

COM303: Digital Signal Processing

Lecture 20: Image Processing

overview

- ► Introduction to Images and Image Processing
- ► Affine Transforms
- ▶ 2D Fourier Analysis
- ► Image Filters

Overview:

- ▶ Images as multidimensional digital signals
- ▶ 2D signal representations
- ► Basic signals and operators

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In the old, non-PC days...



Please meet ...



- two-dimensional signal $x[n_1, n_2], n_1, n_2 \in \mathbb{Z}$
- lacktriangle indices locate a point on a grid ightarrow pixel
- ▶ grid is usually regularly spaced
- ▶ values $x[n_1, n_2]$ refer to the pixel's appearance

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Digital images: grayscale vs color

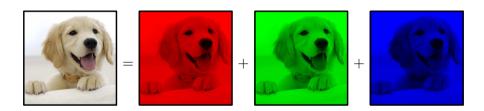
- grayscale images: scalar pixel values
- ▶ color images: multidimensional pixel values in a color space (RGB, HSV, YUV, etc)
- we can consider the single components separately:

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- ► something breaks down
- something is new

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- ▶ new manipulations: affine transforms
- ▶ images are finite-support signals
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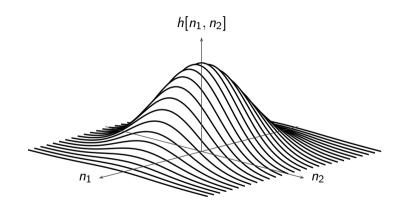
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2D signal processing: the basics

A two-dimensional discrete-space signal:

$$x[n_1, n_2], \qquad n_1, n_2 \in \mathbb{Z}$$

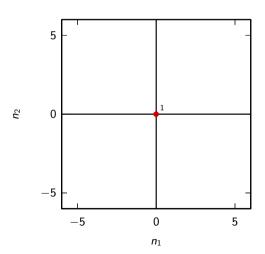
2D signals: Cartesian representation



2D signals: support representation

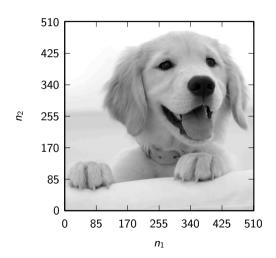
- just show coordinates of nonzero samples
- amplitude may be written along explicitly
- example:

$$\delta[n_1, n_2] = \begin{cases} 1 & \text{if } n_1 = n_2 = 0 \\ 0 & \text{otherwise.} \end{cases}$$



2D signals: image representation

- medium has a certain dynamic range (paper, screen)
- image values are quantized (usually to 8 bits, or 256 levels)
- the eye does the interpolation in space provided the pixel density is high enough



About dynamic ranges...

Images:

- ► human eye: 120dB
- ▶ prints: 12dB to 36dB
- ► LCD: 60dB
- ▶ digital cinema: 90dB

Sounds:

- ▶ human ear: 140dB
- ► speech: 40dB
- ▶ vinyl, tape: 50dB
- ► CD: 96dB

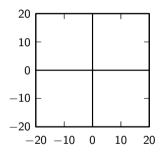
Why 2D?

- ▶ images could be unrolled (printers, fax)
- but what about spatial correlation?

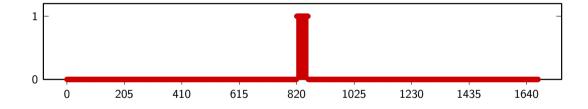
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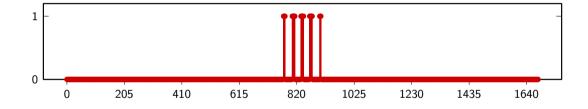
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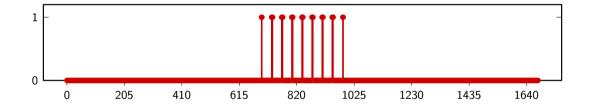
2D vs raster scan

