Lecture 26: Filter Banks to Wavelets

Foundations of Digital Signal Processing

Outline

- DFT Filter Banks [without downsampling]
- DFT Filter Bank [with downsampling]
- General Two-Channel Filter Banks
- Two-Channel Filter Bank Perfect Reconstruction (aliasing canceling)
- Two-Channel Filter Bank Perfect Reconstruction (orthogonal filter bank)
- Multi-Channel Filter Bank Perfect Reconstruction
- Wavelets

News

Schedule / Plan

- <u>Tomorrow:, Nov. 16</u> Homework #10
- <u>Tuesday, Nov. 19:</u> Coding Assignment #6
- Next Week: No Due Dates (except Monday)
- Thursday, Nov. 29th: Homework #11
- Tuesday, Dec. 4th: Exam #3
- Wednesday, Dec. 5th: Coding Assignment #7 (short)
- Wednesday, Dec. 12th: Final Exam
- Friday, Dec. 14th: EEE5502 Reports Due

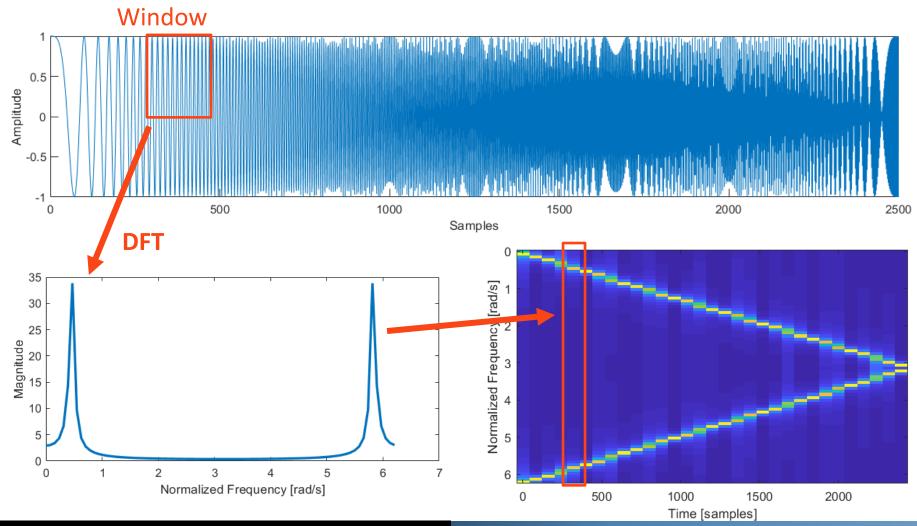
Lecture 26: Filter Banks to Wavelets

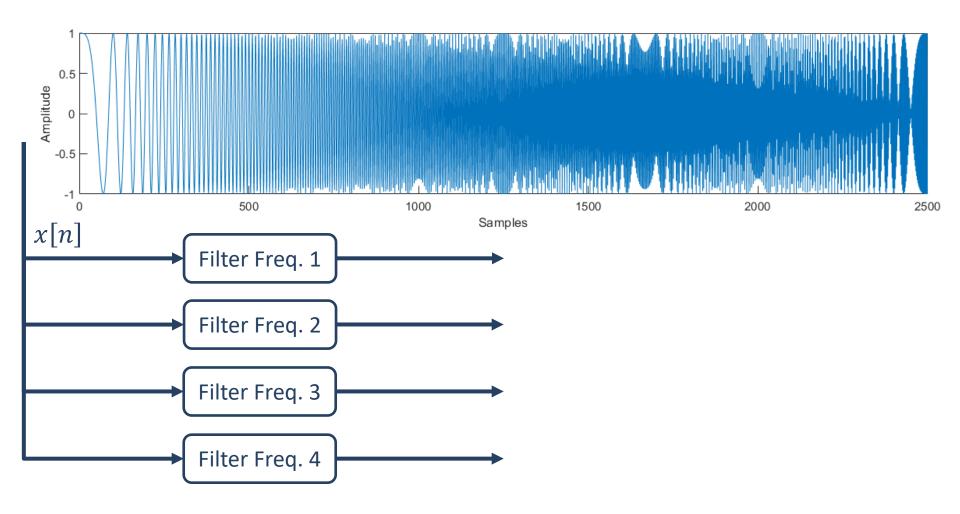
Foundations of Digital Signal Processing

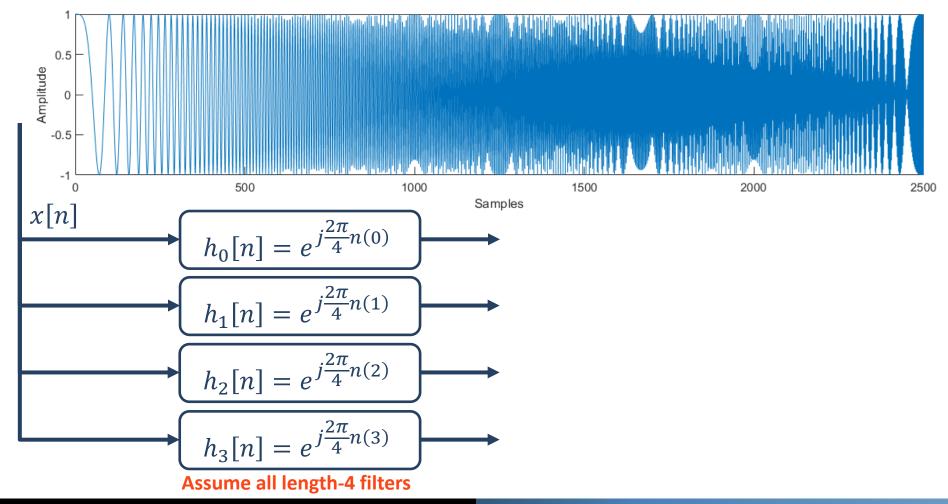
Outline

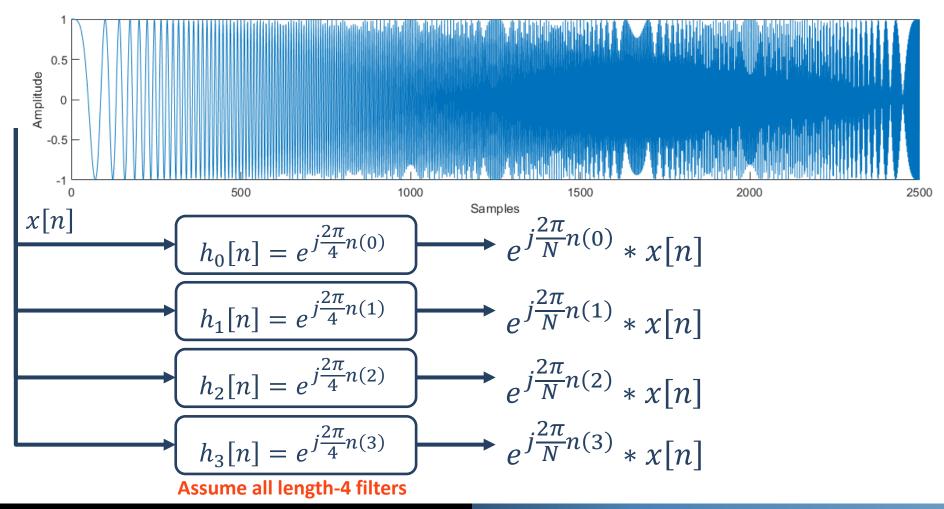
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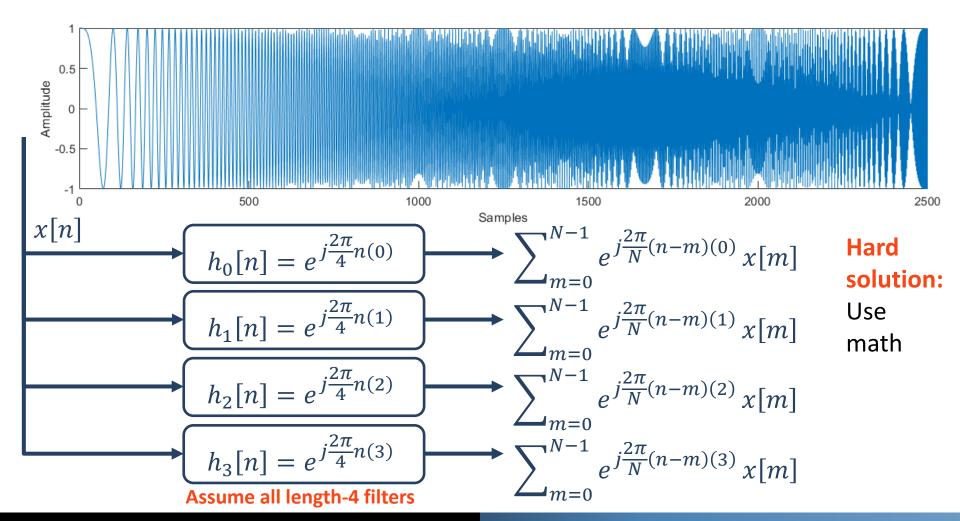
■ The Short Time Fourier Transform Process

