Lecture 25: Filter Bank Reconstruction

Foundations of Digital Signal Processing

Outline

- Homework Questions / Discussions
- Review Short Time Fourier Transform
- Inefficient DFT Filter Banks
- Efficient DFT Filter Bank
- General Two-Channel Filter Bank Reconstruction

News

Schedule / Plan

- Tomorrow:, Nov. 16 Homework #10
- Monday, Nov. 19: Coding Assignment #6
- Next Week: No Due Dates
- Thursday, Nov. 29th: Homework #11
- Tuesday, Dec. 4th: Exam #3
- Wednesday, Dec. 5th: Coding Assignment #7 (short)
- Wednesday, Dec. 12th: Final Exam
- Friday, Dec. 14th: EEE5502 Reports Due

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Filter Design

Impulse Invariant Method

$$\frac{A}{s-p} \leftrightarrow \frac{A}{1 - e^{pT} z^{-1}}$$

What is the time-domain representation?

Filter Design

Impulse Invariant Method

$$\frac{A}{s-p} \leftrightarrow \frac{A}{1-e^{pT}z^{-1}}$$

What is the time-domain representation?

$$\frac{A}{s-p} \leftrightarrow Ae^{pt}u(t)$$

$$\frac{A}{1 - (e^{pT})z^{-1}} \leftrightarrow Ae^{pTn}u[n]$$

Same impulse response!

Lecture 25: Filter Bank Reconstruction

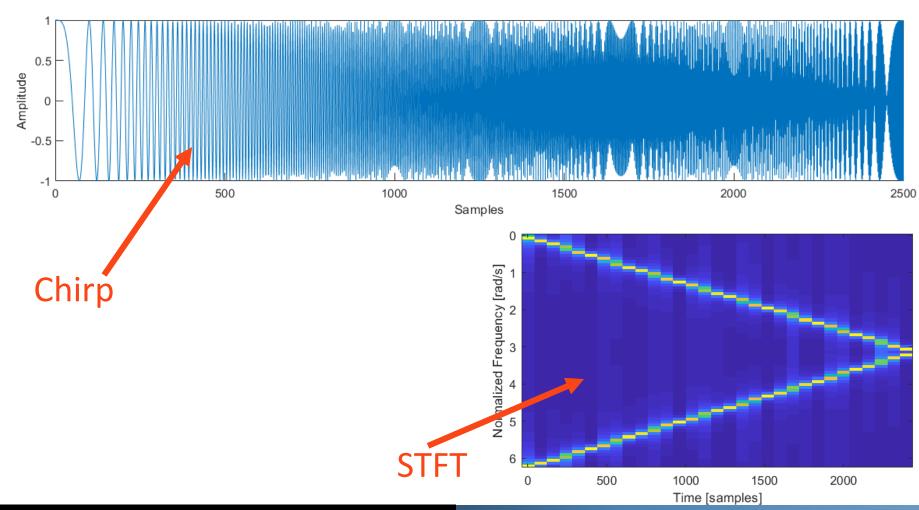
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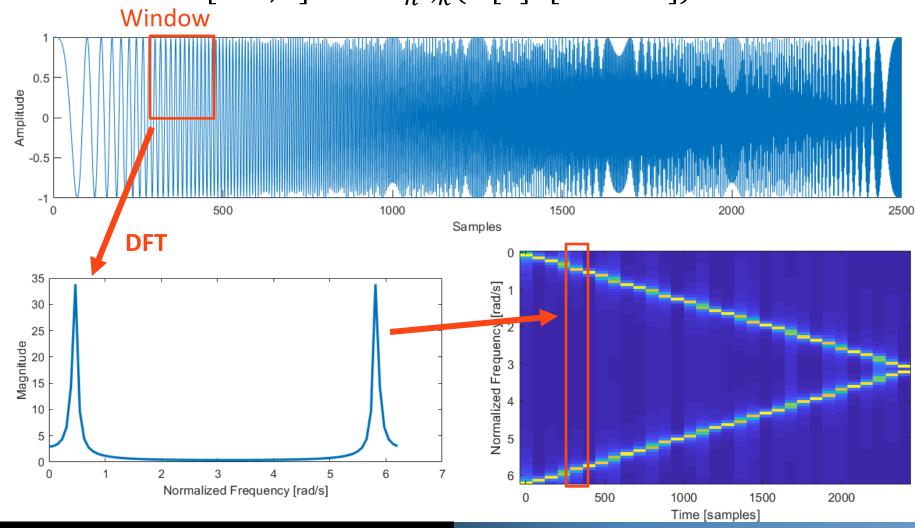
The Definition:

$$X[Mm, k] = DFT_{n \to k}(w[n]x[n - Mm])$$



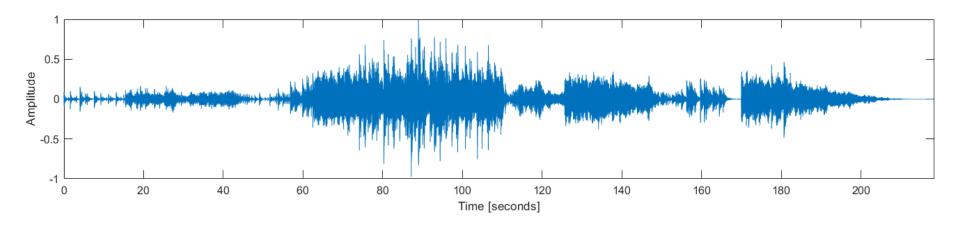
■ The Definition:

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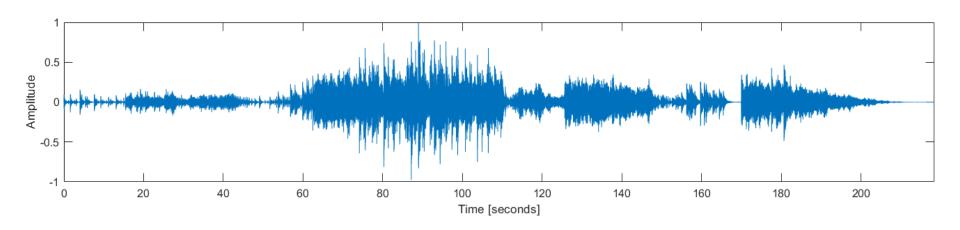
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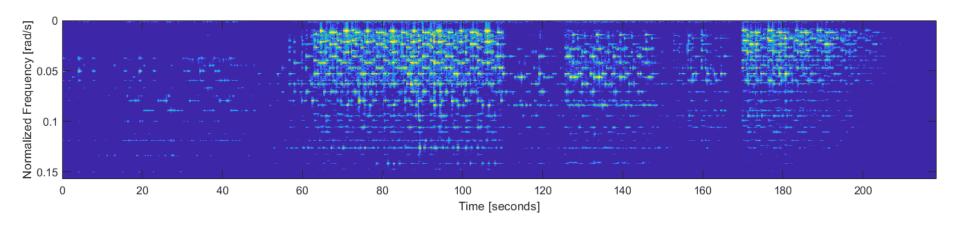


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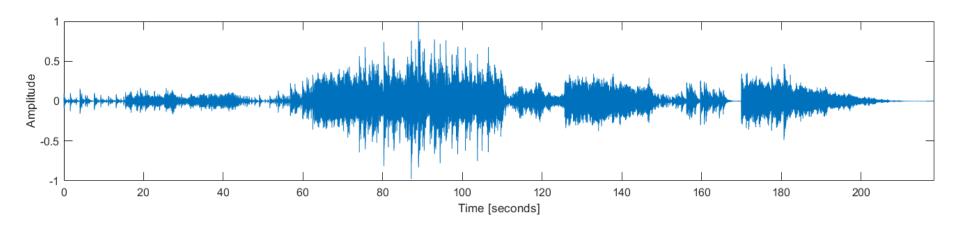
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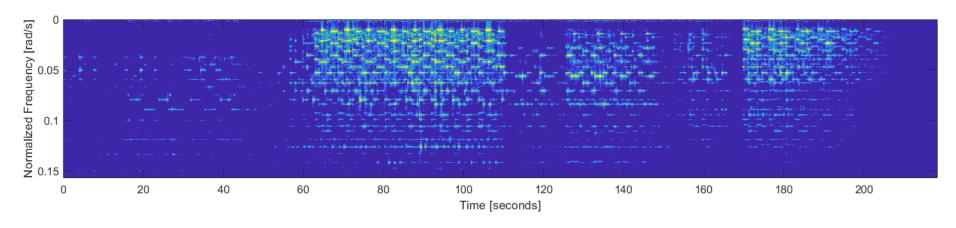