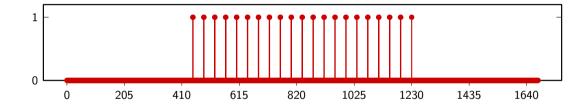


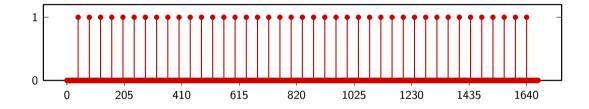
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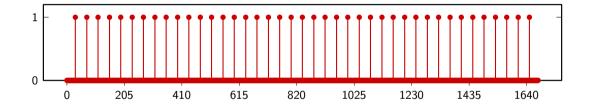


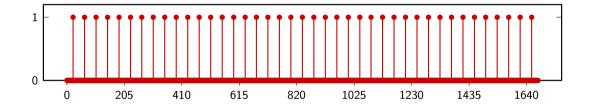


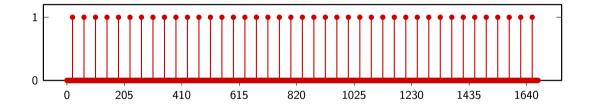


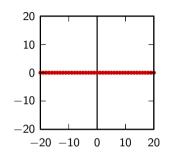


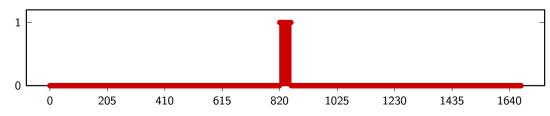


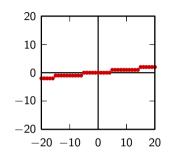


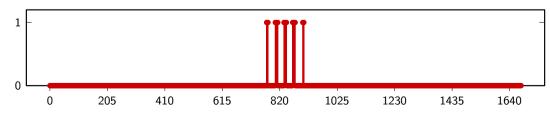


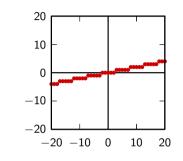


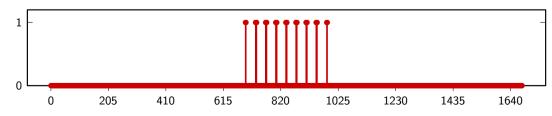


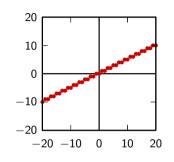


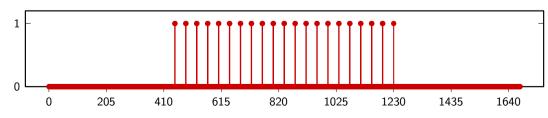


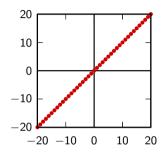


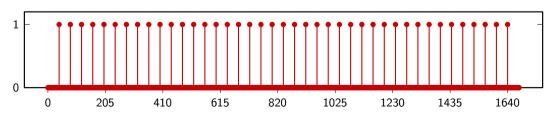


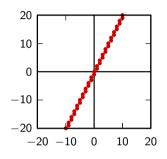


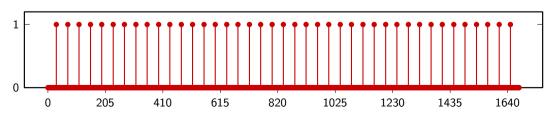


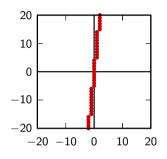


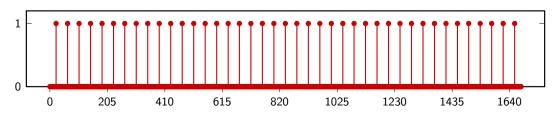


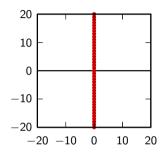


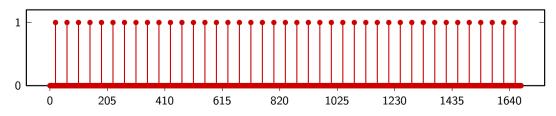






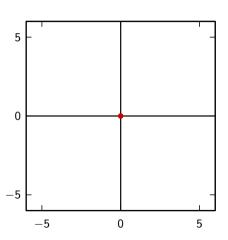






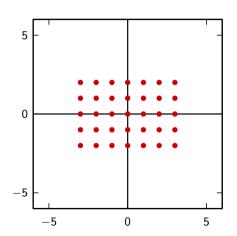
Basic 2D signals: the impulse

$$\delta[n_1, n_2] = \begin{cases} 1 & \text{if } n_1 = n_2 = 0 \\ 0 & \text{otherwise.} \end{cases}$$



