \sin\left(\frac{\pi}{2} - \theta\right) = e^{i(\pi/2-\theta)}$ . Then the product is

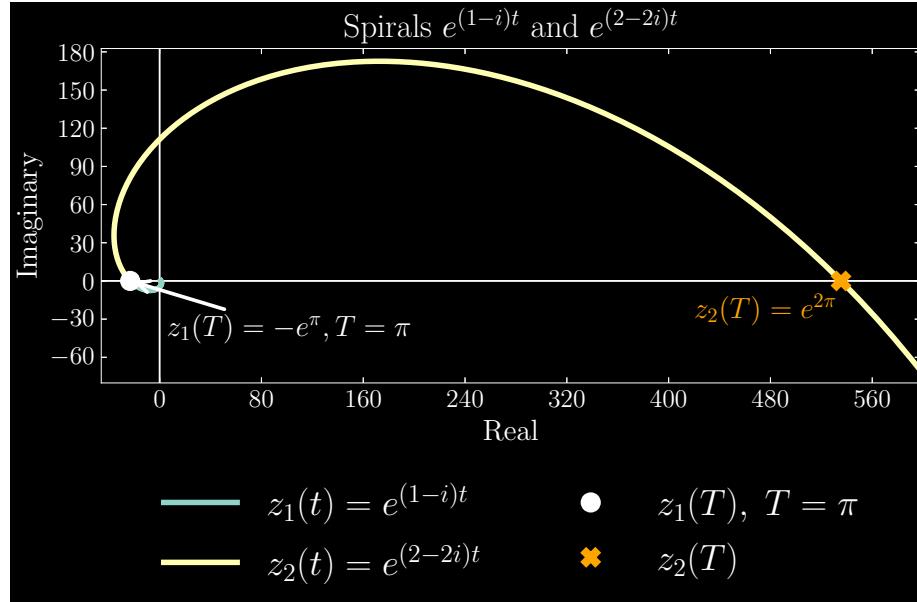
$$e^{i(\pi/2-\theta)} e^{i\theta} = e^{i\pi/2} = i$$

- (b)  $|S| = 1$  and  $\phi = \frac{\pi}{2} - \theta$ .

**Question: 2.2.19**

Draw the spirals  $e^{(1-i)t}$  and  $e^{(2-2i)t}$ . Do those follow the same curves? Do they go clockwise or anticlockwise? When the first one reaches the negative  $x$ -axis, what is the time  $T$ ? What point has the second one reached at that time?

As  $t \rightarrow \infty$ , the spirals progress clockwise. When the imaginary part vanishes, we have  $T = \pi$ , and  $e^{(1-i)\pi} = e^\pi$ . At that point in time, we also have  $e^{(2-2i)\pi} = e^{2\pi}$ .



**Question: 2.2.20**

The solution to  $d^2y/dt^2 = -y$  is  $y = \cos(t)$  if the initial conditions are  $y(0) = \underline{\hspace{2cm}}$  and  $y'(0) = \underline{\hspace{2cm}}$ . The solution is  $y = \sin(t)$  when  $y(0) = \underline{\hspace{2cm}}$  and  $y'(0) = \underline{\hspace{2cm}}$ . Write each of those solutions in the form  $c_1 e^{it} + c_2 e^{-it}$ , to see that real solutions can come from complex  $c_1$  and  $c_2$ .

For the case of  $y = \cos(t)$ , the initial conditions are  $y(0) = 1$  and  $y'(0) = 0$ . In the case of  $y = \sin(t)$ , we have  $y(0) = 0$  and  $y'(0) = 1$ . In the given form, the initial conditions require

$$\underline{y = \cos(t)}$$

$$\begin{aligned} y(0) &= c_1 + c_2 = 1 \\ y'(0) &= i(c_1 - c_2) = 0 \end{aligned}$$

Implying  $c_1 = c_2 = 1/2$ .

$$\underline{y = \sin(t)}$$

$$\begin{aligned} y(0) &= c_1 + c_2 = 0 \\ y'(0) &= i(c_1 - c_2) = 1 \end{aligned}$$

Implying  $c_1 = 1/2i$  and  $c_2 = -1/2i$ .