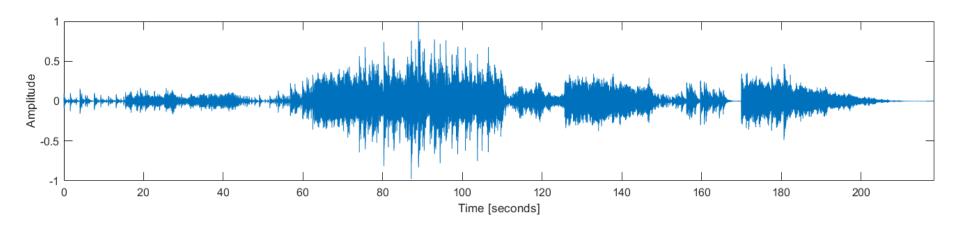


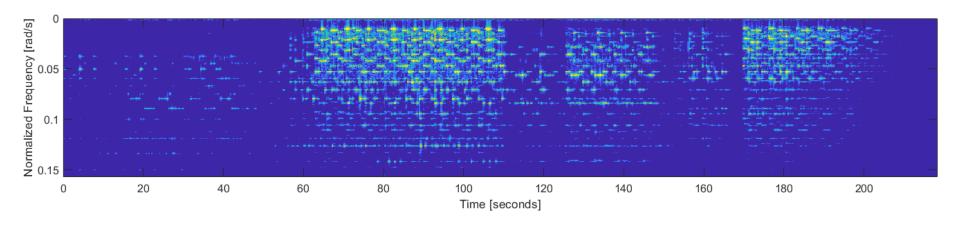
Question: Why do I care about the STFT?



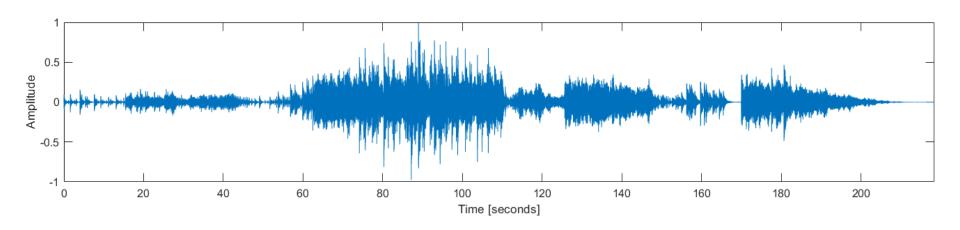


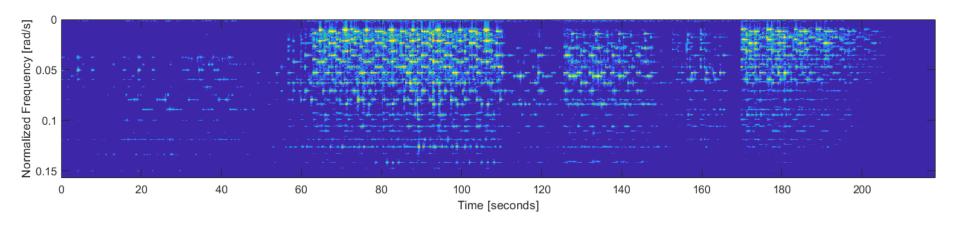
Question: How is this problematic for a real-time system?



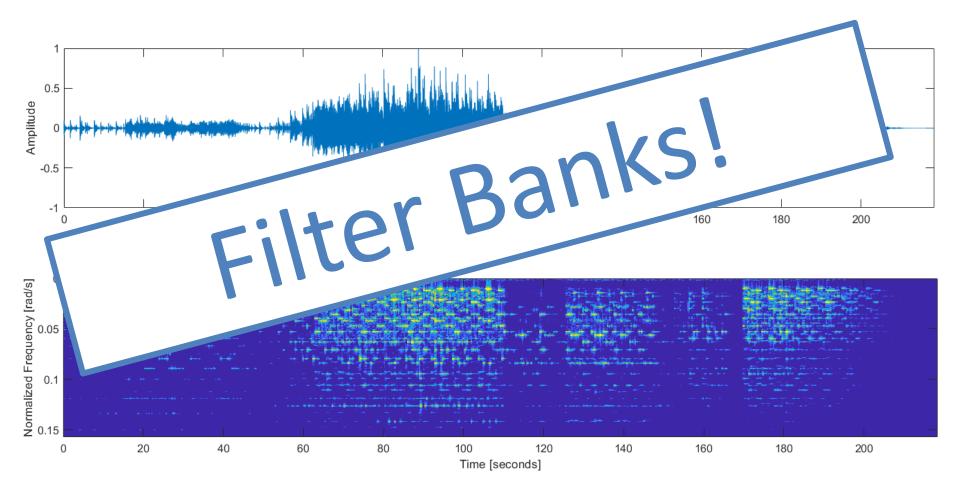


Question: How do I solve this problem??