Basic 2D signals: the rect

$$\operatorname{rect}\left(rac{n_1}{2N_1},rac{n_2}{2N_2}
ight) = egin{cases} 1 & ext{if } |n_1| < N_1 \ & ext{and } |n_2| < N_2 \ 0 & ext{otherwise;} \end{cases}$$



Separability

$$x[n_1, n_2] = x_1[n_1]x_2[n_2]$$

Separable signals

$$\delta[n_1,n_2] = \delta[n_1]\delta[n_2]$$

$$\operatorname{rect}\left(\frac{n_1}{2N_1}, \frac{n_2}{2N_2}\right) = \operatorname{rect}\left(\frac{n_1}{2N_1}\right) \operatorname{rect}\left(\frac{n_2}{2N_2}\right)$$

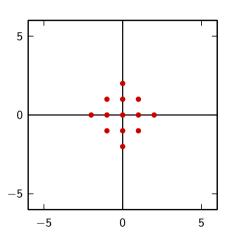
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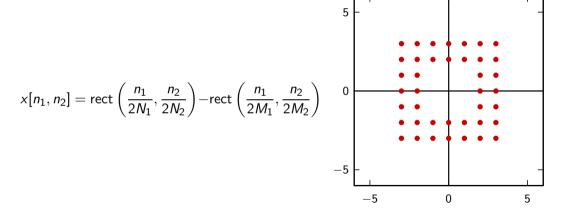
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Nonseparable signal

$$x[n_1, n_2] = \begin{cases} 1 & \text{if } |n_1| + |n_2| < N \\ 0 & \text{otherwise} \end{cases}$$



Nonseparable signal



2D convolution

$$x[n_1, n_2] * h[n_1, n_2] = \sum_{k_1 = -\infty}^{\infty} \sum_{k_2 = -\infty}^{\infty} x[k_1, k_2] h[n_1 - k_1, n_2 - k_2]$$

2D convolution for separable signals

If
$$h[n_1, n_2] = h_1[n_1]h_2[n_2]$$
:

$$x[n_1, n_2] * h[n_1, n_2] = \sum_{k_1 = -\infty}^{\infty} h_1[n_1 - k_1] \sum_{k_2 = -\infty}^{\infty} x[k_1, k_2]h_2[n_2 - k_2]$$

$$= h_1[n_1] * (h_2[n_2] * x[n_1, n_2]).$$

2D convolution for separable signals

If $h[n_1, n_2]$ is an $M_1 \times M_2$ finite-support signal:

- ightharpoonup non-separable convolution: M_1M_2 operations per output sample
- ightharpoonup separable convolution: M_1+M_2 operations per output sample!



Affine transforms

mapping $\mathbb{R}^2 \to \mathbb{R}^2$ that reshapes the coordinate system:

$$\begin{bmatrix} t_1' \\ t_2' \end{bmatrix} = \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix} \begin{bmatrix} t_1 \\ t_2 \end{bmatrix} - \begin{bmatrix} d_1 \\ d_2 \end{bmatrix}$$

$$egin{bmatrix} t_1' \ t_2' \end{bmatrix} = \mathbf{A} egin{bmatrix} t_1 \ t_2 \end{bmatrix} - \mathbf{d}$$

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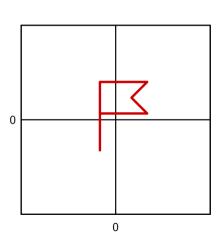
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Translation

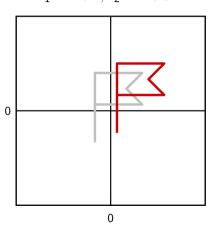
$$\mathbf{A} = egin{bmatrix} 1 & 0 \ 0 & 1 \end{bmatrix} = \mathbf{I}$$
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Translation

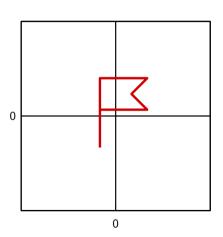
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$$d_1 = -0.7, d_2 = -0.3$$



