

COM303: Digital Signal Processing

Lecture 8: Numerical Fourier Analysis

Overview:

- ▶ the short-time Fourier transform (aka the "spectrogram")
- ▶ the Fast Fourier Transform algorithm

the short-time Fourier transform (STFT)

Overview:

- ▶ Time vs frequency representations
- ▶ The STFT and the spectrogram
- ▶ Time-Frequency tilings

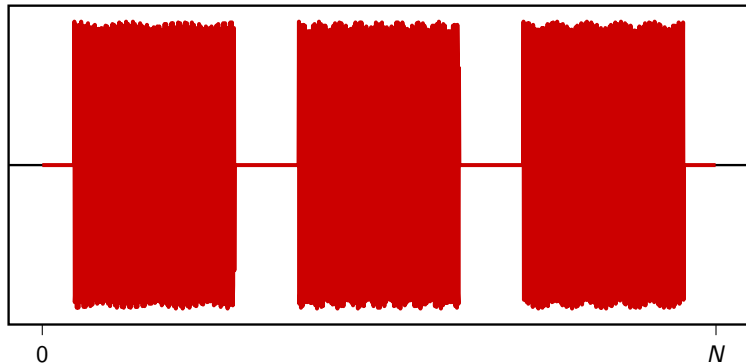
Dual-Tone Multi Frequency dialing



DTMF signaling

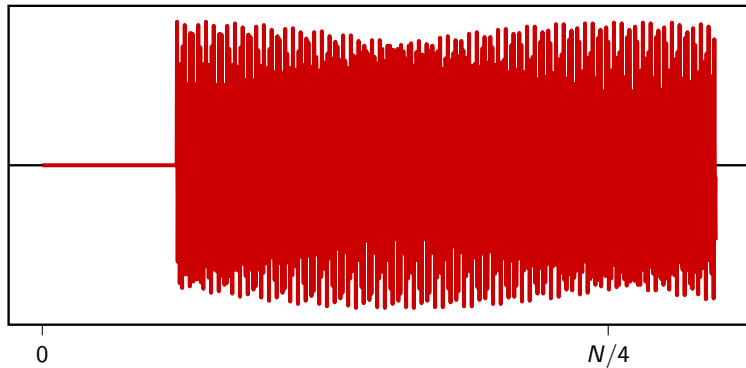
	1209Hz	1336Hz	1477Hz
697Hz	1	2	3
770Hz	4	5	6
852Hz	7	8	9
941Hz	*	0	#

1-5-9 in time

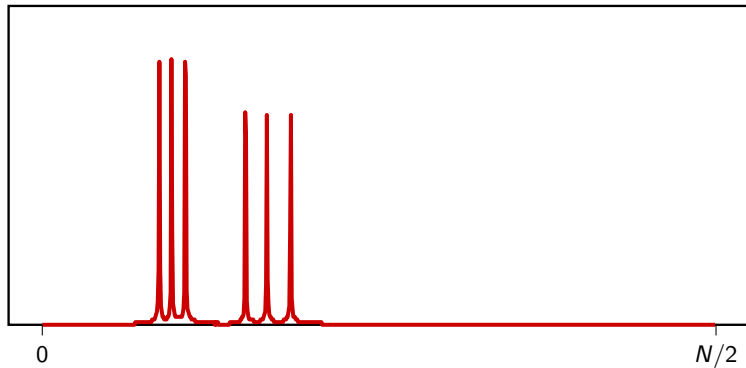


Play

1-5-9 in time (detail)



1-5-9 in frequency (magnitude)



The fundamental tradeoff

- ▶ time representation obfuscates frequency
- ▶ frequency representation obfuscates time

Short-Time Fourier Transform

Idea:

- ▶ take small signal pieces of length L
- ▶ look at the DFT of each piece:

$$X[m; k] = \sum_{n=0}^{L-1} x[m + n] e^{-j\frac{2\pi}{L}nk}$$

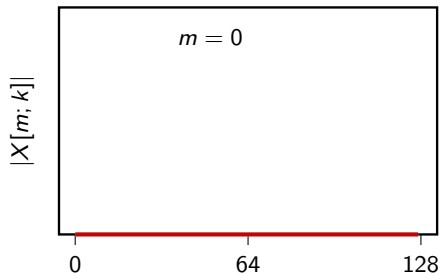
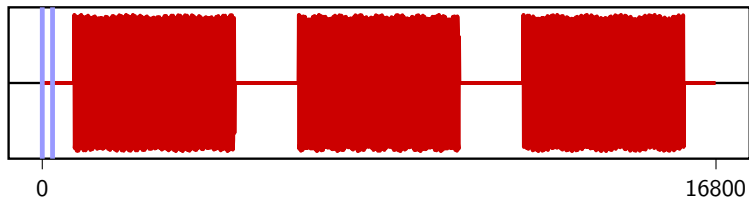
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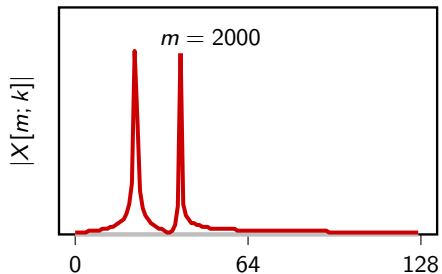
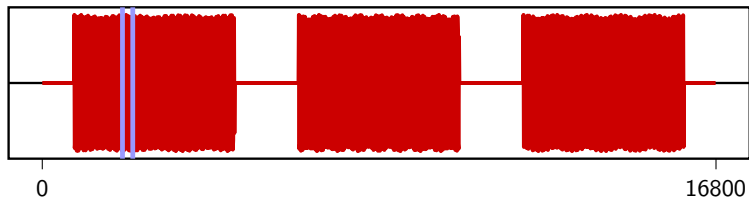
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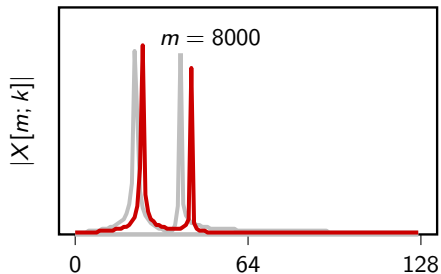
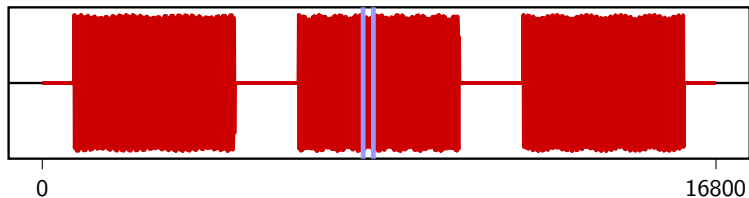
Short-Time Fourier Transform ($L = 256$)