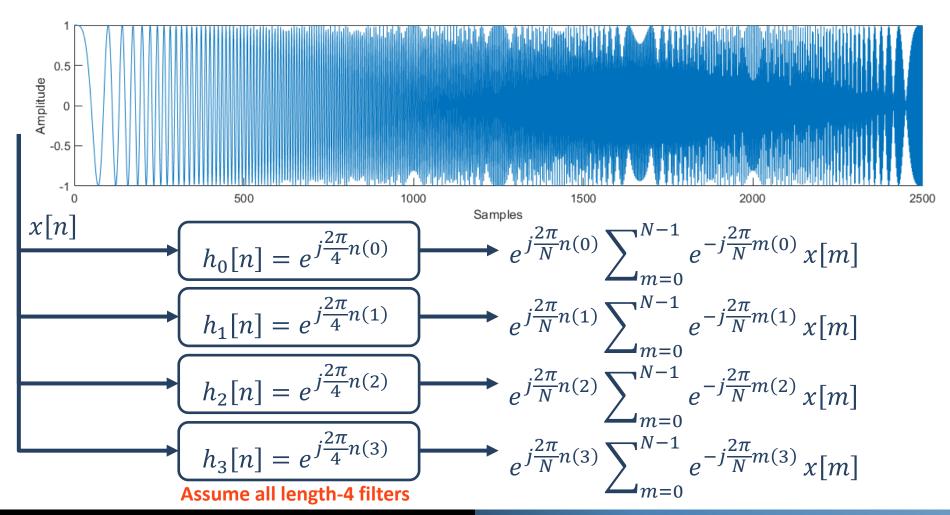


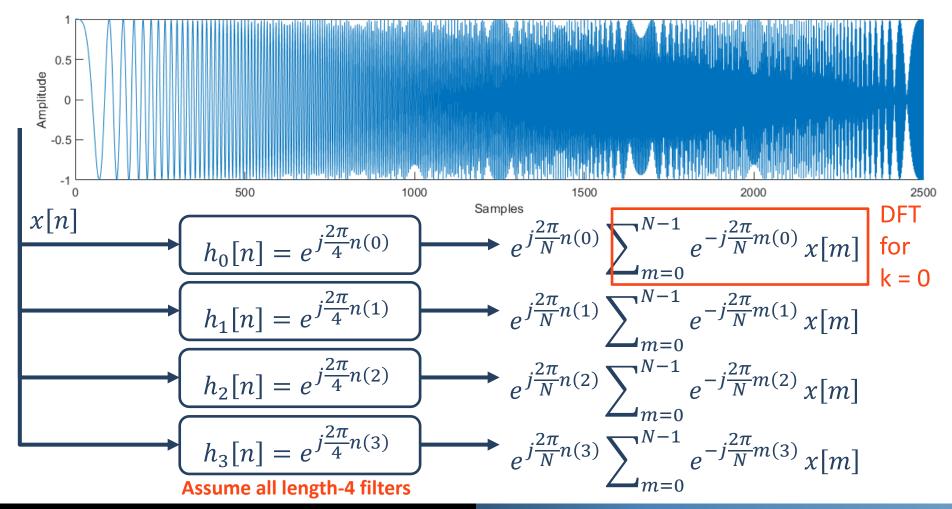


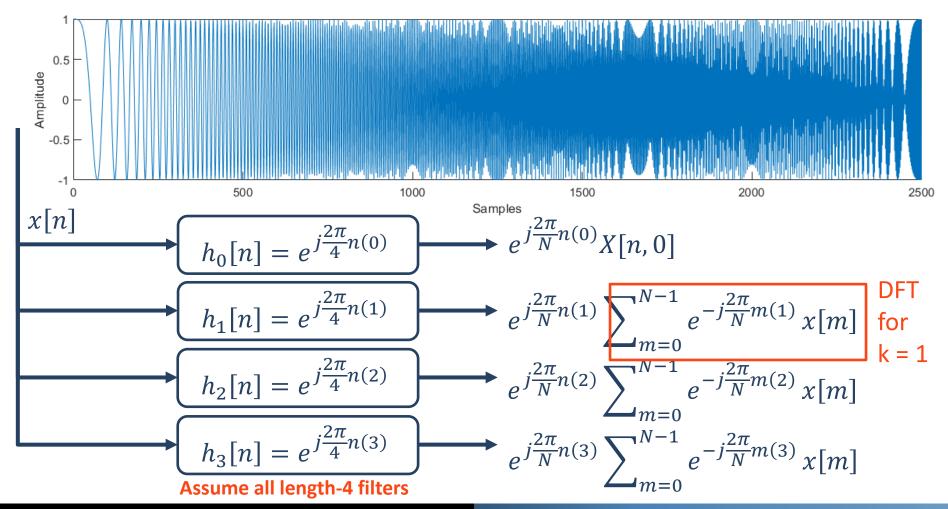


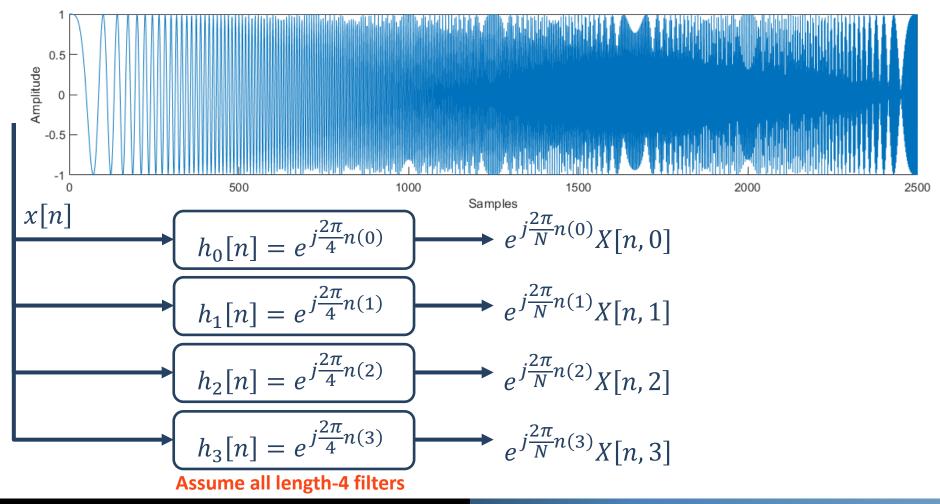


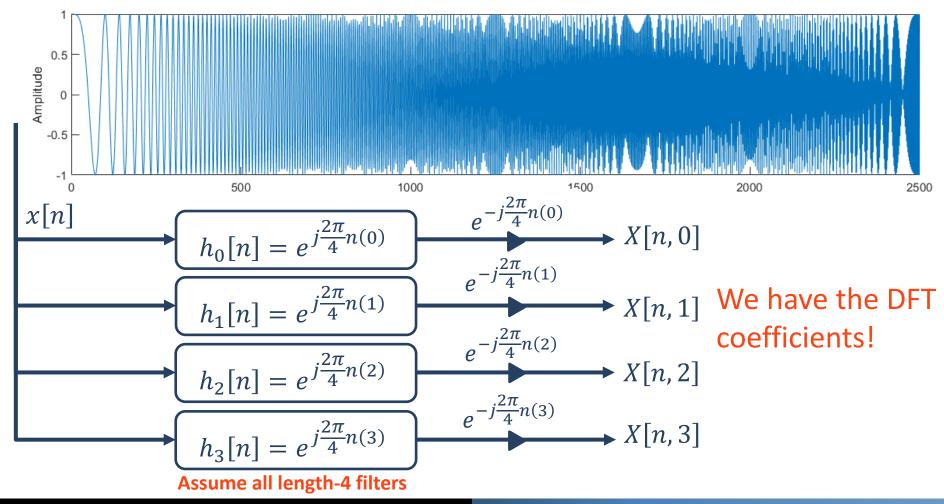


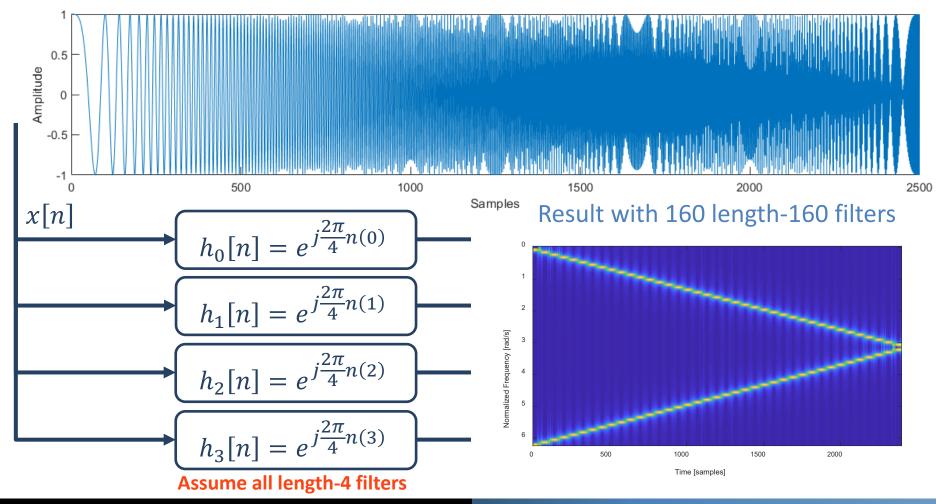




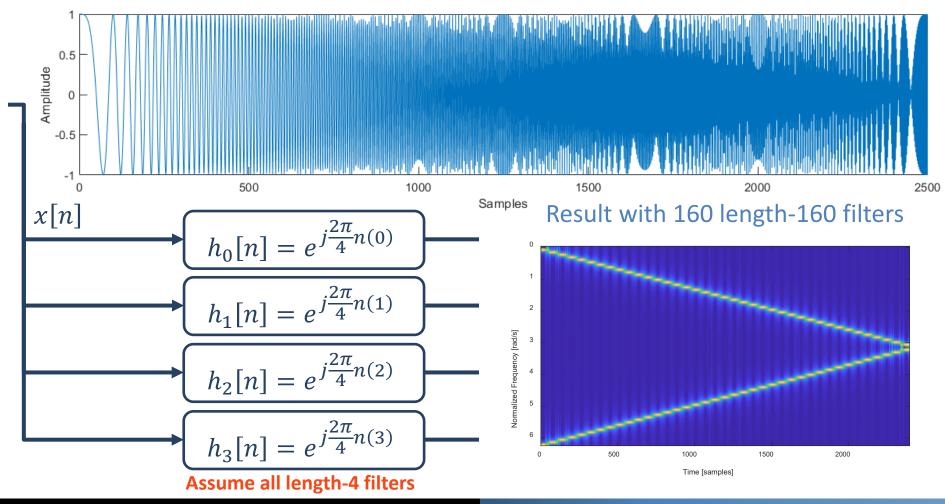




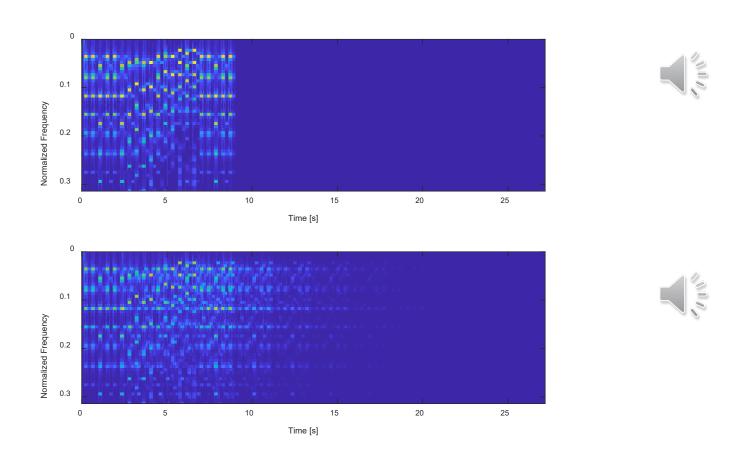




Question: Why is this <u>not</u> a preferred approach?



Example: Example from our coding assignment



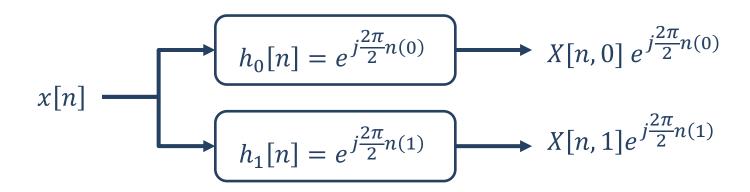
Lecture 26: Filter Banks to Wavelets

Foundations of Digital Signal Processing

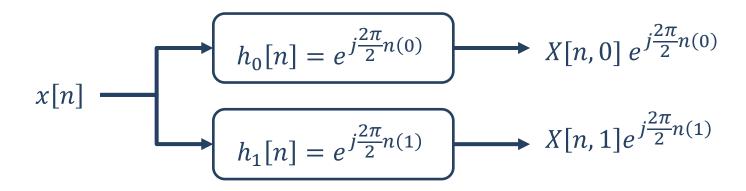
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- Multi-Channel Filter Bank Perfect Reconstruction
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Consider the following filter bank



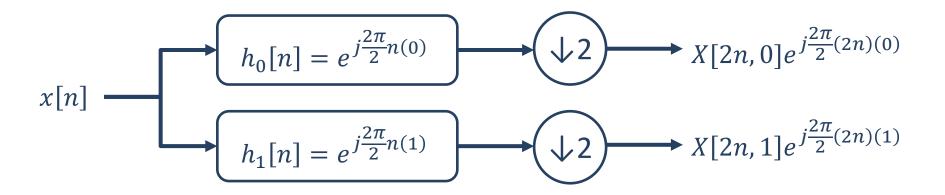
- Consider the following filter bank
 - Question: How do I make this like the STFT?????



- Now recall: The short-time Fourier Transform gave us
 - ♦ X[Mn, 0] <- M = shift amount (often window length)

Consider the following filter bank

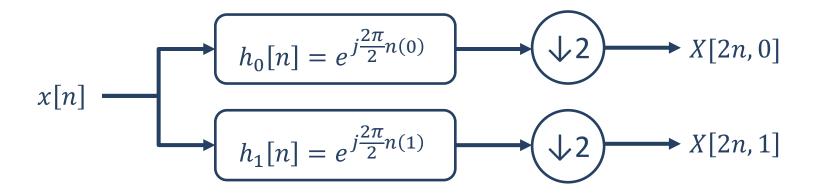
Question: How do I make this like the STFT?????



- Now recall: The short-time Fourier Transform gave us
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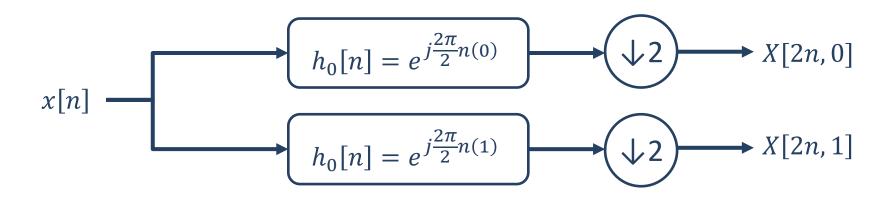
Consider the following filter bank

Question: How do I make this like the STFT?????



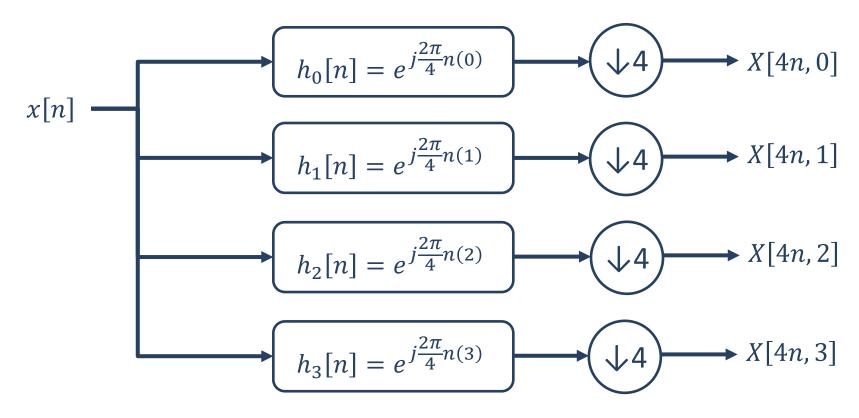
- Now recall: The short-time Fourier Transform gave us
 - ♦ X[Mn, 0] <- M = shift amount (often window length)

Consider the following filter bank

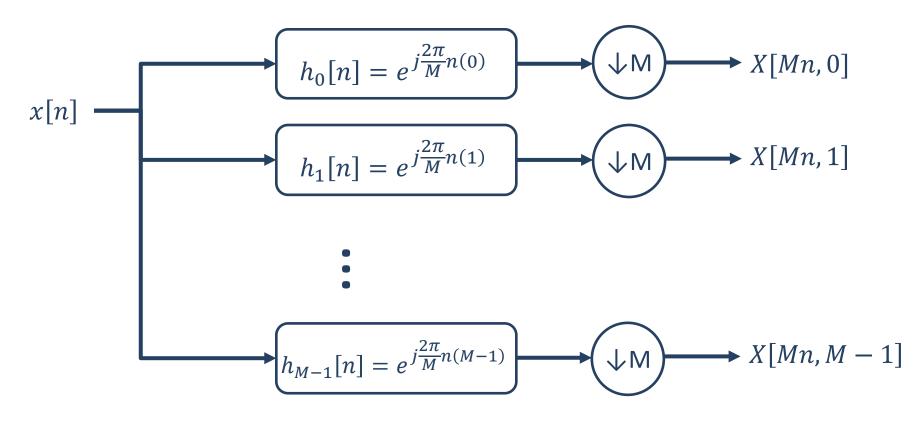


- Result: It is now exactly the same as the STFT with a window of length 2 and shift of 2 between windows
- But, we do not need to buffer x[n]

Consider the following filter bank

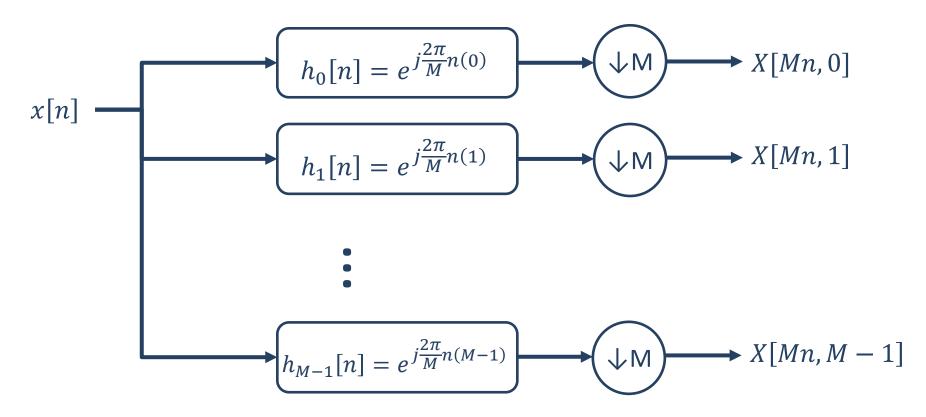


Hence, this is an M-point DFT



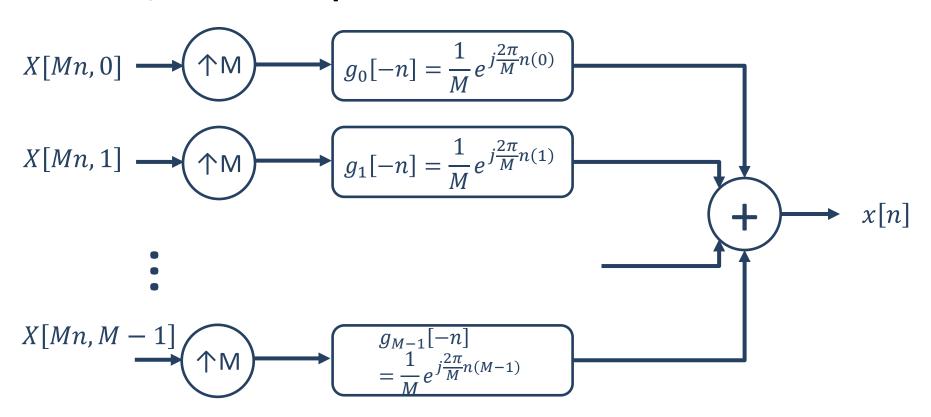
- So, I can implement the STFT as a filter bank...
 - Can I do more?