Thus we can write  $\cos(t)$  and  $\sin(t)$  as

$$\cos(t) = \frac{e^{it} + e^{-it}}{2}, \quad \sin(t) = \frac{e^{it} - e^{-it}}{2i}$$

The general solution is given by

$$y(t) = C_1 \cos(t) + C_2 \sin(t)$$

which we rewrite as

$$\begin{aligned} y(t) &= C_1 \cos(t) + C_2 \sin(t) \\ &= C_1 \left[ \frac{e^{it} + e^{-it}}{2} \right] + C_2 \left[ \frac{e^{it} - e^{-it}}{2i} \right] \\ &= \left[ \frac{C_1 - iC_2}{2} \right] e^{it} + \left[ \frac{C_1 + iC_2}{2} \right] e^{-it} \end{aligned}$$

**Question: 2.2.21**

Suppose  $y(t) = e^{-t}e^{it}$  solves  $y'' + By' + Cy = 0$ . What are  $B$  and  $C$ ? If this equation is solved by  $y = e^{3it}$ , what are  $B$  and  $C$ ?

Note that  $y(t) = e^{-t}e^{it} = e^{(-1+i)t}$ . Then we have

$$y'' + By' + Cy = (-1 + i)^2 e^{(-1+i)t} + B(-1 + i) e^{(-1+i)t} + C e^{(-1+i)t} = 0$$

from which we derive

$$(-1 + i)^2 + B(-1 + i) + C = 0$$

Now wait – this looks suspiciously familiar to the form seen in question 9:

$$s^2 + Bs + C = 0$$

Then we have that  $s + \bar{s} = -B$  and  $s\bar{s} = C$ . Thus

$$s + \bar{s} = -2 = -B \quad s\bar{s} = 2 = C$$

Therefore,  $B = C = 2$ , and the equation is  $y'' + 2y' + 2y = 0$ .

In the case of  $y = e^{3it}$ , we have polynomial

$$-9 + 3Bi + C = 0$$

where it follows that  $s = 3i$ ,  $s + \bar{s} = 0 = -B$ , and  $s\bar{s} = 9 = C$ . The equation is  $y'' + 9y = 0$ .

#### Question: 2.2.22

From the multiplication  $e^{iA}e^{-iB} = e^{i(A-B)}$ , find the “subtraction formulas” for  $\cos(A - B)$  and  $\sin(A - B)$ .

Derive

$$\begin{aligned} e^{iA}e^{-iB} &= [\cos(A) + i\sin(A)][\cos(B) - i\sin(B)] \\ &= [\cos(A)\cos(B) + \sin(A)\sin(B)] + i[\sin(A)\cos(B) - \cos(A)\sin(B)] \\ &= \cos(A - B) + i\sin(A - B) \\ &= e^{i(A-B)} \end{aligned}$$

#### Question: 2.2.23

- (a) If  $r$  and  $R$  are the absolute values of  $s$  and  $S$ , show that  $rR$  is the absolute value of  $sS$ . (Hint: Polar form!)
- (b) If  $\bar{s}$  and  $\bar{S}$  are the complex conjugates of  $s$  and  $S$ , show that  $\bar{s}\bar{S}$  is the complex conjugate of  $sS$ . (Polar form!)

- (a) Write  $s = re^{i\theta}$  and  $S = Re^{i\phi}$ . Then  $sS = rRe^{i(\theta+\phi)}$ . Then  $rR = |sS|$ .

- (b) Using the aforementioned definitions of  $s$  and  $S$ , we have

$$\overline{sS} = rRe^{-i(\theta+\phi)} = (re^{-i\theta})(Re^{-i\phi}) = \bar{s}\bar{S}$$

**Question: 2.2.24**

Suppose a complex number  $s$  solves a real equation  $s^3 + As^2 + Bs + C = 0$  (with  $A, B, C$  real). Why does the complex conjugate  $\bar{s}$  also solve this equation? “*Complex solutions to real equations come in conjugate pairs  $s$  and  $\bar{s}$ .*”

We generalize the results of question 23 and claim that the complex conjugate of a product is equal to the product of the complex conjugates. For instance, suppose  $s = re^{i\theta}$ . Then we prove  $\overline{s^n} = \bar{s}^n$ :

$$\overline{s^n} = \overline{r^n e^{-in\theta}} = \prod_{k=1}^n \overline{re^{-i\theta}} = \bar{s}^n$$

