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El Houssin El Bouchikhi, Vincent Choqueuse, Mohamed Benbouzid, Jean-Frederic Charpentier, Georges Barakat. A Comparative Study of Time-Frequency Representations for Fault Detection in Wind Turbine. IECON 2011, Nov 2011, Melbourne, Australia. pp.3584 - 3589, 10.1109/IECON.2011.6119891 . hal-00657624

HAL Id: hal-00657624

<https://hal.archives-ouvertes.fr/hal-00657624>

Submitted on 9 Jan 2012

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A Comparative Study of Time-Frequency Representations for Fault Detection in Wind Turbine

El.H. Bouchikhi, V. Choqueuse, M.E.H. Benbouzid, J.F. Charpentier and G. Barakat

Abstract—To reduce the cost of wind energy, minimization and prediction of maintenance operations in wind turbine is of key importance. In variable speed turbine generator, advanced signal processing tools are required to detect and diagnose the generator faults from the stator current. To detect a fault in non-stationary conditions, previous studies have investigated the use of time-frequency techniques such as the Spectrogram, the Wavelet transform, the Wigner-Ville representation and the Hilbert-Huang transform. In this paper, these techniques are presented and compared for broken-rotor bar detection in squirrel-cage generators. The comparison is based on several criteria such as the computational complexity, the readability of the representation and the easiness of interpretation.

Index Terms—Wind turbine, fault detection, broken-rotor bars, signal processing, time-frequency representations.

I. INTRODUCTION

Wind energy is the fastest growing energy source in the world and it is expected to remain so for the foreseeable future. To reduce the cost of wind energy, there is a need to improve the wind turbines availability and to reduce the maintenance cost [1]. This need is even more important in the offshore context due to the aggressive environmental conditions. To answer this need, condition monitoring systems based on current, vibration, oil and temperature analysis have been developed. As compared to other techniques, generator current analysis has several advantages since sensors are low-cost and easy to install. Furthermore, the fault-detection techniques based on stator-current, initially developed for electric motor [2], can be easily adapted to wind turbine generator.

In the field of electric motor, many signal processing techniques have been employed to detect a fault from the stator current. Most of them perform spectral analysis, such as Fourier or MUSIC techniques. Although these techniques exhibit good results in stationary conditions, they are badly suited for wind turbines since the environment is predominantly non-stationary due to transient or variable speed conditions. In non-stationary conditions, fault detection is usually performed by using sophisticated approaches, such as time-frequency or time-scale representations [3]. These approaches include the

Short-Time Fourier Transform (STFT) [4]–[8], the Continuous Wavelet Transform (CWT) [1], [9], [10], the Wigner-Ville and other quadratic Distributions [6], [11]–[15] and the Hilbert-Huang Transform (HHT) [15]–[17]. However, despite of the rich literature, the choice of a particular representation is not an easy task since several criteria must be taken into account.

In this study, several time-frequency representations are considered for fault detection and are applied to a squirrel-cage wind-turbine. These representations are compared according to their computational complexity, readability of the representation and easiness of interpretation. This paper is organized as follows: section II describes the generator stator current model, section III presents the most commonly used time-frequency representations and section IV reports on their performances for broken rotor bar faults detection in transient condition.

II. COUPLED MAGNETIC CIRCUIT GENERATOR MODEL (CMCM)

This section describes the generator model used to simulate the stator current under healthy and faulty conditions. For most of the faults, the harmonic contents of the stator current can be approximated using linear models of the machine such as the Coupled Magnetic Circuit Method (CMCM) [18]. This approach is based on the analytical equations of the induction machine. All parameters are calculated from the actual geometry and winding layout of the machines rather than from transformed or equivalent variables. This analysis is based on the following assumptions: i) infinite iron permeability, ii) non conductive magnetic circuit, iii) no inter-bars currents and iv) negligible saturation. The equations system governing the operation of induction machine in the phase space is given by (1).