Hence, this is an M-point DFT

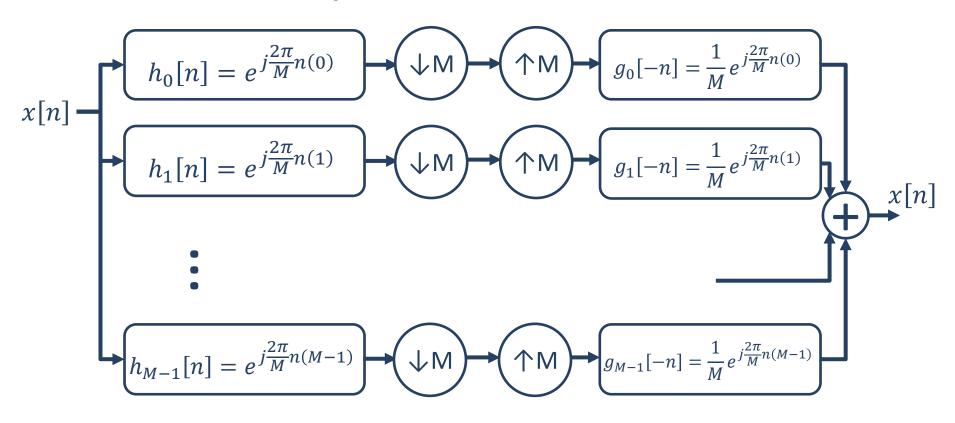


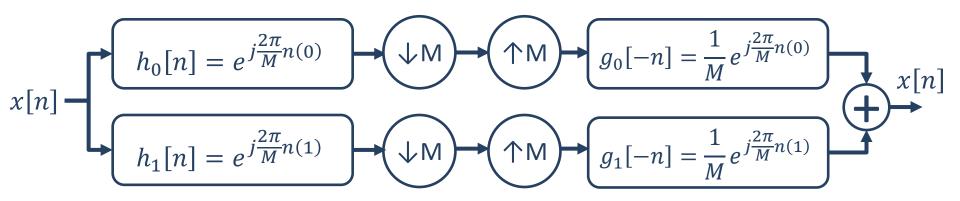
- So, I can implement the STFT as a filter bank...
 - Question: How do I get back into the time domain?

Hence, this is an M-point IDFT

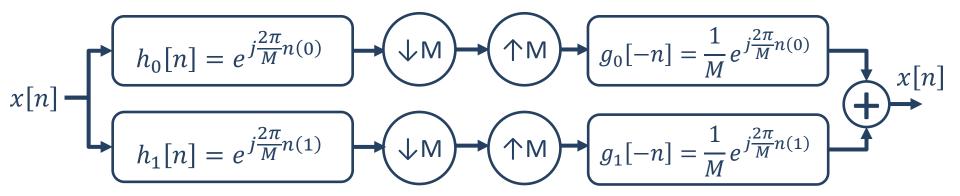


Hence, this is an M-point DFT and IDFT

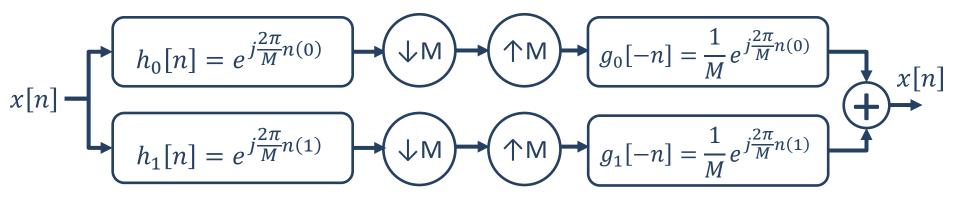




Example: The M=2 point DFT and IDFT



• What are the filter coefficients for $h_0[n]$ and $h_1[n]$?



$$h_0[n] = e^{j\frac{2\pi}{2}n(0)} = 1$$

for
$$0 \le n \le 1$$

$$h_1[n] = e^{j\frac{2\pi}{2}n(1)} = e^{\pi n}$$

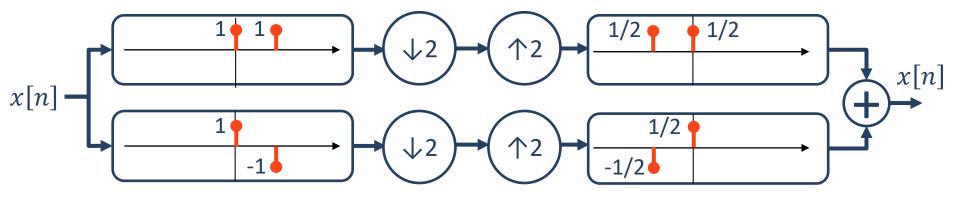
for
$$0 \le n \le 1$$

$$g_0[n] = \frac{1}{M}e^{-j\frac{2\pi}{2}n(0)} = \frac{1}{M}$$

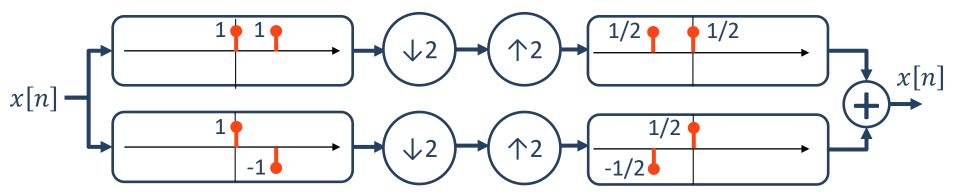
for
$$-1 \le n \le 0$$

$$g_1[n] = \frac{1}{M}e^{-j\frac{2\pi}{2}n(1)} = \frac{1}{M}e^{\pi n}$$

for
$$-1 \le n \le 0$$



- $h_0[n] = \delta[n] + \delta[n-1]$
- $h_1[n] = \delta[n] \delta[n-1]$
- $g_0[n] = \frac{1}{2} [\delta[n+1] + \delta[n]]$
- $g_1[n] = \frac{1}{2} \left[-\delta[n+1] + \delta[n] \right]$

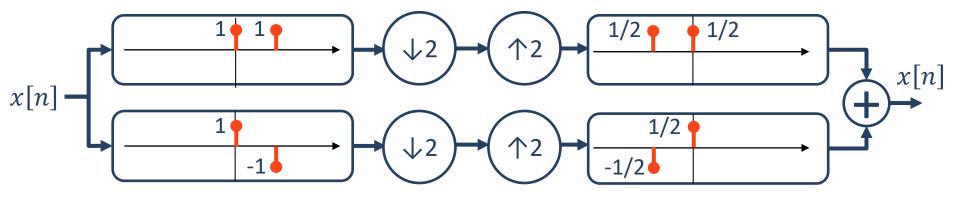


•
$$h_0[n] = \delta[n] + \delta[n-1]$$

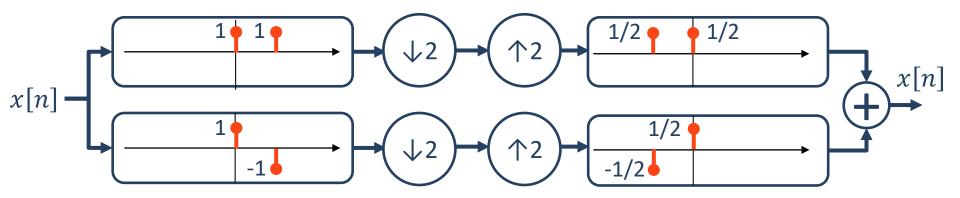
$$H(\omega) = e^{-j\frac{2\pi}{2}\omega(0)} + e^{-j\frac{2\pi}{2}\omega(1)}$$

•
$$h_1[n] = \delta[n] - \delta[n-1]$$

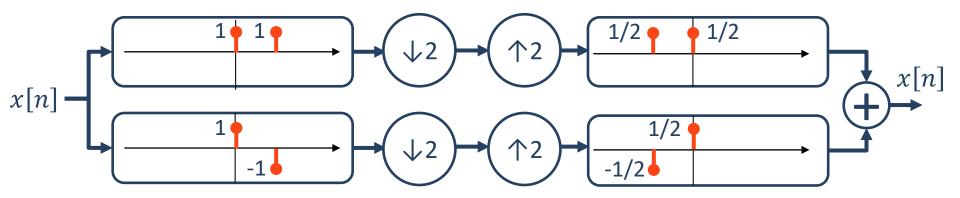
$$H(\omega) = e^{-j\frac{2\pi}{2}\omega(0)} - e^{-j\frac{2\pi}{2}\omega(1)}$$



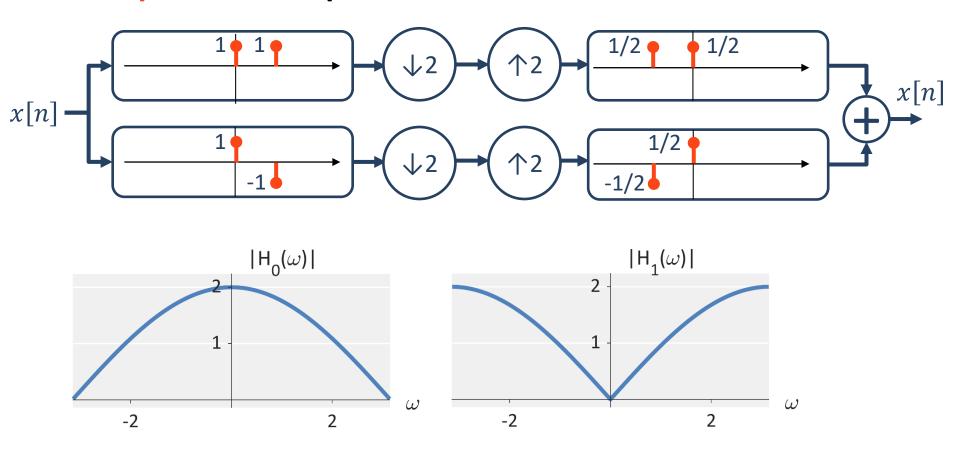
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $H_0(\omega) = 1 + e^{-j\pi\omega} = 2\cos(\omega/2) e^{-j\frac{\pi}{2}\omega}$
- $h_1[n] = \delta[n] \delta[n-1]$
 - $H_1(\omega) = 1 e^{-j\pi\omega} = 2j\sin(\omega/2)e^{-j\frac{\pi}{2}\omega}$



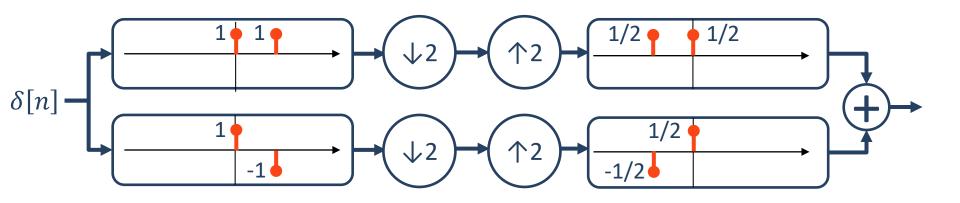
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $|H_0(\omega)| = 2|\cos(\omega/2)|$
- $h_1[n] = \delta[n] \delta[n-1]$
 - $|H_1(\omega)| = 2|\sin(\omega/2)|$



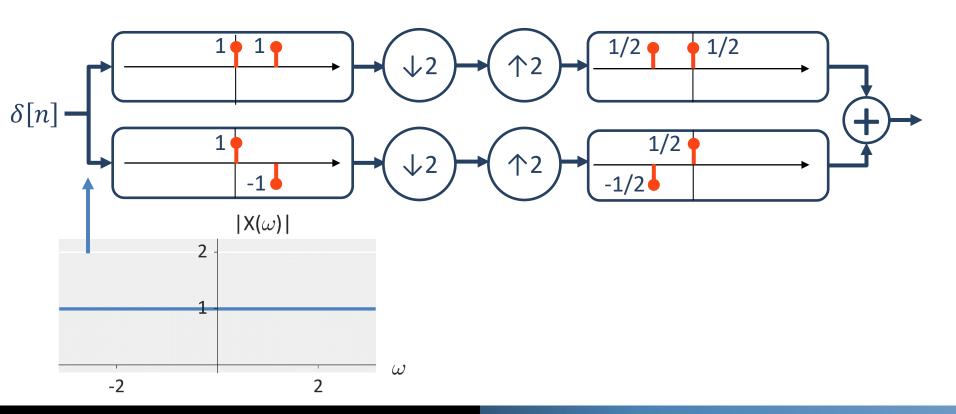
- $h_0[n] = \delta[n] + \delta[n-1]$
 - $|H_0(\omega)| = 2|\cos(\omega/2)|$
 - $|G_0(\omega)| = |\cos(\omega/2)|$
- $h_1[n] = \delta[n] \delta[n-1]$
 - $|H_1(\omega)| = 2|\sin(\omega/2)|$
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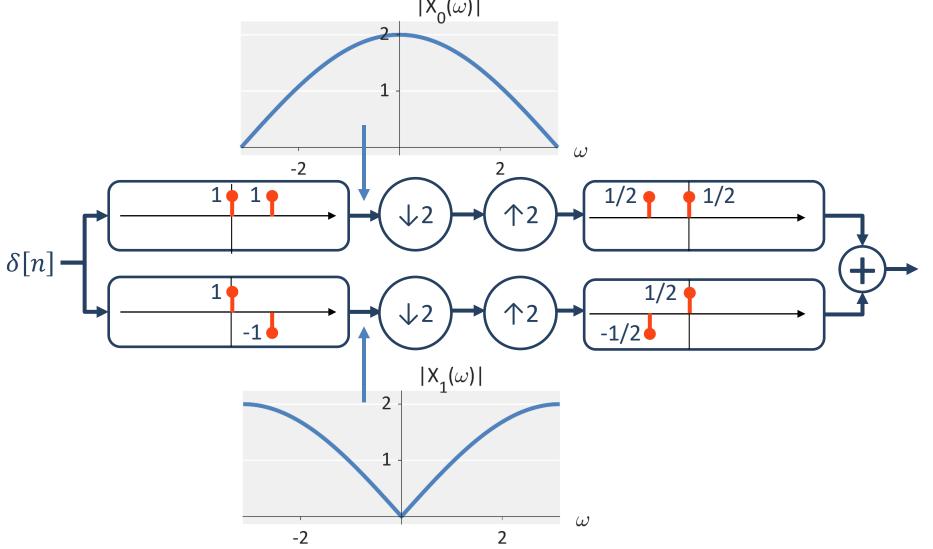
Example: Plot of the intermediate frequency magnitudes