Scaling

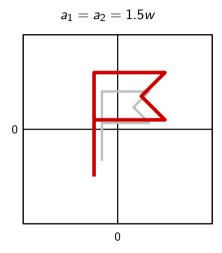
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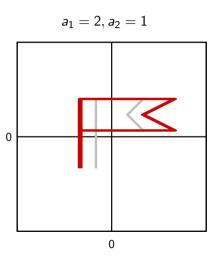
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if $a_1 = a_2$ the aspect ratio is preserved



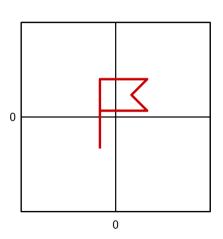
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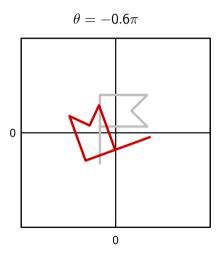
Rotation

$$\mathbf{A} = \begin{bmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{bmatrix}$$
$$\mathbf{d} = 0$$



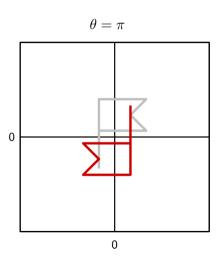
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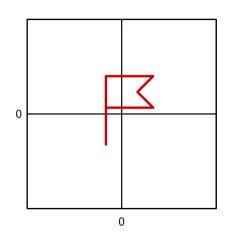


Flips

Horizontal:

$$\mathbf{A} = \begin{bmatrix} -1 & 0 \\ 0 & 1 \end{bmatrix}$$
$$\mathbf{d} = 0$$

$$\mathbf{A} = egin{bmatrix} 1 & 0 \ 0 & -1 \end{bmatrix}$$
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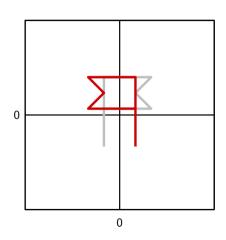


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Shear

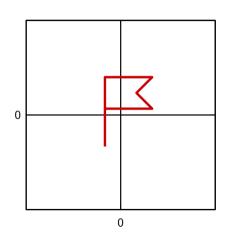
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Shear

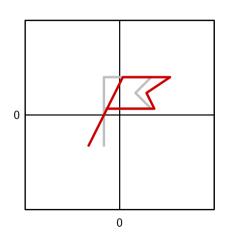
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Affine transforms in discrete-space

$$\begin{bmatrix} t_1' \\ t_2' \end{bmatrix} = \mathbf{A} \begin{bmatrix} n_1 \\ n_2 \end{bmatrix} - \mathbf{d} \quad \in \mathbb{R}^2 \neq \mathbb{Z}^2$$

Solution for images

- ▶ take each output point $y[m_1, m_2]$
- ▶ apply the *inverse* transform to $[m_1, m_2]$ and find the *source* point's coordinates:

$$\begin{bmatrix} t_1 \\ t_2 \end{bmatrix} = \mathbf{A}^{-1} \begin{bmatrix} m_1 + d_1 \\ m_2 + d_2 \end{bmatrix};$$

▶ if source point not on source grid, write

$$(t_1, t_2) = (\eta_1 + \tau_1, \eta_2 + \tau_2), \qquad \eta_{1,2} \in \mathbb{Z}, \quad 0 \le \tau_{1,2} < 1$$