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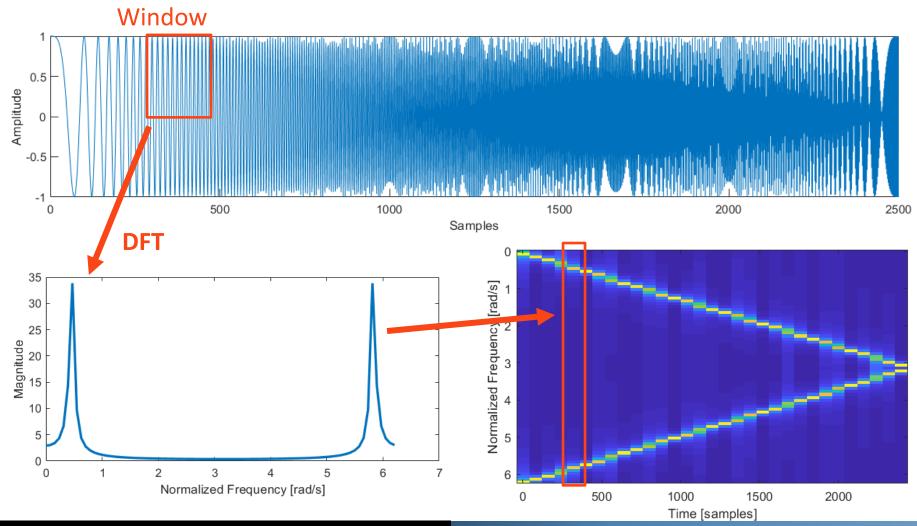
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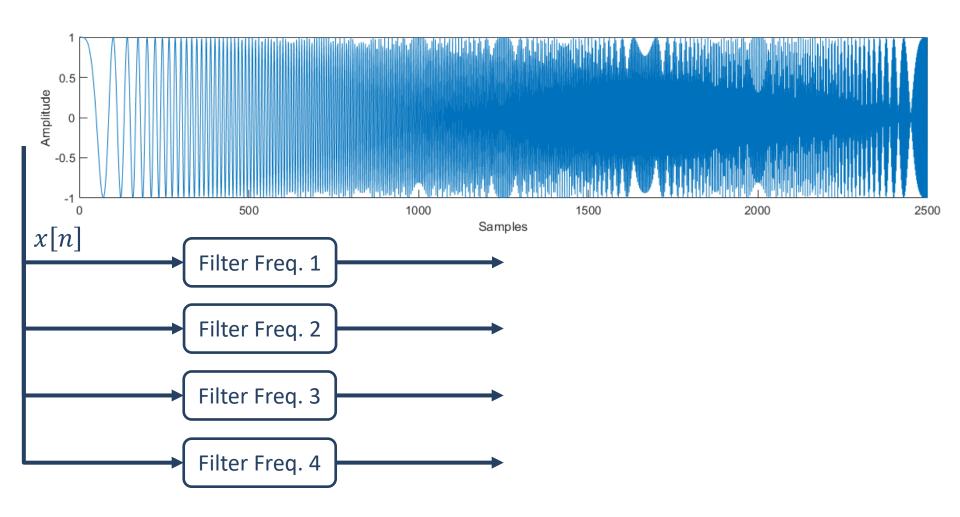
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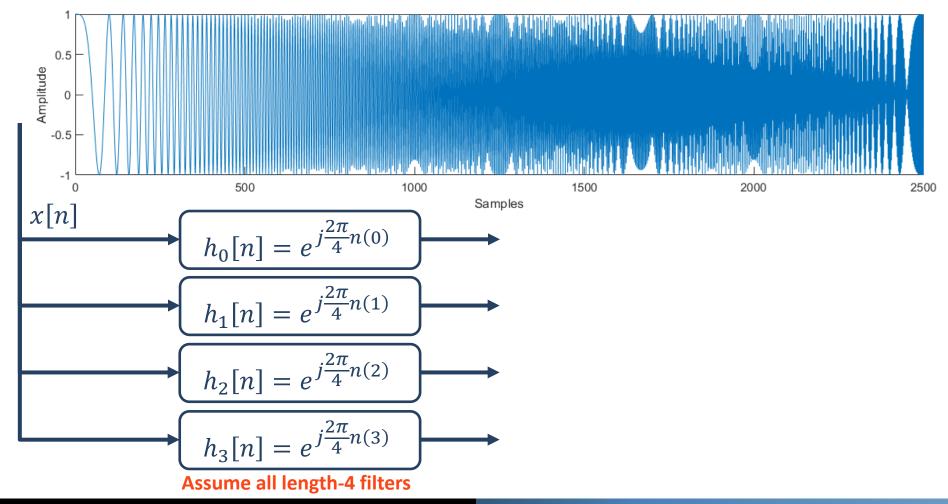
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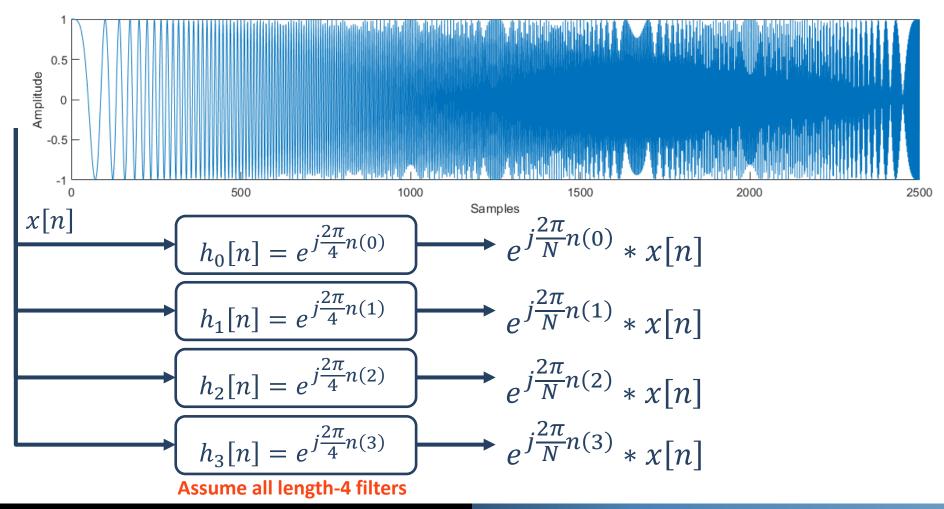
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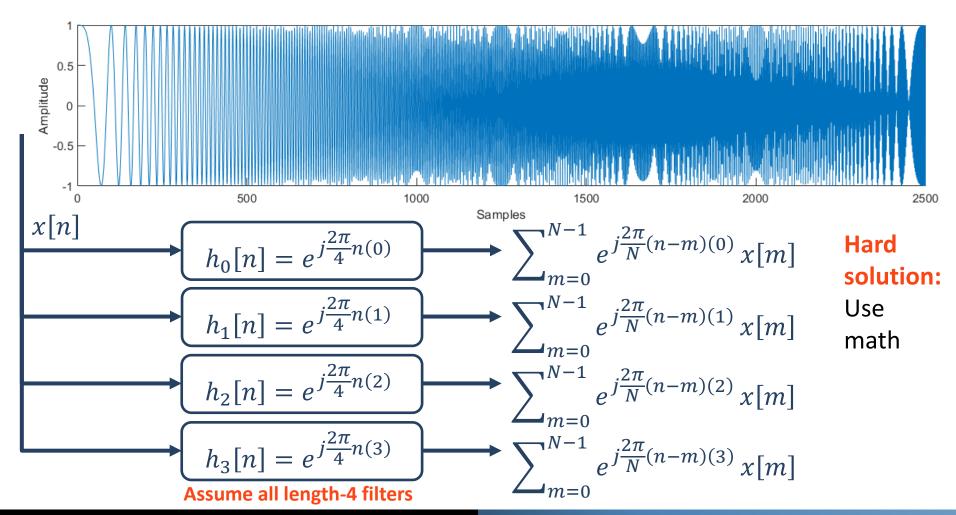
■ The Short Time Fourier Transform Process