We can also prove that the complex conjugate of a sum is the sum of the complex conjugates. Define  $z_k = a_k + ib_k$  for  $k \in \{1, \dots, n\}$ . Then

$$\overline{\sum_{k=1}^n z_k} = \overline{\sum_{k=1}^n (a_k + ib_k)} = \overline{\sum_{k=1}^n a_k + i \sum_{k=1}^n b_k} = \sum_{k=1}^n a_k - i \sum_{k=1}^n b_k = \sum_{k=1}^n \overline{z_k}$$

With these facts, the complex conjugate of the left side simply yields that  $\bar{s}$  is too a solution of the real equation.

**Question: 2.2.25**

- (a) If two complex numbers add to  $s + S = 6$  and multiply to  $sS = 10$ , what are  $s$  and  $S$ ? (They are complex conjugates.)
- (b) If two numbers add to  $s + S = 6$  and multiply to  $sS = -16$ , what are  $s$  and  $S$ ? (Now they are real.)

(a)  $s = 3 + i, S = 3 - i$

(b)  $s = 8, S = -2$

**Question: 2.2.26**

If two numbers  $s$  and  $S$  add to  $s + S = -B$  and multiply to  $sS = C$ , show that  $s$  and  $S$  solve the quadratic equation  $s^2 + Bs + C = 0$ .

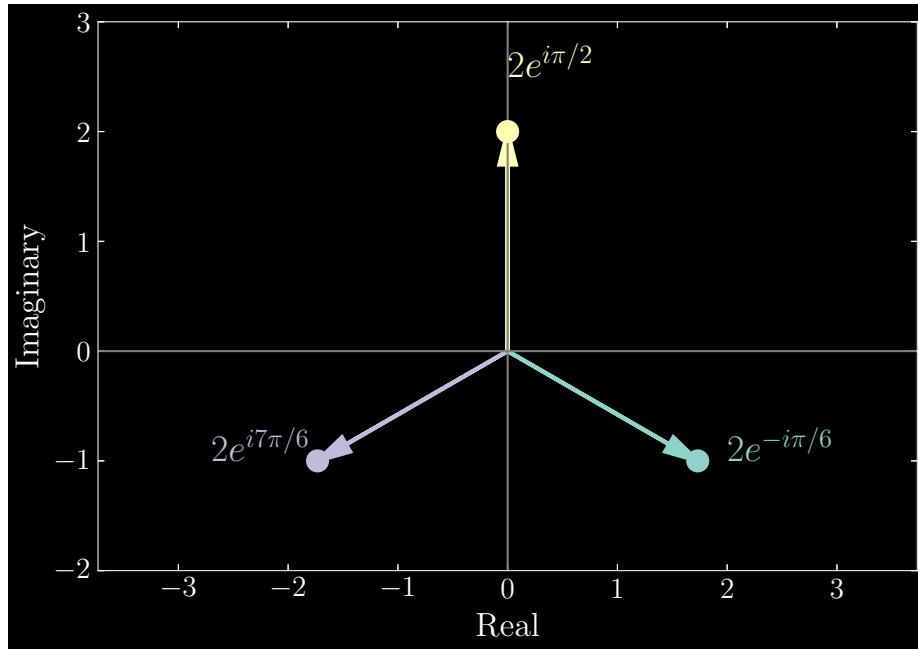
This is an interesting question because we relax the condition that  $S = \bar{s}$ . Substituting for  $B$  and  $C$ , we have

$$s^2 - (s + S)s + sS = 0, \quad S^2 - (s + S)S + sS = 0$$

**Question: 2.2.27**

Find three solutions to  $s^3 = -8i$  and plot the three points in the complex plane. What is the sum of the three solutions?

In polar form,  $s^3 = 8e^{-i\pi/2}$ . Its cube roots are  $2e^{-i\pi/6}, 2e^{i\pi/2}$ , and  $2e^{i7\pi/6}$ . By vector superposition, the roots sum to zero.

**Question: 2.2.28**

- (a) For which complex numbers  $s = a + i\omega$  does  $e^{st}$  approach 0 as  $t \rightarrow \infty$ ? Those numbers  $s$  fill which “half-plane” in the complex plane?
- (b) For which complex numbers  $s = a + i\omega$  does  $s^n$  approach 0 as  $n \rightarrow \infty$ ? Those numbers  $s$  fill which part of the complex plane? Not a half-plane!