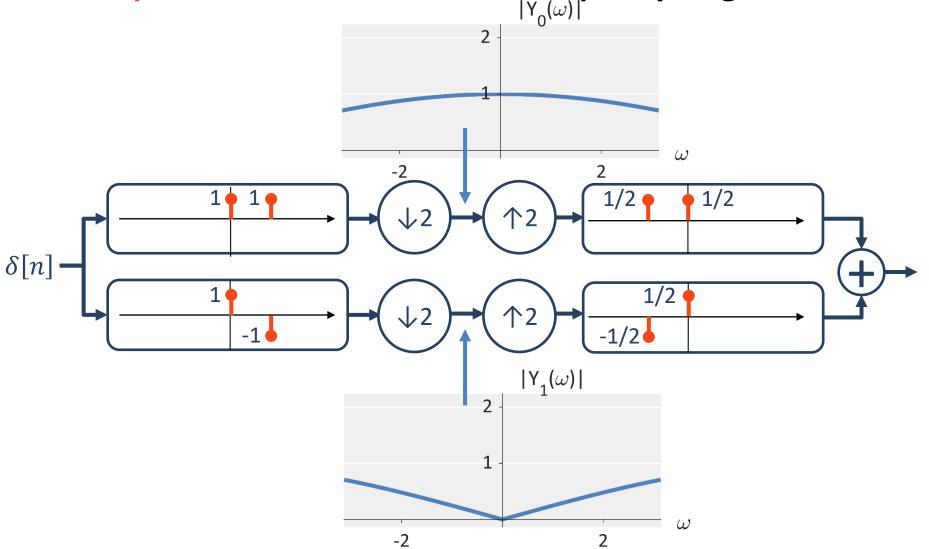
Example: Plot of the intermediate frequency magnitudes



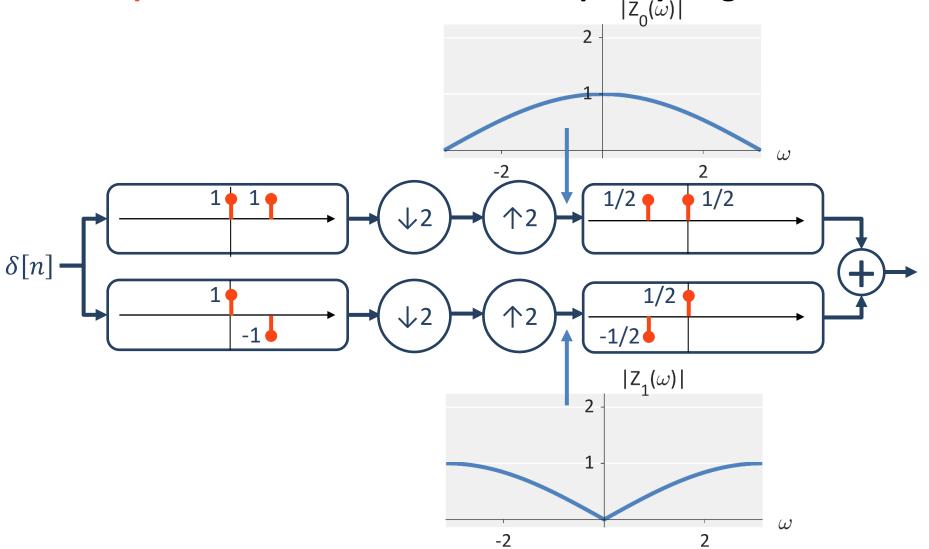
Example: Plot of the intermediate frequency magnitudes $|X_0(\omega)|$



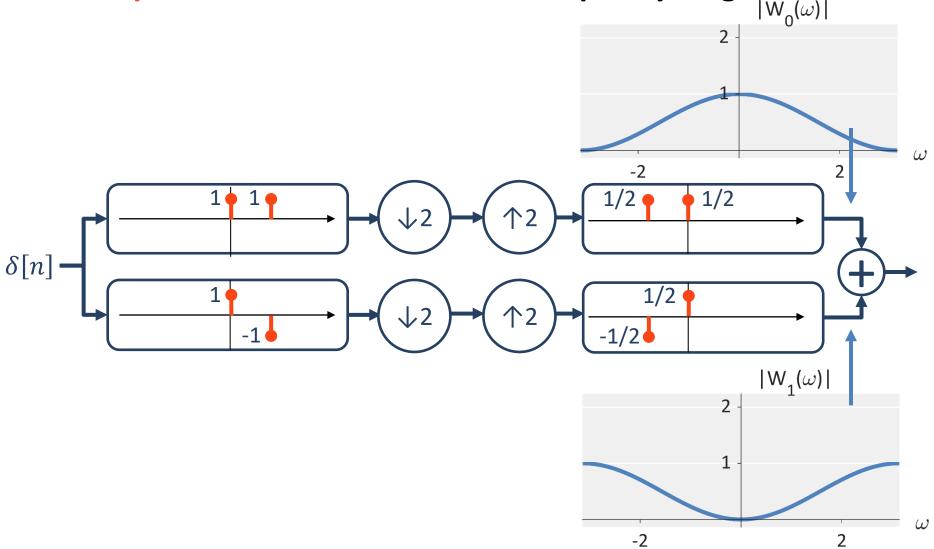
Example: Plot of the intermediate frequency magnitudes $|Y_0(\omega)|$



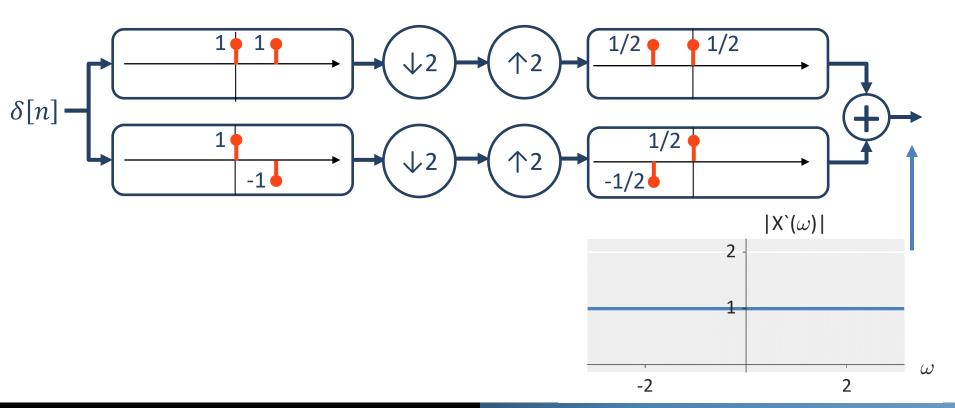
Example: Plot of the intermediate frequency magnitudes $|Z_0(\omega)|$



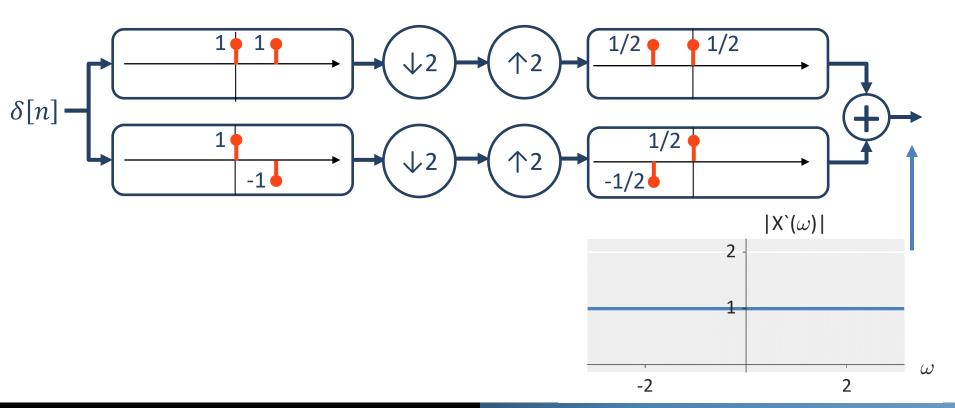
Example: Plot of the intermediate frequency magnitudes $|W_0(\omega)|$



Example: Plot of the intermediate frequency magnitudes



Question: Can we generalize this?

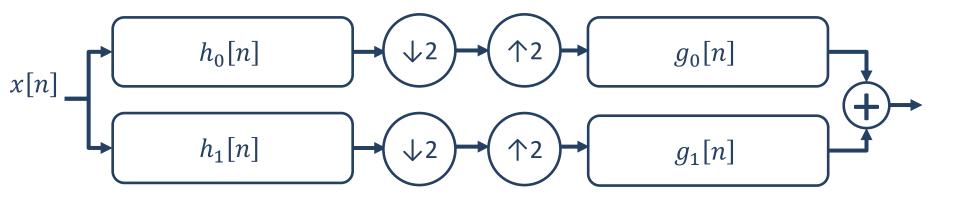


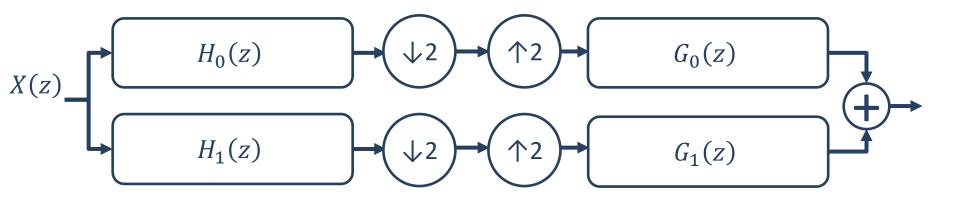
Lecture 26: Filter Banks to Wavelets

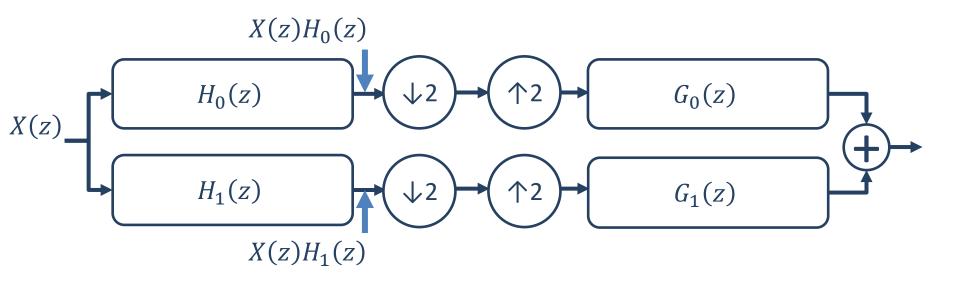
Foundations of Digital Signal Processing

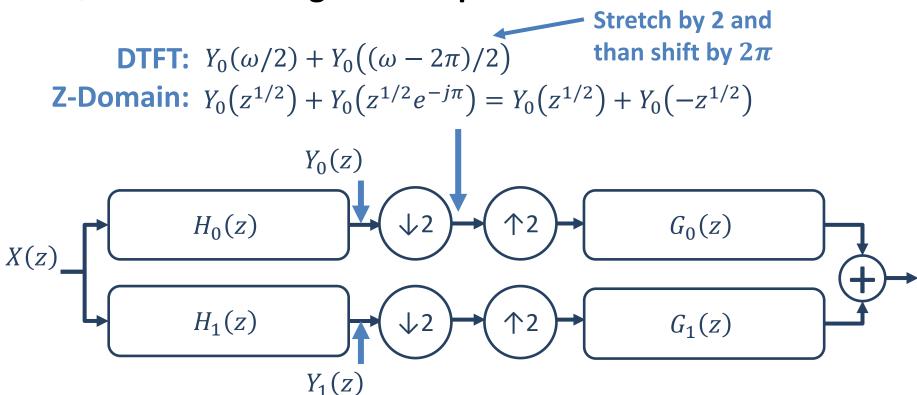
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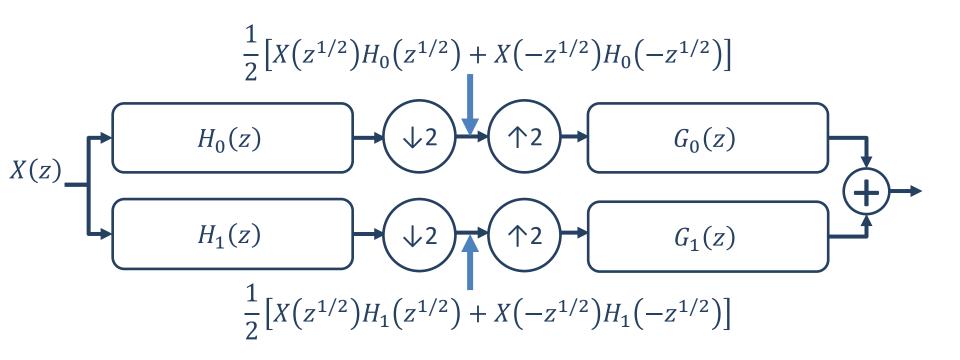
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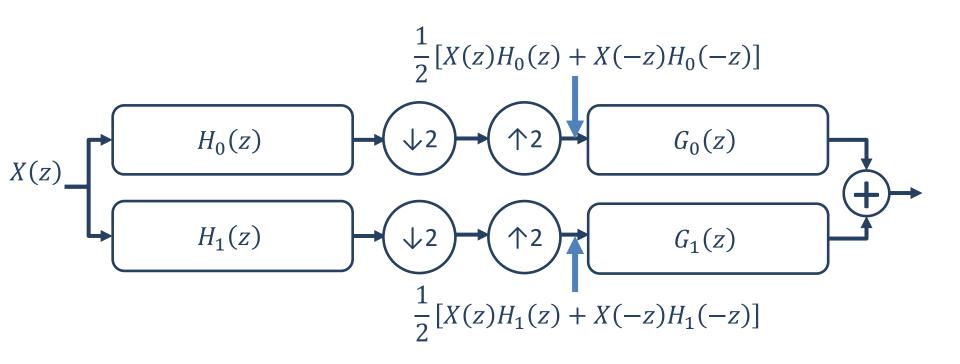