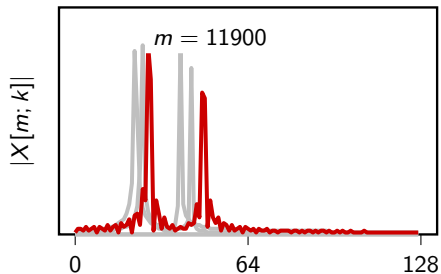
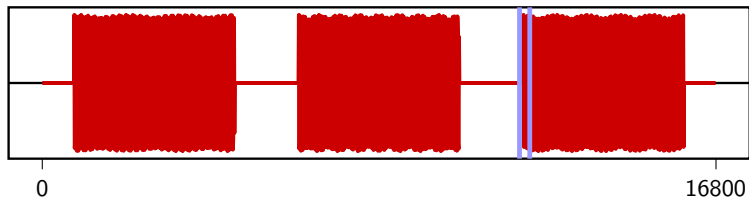
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Short-Time Fourier Transform ($L = 256$)



The Spectrogram

Idea:

- ▶ color-code the magnitude: dark is small, white is large
- ▶ use $10 \log_{10}(|X[m; k]|)$ to see better (power in dBs)
- ▶ plot spectral slices one after another

The Spectrogram

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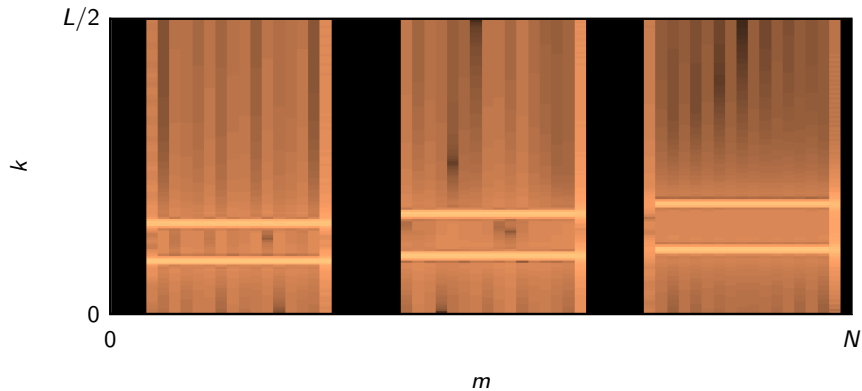
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DTMF spectrogram



Labeling the Spectrogram

If we know the “system clock” $F_s = 1/T_s$ we can label the axis

- ▶ highest positive frequency: $F_s/2$ Hz
- ▶ frequency resolution: F_s/L Hz
- ▶ width of time slices: LT_s seconds

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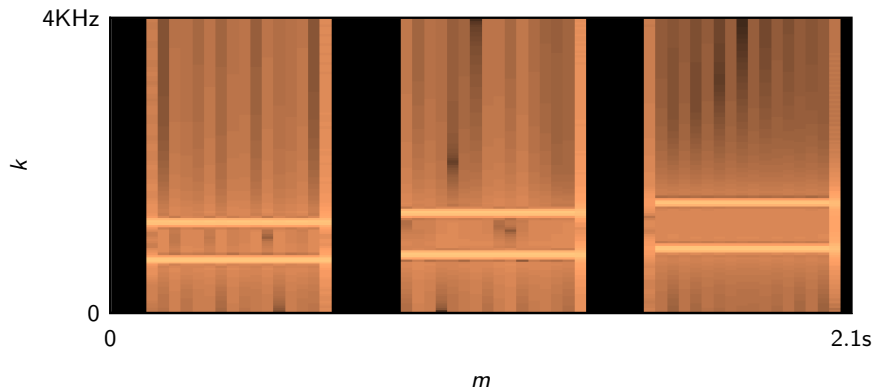
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DTMF spectrogram ($F_s = 8000$)



The Spectrogram

Questions:

- ▶ width of the analysis window?
- ▶ position of the windows (overlapping?)
- ▶ shape of the window (weighing the samples)

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Wideband vs Narrowband

Long window: narrowband spectrogram

- ▶ long window \Rightarrow more DFT points \Rightarrow more frequency resolution
- ▶ long window \Rightarrow more “things can happen” \Rightarrow less precision in time

Short window: wideband spectrogram

- ▶ short window \Rightarrow many time slices \Rightarrow precise location of transitions
- ▶ short window \Rightarrow fewer DFT points \Rightarrow poor frequency resolution