$$\begin{cases} \frac{d}{dt}[I] = -[L]^{-1} \left([T] + \Omega \frac{d}{d\theta_M} [L] \right) [I] + [L]^{-1} [V] \\ \frac{d}{dt}\omega = \frac{1}{2J} [I]^T \left(\frac{d}{d\theta_M} [L] \right) [I] - \frac{f}{J} \Omega - \frac{1}{J} \Gamma_L \\ \frac{d}{dt}\theta_m = \omega \end{cases} \quad (1)$$

The expression of the inductance matrix $[L]$ of the induction machine can be extracted from the flux or from the magnetic energy stored on the air-gap. The computation of this matrix containing all magnetizing and mutual inductances is the key to a successful simulation of the squirrel cage induction machine. In order to model the machine in the case of broken bar, the equivalent circuit is opened and the inductances matrix are recalculated.

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III. FAULT DETECTION BASED ON TIME-FREQUENCY ANALYSIS

The following subsections describes several techniques for representing time-frequency content of the stator current, $x(t)$.

A. Short-Time Fourier Transform (STFT)

The Short-Time Fourier Transform (STFT) is based on the assumption that the signal is quasi-stationary over a short period of time. It is composed of two steps: first, the signal is divided into time segments and then, the spectrum of each segment is obtained via the Fourier Transform. This procedure, called the STFT, leads to a 3D representation which displays the frequency content over time.

Mathematically, the Short-Time Fourier Transform can be expressed as [3]

$$S_x(t, f) = \int_{-\infty}^{\infty} x(\tau) h^*(\tau - t) e^{-2j\pi f \tau} d\tau \quad (2)$$

where $h(t)$ is a time-window centered in $t = 0$ which is used to extract the time segments. The STFT is a linear transform i.e. $S_{x+y}(t, f) = S_x(t, f) + S_y(t, f)$. To obtain an admissible representation, $h(t)$ must have unit energy i.e.

$$\int_{-\infty}^{\infty} |h(t)|^2 dt = 1 \quad (3)$$

A classical choice for $h(t)$ is the rectangular, Hanning, Hamming or Gaussian window. The length of the window $h(t)$ determines the time and frequency resolution of the representation. This resolution is kept constant over the time-frequency plane. A short window leads to a representation which is fine in time but coarse in the the frequency domain. Conversely, a long window leads to a representation which is coarse in time but fine in the the frequency domain. This trade-off is formalized by the Heisenberg-Gabor uncertainty principle [3].

B. Continuous Wavelet Transform (CWT)

The Continuous Wavelet Transform (CWT) is obtained by breaking up the signal into shifted and scaled versions of a mother wavelet.

Mathematically, the Continuous Wavelet Transform can be expressed as [3]

$$T_x(t, a) = \frac{1}{\sqrt{a}} \int_{-\infty}^{\infty} x(\tau) w^* \left(\frac{\tau - t}{a} \right) d\tau \quad (4)$$

where a is the scale and $w(t)$ is the mother wavelet. The CWT is a linear transform i.e. $T_{x+y}(t, f) = T_x(t, f) + T_y(t, f)$. To obtain an admissible representation, $w(t)$ must have zero-mean i.e.

$$\int_{-\infty}^{\infty} w(t) dt = 0 \quad (5)$$

A classical choice for $w(t)$ is the Mexican-Hat, Morlet or Daubechies wavelets. Rigorously speaking, the CWT leads to a time-scale representation since it displays the signal time-evolution at different scales. However, there is a direct link between scale and frequency. Indeed, if the central frequency

of the mother wavelet $w(t)$ is f_0 , the scale a corresponds to the frequency $f = f_0/a$. As opposed to the STFT, the CWT is a multi-resolution technique which favors the time-resolution at high-frequencies and the frequency-resolution at low-frequencies.

C. Wigner-Ville and other quadratic distributions

As opposed to the previous techniques, which focus on the decomposition of the signal itself, the Wigner-Ville distribution (WVD) focuses on the decomposition of the signal energy in the time-frequency plane.