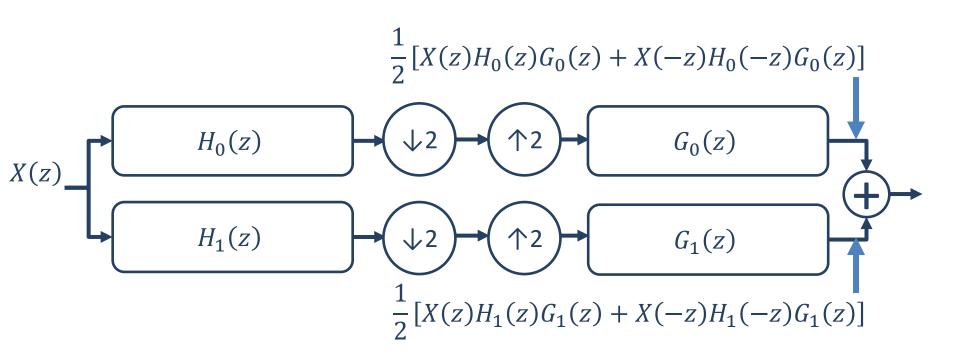


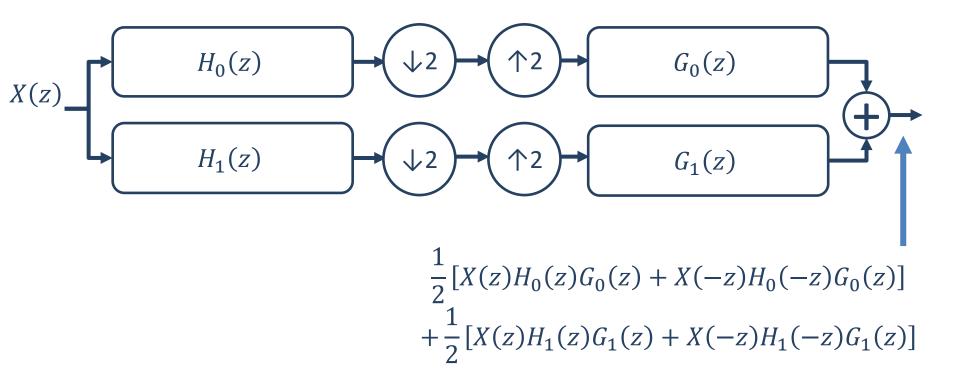


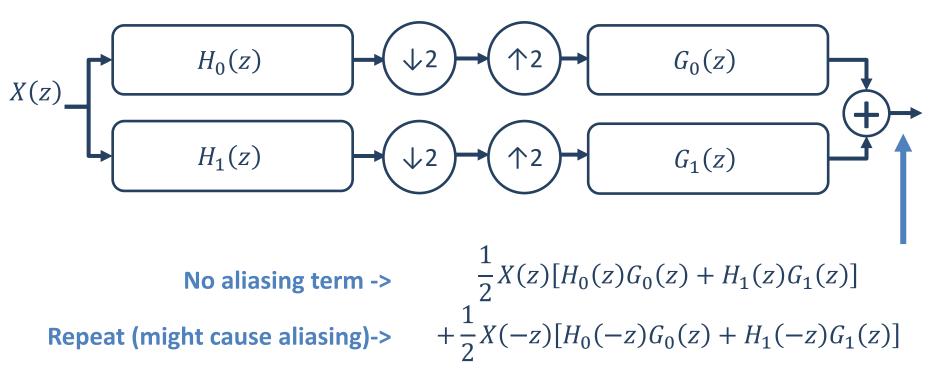












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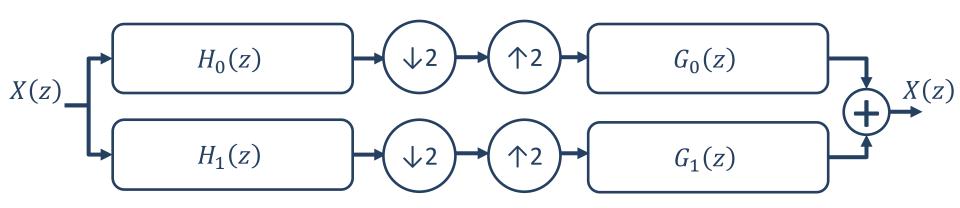
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Question: Can we generalize perfect reconstruction?

$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+ $\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$

Option 1: Alias canceling



Question: Can we generalize perfect reconstruction?

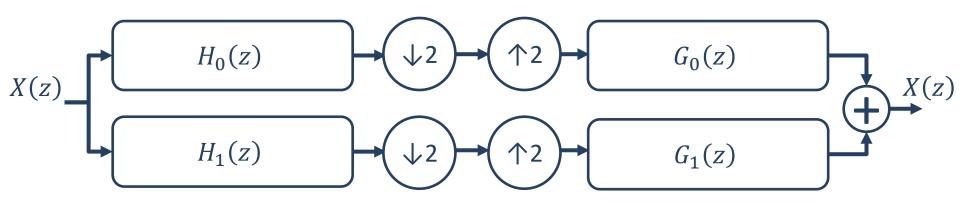
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Option 1: Alias canceling

$$\bullet H_0(z)G_0(z) + H_1(z)G_1(z) = 2$$

$$\phi H_0(-z)G_0(z) + H_1(-z)G_1(z) = 0$$



Question: Can we generalize perfect reconstruction?

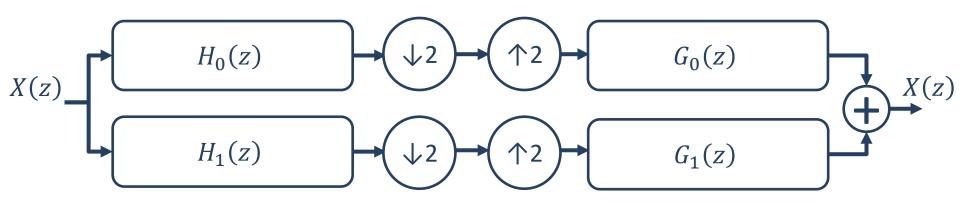
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Option 1: Alias canceling (what is this in frequency?)

$$\bullet H_0(z)G_0(z) + H_1(z)G_1(z) = 2$$

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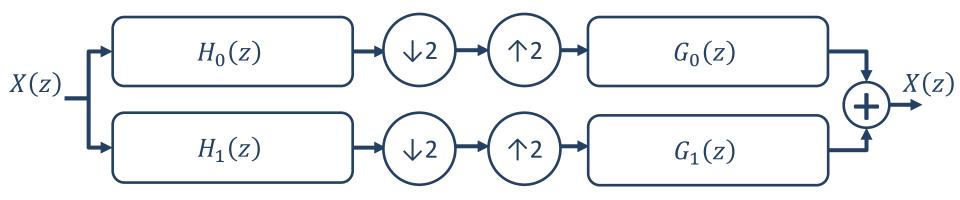
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Option 1: Alias canceling (what is this in frequency?)

$$\bullet \ H_0(\omega)G_0(\omega) + H_1(\omega)G_1(\omega) = 2$$

$$\diamond \ H_0(\omega-\pi)G_0(\omega) + H_1(\omega-\pi)G_1(\omega) = 0$$

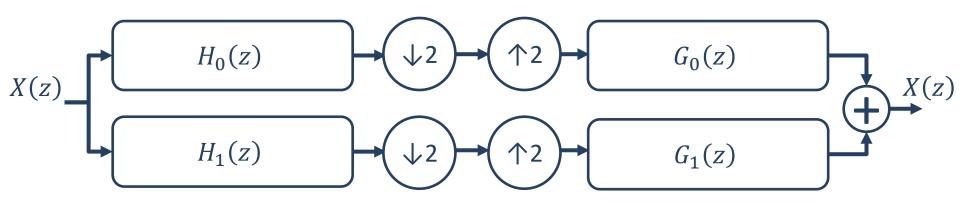


Question: Can we generalize perfect reconstruction?

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Option 1: Alias canceling (what is this in time?)



Question: Can we generalize perfect reconstruction?

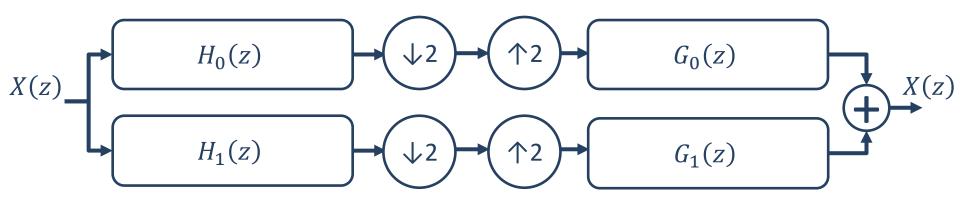
$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+ $\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$

Option 1: Alias canceling (what is this in time?)

$$b_0[n] * g_0[n] + h_1[n] * g_1[n] = 2\delta[n]$$

$$(-1)^n h_0[n] * g_0[n] + [(-1)^n h_1[n]] * g_1[n] = 0$$



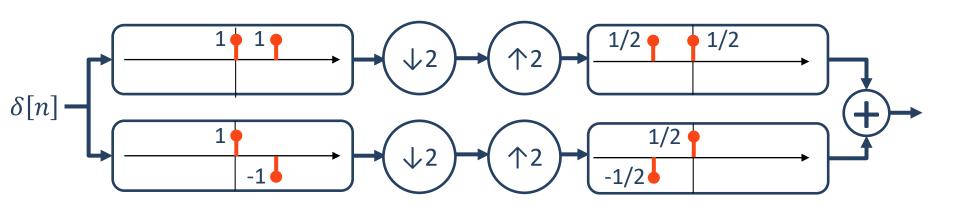
Example: Alias Canceling

$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+ $\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$

$$\bullet H_0(z)G_0(z) + H_1(z)G_1(z) = 2$$

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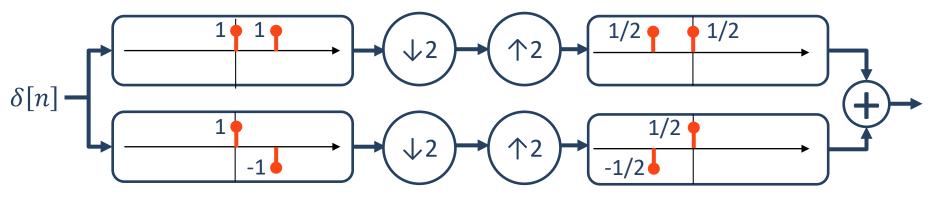
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$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[H_0(-z)G_0(z) + H_1(-z)G_1(z) \right]$$

$$^{\circ} \frac{1}{2} (1+z^{-1})(z^{+1}+1) + \frac{1}{2} (1-z^{-1})(-z^{+1}+1) = 2$$

$$^{\circ} \frac{1}{2} (1-z^{-1})(z^{+1}+1) + \frac{1}{2} (1+z^{-1})(-z^{+1}+1) = 0$$



Example: Alias Canceling

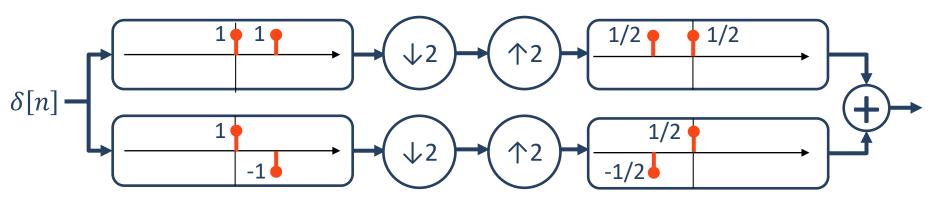
$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[H_0(-z)G_0(z) + H_1(-z)G_1(z) \right]$$

$$^{\circ} \frac{1}{2}(2+z^{+1}+z^{-1}) + \frac{1}{2}(2-z^{+1}-z^{-1}) = 2$$

$$^{\circ} \frac{1}{2}(z^{+1}-z^{-1}) + \frac{1}{2}(-z^{+1}+z^{-1}) = 0$$

$$^{\circ} \frac{1}{2}(z^{+1} - z^{-1}) + \frac{1}{2}(-z^{+1} + z^{-1}) = 0$$

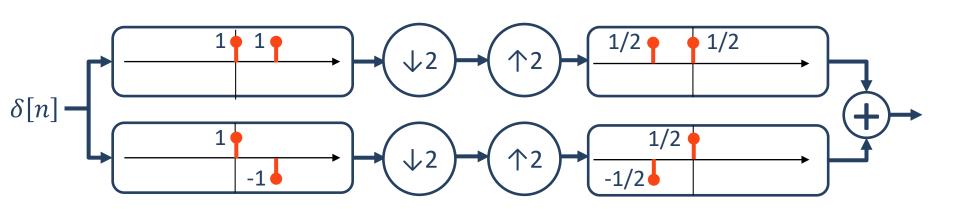


Example: Alias Canceling

$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+
$$\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$$

- $\diamond 2 = 2$
- 0 = 0



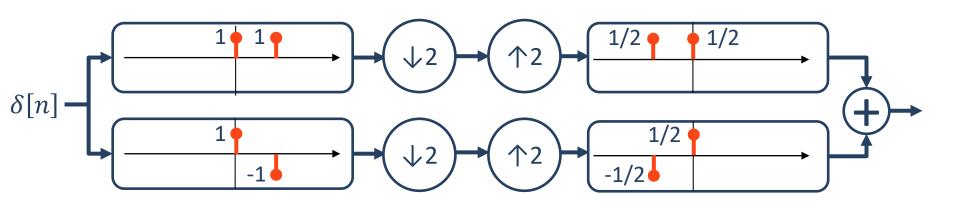
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+ $\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$

$$b_0[n] * g_0[n] + h_1[n] * g_1[n] = 2\delta[n]$$

$$(-1)^n h_0[n] * g_0[n] + [(-1)^n h_1[n]] * g_1[n] = 0$$



Lecture 26: Filter Banks to Wavelets

Foundations of Digital Signal Processing

Outline

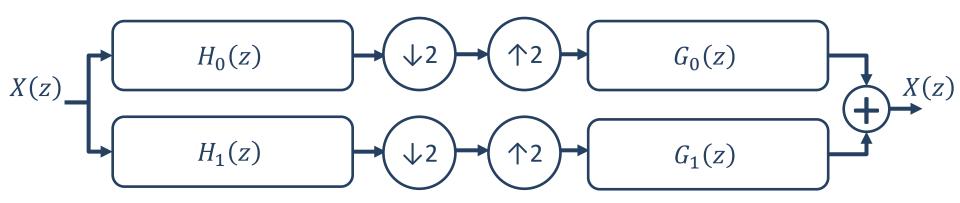
- DFT Filter Banks [without downsampling]
- DFT Filter Bank [with downsampling]
- General Two-Channel Filter Banks
- Two-Channel Filter Bank Perfect Reconstruction (aliasing canceling)
- Two-Channel Filter Bank Perfect Reconstruction (orthogonal filter bank)
- Polyphase Filters
- Multi-Channel Filter Bank Perfect Reconstruction
- Wavelets

Question: Can we generalize perfect reconstruction?

$$X(z) = \frac{1}{2}X(z) \left[H_0(z)G_0(z) + H_1(z)G_1(z) \right]$$

+
$$\frac{1}{2}X(-z)[H_0(-z)G_0(z) + H_1(-z)G_1(z)]$$

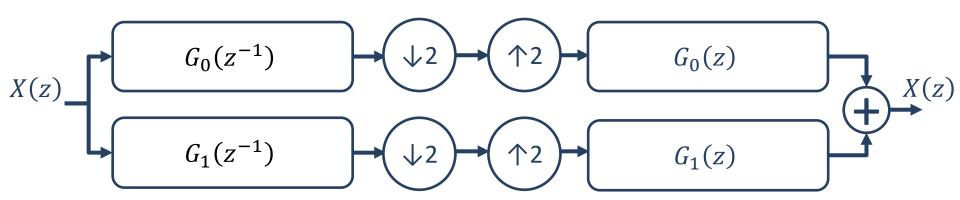
Option 2: Orthogonal Filter Bank



$$X(z) = \frac{1}{2}X(z) \left[G_0(z^{-1})G_0(z) + G_1(z^{-1})G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[G_0(-z^{-1})G_0(z) + G_1(-z^{-1})G_1(z) \right]$$