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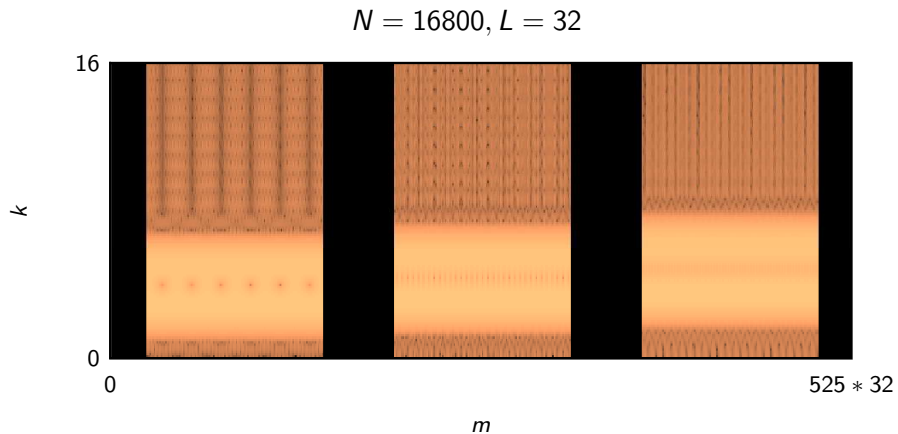
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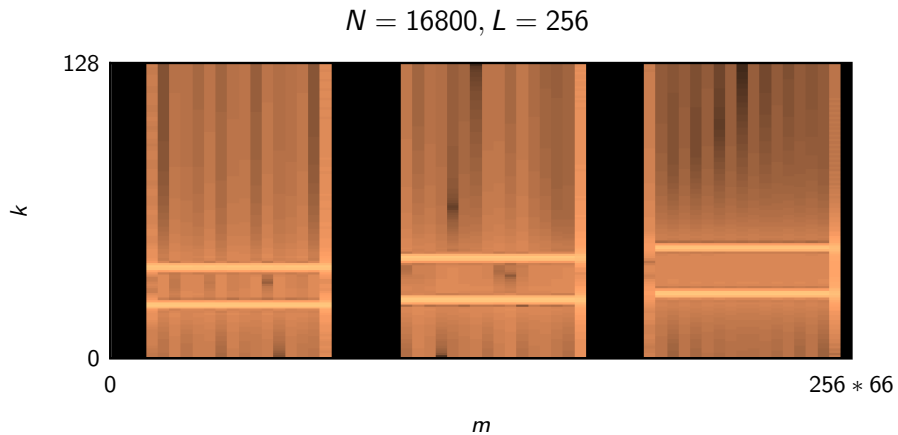
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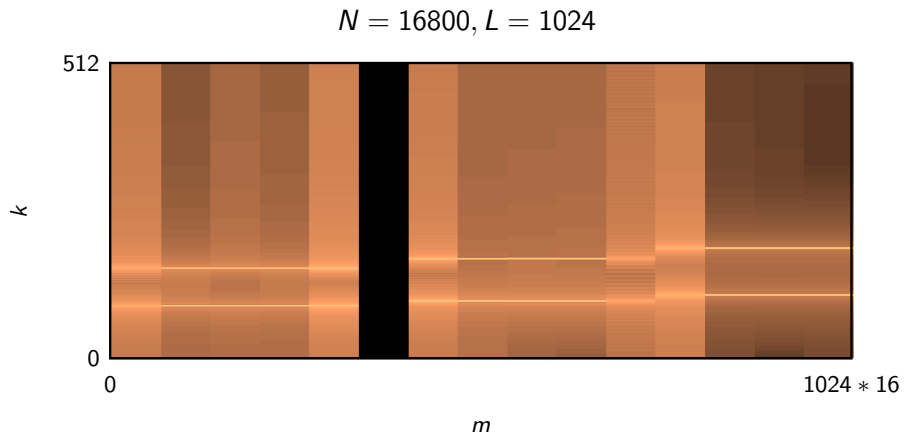
DTMF spectrogram (wideband)



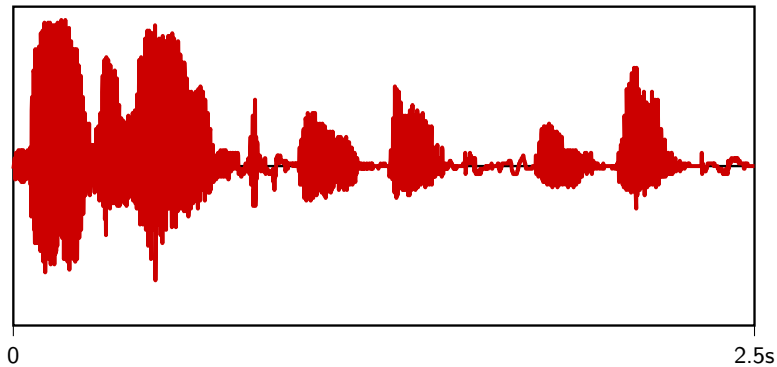
DTMF spectrogram



DTMF spectrogram (narrowband)



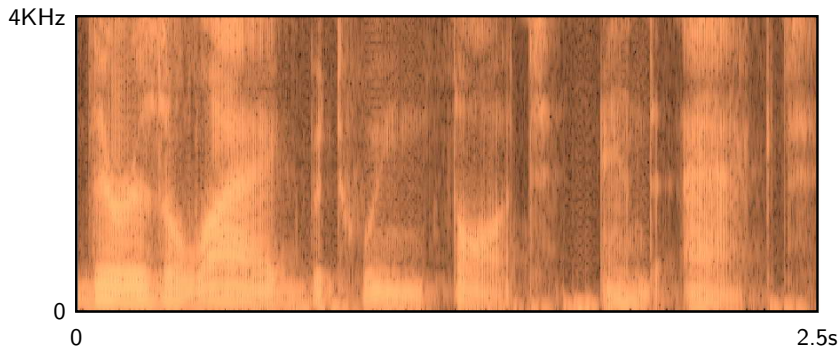
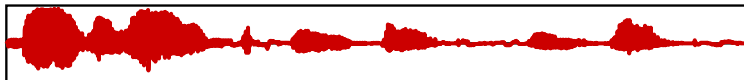
Speech analysis