and interpolate from the surrounding original grid points

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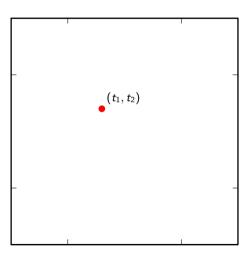
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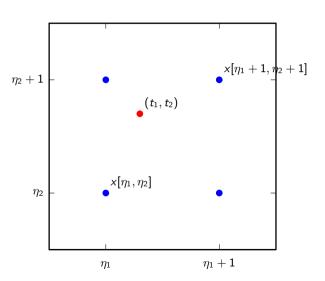
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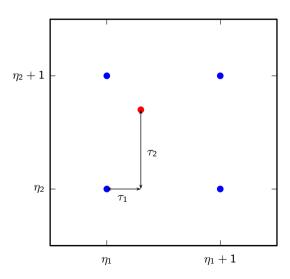
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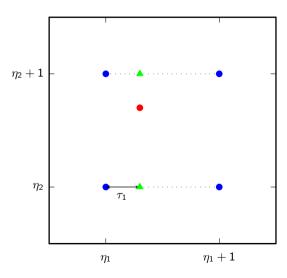
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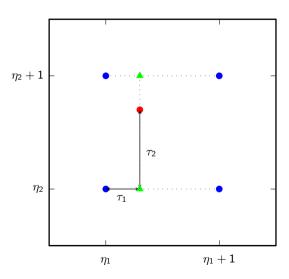
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If we use a first-order interpolator:

$$y[m_1, m_2] = (1 - \tau_1)(1 - \tau_2)x[\eta_1, \eta_2] + \tau_1(1 - \tau_2)x[\eta_1 + 1, \eta_2]$$

+ $(1 - \tau_1)\tau_2x[\eta_1, \eta_2 + 1] + \tau_1\tau_2x[\eta_1 + 1, \eta_2 + 1]$

Shearing



2D Fourier Analysis

$$X[k_1, k_2] = \sum_{n_1=0}^{N_1-1} \sum_{n_2=0}^{N_2-1} x[n_1, n_2] e^{-j\frac{2\pi}{N_1}n_1k_1} e^{-j\frac{2\pi}{N_2}n_2k_2}$$

$$x[n_1, n_2] = \frac{1}{N_1 N_2} \sum_{k_1=0}^{N_1-1} \sum_{k_2=0}^{N_2-1} X[k_1, k_2] e^{j\frac{2\pi}{N_1} n_1 k_1} e^{j\frac{2\pi}{N_2} n_2 k_2}$$

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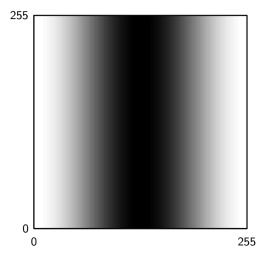
2D-DFT Basis Vectors

There are N_1N_2 orthogonal basis vectors for an $N_1 \times N_2$ image:

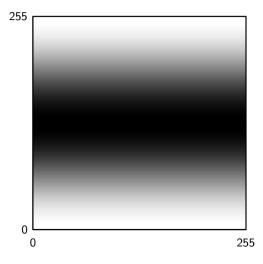
$$w_{k_1,k_2}[n_1,n_2] = e^{j\frac{2\pi}{N_1}n_1k_1}e^{j\frac{2\pi}{N_2}n_2k_2}$$

for
$$n_1, k_1 = 0, 1, \dots, N_1 - 1$$
 and $n_2, k_2 = 0, 1, \dots, N_2 - 1$

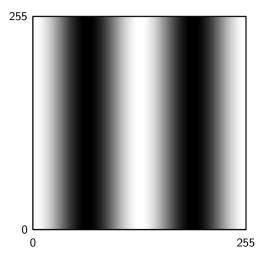
2D-DFT basis vectors for $k_1 = 1, k_2 = 0$ (real part)



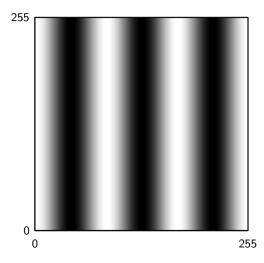
2D-DFT basis vectors for $k_1 = 0, k_2 = 1$ (real part)



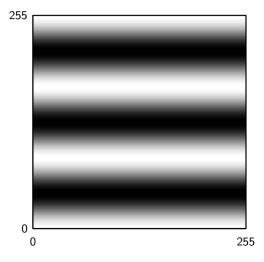
2D-DFT basis vectors for $k_1 = 2, k_2 = 0$ (real part)



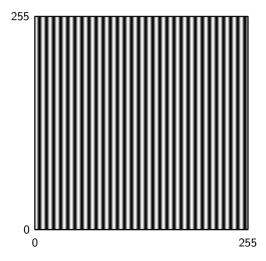
2D-DFT basis vectors for $k_1 = 3$, $k_2 = 0$ (real part)