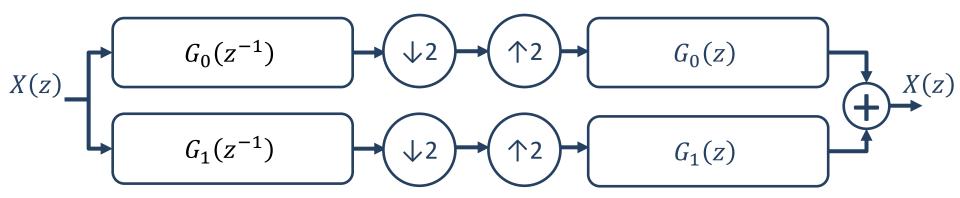
- Option 2: Orthogonal Filter Bank
 - Let $H_0(z) = G_0(z^{-1})$
 - \bullet Let $H_1(z) = G_1(z^{-1})$



$$X(z) = \frac{1}{2}X(z) \left[G_0(z^{-1})G_0(z) + G_1(z^{-1})G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[G_0(-z^{-1})G_0(z) + G_1(-z^{-1})G_1(z) \right]$$

- Option 2: Orthogonal Filter Bank
 - Let $H_0(z) = G_0(z^{-1})$
 - Let $H_1(z) = G_1(z^{-1})$
 - Assume $X(z) = \alpha G_0(z) + \beta G_1(z)$



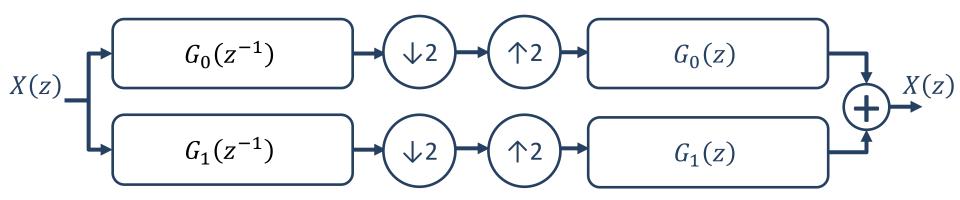
Question: Can we generalize perfect reconstruction?

$$X(z) = \frac{1}{2} \left[\alpha G_0(z) + \beta G_1(z) \right] \left[G_0(z^{-1}) G_0(z) + G_1(z^{-1}) G_1(z) \right]$$

+
$$\frac{1}{2} \left[\alpha G_0(-z) + \beta G_1(-z) \right] \left[G_0(-z^{-1}) G_0(z) + G_1(-z^{-1}) G_1(z) \right]$$

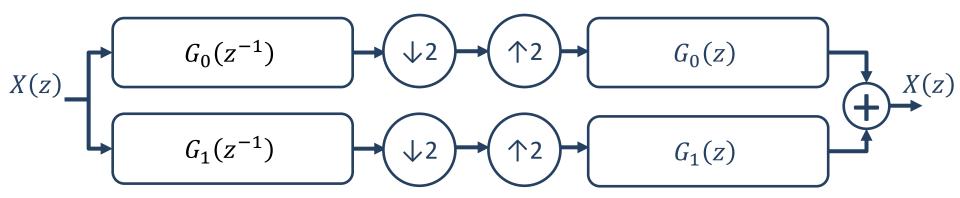
- Option 2: Orthogonal Filter Bank
 - \bullet Let $H_0(z) = G_0(z^{-1})$
 - Let $H_1(z) = G_1(z^{-1})$
 - Assume $X(z) = \alpha G_0(z) + \beta G_1(z)$

i.e., $G_0(z)$ and $G_1(z)$ are bases of X(z)



$$\begin{split} &X(z)\\ &=\frac{1}{2}\alpha\big[G_0(z)[G_0(z)G_0(z^{-1})+G_0(-z)G_0(-z^{-1})]+G_1(z)[G_0(z)G_1(z^{-1})+G_0(-z)G_1(-z^{-1})]\big]\\ &+\frac{1}{2}\beta\big[G_0(z)[G_1(z)G_0(z^{-1})+G_1(-z)G_0(-z^{-1})]+G_1(z)[G_1(z)G_1(z^{-1})+G_1(-z)G_1(-z^{-1})]\big] \end{split}$$

- Option 2: Orthogonal Filter Bank
 - \bullet Let $H_0(z) = G_0(z^{-1})$
 - \bullet Let $H_1(z) = G_1(z^{-1})$
 - Assume $X(z) = \alpha G_0(z) + \beta G_1(z)$



Question: Can we generalize perfect reconstruction?

$$\begin{split} X(z) & 2 & 0 \\ &= \frac{1}{2}\alpha \big[G_0(z) \big[G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1})\big] + G_1(z) \big[G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1})\big] \big] \\ &+ \frac{1}{2}\beta \big[G_0(z) \big[G_1(z)G_0(z^{-1}) + G_1(-z)G_0(-z^{-1})\big] + G_1(z) \big[G_1(z)G_1(z^{-1}) + G_1(-z)G_1(-z^{-1})\big] \big] \\ & 0 & 2 \end{split}$$

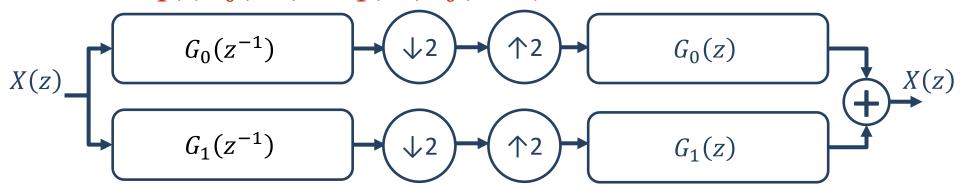
Option 2: Orthogonal Filter Bank

$$\circ G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$$

$$G_1(z)G_1(z^{-1}) + G_1(-z)G_1(-z^{-1}) = 2$$

$$\circ G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1}) = 0$$

$$\circ G_1(z)G_0(z^{-1}) + G_1(-z)G_0(-z^{-1}) = 0$$



Question: Can we generalize perfect reconstruction?

$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

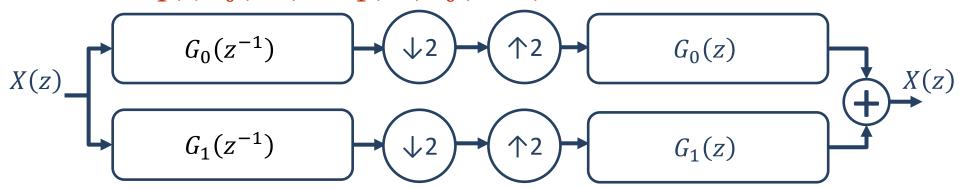
Option 2: Orthogonal Filter Bank

$$\circ$$
 $G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$

$$\circ$$
 $G_1(z)G_1(z^{-1}) + G_1(-z)G_1(-z^{-1}) = 2$

$$G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1}) = 0$$

$$\circ G_1(z)G_0(z^{-1}) + G_1(-z)G_0(-z^{-1}) = 0$$



Question: Can we generalize perfect reconstruction?

$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

Option 2: Orthogonal Filter Bank

$$\circ G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$$

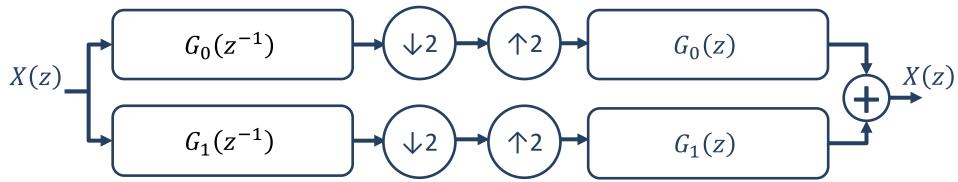
$$\circ G_1(z)G_1(z^{-1}) + G_1(-z)G_1(-z^{-1}) = 2$$

$$\circ G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1}) = 0$$

Equations are same if:

$$z \leftarrow z^{-1}$$

So we only need one.



Question: Can we generalize perfect reconstruction?

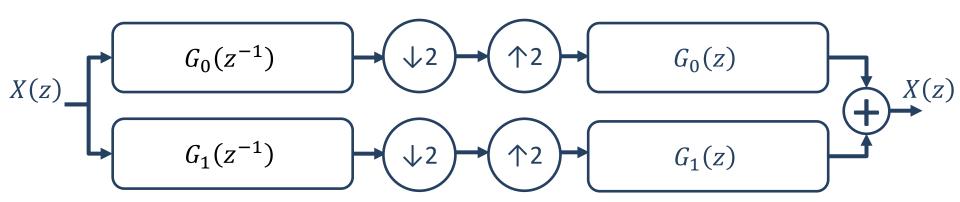
$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

Option 2: Orthogonal Filter Bank

$$\circ$$
 $G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$

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Question: Can we generalize perfect reconstruction?

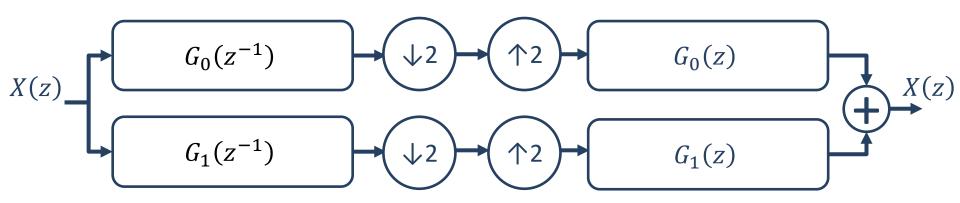
$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

Option 2: Orthogonal Filter Bank (what is this in frequency?)

$$\circ G_0(\omega)G_0(-\omega) + G_0(\omega - \pi)G_0(-\omega - \pi) = 2$$

•
$$G_1(\omega)G_1(-\omega) + G_1(\omega - \pi)G_1(-\omega - \pi) = 2$$

$$\circ G_0(\omega)G_1(-\omega) + G_0(\omega - \pi)G_1(-\omega - \pi) = 0$$



Question: Can we generalize perfect reconstruction?

$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

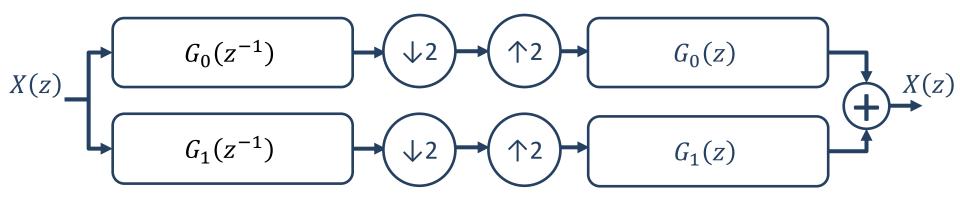
Option 2: Orthogonal Filter Bank (what is this in frequency?)

$$\circ G_0(\omega)G_0^*(\omega) + G_0(\omega - \pi)G_0^*(\omega - \pi) = 2$$

$$\circ G_1(\omega)G_1^*(\omega) + G_1(\omega - \pi)G_1^*(\omega - \pi) = 2$$

$$G_0(\omega)G_1^*(\omega) + G_0(\omega - \pi)G_1^*(\omega - \pi) = 0$$

Assuming real filter coefficients

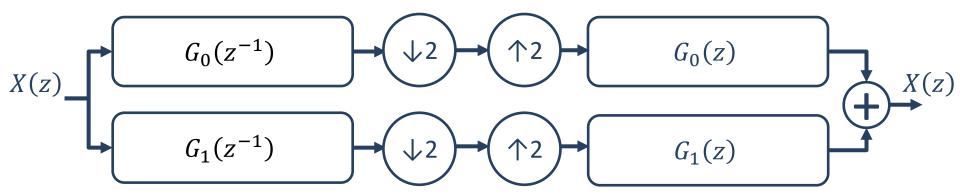


Question: Can we generalize perfect reconstruction?

$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

- Option 2: Orthogonal Filter Bank (what is this in frequency?)
 - $|G_0(\omega)|^2 + |G_0(\omega \pi)|^2 = 2$
 - $|G_1(\omega)|^2 + |G_1(\omega \pi)|^2 = 2$
 - $\circ G_0(\omega)G_1^*(\omega) + G_0(\omega \pi)G_1^*(\omega \pi) = 0$

Assuming real filter coefficients



Question: Can we generalize perfect reconstruction?

$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

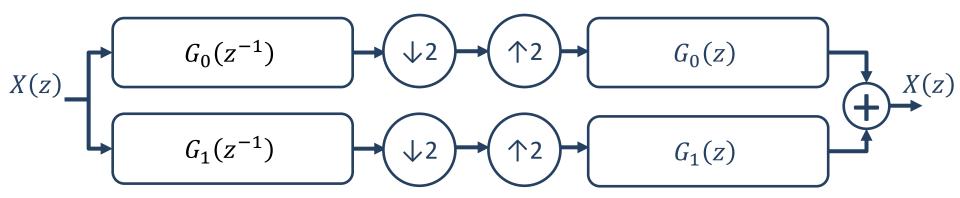
Option 2: Orthogonal Filter Bank (what is this in time?)

$$|G_0(\omega)|^2 + |G_0(\omega - \pi)|^2 = 2$$

$$|G_1(\omega)|^2 + |G_1(\omega - \pi)|^2 = 2$$

$$G_0(\omega)G_1^*(\omega) + G_0(\omega - \pi)G_1^*(\omega - \pi) = 0$$

Assuming real filter coefficients



Question: Can we generalize perfect reconstruction?