- (a) For  $a < 0$ ,  $e^{st} = e^{at}e^{i\omega t}$  approaches zero.
- (b) In polar form,  $s = \sqrt{a^2 + \omega^2}e^{i\tan^{-1}(\omega/a)}$ . We must require the modulus to be less than zero, or  $\sqrt{a^2 + \omega^2} < 0$ , for  $s^n$  to tend towards zero as  $n$  increases.

### 2.3 - Constant Coefficients $A, B, C$

**Question: 2.3.1**

Substitute  $y = e^{st}$  and solve the characteristic equation for  $s$ :

- (a)  $2y'' + 8y' + 6y = 0$
- (b)  $y''' - 2y'' + y = 0$

- (a) The characteristic equation is  $2s^2 + 8s + 6 = 0$  which factorizes into  $2(s+3)(s+1) = 0$ . Then  $s_1 = -3$  and  $s_2 = -1$ .
- (b) The characteristic equation is  $s^4 - 2s^2 + 1 = 0$  which factorizes into  $(s^2 - 1)^2 = (s-1)^2(s+1)^2$ . It has repeated solutions:  $s_1, s_2 = 1$  and  $s_3, s_4 = -1$ .

**Question: 2.3.2**

Substitute  $y = e^{st}$  and solve the characteristic equation for  $s = a + i\omega$ :

- (a)  $y'' + 2y' + 5y = 0$
- (b)  $y''' + 2y'' + y = 0$

- (a) The characteristic equation is  $s^2 + 2s + 5 = 0$ , which by the quadratic formula has roots

$$s_1, s_2 = \frac{-2 \pm \sqrt{-16}}{2} = -1 \pm 2i$$

- (b) The characteristic equation is  $s^4 + 2s^2 + 1 = 0$ , which by the quadratic formula has double root

$$s^2 = \frac{-2}{2} = -1$$

Rooting once more, we find repeated roots

$$s_1, s_2 = i, \quad s_3, s_4 = -i$$

**Question: 2.3.3**

Which second order equation is solved by  $y = c_1 e^{-2t} + c_2 e^{-4t}$ ? Or  $y = te^{5t}$ ?

In the first instance, the roots are  $s_1 = -2$  and  $s_2 = -4$ . Then the characteristic polynomial is  $(s+2)(s+4) = s^2 + 6s + 8$ , corresponding to second order equation  $y'' + 6y' + 8y = 0$ . In the second case, we have repeated roots  $s_1, s_2 = 5$ , which has characteristic polynomial  $(s-5)^2 = s^2 - 10s + 25$ . This corresponds to second order equation  $y'' - 10y' + 25y = 0$ .

**Question: 2.3.4**

Which second order equation has solutions  $y = c_1 e^{-2t} \cos(3t) + c_2 e^{-2t} \sin(3t)$ ?

The complex roots are  $s_1 = -2 + 3i$  and  $s_2 = -2 - 3i$ , with characteristic polynomial  $s^2 - 4s + 13$ . This corresponds to second order equation

$$y'' - 4y' + 13 = 0$$

**Question: 2.3.5**

Which numbers  $B$  give (under)(critical)(over) damping in  $4y'' + By' + 16y = 0$ ?

**Underdamping:** We must have  $B^2 < 4AC$ , or in this case,  $B^2 < 256$ . Then we must have  $|B| < 16$ .

**Critical damping:** We have  $B^2 = 4AC$ , or  $B^2 = 256$ , implying  $|B| = 16$ .

**Overdamping:** We have  $B^2 > 4AC$ , or  $B^2 > 256$ , implying  $|B| > 16$ .

**Question: 2.3.6**

If you want oscillation from  $my'' + by' + ky = 0$ , then  $b$  must stay below \_\_\_\_\_.

We must have  $b^2 < 4mk$  or  $|b| < 2\sqrt{mk}$ .