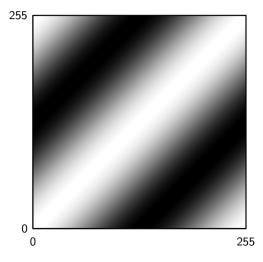
2D-DFT basis vectors for $k_1 = 0, k_2 = 3$ (real part)



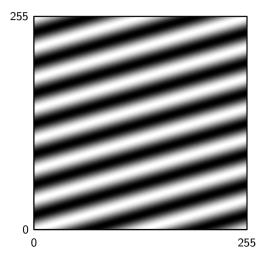
2D-DFT basis vectors for $k_1 = 30, k_2 = 0$ (real part)



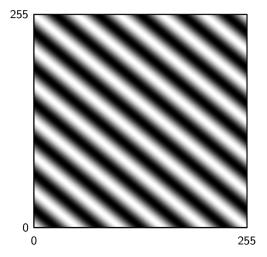
2D-DFT basis vectors for $k_1 = 1, k_2 = 1$ (real part)



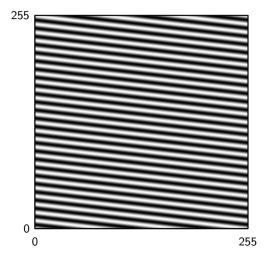
2D-DFT basis vectors for $k_1 = 2, k_2 = 7$ (real part)



2D-DFT basis vectors for $k_1 = 5$, $k_2 = 250$ (real part)



2D-DFT basis vectors for $k_1 = 3$, $k_2 = 230$ (real part)



2D-DFT basis functions are separable, and so is the 2D-DFT:

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2D DFT in matrix form

- ▶ finite-support 2D signal can be written as a matrix **x**
- ▶ $N_1 \times N_2$ image is an $N_2 \times N_1$ matrix (n_1 spans the columns, n_2 spans the rows)
- recall also the N × N DFT matrix

$$\mathbf{W}_{N} = \begin{bmatrix} 1 & 1 & 1 & 1 & \dots & 1 \\ 1 & W_{N}^{1} & W_{N}^{2} & W^{3} & \dots & W_{N}^{N-1} \\ 1 & W_{N}^{2} & W^{4} & W_{N}^{6} & \dots & W_{N}^{2(N-1)} \\ & & & & \dots & & \\ 1 & W_{N}^{N-1} & W_{N}^{2(N-1)} & W_{N}^{3(N-1)} & \dots & W_{N}^{(N-1)^{2}} \end{bmatrix}$$

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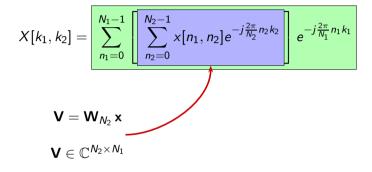
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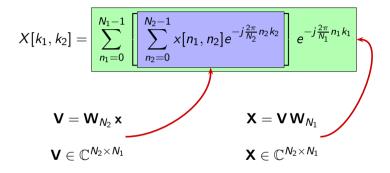
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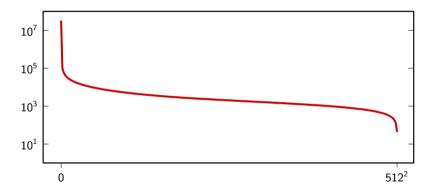
$$X[k_1, k_2] = \begin{bmatrix} \sum_{n_1=0}^{N_1-1} \begin{bmatrix} \sum_{n_2=0}^{N_2-1} x[n_1, n_2] e^{-j\frac{2\pi}{N_2}n_2k_2} \end{bmatrix} e^{-j\frac{2\pi}{N_1}n_1k_1} \\ \mathbf{V} = \mathbf{W}_{N_2} \mathbf{x} \\ \mathbf{V} \in \mathbb{C}^{N_2 \times N_1} \\ \mathbf{X} = \mathbf{W}_{N_2} \mathbf{x} \mathbf{W}_{N_1} \end{bmatrix}$$

54

How does a 2D-DFT look like?