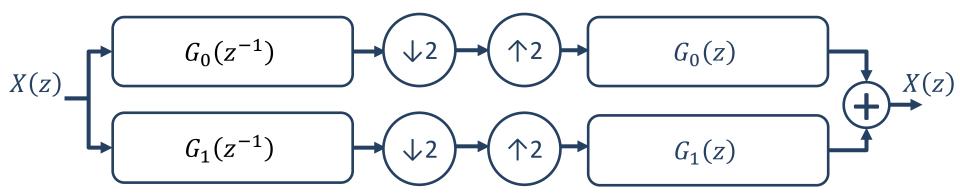
$$X(z) = \alpha G_0(z) + \beta G_1(z)$$

Option 2: Orthogonal Filter Bank (what is this in time?)

$$\circ g_0[n] * g_0[-n] + [(-1)^n g_0[n]] * [(-1)^n g_0[-n]] = 2\delta[n]$$

$$\circ g_1[n] * g_1[-n] + [(-1)^n g_1[n]] * [(-1)^n g_1[-n]] = 2\delta[n]$$

$$\diamond \ g_0[n] * g_1[-n] + \left[(-1)^n g_0[n] \right] * \left[(-1)^n g_1[-n] \right] = 0$$

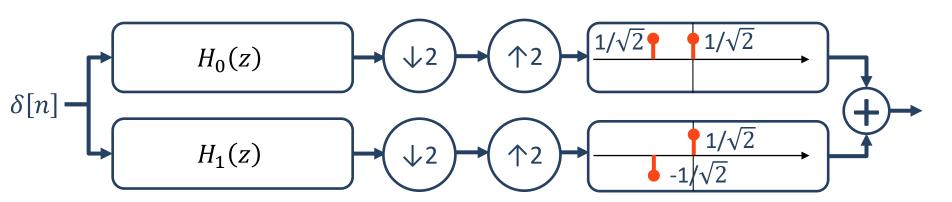


Question: Can we generalize perfect reconstruction?

$$X(z) = \frac{1}{2}X(z) \left[G_0(z^{-1})G_0(z) + G_1(z^{-1})G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[G_0(-z^{-1})G_0(z) + G_1(-z^{-1})G_1(z) \right]$$

- Do the following filters satisfy an orthogonal filter bank?
 - Let $H_0(z) = G_0(z^{-1})$
 - \bullet Let $H_1(z) = G_1(z^{-1})$

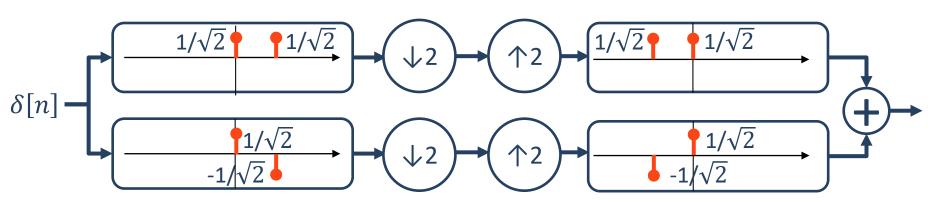


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- Do the following filters satisfy an orthogonal filter bank?
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Question: Can we generalize perfect reconstruction?

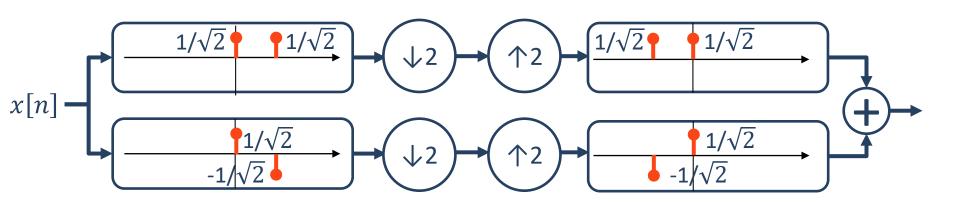
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$$\circ$$
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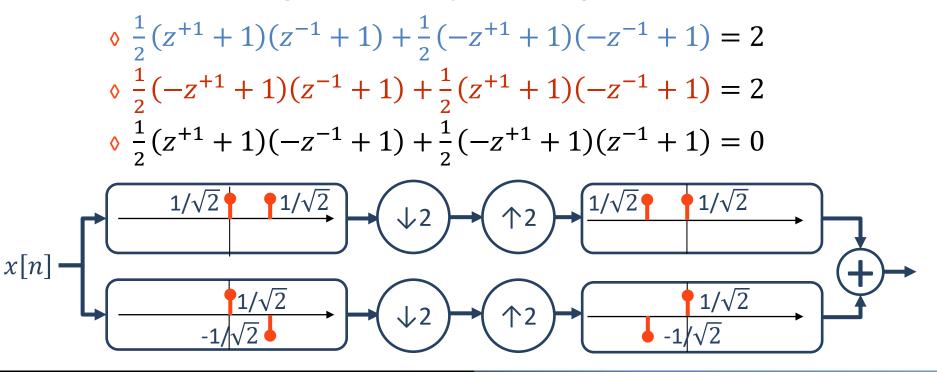
$$\circ G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1}) = 0$$



Question: Can we generalize perfect reconstruction?

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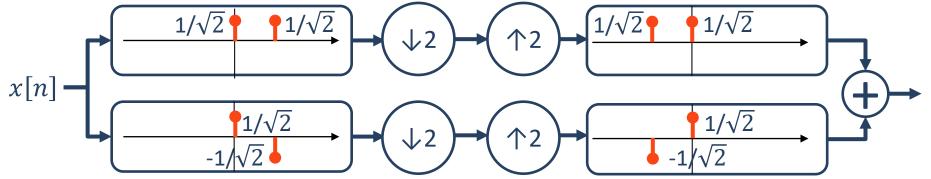
+
$$\frac{1}{2}X(-z) \left[G_0(-z^{-1})G_0(z) + G_1(-z^{-1})G_1(z) \right]$$



Question: Can we generalize perfect reconstruction?

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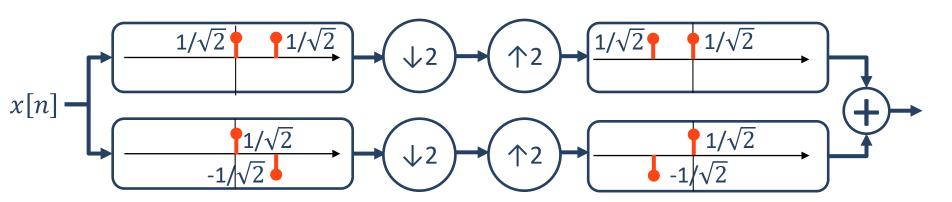


Question: Can we generalize perfect reconstruction?

$$X(z) = \frac{1}{2}X(z) \left[G_0(z^{-1})G_0(z) + G_1(z^{-1})G_1(z) \right]$$

+
$$\frac{1}{2}X(-z) \left[G_0(-z^{-1})G_0(z) + G_1(-z^{-1})G_1(z) \right]$$

- Do the following filters satisfy an orthogonal filter bank?
 - \diamond 2 = 2
 - $\diamond 2 = 2$
 - 0 = 0



Question: Can we generalize perfect reconstruction?

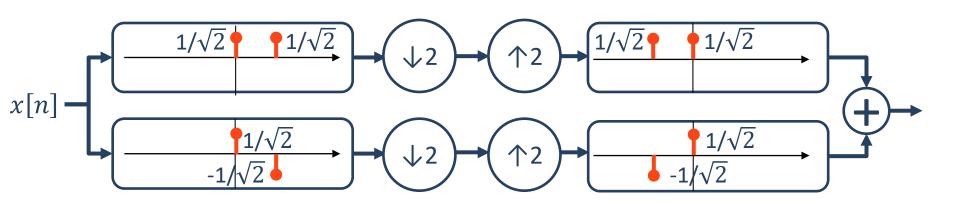
$$X(z) = \frac{1}{2}X(z) \left[G_0(z^{-1})G_0(z) + G_1(z^{-1})G_1(z) \right]$$

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$$\circ g_0[n] * g_0[-n] + [(-1)^n g_0[n]] * [(-1)^n g_0[-n]] = 2\delta[n]$$

$$\circ g_1[n] * g_1[-n] + [(-1)^n g_1[n]] * [(-1)^n g_1[-n]] = 2\delta[n]$$

$$\diamond \ g_0[n] * g_1[-n] + \left[(-1)^n g_0[n] \right] * \left[(-1)^n g_1[-n] \right] = 0$$



Question:

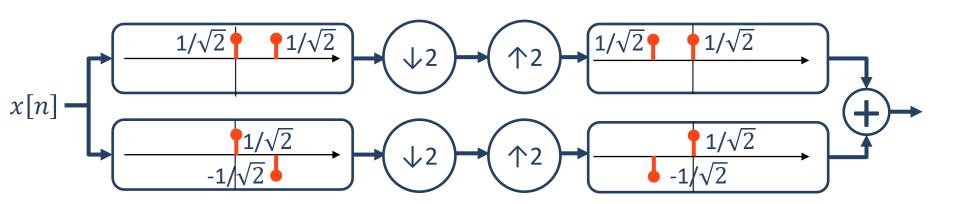
What are the benefits of this approach over alias canceling?

Option 2: Orthogonal Filter Bank

$$\circ G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$$

$$\circ$$
 $G_1(z)G_1(z^{-1}) + G_1(-z)G_1(-z^{-1}) = 2$

$$\circ G_0(z)G_1(z^{-1}) + G_0(-z)G_1(-z^{-1}) = 0$$



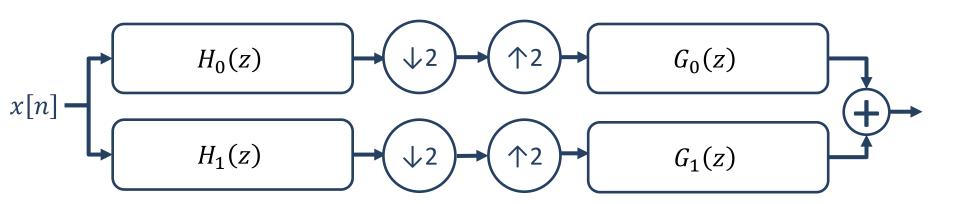
Lecture 26: Filter Banks to Wavelets

Foundations of Digital Signal Processing

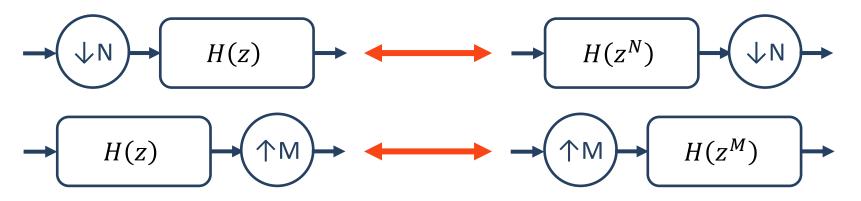
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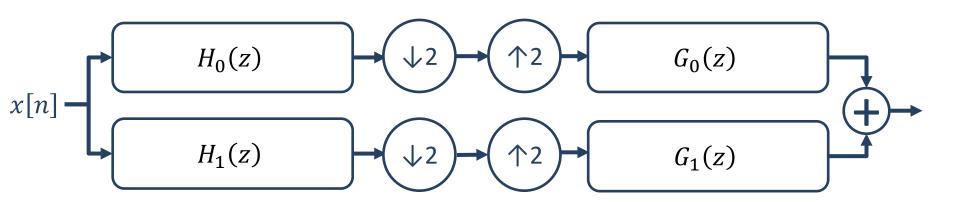
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- Multi-Channel Filter Bank Perfect Reconstruction
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Question: Can we make this more efficient?

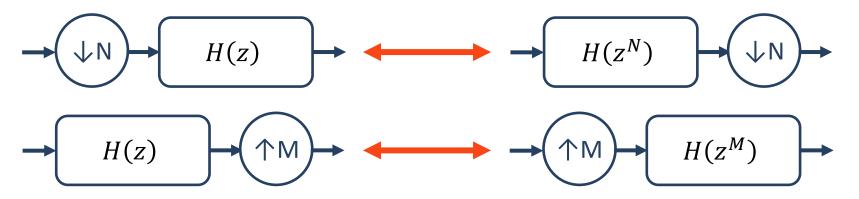


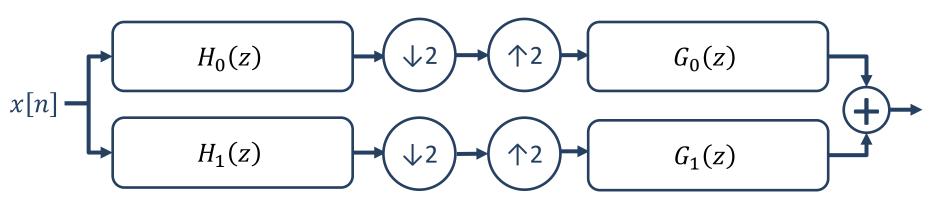
Noble Properties



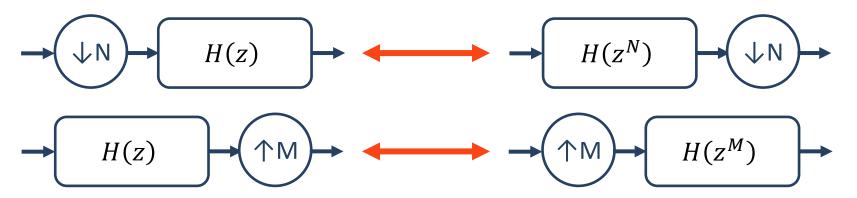


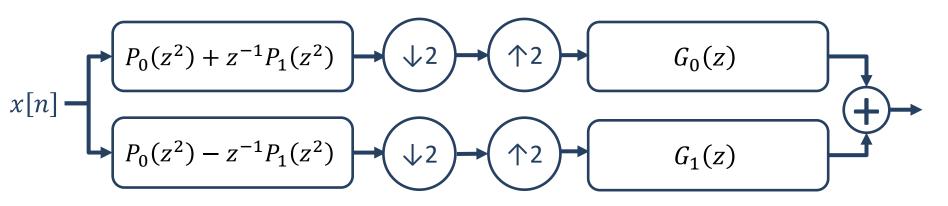
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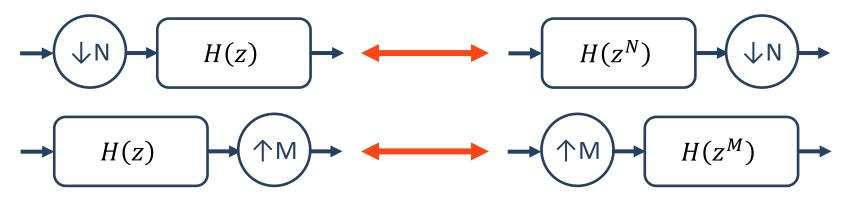


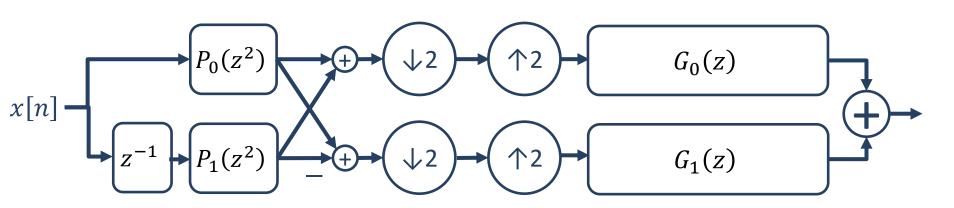
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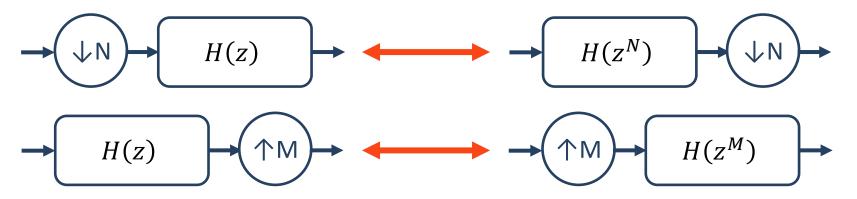