Mathematically, the WVD can be expressed as [3]

$$W_{x,x}(t, f) = \int_{-\infty}^{\infty} x \left(t + \frac{\tau}{2} \right) x^* \left(t - \frac{\tau}{2} \right) e^{-2j\pi f \tau} d\tau \quad (6)$$

The time-frequency resolution of WVD is not constrained by the Heisenberg Gabor inequality. However, the WVD is non-linear since $W_{(x+y), (x+y)}(t, f) \neq W_{x,x}(t, f) + W_{y,y}(t, f)$. This non-linearity is responsible for the introduction of interference terms. These interference terms can render the time-frequency representation difficult to interpret. To reduce interference terms, the analytical signal of $x(t)$ is usually employed instead of the signal itself. Furthermore, several authors have proposed extensions of the WVD for interference reduction at the expense of reduced resolution. These extensions have been unified by the Cohen's class of time-frequency distributions [19]. This class is given by

$$C_{x,x}^{\phi}(t, f) = \iint_{\mathbb{R}^3} \phi(\xi, \tau) e^{2j\pi \xi(s-t)} \times x \left(s + \frac{\tau}{2} \right) x^* \left(s - \frac{\tau}{2} \right) e^{-2j\pi f \tau} d\xi ds d\tau \quad (7)$$

where $\phi(\xi, \tau)$ is the kernel of the distribution. The Cohen's class includes many commonly used time-frequency distributions such as the Pseudo Wigner-Ville, the Choi-Williams and the Zhao-Atlas-Marks Distributions.

D. Hilbert-Huang Transform

The Hilbert-Huang Transform is a nonlinear technique which extract the time-frequency content of a non-stationary signal. This technique is composed of 2 steps: i) first, the signal is decomposed into a sum of amplitude- and frequency- modulated sine waves using an Empirical Mode Decomposition, and then ii) the instantaneous amplitude and frequency are extracted using a demodulation technique. Finally the time-frequency representation is obtained by displaying the time evolution of the instantaneous amplitude and frequency for each sine wave.

1) *Empirical Mode Decomposition*: The Empirical Mode Decomposition (EMD) has been originally proposed by Huang [20]. As opposed to the previous techniques, the EMD is essentially defined by an algorithm and does not admit an analytical definition [21]. The algorithm is described by the following steps [20]:

- Identification of all extrema of $x(t)$

- Interpolation between minimal (resp. maxima) ending up with some envelope $e_{min}(t)$ (resp. $e_{max}(t)$).
- computation of the mean:

$$m(t) = \frac{e_{min}(t) + e_{max}(t)}{2} \quad (8)$$

- extraction of the detail:

$$d(t) = x(t) - m(t) \quad (9)$$

- Iteration on the residual $m(t)$.

In practice, this algorithm has to be refined by a *shifting process* until $d(t)$ can be considered as zero-mean. After this procedure, the detail $d(t)$ corresponds to an amplitude- and frequency- modulated (AM/FM) sine wave called Intrinsic Mode Function (IMF). By iterating the algorithm on the residual $m(t)$, the EMD extracts several IMFs until a stopping criterion is reached.

2) *Instantaneous Amplitude (IA) and frequency (IF) extraction*: The IA and IF of each sine wave can be extracted using a demodulation technique. To achieve this task, a popular technique is based on the Hilbert Transform. Let us denote $x_k(t)$ the k^{th} IMF, the analytical signal of $x_k(t)$, denoted $z_k(t)$, is defined as

$$z_k(t) = x_k(t) + j\mathcal{H}[x_k(t)] \quad (10)$$

where $\mathcal{H}[\cdot]$ denotes the Hilbert Transform. The instantaneous amplitude and frequency can be extracted from the analytical signal as follows

$$a_k(t) = |z_k(t)| \quad (11)$$

$$f_k(t) = \frac{1}{2\pi} \frac{d \arg[z_k(t)]}{dt} \quad (12)$$

where $|\cdot|$ and $\arg[\cdot]$ denote respectively the modulus and the argument of a complex number. Finally, the time-frequency representation is obtained by displaying the evolution of $a_k(t)$ and $f_k(t)$ for each IMF in the time-frequency plane.