Play

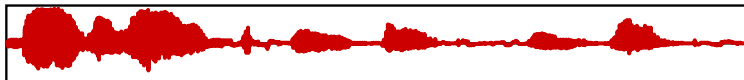
Speech analysis

8ms analysis window (125Hz frequency bins) , 4ms shifts

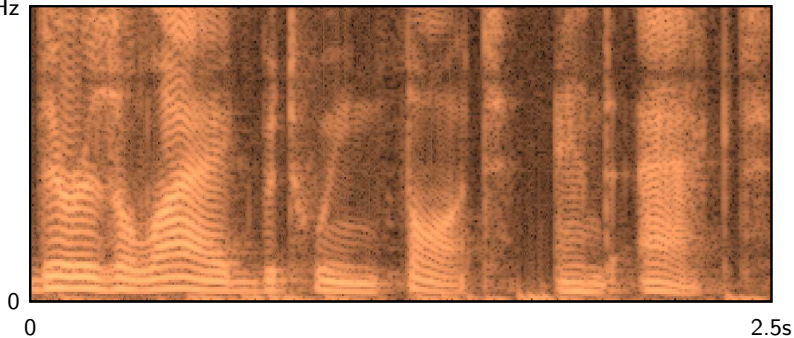


Speech analysis

32ms analysis window (31Hz frequency bins), 4ms shifts

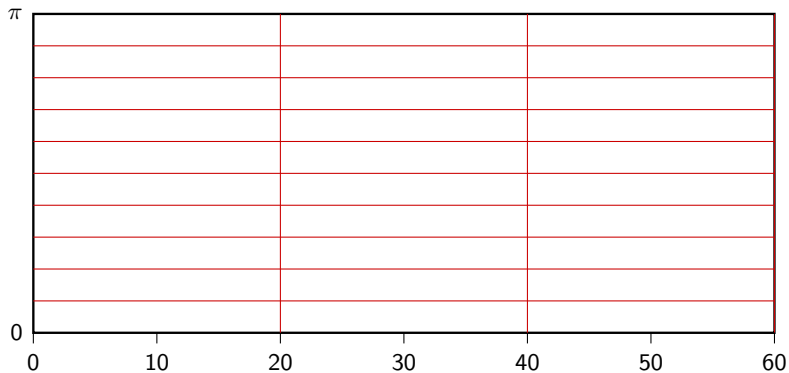


4KHz



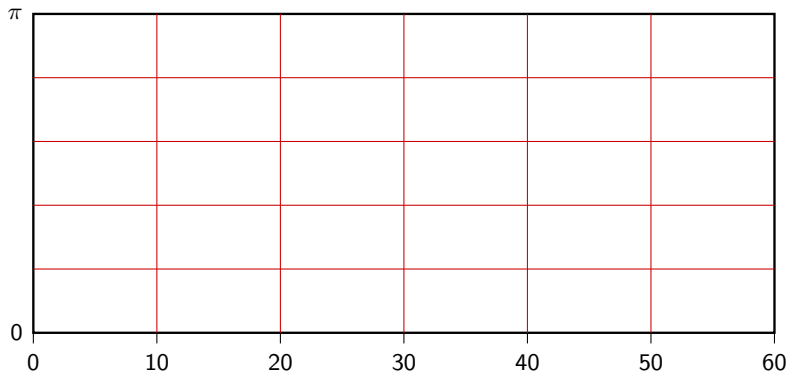
Time-Frequency tiling

$$L = 20$$



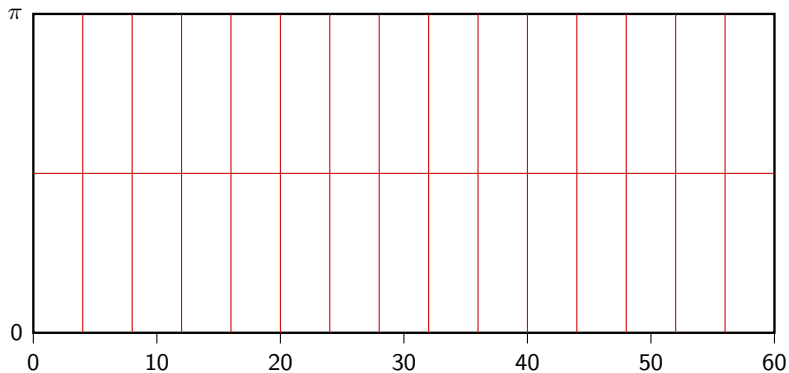
Time-Frequency tiling

$$L = 10$$



Time-Frequency tiling

$$L = 4$$



Food for thought

- ▶ time “resolution” $\Delta t = L$
- ▶ frequency “resolution” $\Delta f = 2\pi/L$
- ▶ $\Delta t \Delta f = 2\pi$

uncertainty principle!

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Even more food for thought

more sophisticated tilings of the time-frequency planes
can be obtained with the *wavelet* transform

the Fast Fourier Transform (FFT)

Overview

- ▶ A bit of history: From Gauss to the fastest FFT in the west
- ▶ Small DFT matrices
- ▶ The Cooley-Tukey FFT
- ▶ Decimation-in-Time FFT for length 2^N FFTs
- ▶ Conclusions: There are FFTs for any length!