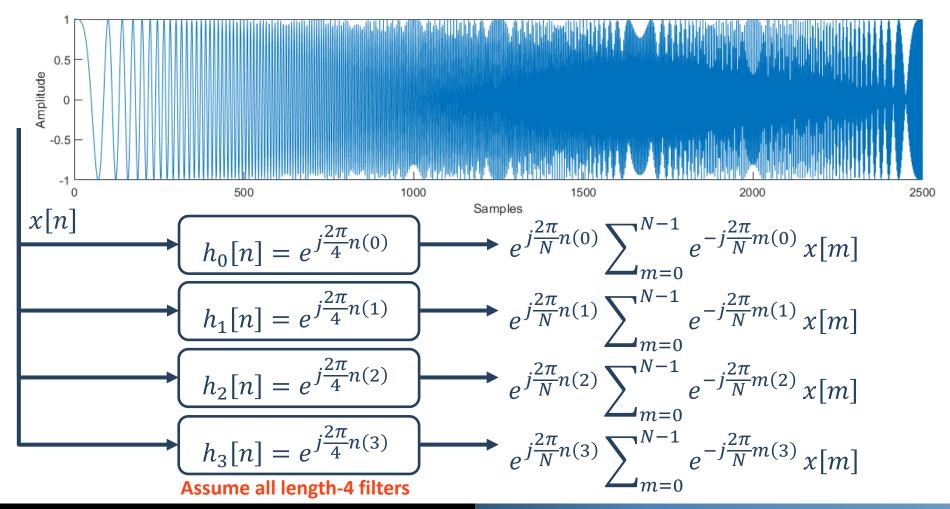


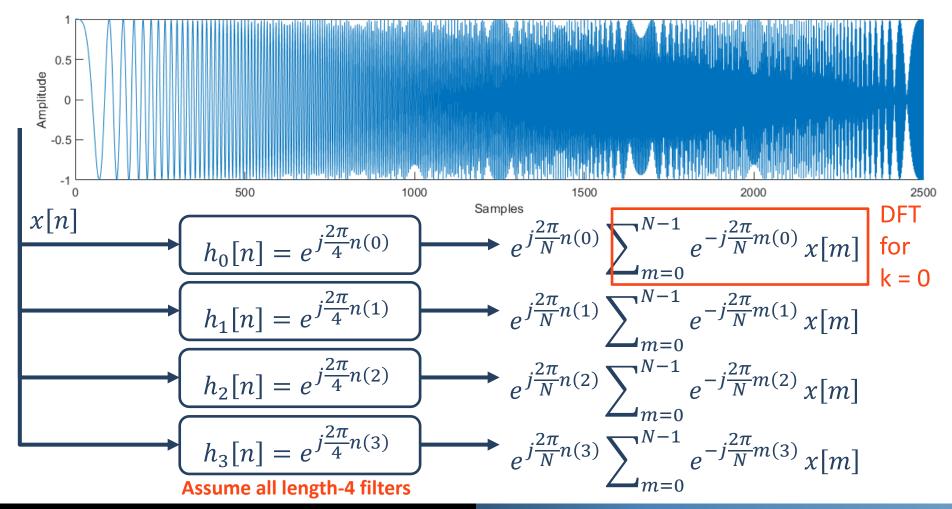


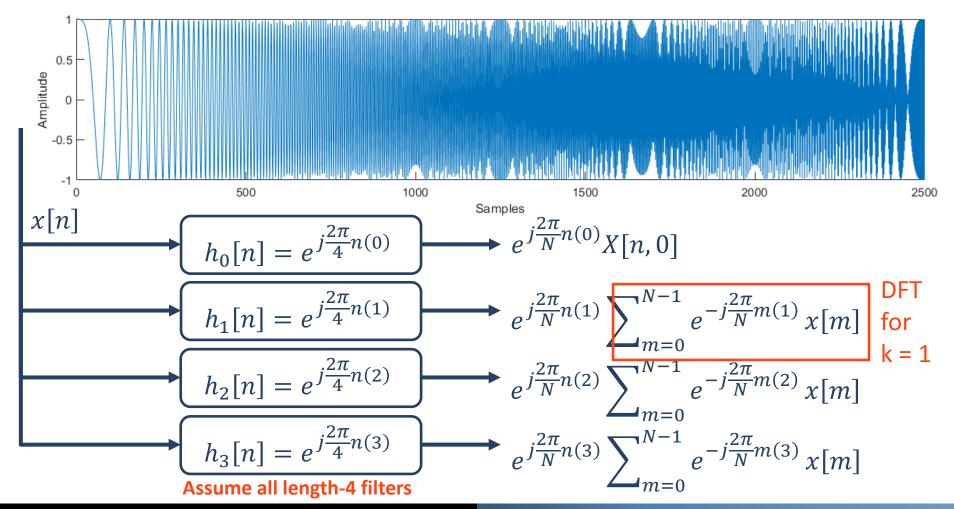


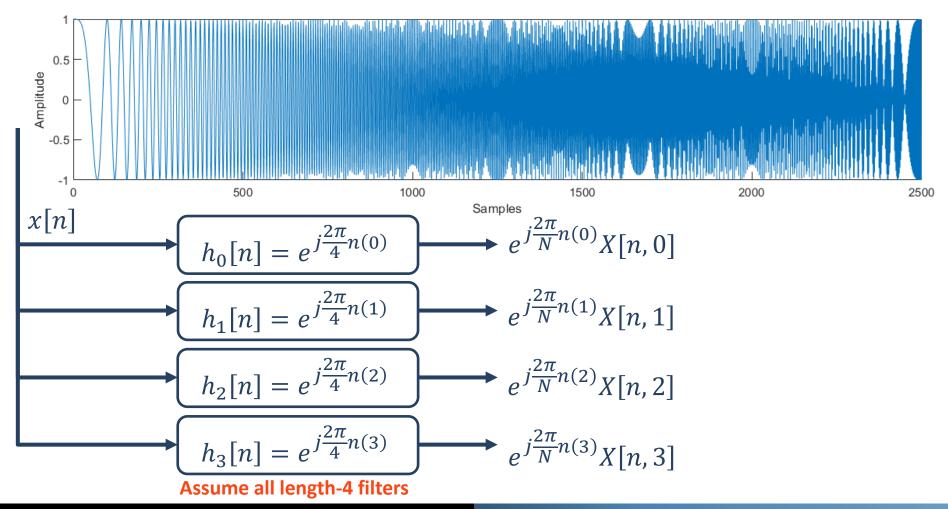


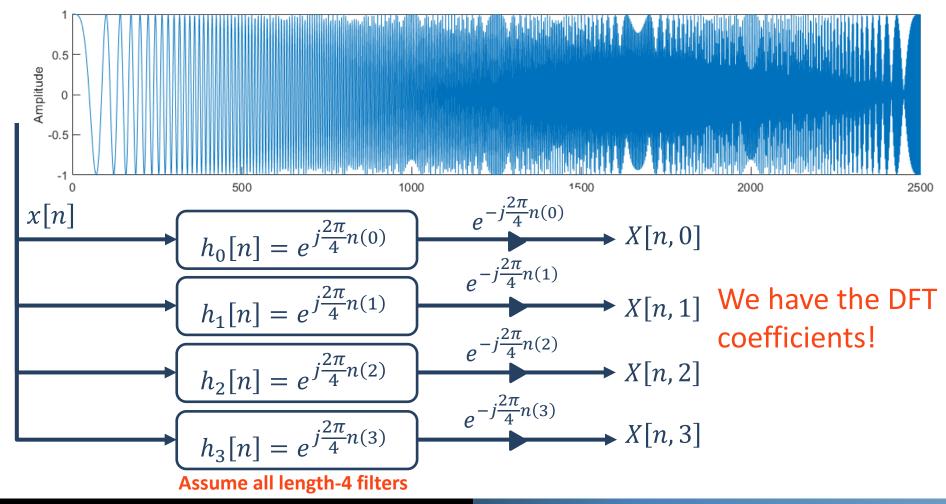


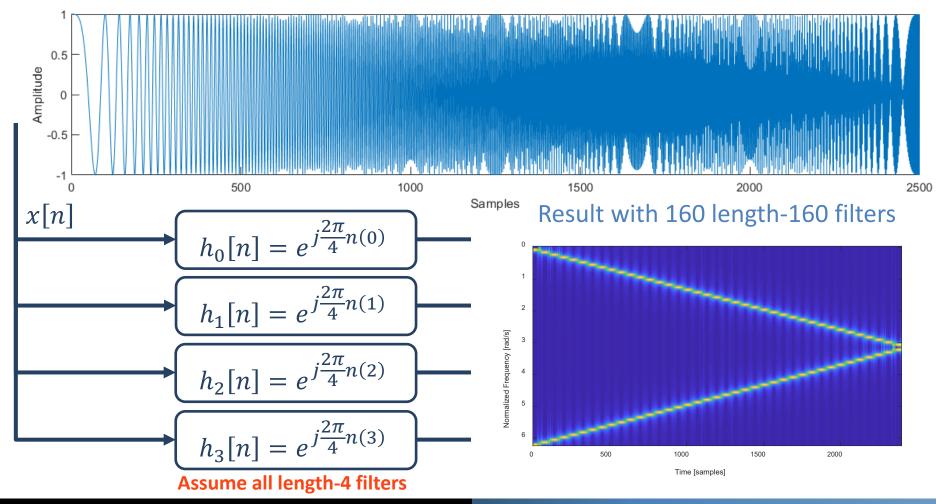




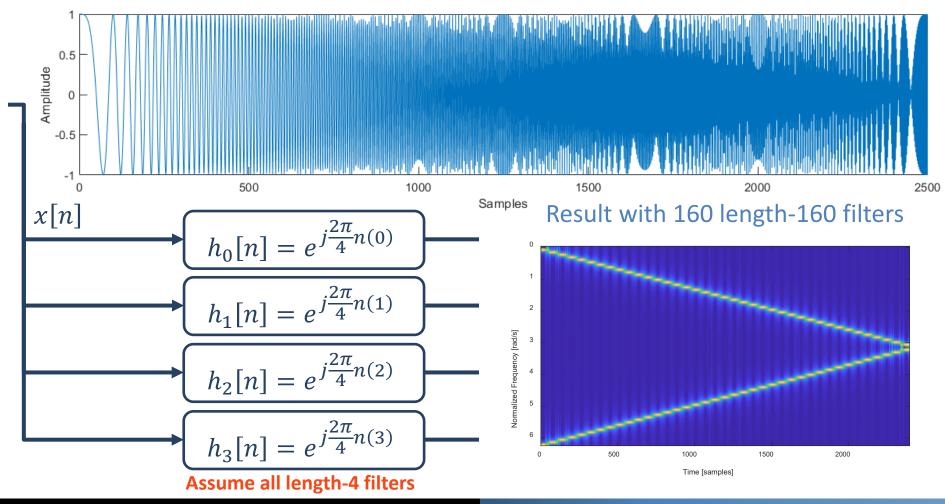








Question: Why is this not a preferred approach?