IV. SIMULATION RESULTS

In this section, we compare several time-frequency representations for the detection of broken rotor bar faults. The stator current have been simulated using a coupled magnetic circuit method (CMCM). CMCM has been performed with a healthy generator and a faulty one with two broken bars. The stator-current signals have been simulated during 1 second at a $1kHz$ sampling frequency. Signals have been analyzed with the following time-frequency representations:

- A Short-Time Fourier Transform with a Hamming window $h(t)$ of 256 samples length.
- A Continuous Wavelet Transform with a Morlet mother wavelet $w(t)$.
- The Pseudo Wigner distribution computed from the analytical signal.
- The Hilbert-Huang transform.

All these representations, except the last one, have been implemented using the Time-Frequency Toolbox for Matlab [22]. The HHT has been implemented using the G. Rilling's sub-routines for Matlab. The following subsections compare the computational complexity and the readability of these time-frequency representations.

A. Computational Complexity

Matlab programs were run offline on a Pentium dual core PC at 1.6 Ghz with 1 Go of RAM. The computational requirement of each algorithm is reported in Table I. This table shows that the CWT is the most computationally demanding technique as compared to the STFT, the Pseudo WV and the HHT. By comparing the computational time with the length of the recorded signal (1s), we can observe that the STFT and the Pseudo WV are the only two candidates that could be implemented in real-time under our configuration. In particular, the STFT, which has the lowest computational cost, is an attractive technique since it can be efficiently implemented using Fast Fourier Transform.

TABLE I
COMPUTATIONAL COMPLEXITY

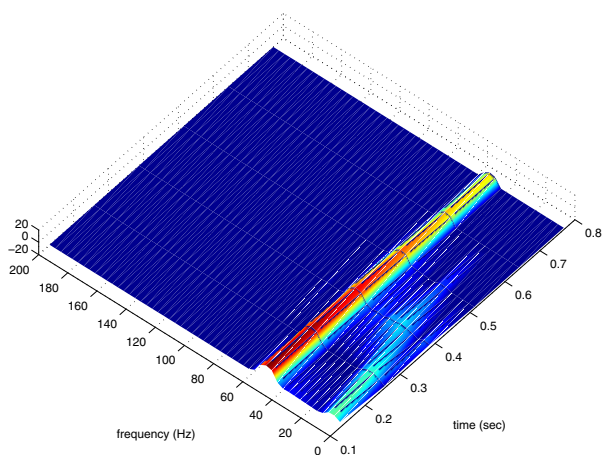
Representation	(Average) Computational time
STFT	0.2 s
CWT	25.2 s
Pseudo WV	0.4s
HHT	1.57s

B. Representation readability and easiness of interpretation

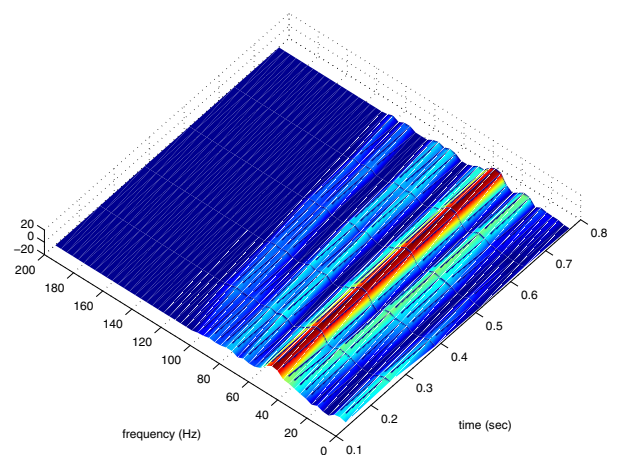
Figs.1 to 4 display the time-frequency representations for healthy and faulty generators. For each representation, we can observe a strong difference between the time-frequency representations of healthy and faulty generators. For the STFT, CWT and Pseudo WV, the faulty generator exhibits singular time-varying components at frequencies below 40Hz and above 70Hz. For the HHT, the fault generator introduces frequency modulation of the 50Hz component and some components below 30Hz.