Fourier had the Fourier transform



But Gauss had the FFT all along ;)



History

- ▶ Gauss computes trigonometric series efficiently in 1805
- ▶ Fourier invents Fourier series in 1807
- ▶ People start computing Fourier series, and develop tricks
- ▶ Good comes up with an algorithm in 1958
- ▶ Cooley and Tukey (re)-discover the fast Fourier transform algorithm in 1965 for N a power of a prime
- ▶ Winograd combines all methods to give the most efficient FFTs in 1978
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The DFT matrix

- ▶ $W_N = e^{-j\frac{2\pi}{N}}$ (or simply W when N is clear from the context)
- ▶ powers of N can be taken modulo N , since $W^N = 1$.
- ▶ DFT Matrix of size N by N :

$$\mathbf{W} = \begin{bmatrix} 1 & 1 & 1 & 1 & \dots & 1 \\ 1 & W^1 & W^2 & W^3 & \dots & W^{N-1} \\ 1 & W^2 & W^4 & W^6 & \dots & W^{2(N-1)} \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W^{N-1} & W^{2(N-1)} & W^{3(N-1)} & \dots & W^{(N-1)^2} \end{bmatrix}$$

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Small DFT matrices: $N = 2$

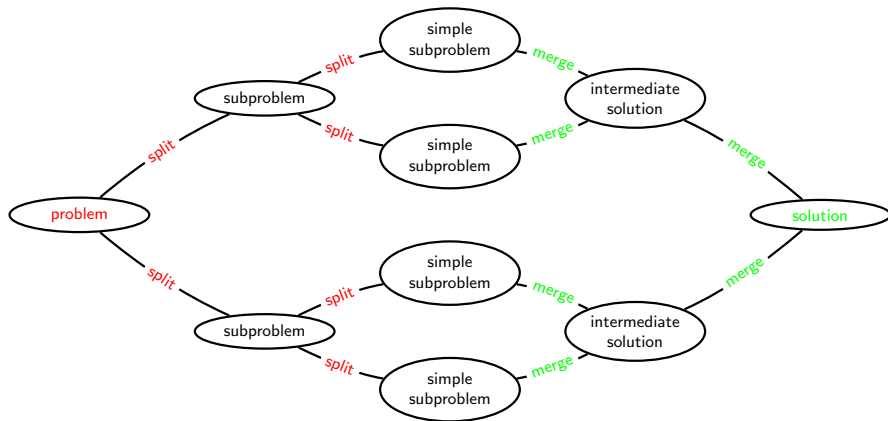
$$\mathbf{W}_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

Small DFT matrices: $N = 4$

$$\mathbf{W}_4 = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W & W^2 & W^3 \\ 1 & W^2 & W^4 & W^6 \\ 1 & W^3 & W^6 & W^9 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W & W^2 & W^3 \\ 1 & W^2 & 1 & W^2 \\ 1 & W^3 & W^2 & W \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

Divide et impera - Divide and Conquer (Julius Caesar)

Divide and conquer is a standard attack for developing fast algorithms.



Divide and Conquer for DFT - One step

Recall: computing $\mathbf{X} = \mathbf{W}_N \mathbf{x}$ has complexity $O(N^2)$.

Idea:

- ▶ Assume N even
- ▶ Split the problem into two subproblems of size $N/2$; cost is $N^2/4$ each
- ▶ If the cost to recover the full solution is linear $N \dots$
- ▶ ... the divide-and-conquer solution costs $N^2/2 + N$ for one step
- ▶ For $N \geq 4$ this is better than N^2

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Divide and Conquer for DFT - One step

Graphically

- ▶ Split DFT input into 2 pieces of size $N/2$

Divide and Conquer for DFT - One step

Graphically

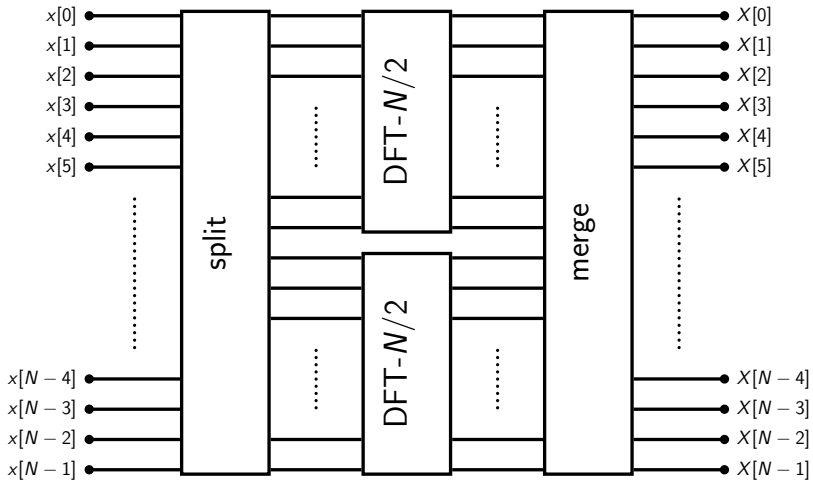
- ▶ Split DFT input into 2 pieces of size $N/2$
- ▶ Compute two DFT's of size $N/2$

Divide and Conquer for DFT - One step

Graphically

- ▶ Split DFT input into 2 pieces of size $N/2$
- ▶ Compute two DFT's of size $N/2$
- ▶ Merge the two results

Divide and Conquer for DFT - One step



Divide and Conquer for DFT - Multiple steps

Idea: if $N = 2^K$, divide and conquer can be reapplied!