Question: Why is this <u>not</u> a preferred approach?

It is really expensive!

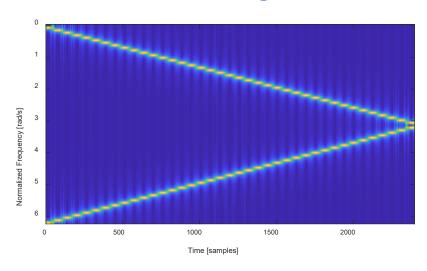
STFT Approach

- W² multiplications for every W samples
- $W^2 = 160^2 = 25,600$

Filter Bank Approach

- W³ multiplications for every W samples
- $W^3 = 160^3 = 4,086,000$

Result with 160 length-160 filters



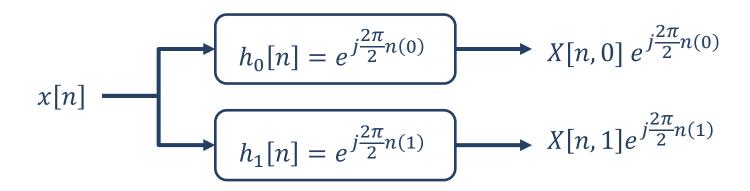
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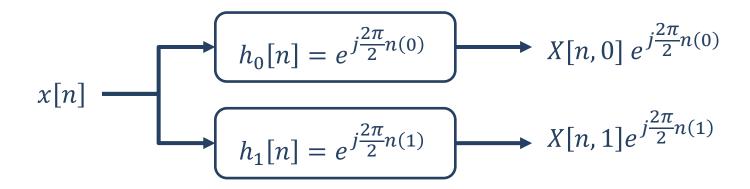
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Consider the following filter bank



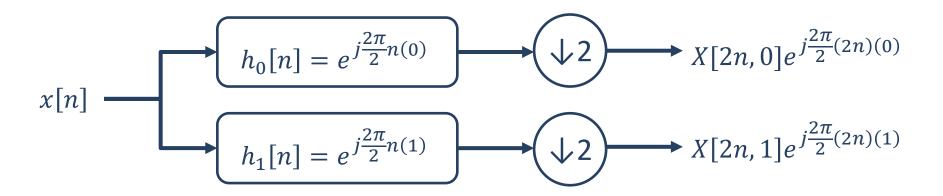
- Consider the following filter bank
 - Question: How do I make this like the STFT?????



- Now recall: The short-time Fourier Transform gave us
 - ♦ X[Mn, 0] <- M = shift amount (often window length)

Consider the following filter bank

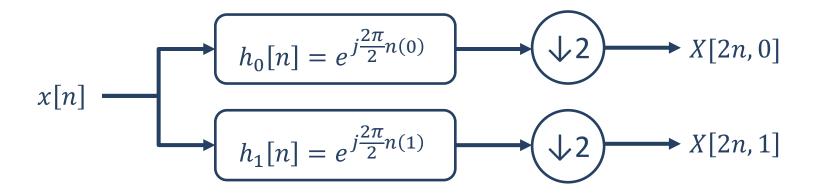
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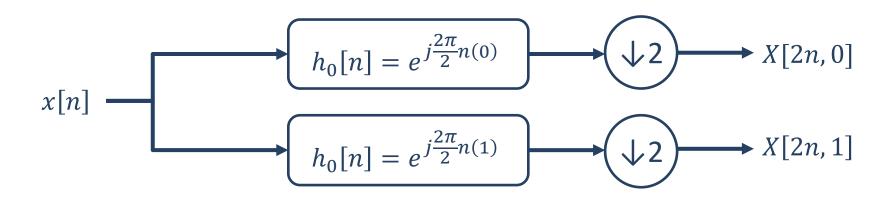
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Question: How do I make this like the STFT?????



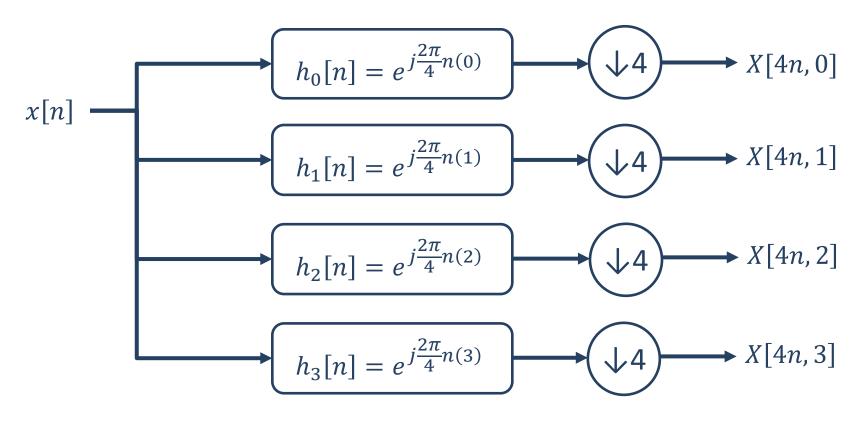
- Now recall: The short-time Fourier Transform gave us
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Consider the following filter bank

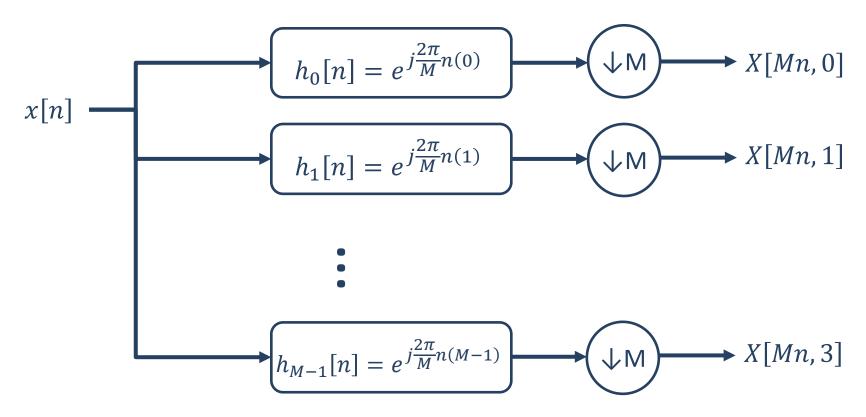


- Result: It is now exactly the same as the STFT with a window of length 2 and shift of 2 between windows
- But, we do not need to buffer x[n]