Concerning the time-frequency resolution, figures show that the resolution of the STFT and CWT is lower than the one of the other representations. Indeed, as opposed to the Pseudo WV and HHT, the resolution of the STFT and CWT are limited by the Heisenberg-Gabor uncertainty principle. This constraint makes the STFT and CWT bad-suited for analysis of signals with rapidly evolving frequency content.

We can observe that the readability of the Pseudo-WV and HHT is lower than the one of the STFT and CWT. First, the Pseudo-WV exhibits interference terms which can render difficult the interpretation. Then, the HHT presents several low-frequency components which are difficult to analyse (below 30Hz). Even if these components have low-energy, they negatively impact the readability of the time-frequency representation.

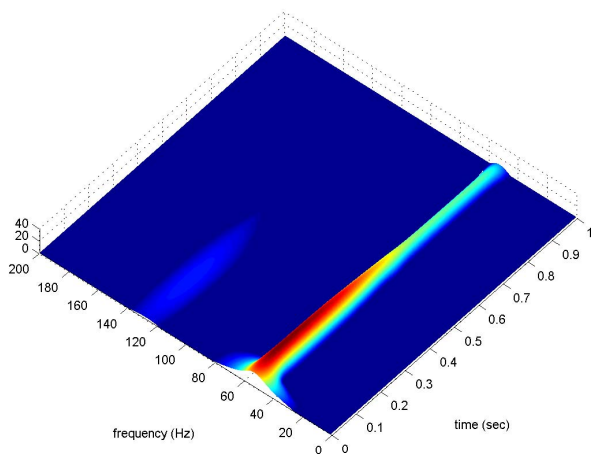


(a) Healthy Generator

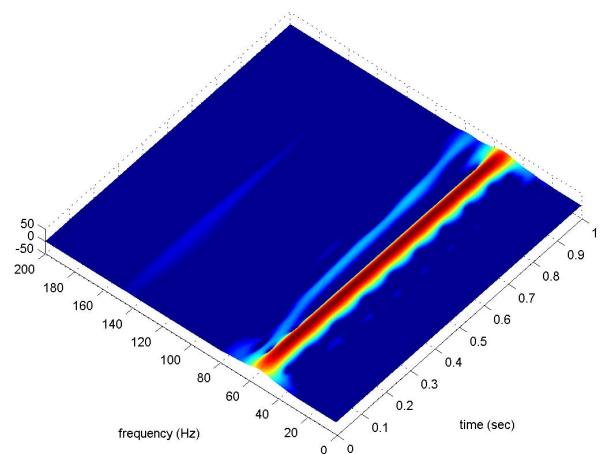


(b) Faulty Generator

Fig. 1. Modulus of the STFT for a) healthy and b) faulty generator.