- ▶ Cut the two problems of size $N/2$ into 4 problems of size $N/4$
- ▶ Assume complexity to recover the full solution still linear, e.g. N at each step
- ▶ You can do this $\log_2 N - 1 = K - 1$ times, until problem of size 2 is obtained
- ▶ The divide-and-conquer solution has therefore complexity of order $N \log_2 N$
- ▶ For $N \geq 4$ this is much better than N^2 !

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Divide and Conquer for DFT - Multiple steps

N	N^2	$N \log N$
10	100	10
100	10,000	200
1000	1M	3000
10,000	100M (10^8)	40,000 ($4 \cdot 10^4$)
100,000	10B (10^{10})	500,000 ($5 \cdot 10^5$)

Divide and Conquer for DFT - Multiple steps

Graphically

- ▶ Split DFT input into 2, 4 and 8 pieces of sizes $N/2$, $N/4$ and $N/8$, respectively
- ▶ Compute 8 DFT's of size $N/8$
- ▶ Merge the results successively into DFT's of size $N/4$, $N/2$ and finally N

Divide and Conquer for DFT - Multiple steps

Graphically

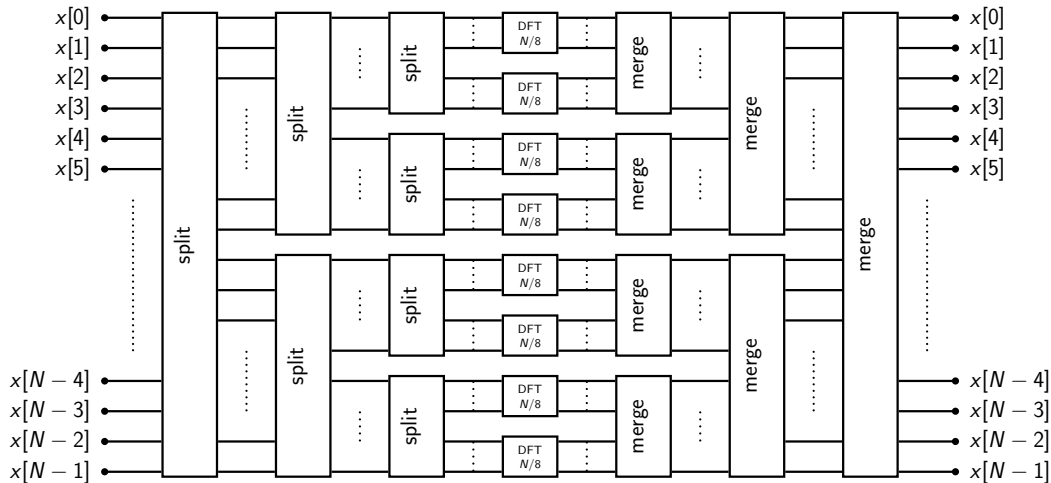
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Divide and Conquer for DFT - Multiple steps

Graphically

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Divide and Conquer for DFT - Multiple steps



Divide and Conquer for DFT- Analysis of DIT

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}, \quad k = 0, 1, \dots, N-1, \quad W_N = e^{-j\frac{2\pi}{N}}$$

Idea (a good guess is half of the answer!):

- ▶ break input into even and odd indexed terms (so-called "decimation in time"):

$$x[n], \quad n = 0, 1, \dots, N-1 \longrightarrow x[2n] \text{ and } x[2n+1], \quad n = 0, \dots, N/2-1$$

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Divide and Conquer for DFT- Analysis of DIT

Consider even and odd inputs separately:

$$\begin{aligned} X[k] &= \sum_{n=0}^{N/2-1} x[2n] W_N^{2nk} + \sum_{n=0}^{N/2-1} x[2n+1] W_N^{(2n+1)k} \\ &= \sum_{n=0}^{N/2-1} x[2n] W_N^{2nk} + \sum_{n=0}^{N/2-1} x[2n+1] W_N^{2nk+k} \\ &= \sum_{n=0}^{N/2-1} x[2n] W_{N/2}^{nk} + W_N^k \sum_{n=0}^{N/2-1} x[2n+1] W_{N/2}^{nk} \\ &= X_A[k] + W_N^k X_B[k], \quad k = 0, 1, \dots, N-1 \end{aligned}$$

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Divide and Conquer for DFT- Analysis of DIT

hmmm, we haven't gained much so far:

- ▶ both $X_A[k]$ and $X_B[k]$ require $N/2$ multiplications
- ▶ multiplying the second DFT by W_N^k requires another multiplication
- ▶ to compute for all k we need $N(N/2 + N/2 + 1) \approx N^2$
- ▶ but here comes the trick!

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Divide and Conquer for DFT- Analysis of DIT

Consider now the first and second half of the outputs separately:

$$\begin{aligned}X[k] &= X_A[k] + W^k X_B[k] \\&= \sum_{n=0}^{N/2-1} x[2n] W_{N/2}^{nk} + W^k \sum_{n=0}^{N/2-1} x[2n+1] W_{N/2}^{nk} \\X[k + N/2] &= \sum_{n=0}^{N/2-1} x[2n] W_{N/2}^{n(k+N/2)} + W^{k+N/2} \sum_{n=0}^{N/2-1} x[2n+1] W_{N/2}^{n(k+N/2)} \\&= \sum_{n=0}^{N/2-1} x[2n] W_{N/2}^{nk} - W^k \sum_{n=0}^{N/2-1} x[2n+1] W_{N/2}^{nk} \\&= X_A[k] - W_N^k X_B[k], \quad k = 0, 1, \dots, N/2 - 1\end{aligned}$$

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Divide and Conquer for DFT- Analysis of DIT

so the trick is that we only need to compute for half the range of k :