Consider the following filter bank

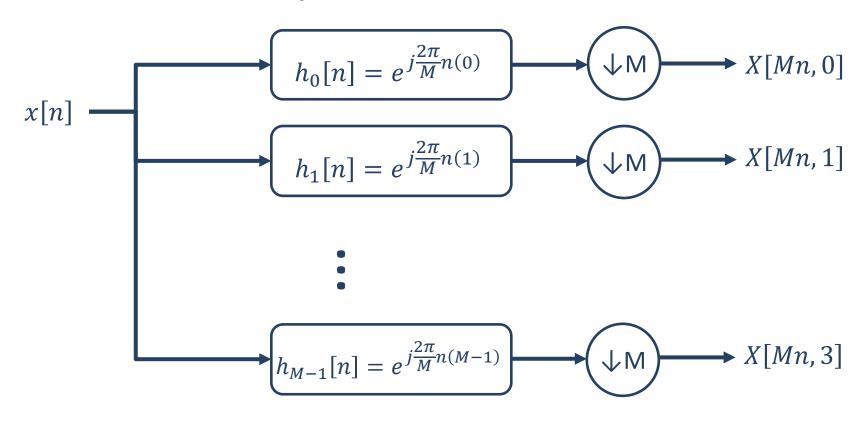


Hence, this is an M-point DFT



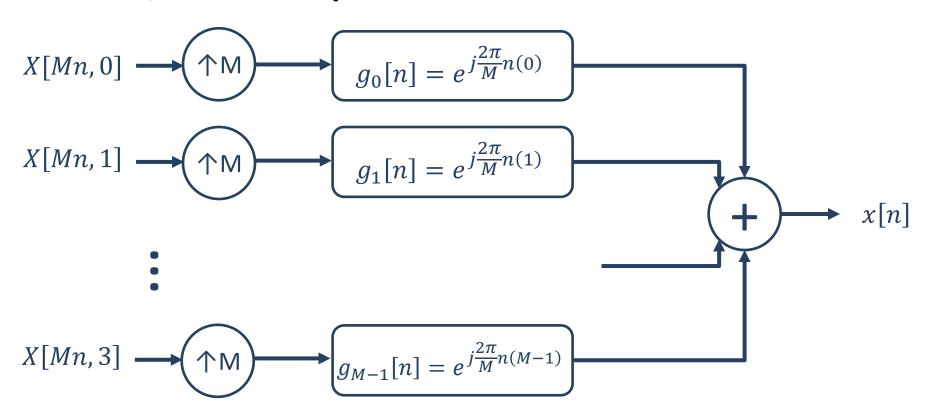
- So, I can implement the STFT as a filter bank...
 - Can I do more?

Hence, this is an M-point DFT

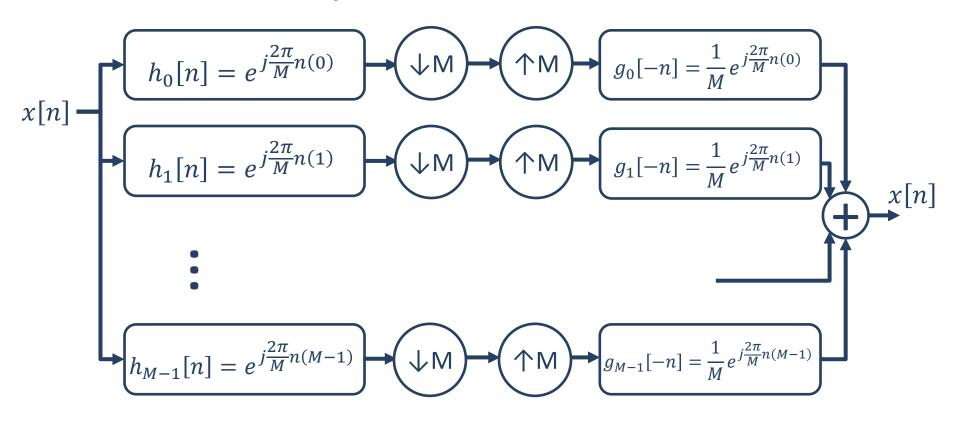


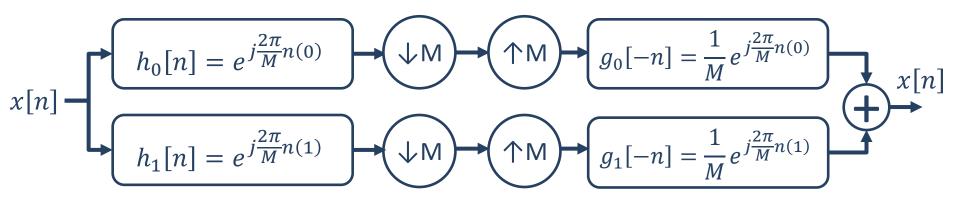
- So, I can implement the STFT as a filter bank...
 - Question: How do I get back into the time domain?

Hence, this is an M-point IDFT

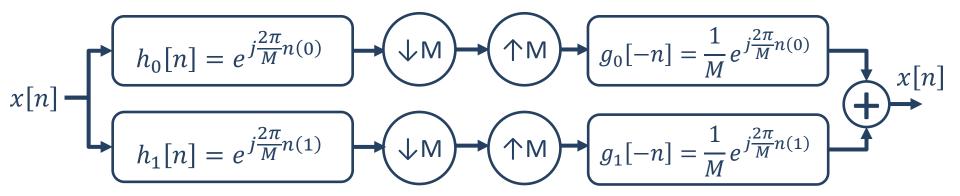


Hence, this is an M-point DFT and IDFT

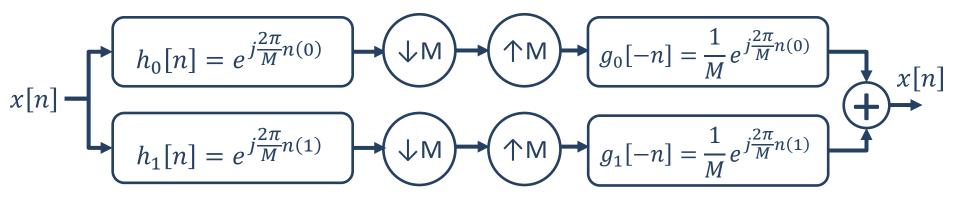




Example: The M=2 point DFT and IDFT



• What are the filter coefficients for $h_0[n]$ and $h_1[n]$?



$$h_0[n] = e^{j\frac{2\pi}{2}n(0)} = 1$$

for
$$0 \le n \le 1$$

$$h_1[n] = e^{j\frac{2\pi}{2}n(1)} = e^{\pi n}$$

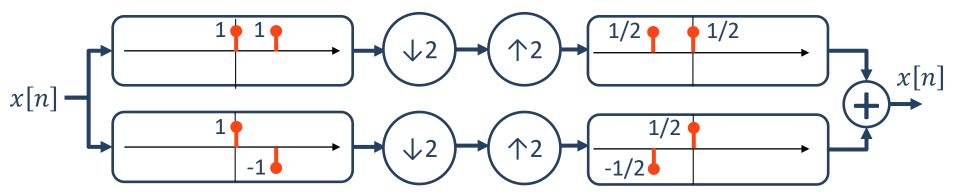
for
$$0 \le n \le 1$$

$$g_0[n] = \frac{1}{M}e^{-j\frac{2\pi}{2}n(0)} = \frac{1}{M}$$

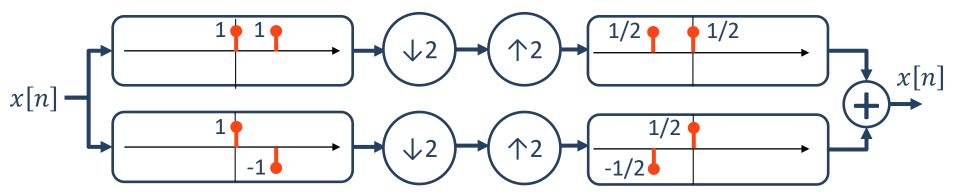
for
$$-1 \le n \le 0$$

$$g_1[n] = \frac{1}{M}e^{-j\frac{2\pi}{2}n(1)} = \frac{1}{M}e^{\pi n}$$

for
$$-1 \le n \le 0$$



- $h_0[n] = \delta[n] + \delta[n-1]$
- $h_1[n] = \delta[n] \delta[n-1]$
- $g_0[n] = \frac{1}{2} [\delta[n+1] + \delta[n]]$
- $g_1[n] = \frac{1}{2} \left[-\delta[n+1] + \delta[n] \right]$