- ▶ try to show the magnitude as an image
- ▶ problem: the range is too big for the grayscale range of paper or screen
- ▶ try to normalize: $|X'[n_1, n_2]| = |X[n_1, n_2]| / \max |X[n_1, n_2]|$
- ▶ but it doesn't work...

DFT coefficients sorted by magnitude



Dealing with HDR images

if the image is high dynamic range we need to compress the levels

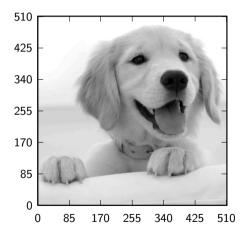
- ▶ remove flagrant outliers (e.g. $X[0,0] = \sum \sum x[n_1,n_2]$)
- use a nonlinear mapping: e.g. $y = x^{1/3}$ after normalization $(x \le 1)$

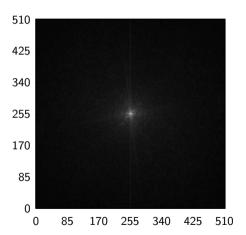
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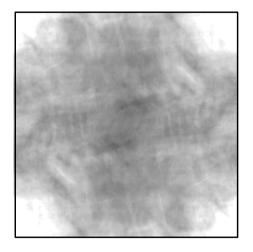
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How does a 2D-DFT look like?





DFT magnitude doesn't carry much information



DFT phase, on the other hand...



Image frequency analysis

- ▶ most of the information is contained in image's edges
- edges are points of abrupt change in signal's values
- lacktriangle edges are a "space-domain" feature ightarrow not captured by DFT's magnitude
- phase alignment is important for reproducing edges



Overview:

- ► Filters for image processing
- Classification
- Examples

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Analogies with 1D filters

- ► linearity
- ► *space* invariance
- ▶ impulse response
- ► frequency response
- stability
- ▶ 2D CCDE

- ▶ interesting images contain lots of *semantics*: different information in different areas
- space-invariant filters process everything in the same way
- but we should process things differently
 - edges
 - gradients
 - textures
 - ...

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- ► highpass, lowpass, ...
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- ► nonlinear phase (edges!)
- border effects
- ▶ stability: the fundamental theorem of algebra doesn't hold in multiple dimensions!
- computability

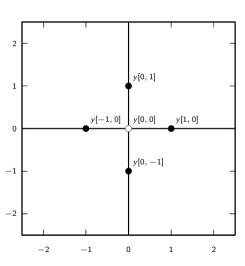
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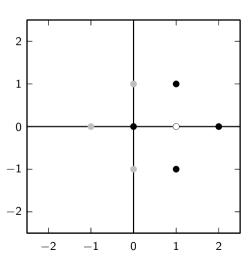
A noncomputable CCDE

$$y[n_1,n_2] = a_0y[n_1+1,n_2] + a_1y[n_1,n_2-1] + a_2y[n_1-1,n_2] + a_3y[n_1,n_2+1] + x[n_1,n_2];$$



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Practical FIR filters

- ▶ generally zero centered (causality not an issue) ⇒ odd number of taps in both directions
- per-sample complexity:
 - $M_1 M_2$ for nonseparable impulse responses
 - $M_1 + M_2$ for separable impulse responses
- obviously always stable

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$$y[n_1, n_2] = \frac{1}{(2N+1)^2} \sum_{k_1=-N}^{N} \sum_{k_2=-N}^{N} x[n_1 - k_1, n_2 - k_2]$$

$$h[n_1, n_2] = \frac{1}{(2N+1)^2} \operatorname{rect}\left(\frac{n_1}{2N}, \frac{n_2}{2N}\right)$$

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$$h[n_1, n_2] = \frac{1}{9} \begin{bmatrix} 1 & 1 & 1 \\ 1 & 1 & 1 \\ 1 & 1 & 1 \end{bmatrix}$$