- ▶ both $X_A[k]$ and $X_B[k]$ require $N/2$ multiplications
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- ▶ to compute for all k we need $(N/2)(N/2 + N/2 + 1) \approx N^2/2$
- ▶ the rest is just sums and differences

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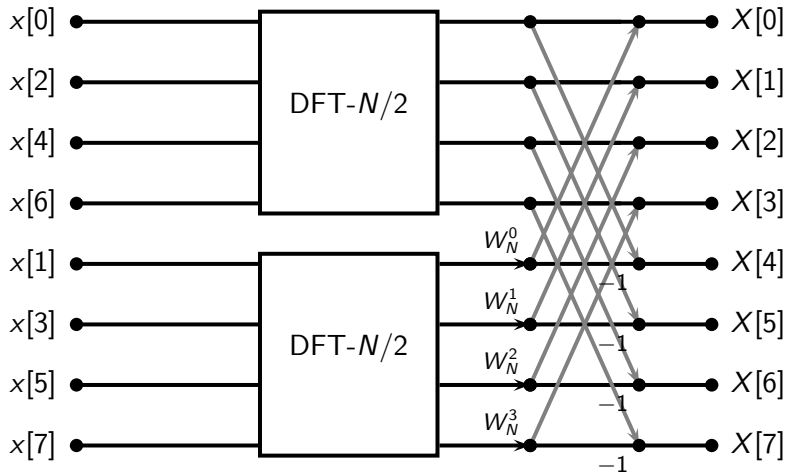
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Divide and Conquer for DFT- Analysis of DIT



Divide and Conquer for DFT- Analysis of DIT

So, what is the complexity now?

- ▶ Split DFT input into 2 pieces of size $N/2$: free!
- ▶ Compute 2 DFT- $N/2$: twice $(N/2)^2$, or $N^2/2$
- ▶ Merge the two results: multiplication by $N/2$ complex numbers W^k
- ▶ Total: $N^2/2 + N/2$ which is indeed smaller than N^2 for any $N \geq 4$,
- ▶ In general, about half the complexity of the initial problem!

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So, what if we repeat the process?

- ▶ Go until DFT-2, since that is trivial (sum and difference)
- ▶ Requires $\log_2 N - 1$ steps
- ▶ Each step requires a merger of order $N/2$ multiplications and N additions
- ▶ Total: $(N/2)(\log_2 N - 1)$ multiplications and $N \log_2 N$ additions

Key Result: **A DFT of size N requires order $N \log_2 N$ operations!**

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Matrix factorization view of DFT, $N = 4$

- ▶ Separate even and odd samples
- ▶ Compute two DFT's of size 2 having output $X'[k]$ and $X''[k]$
- ▶ Compute sum and difference of $X'[k]$ and $W^k X''[k]$

$$\mathbf{W}_4 = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} = \begin{bmatrix} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & -j \\ 1 & 0 & -1 & 0 \\ 0 & 1 & 0 & j \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 & 0 \\ 1 & -1 & 0 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & -1 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

This uses 8 additions and no multiplications!

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Matrix factorization view of DFT, $N = 8$, $1/8$

Now this is going to be big...

Too big for a single slide!

$$\mathbf{W}_8 = \begin{bmatrix} 1 & 1 & 1 & 1 & \dots & 1 \\ 1 & W^1 & W^2 & W^3 & \dots & W^7 \\ 1 & W^2 & W^4 & W^6 & \dots & W^{14} \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W^7 & W^{14} & W^{21} & \dots & W^{49} \end{bmatrix} = \dots$$

Matrix factorization view of DFT, $N = 8$, 2/8

Step 1: separate even from odd indexed samples

Call this \mathbf{D}_8 for decimation of size 8

$$\mathbf{D}_8 = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}$$

This requires no arithmetic operations!

Matrix factorization view of DFT, $N = 8$, 3/8

Step 2: Compute two DFTs of size $N/2$ on the even and on the odd indexed samples
Each submatrix is \mathbf{W}_4 , and the matrix is block diagonal, where $\mathbf{0}_4$ stands for a matrix of 0's

$$\begin{bmatrix} \mathbf{W}_4 & \mathbf{0}_4 \\ \mathbf{0}_4 & \mathbf{W}_4 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 & & & & \\ 1 & j & -1 & -j & & & & \\ 1 & -1 & 1 & -1 & & & & \\ 1 & -j & -1 & j & & & & \\ & & & & 1 & 1 & 1 & 1 \\ & & & & 1 & j & -1 & -j \\ & & & & 1 & -1 & 1 & -1 \\ & & & & 1 & -j & -1 & j \end{bmatrix}$$

This requires two DFT-4, or a total of 16 additions!

Matrix factorization view of DFT, $N = 8$, 4/8

Step 3: Multiply output of second DFT of size 4 by W^k

This is a diagonal matrix, with \mathbf{I}_4 for the identity of size 4,

$$\begin{bmatrix} \mathbf{I}_4 & \mathbf{0}_4 \\ \mathbf{0}_4 & \mathbf{\Lambda}_4 \end{bmatrix} = \begin{bmatrix} 1 & & & & & & & \\ & 1 & & & & & & \\ & & 1 & & & & & \\ & & & 1 & & & & \\ & & & & 1 & & & \\ & & & & & W & & \\ & & & & & & W^2 & \\ & & & & & & & W^3 \end{bmatrix} \text{ where } \mathbf{\Lambda}_4 = \begin{bmatrix} 1 & & & \\ & W & & \\ & & W^2 & \\ & & & W^3 \end{bmatrix}$$

This requires 2 multiplications ($W^2 = -j$ is free)

Matrix factorization view of DFT, $N = 8$, 5/8

Step 4: Recombine final output $X[k]$ and $X[k + N/2]$ by sum and difference, \mathbf{S}_8

$$\mathbf{S}_8 = \begin{bmatrix} \mathbf{I}_4 & \mathbf{I}_4 \\ \mathbf{I}_4 & -\mathbf{I}_4 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 & -1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & -1 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & -1 \end{bmatrix}$$

This requires 8 additions!