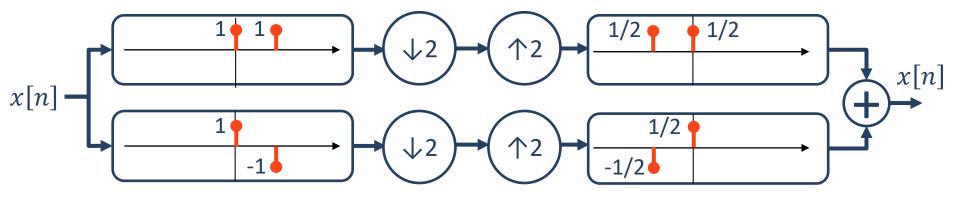


•
$$h_0[n] = \delta[n] + \delta[n-1]$$

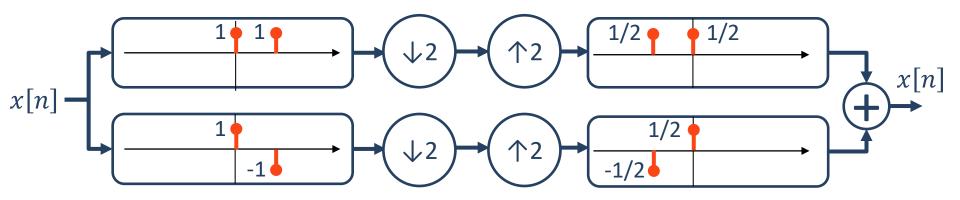
$$H(\omega) = e^{-j\frac{2\pi}{2}\omega(0)} + e^{-j\frac{2\pi}{2}\omega(1)}$$

•
$$h_1[n] = \delta[n] - \delta[n-1]$$

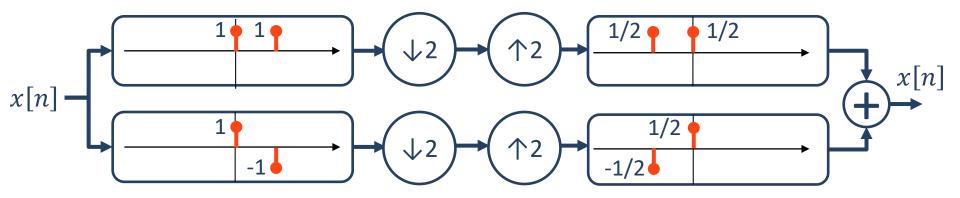
$$H(\omega) = e^{-j\frac{2\pi}{2}\omega(0)} - e^{-j\frac{2\pi}{2}\omega(1)}$$



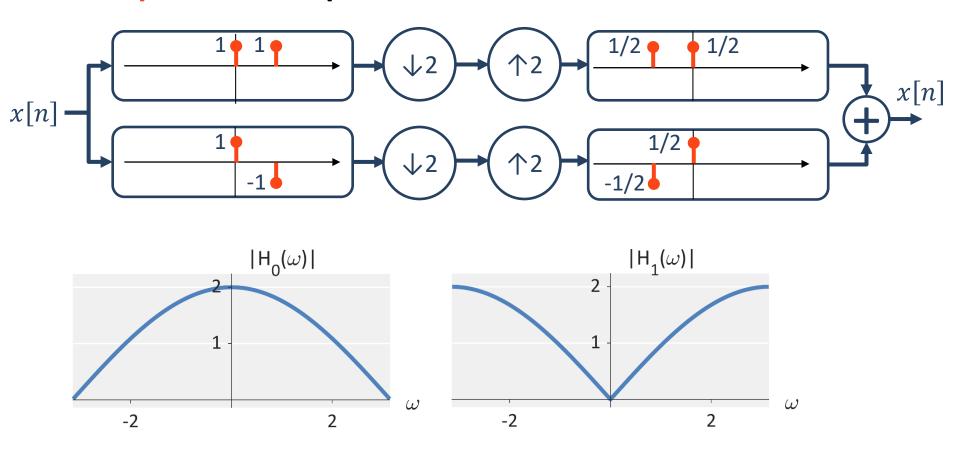
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $H_0(\omega) = 1 + e^{-j\pi\omega} = 2\cos(\omega/2) e^{-j\frac{\pi}{2}\omega}$
- $b_1[n] = \delta[n] \delta[n-1]$
 - $H_1(\omega) = 1 e^{-j\pi\omega} = 2j\sin(\omega/2)e^{-j\frac{\pi}{2}\omega}$



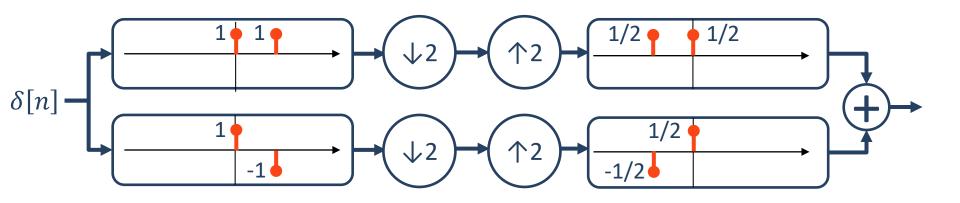
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $|H_0(\omega)| = 2|\cos(\omega/2)|$
- $h_1[n] = \delta[n] \delta[n-1]$
 - $|H_1(\omega)| = 2|\sin(\omega/2)|$



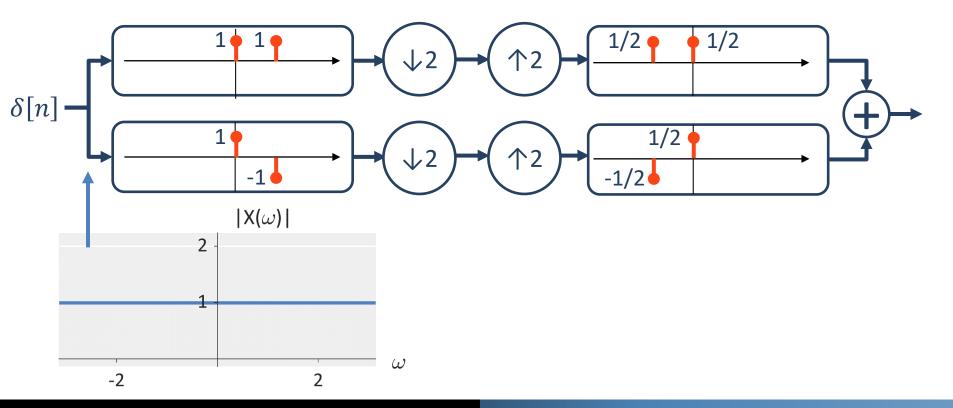
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $|H_0(\omega)| = 2|\cos(\omega/2)|$
 - $|G_0(\omega)| = |\cos(\omega/2)|$
- $h_1[n] = \delta[n] \delta[n-1]$
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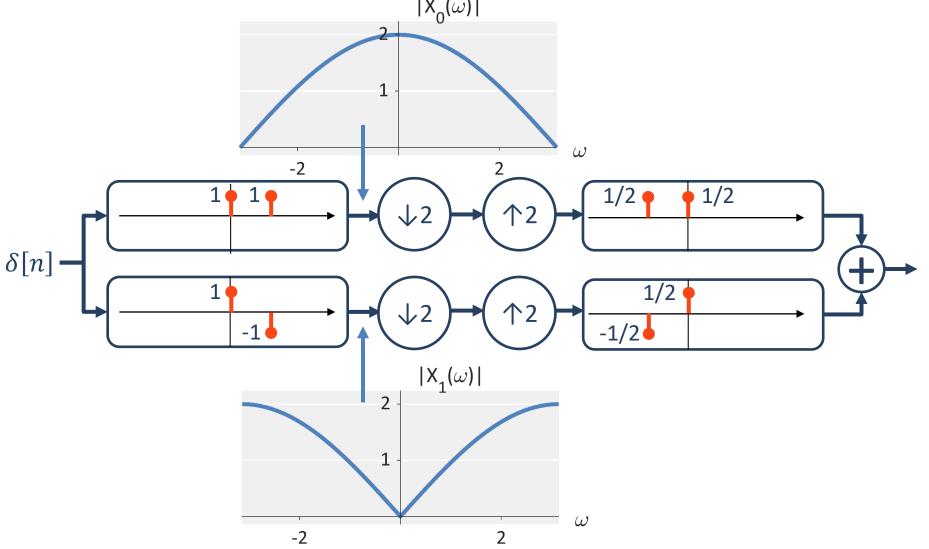
Example: Plot of the intermediate frequency magnitudes