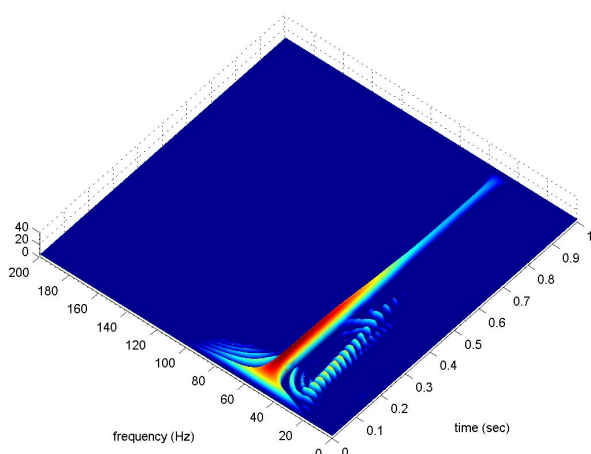


(a) Healthy Generator

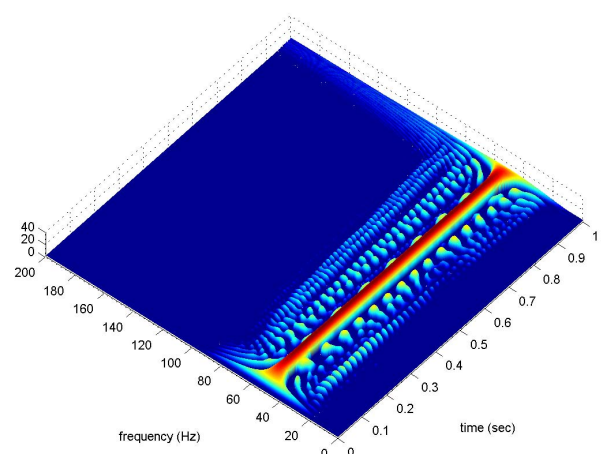


(b) Faulty Generator

Fig. 2. Modulus of the Continuous Wavelet Transform for a) healthy and b) faulty generator.



(a) Healthy Generator



(b) Faulty Generator

Fig. 3. Pseudo Wigner-Ville Distribution for a) healthy and b) faulty generator.

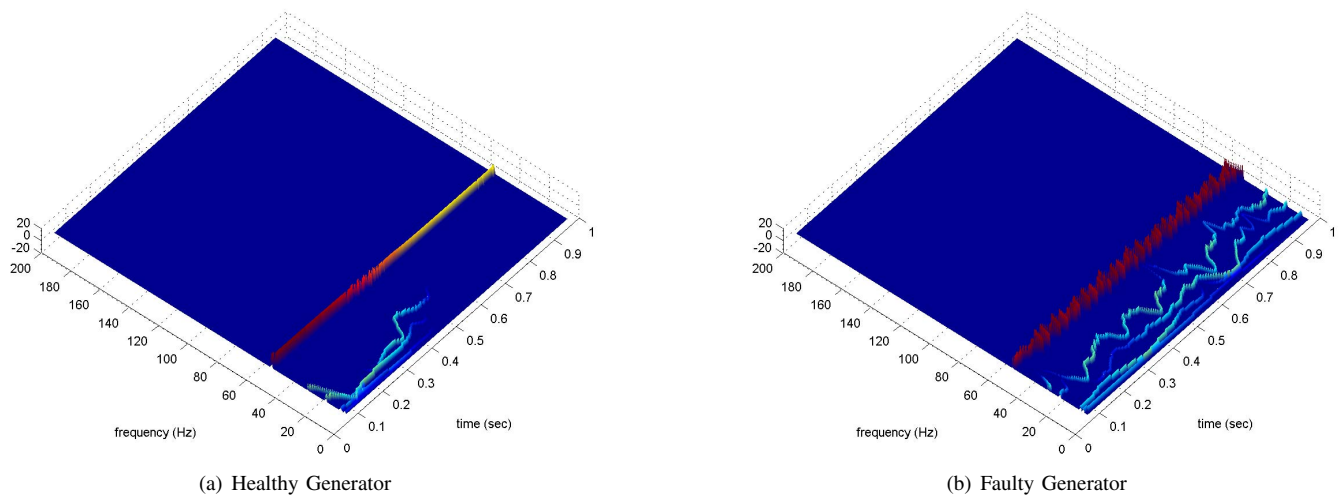


Fig. 4. Hilbert Huang Transform for a) healthy and b) faulty generator.

V. CONCLUSION

This study has presented a comparison of several time-frequency representations for broken rotor bars fault detection in a squirrel-cage wind turbine generator. The Short-Time Fourier, the Continuous Wavelet Transform, the Pseudo Wigner-Ville and the Hilbert-Huang transform have been compared according to their computational complexity, time-frequency resolutions and readability. Simulation results have showed that each technique allows detection of the broken rotor bars fault. For our simulations, Table II reports the advantages and disadvantages of each representation. This table shows that the choice of a particular representation depends on the user's constraints and priorities in terms of computation complexity and easiness of interpretation.

TABLE II
COMPARISON OF SEVERAL TIME-FREQUENCY REPRESENTATIONS

Representation	Computational Cost	Resolution	Readability
STFT	++	--	++
CWT	--	+	++
Pseudo WV	+	+	-
HHT	-	++	+

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