Matrix factorization view of DFT, $N = 8$, 6/8

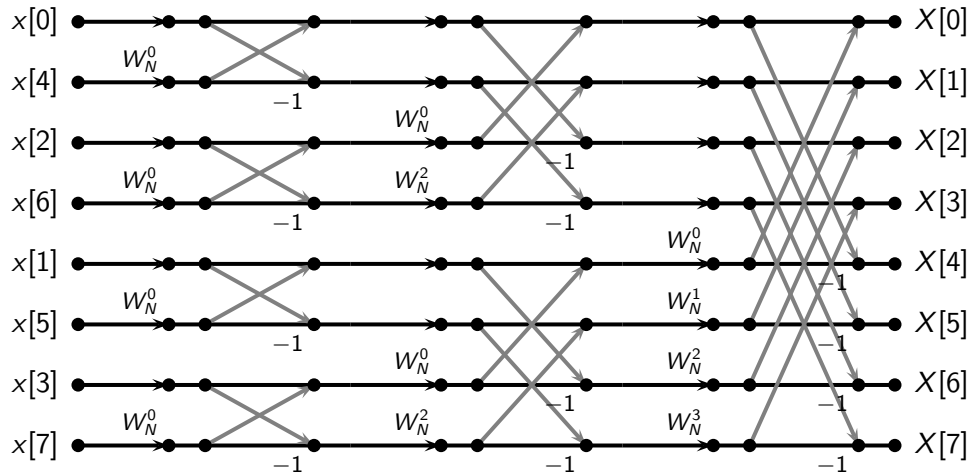
In total:

Product of 4 matrices

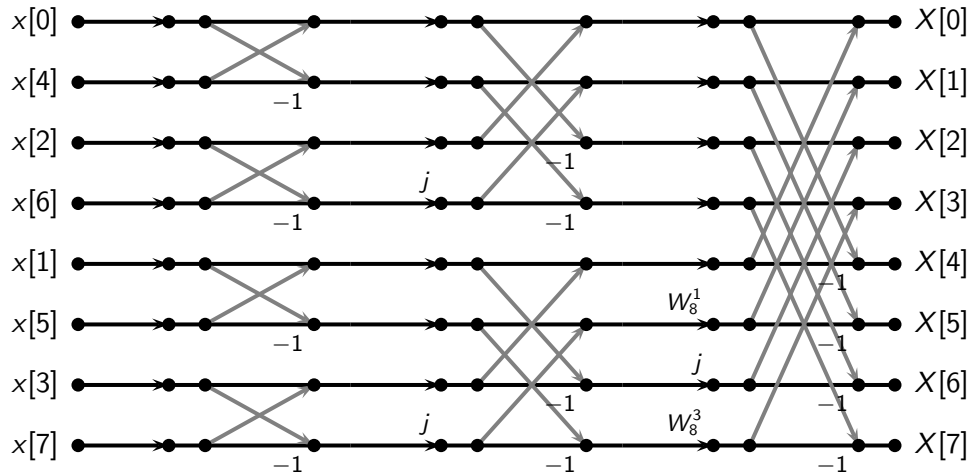
$$\mathbf{W}_8 = \begin{bmatrix} \mathbf{I}_4 & \mathbf{I}_4 \\ \mathbf{I}_4 & -\mathbf{I}_4 \end{bmatrix} \cdot \begin{bmatrix} \mathbf{I}_4 & \mathbf{0}_4 \\ \mathbf{0}_4 & \mathbf{A}_4 \end{bmatrix} \cdot \begin{bmatrix} \mathbf{W}_4 & \mathbf{0}_4 \\ \mathbf{0}_4 & \mathbf{W}_4 \end{bmatrix} \cdot \mathbf{D}_8$$

This requires 24 additions and 2 multiplications!

Flowgraph view of FFT, $N = 8$



Flowgraph view of FFT, $N = 8$



Matrix factorization view of DFT, $N = 8$, 8/8

Is this a big deal?

- ▶ In image processing (e.g. digital photography) one takes block of 8 by 8 pixels
- ▶ One computes a transform (called DCT) similar to a DFT
- ▶ It has a fast algorithm inspired by what we just saw

Matrix factorization view of DFT, $N = 8$, 8/8

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Matrix factorization view of DFT, $N = 8$, 8/8

Is this a big deal?

- ▶ In image processing (e.g. digital photography) one takes block of 8 by 8 pixels
- ▶ One computes a transform (called DCT) similar to a DFT
- ▶ It has a fast algorithm inspired by what we just saw

Some examples

image processing (JPEG compression)

- ▶ image is divided into 8×8 -pixel blocks
- ▶ DFT performed on rows and columns: 16 8-point DFTs
- ▶ direct computation: $16 \times 8^2 = 1024$ multiplications
- ▶ FFT: $16 \times 2 = 32$ multiplications

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audio processing (MP3 compression)

- ▶ audio is split into 1152-point frames
- ▶ direct DFT computation: $1.3 \cdot 10^6$ multiplications
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Conclusions

Don't worry, be happy!

- ▶ The Cooley-Tukey is the most popular algorithm, mostly for $N = 2^N$
- ▶ Note that there is always a good FFT algorithm around the corner
- ▶ Do not zero-pad to lengthen a vector to have a size equal to a power of 2
- ▶ There are good packages out there (e.g. Fastest Fourier Transform in the West, SPIRAL)
- ▶ It does make a BIG difference!



FFTW