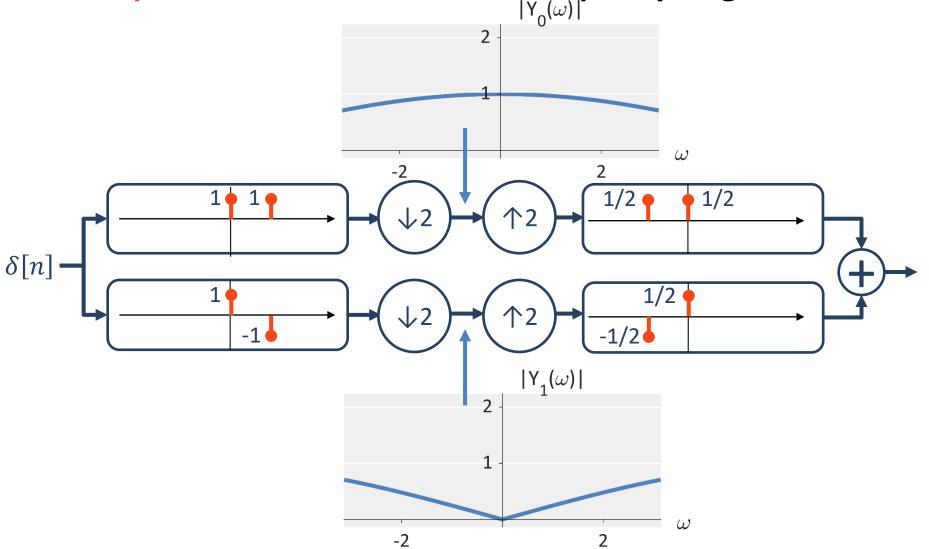
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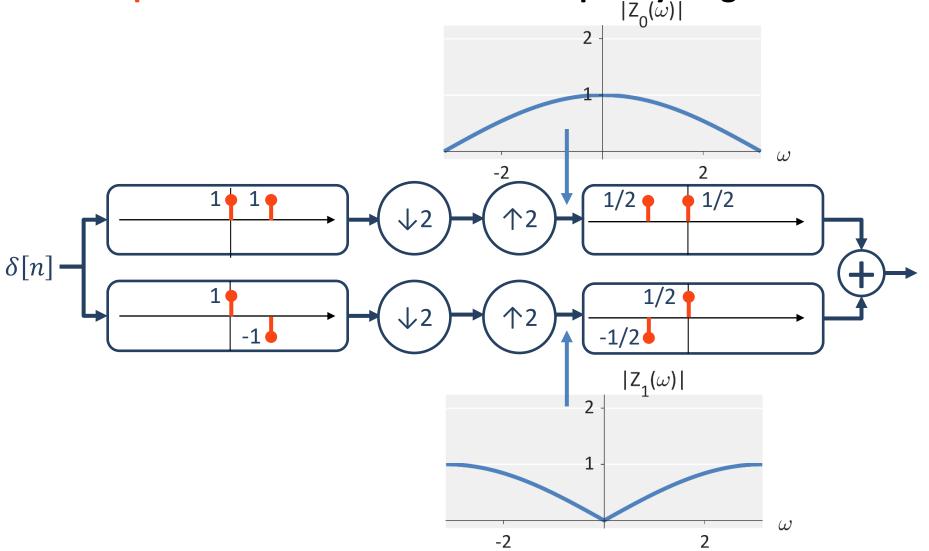
Example: Plot of the intermediate frequency magnitudes $|X_0(\omega)|$



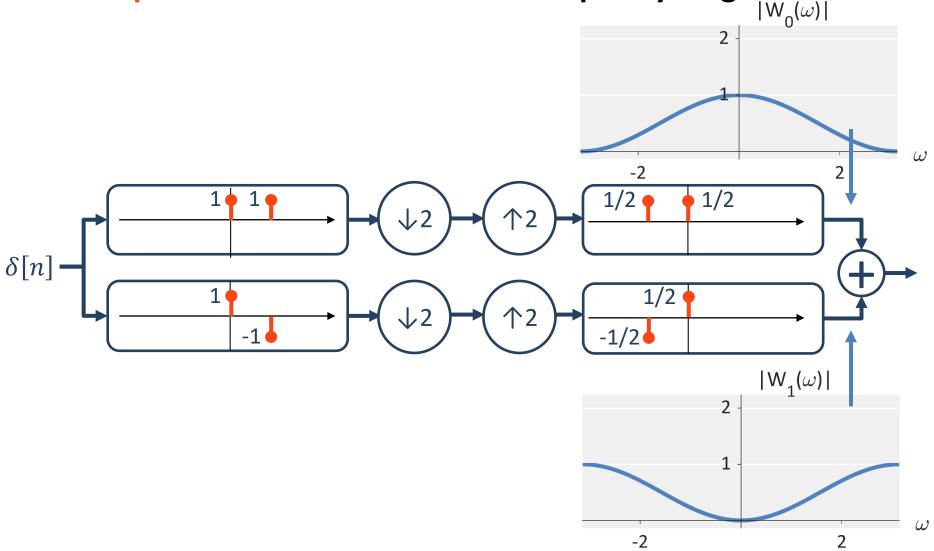
Example: Plot of the intermediate frequency magnitudes $|Y_0(\omega)|$



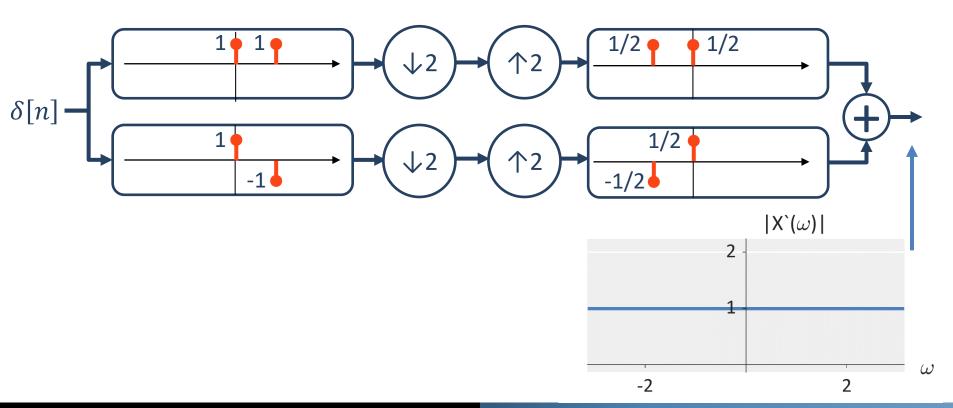
Example: Plot of the intermediate frequency magnitudes $|Z_0(\omega)|$



Example: Plot of the intermediate frequency magnitudes $|W_0(\omega)|$



Example: Plot of the intermediate frequency magnitudes



Question: Can we generalize this?