**Question: 2.3.7**

The roots  $s_1$  and  $s_2$  satisfy  $s_1 + s_2 = -2p = -B/A$  and  $s_1 s_2 = p^2 + \omega_n^2 = C/A$ . Show this two ways:

- (a) Start from  $A^2 + Bs + C = A(s - s_1)(s - s_2)$ . Multiply to see  $s_1 s_2$  and  $s_1 + s_2$ .
- (b) Start from  $s_1 = -p + i\omega_d$ ,  $s_2 = -p - i\omega_d$

- (a) We write

$$\begin{aligned} A^2 + Bs + C &= A(s - s_1)(s - s_2) \\ &= A(s^2 - s(s_1 + s_2) + s_1 s_2) \end{aligned}$$

Concluding that  $B = -A(s_1 + s_2)$  and  $C = As_1 s_2$ . Then

$$-\frac{B}{A} = s_1 + s_2, \quad \frac{C}{A} = s_1 s_2$$

- (b) Let  $s_1 = -p + i\omega_d$  and  $s_2 = -p - i\omega_d$ . Then  $s_1 + s_2 = -2p$  and

$$s_1 s_2 = (-p + i\omega_d)(-p - i\omega_d) = p^2 + \omega_d^2$$

**Note:** There are two errors in the problem statement. The textbook erroneously has the equalities  $s_1 + s_2 = -B/2A$  and  $s_1 s_2 = p^2 + \omega_n^2$ . The corrections have been made in my writing of the problem above.

**Question: 2.3.8**

Find  $s$  and  $y$  at the bottom point of the graph of  $y = As^2 + Bs + C$ . At that minimum point  $s = s_{\min}$  and  $y = y_{\min}$ , the slope is  $dy/ds = 0$ .

Differentiate with respect to  $s$  to find

$$\frac{dy}{ds} = 2As + B = 0 \implies s_{\min} = -\frac{B}{2A}$$

The corresponding  $y_{\min}$  is

$$\begin{aligned} y_{\min} &= A\left(-\frac{B}{2A}\right)^2 + B\left(-\frac{B}{2A}\right) + C \\ &= \frac{B^2}{4A} - \frac{B^2}{2A} + C \\ &= -\frac{B^2}{4A} + C \end{aligned}$$

**Question: 2.3.9**

The parabolas in Figure 2.10 show how the graph of  $y = As^2 + Bs + C$  is raised by increasing  $B$ . Using Problem 8, show that the bottom point of the graph moves left (change in  $s_{\min}$ ) and down (change in  $y_{\min}$ ) when  $B$  is increased by  $\Delta B$ .

Assume  $\Delta B > 0$ . By Problem 8, the new  $s_{\min}$  becomes

$$s_{\min} = -\frac{(B + \Delta B)}{2A} < -\frac{B}{2A}$$

and the new  $y_{\min}$

$$y_{\min} = -\frac{(B + \Delta B)^2}{4A} + C = -\frac{(B^2 + 2B\Delta B + \Delta B^2)}{4A} + C < -\frac{B^2}{4A} + C$$

both of which correspond to a leftward and downward shift, respectively.