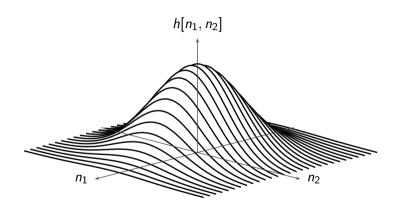


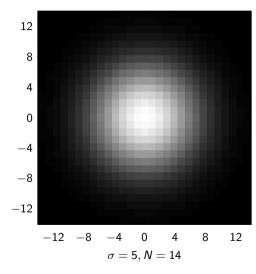
11 imes 11 MA

 $51 \times 51 \text{ MA}$

$$h[n_1, n_2] = \frac{1}{2\pi\sigma^2} e^{-\frac{n_1^2 + n_2^2}{2\sigma^2}}, \qquad |n_1, n_2| < N$$

with
$$N \approx 3\sigma$$







 $\sigma = 1.8, 11 \times 11 \text{ blur}$



 $\sigma = 8.7, 51 \times 51$ blur

approximation of the first derivative in the horizontal direction:

$$s_o[n_1, n_2] = \begin{bmatrix} -1 & 0 & 1 \\ -2 & 0 & 2 \\ -1 & 0 & 1 \end{bmatrix}$$

separability and structure:

$$S_o[n_1, n_2] = \begin{bmatrix} 1 \\ 2 \\ 1 \end{bmatrix} \begin{bmatrix} -1 & 0 & 1 \end{bmatrix}$$

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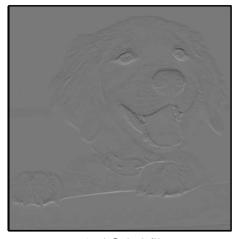
$$s_o[n_1, n_2] = \begin{bmatrix} 1 \\ 2 \\ 1 \end{bmatrix} \begin{bmatrix} -1 & 0 & 1 \end{bmatrix}$$

approximation of the first derivative in the vertical direction:

$$s_{\nu}[n_1, n_2] = \begin{bmatrix} -1 & -2 & 1 \\ 0 & 0 & 0 \\ 1 & 2 & 1 \end{bmatrix}$$



horizontal Sobel filter



vertical Sobel filter

Sobel operator

approximation for the square magnitude of the gradient:

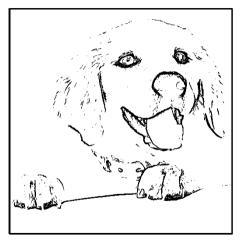
$$|\nabla x[n_1,n_2]|^2 = |s_o[n_1,n_2] * x[n_1,n_2]|^2 + |s_v[n_1,n_2] * x[n_1,n_2]|^2$$

("operator" because it's nonlinear)

Gradient approximation for edge detection



Sobel operator



thresholeded Sobel operator

Laplacian operator

Laplacian of a function in continuous-space:

$$\Delta f(t_1, t_2) = \frac{\partial^2 f}{\partial t_1^2} + \frac{\partial^2 f}{\partial t_2^2}$$

Laplacian operator

approximating the Laplacian; start with a Taylor expansion

$$f(t+\tau) = \sum_{n=0}^{\infty} \frac{f^{(n)}(t)}{n!} \tau^n$$

and compute the expansion in $(t + \tau)$ and $(t - \tau)$:

$$f(t + \tau) = f(t) + f'(t)\tau + \frac{1}{2}f''(t)\tau^{2}$$
$$f(t - \tau) = f(t) - f'(t)\tau + \frac{1}{2}f''(t)\tau^{2}$$

Laplacian operator

by rearranging terms:

$$f''(t) = \frac{1}{\tau^2}(f(t-\tau) - 2f(t) + f(t+\tau))$$

which, on the discrete grid, is the FIR $h[n] = \begin{bmatrix} 1 & -2 & 1 \end{bmatrix}$

Laplacian

summing the horizontal and vertical components:

$$h[n_1, n_2] = \begin{bmatrix} 0 & 1 & 0 \\ 1 & -4 & 1 \\ 0 & 1 & 0 \end{bmatrix}$$

Laplacian

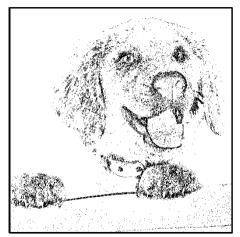
If we use the diagonals too:

$$h[n_1, n_2] = \begin{bmatrix} 1 & 1 & 1 \\ 1 & -8 & 1 \\ 1 & 1 & 1 \end{bmatrix}$$

Laplacian for Edge Detection



Laplacian operator



thresholeded Laplacian operator