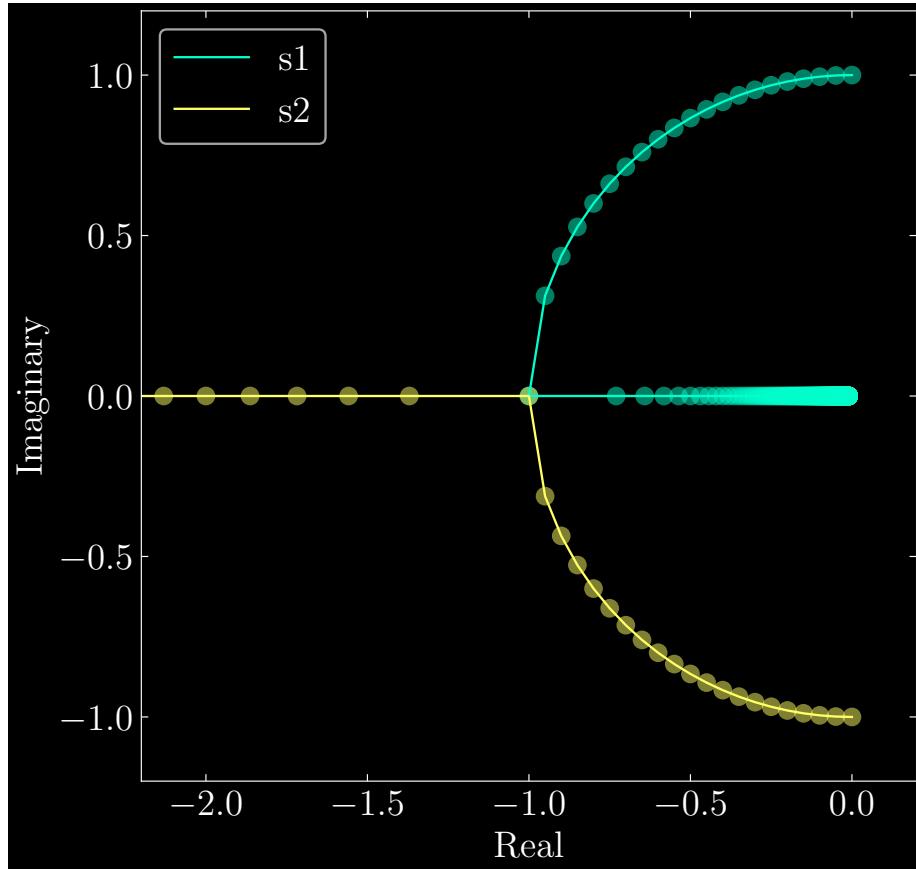
**Question: 2.3.10**

(recommended) Draw a picture to show the paths of  $s_1$  and  $s_2$  when  $s^2 + Bs + 1 = 0$  and the damping increases from  $B = 0$  to  $B = \infty$ . At  $B = 0$ , the roots are on the \_\_ axis. As  $B$  increases, the roots travel on a circle (why?). At  $B = 2$ , the roots meet on the real axis. For  $B > 2$  the roots separate to approach 0 and  $-\infty$ . *Why is their product  $s_1 s_2$  always equal to 1?*

The roots are

$$s = \frac{-B \pm \sqrt{B^2 - 4}}{2}$$

When  $B = 0$ , the roots are  $s_{1,2} = \pm i$ , which lie on the imaginary axis. The roots travel on a circle as  $B$  grows because they are complex conjugates, and given that the modulus of the roots remains constant, a quarter-circle traced out by the complex vector for one root will be reflected over the real axis for the other root, conforming to the geometry of a circle.



Recall from problem 7 that  $s_1 s_2 = C/A = 1$ . Another, more intuitive approach is to realize that,  $s_1 = \overline{s_2}$ , so they are complex conjugates of one another, and their product is the modulus (in this case, unity).

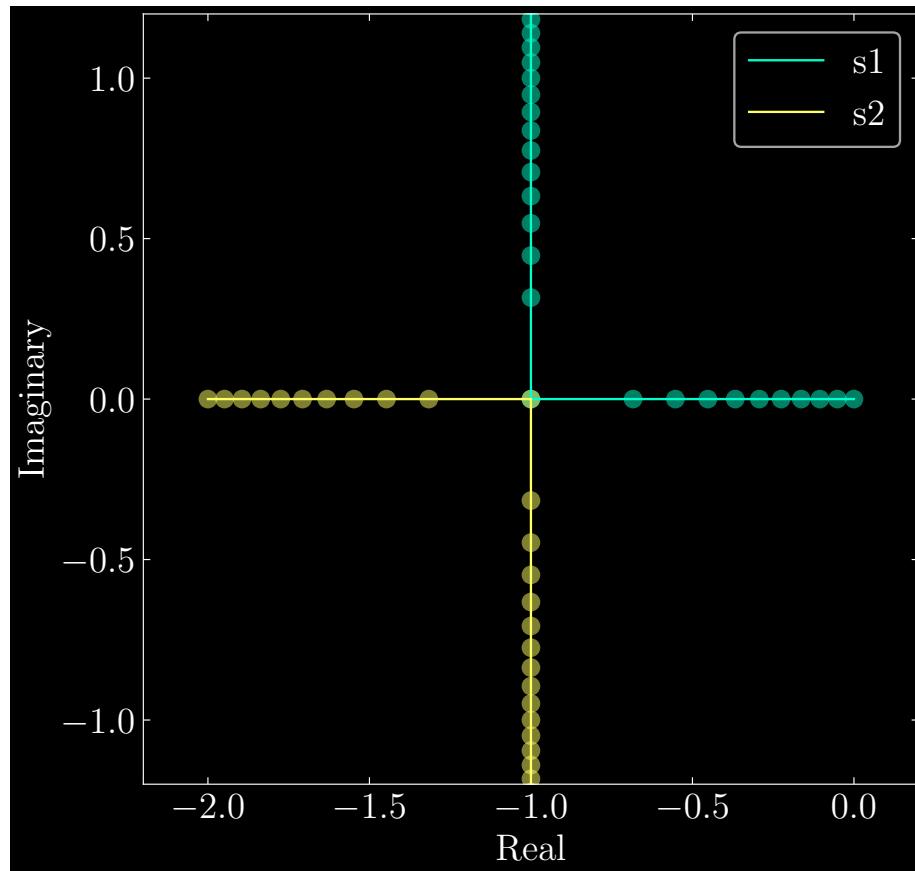
**Question: 2.3.11**

(this too if possible) Draw the paths of  $s_1$  and  $s_2$  when  $s^2 + 2s + k = 0$  and the stiffness increases from  $k = 0$  to  $k = \infty$ . When  $k = 0$ , the roots are \_\_\_\_\_. At  $k = 1$ , the roots meet at  $s = _____$ . For  $k \rightarrow \infty$  the two roots travel up/down on a \_\_\_\_\_ in the complex plane. *Why is their sum  $s_1 + s_2$  always equal to  $-2$ ?*

The roots are

$$s = \frac{-2 \pm \sqrt{2^2 - 4k}}{2} = -1 \pm \sqrt{1 - k}$$

When  $k = 0$ , the roots are  $s_{1,2} = 0, -2$ . Once we reach  $k = 1$ , we have double roots with  $s_{1,2} = -1$ . As  $k$  increases to infinity, the real component remains fixed, while the imaginary component grows in both directions on the vertical axis. Altogether, the trajectory of the roots forms a cross.



As for their sum, observe that the imaginary components of  $s_{1,2}$  cancel, and the real parts add to  $-2$ .

**Question: 2.3.12**

If a polynomial  $P(s)$  has a double root at  $s = s_1$ , then  $(s - s_1)$  is a double factor and  $P(s) = (s - s_1)^2 Q(s)$ . Certainly  $P = 0$  at  $s = s_1$ . Show that also  $dP/ds = 0$  at  $s = s_1$ . Use the product rule to find  $dP/ds$ .

By the product rule, we have

$$\frac{dP}{ds}_{(s=0)} = 2(s - s_1)Q(s) + (s - s_1)^2 \frac{dQ}{ds} = 0$$

**Question: 2.3.13**

Show that  $y'' = 2ay' - (a^2 + \omega^2)y$  leads to  $s = a \pm i\omega$ . Solve  $y'' - 2y' + 10y = 0$ .

The characteristic equation is given by

$$s^2 - 2as + (a^2 + \omega^2) = 0$$

which has roots

$$s = \frac{2a \pm \sqrt{4a^2 - 4(a^2 + \omega^2)}}{2} = a \pm i\omega$$

For the given equation, we have  $a = 1$  and  $\omega = 3$ . Then we have solution

$$\begin{aligned} y(t) &= e^{(1+3i)t} + e^{(1-3i)t} \\ &= 2e^t \cos(3t) \end{aligned}$$

**Question: 2.3.14**

The undamped *natural frequency* is  $\omega_n = \sqrt{k/m}$ . The two roots of  $ms^2 + k = 0$  are  $s = \pm i\omega_n$  (pure imaginary). With  $p = b/2m$ , the roots of  $ms^2 + bs + k = 0$  are  $s_1, s_2 = -p \pm \sqrt{p^2 - \omega_n^2}$ . The coefficient  $p = b/2m$  has the units of 1 / time.

Solve  $s^2 + 0.1s + 1 = 0$  and  $s^2 + 10s + 1 = 0$  with numbers correct to two decimals.

In the first equation, we have  $m = 1$ ,  $b = 0.1$ , and  $k = 1$ . Then  $p = 0.05$  and  $\omega_n = 1$ . This yields roots  $s_{1,2} = -0.05 \pm 0.99i$ .

The second equation gives us  $p = 5$  and  $\omega_n = 1$ , which has roots  $s_{1,2} = -5 \pm \sqrt{24} = -0.10, -9.89$ .

**Question: 2.3.15**