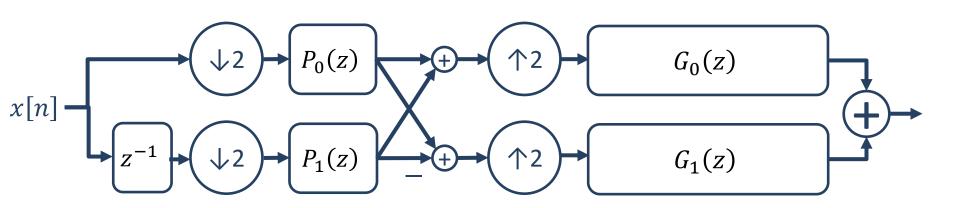
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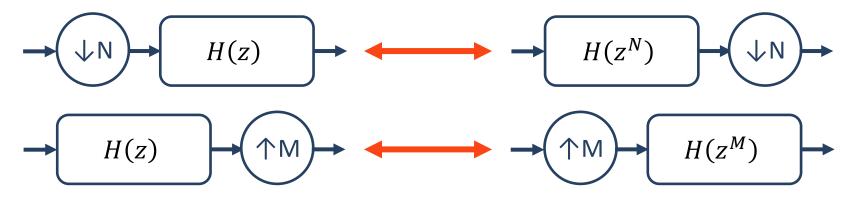


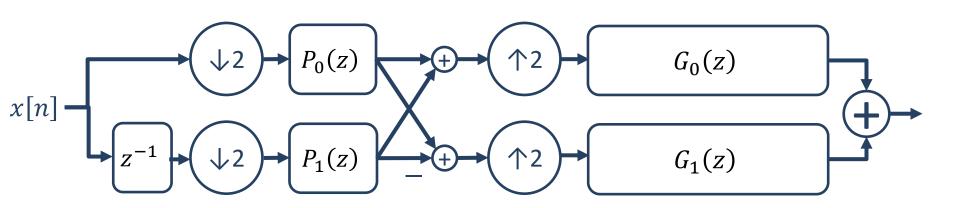
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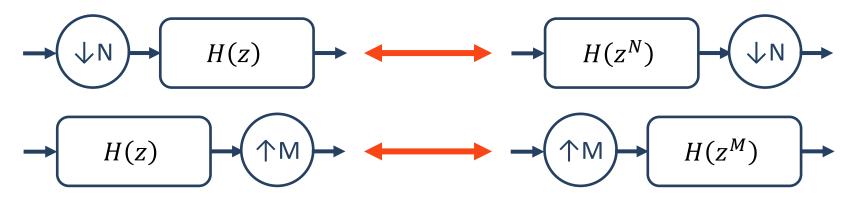


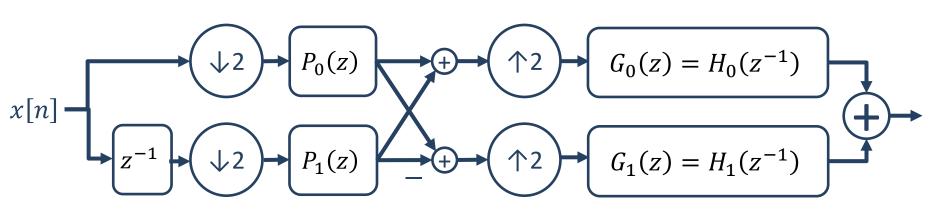
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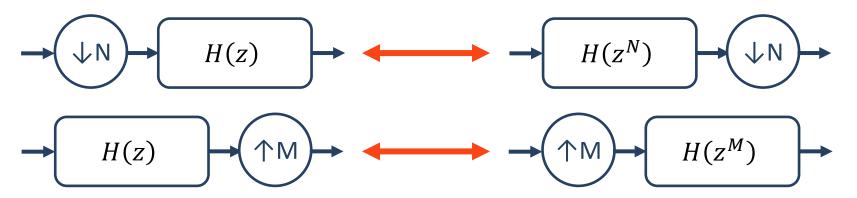


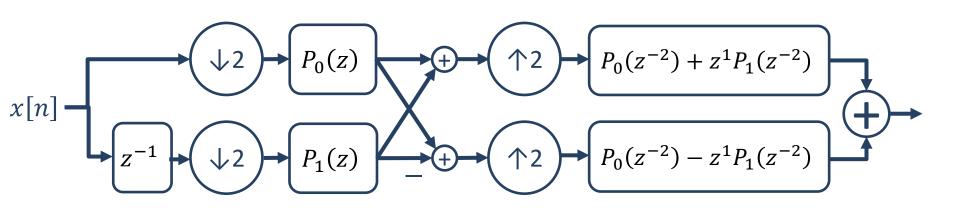
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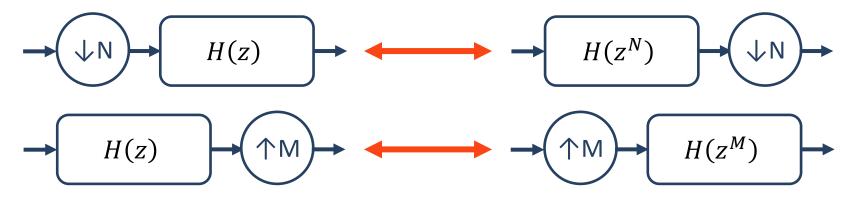


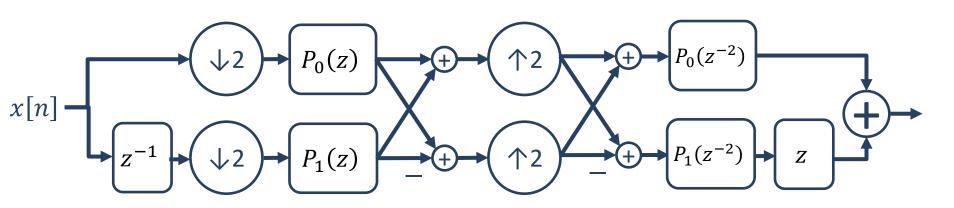
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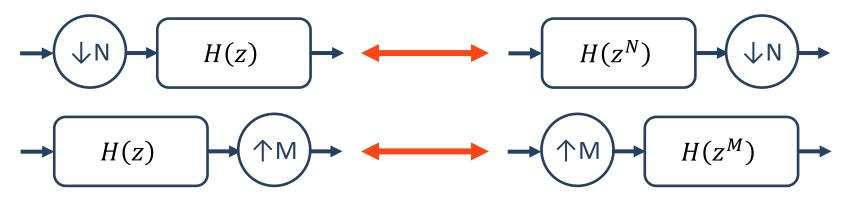


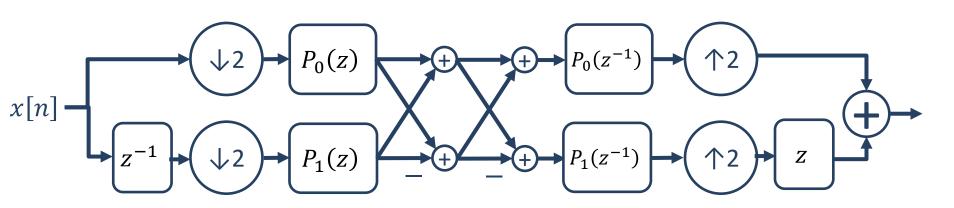
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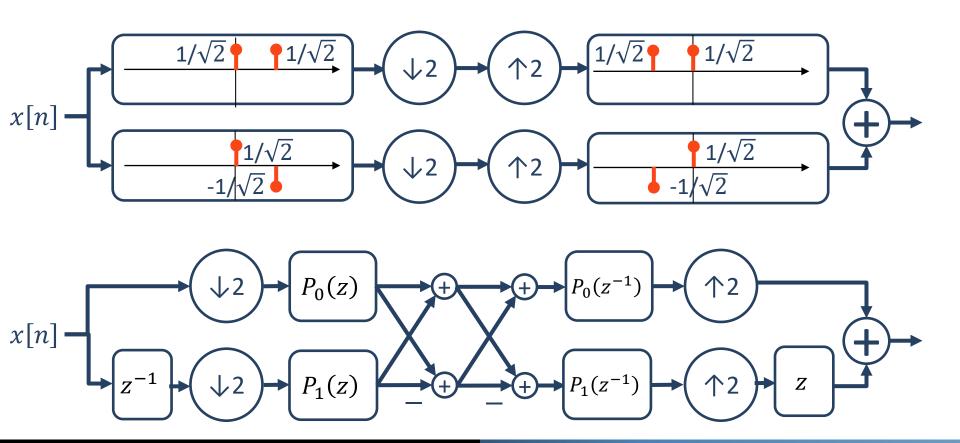




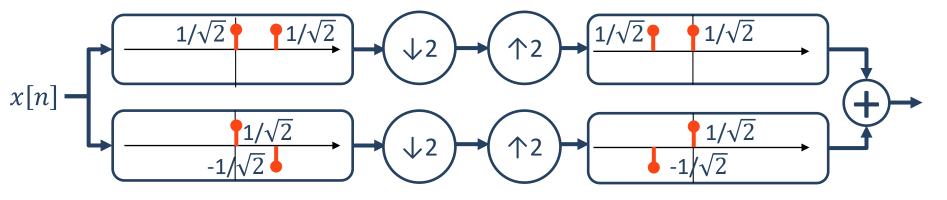
Noble Properties

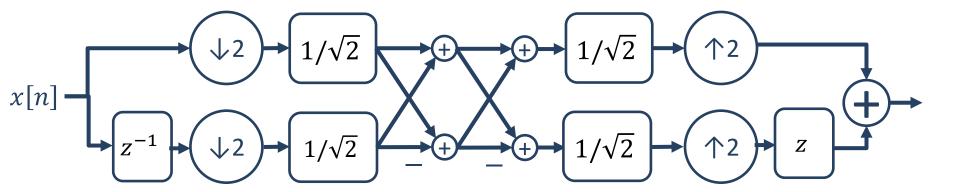


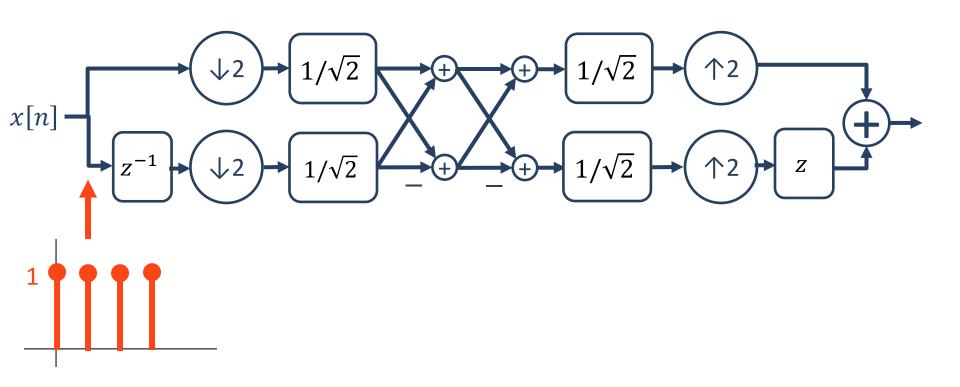


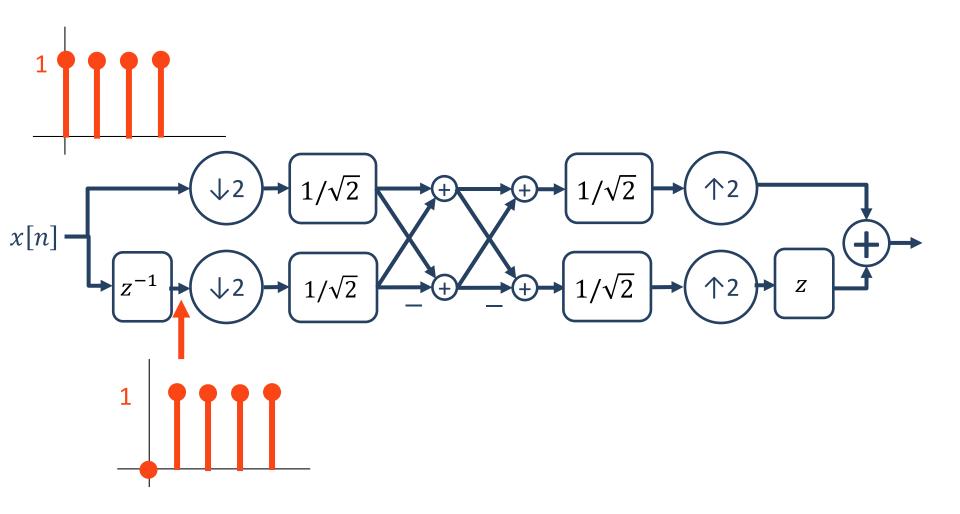


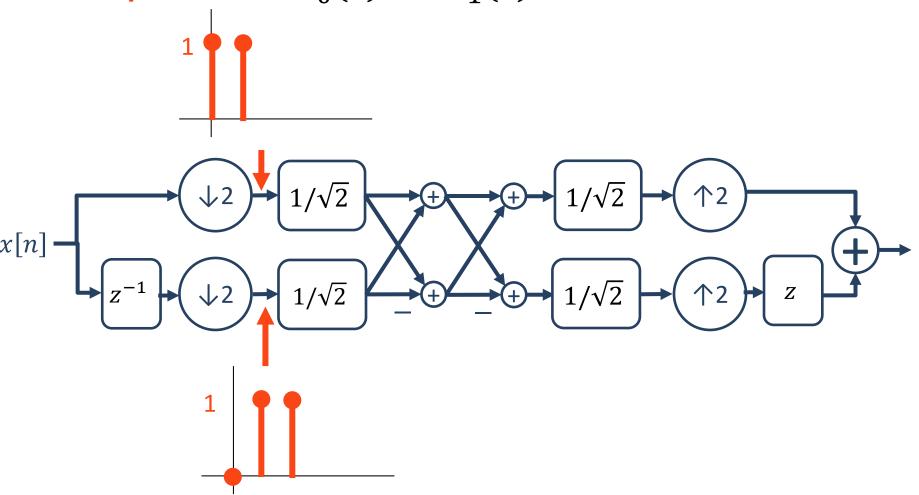
- **Example:** What is $P_0(z)$ and $P_1(z)$ for the filters below?
 - $P_0(z) = 1/\sqrt{2}$
 - $P_1(z) = 1/\sqrt{2}$

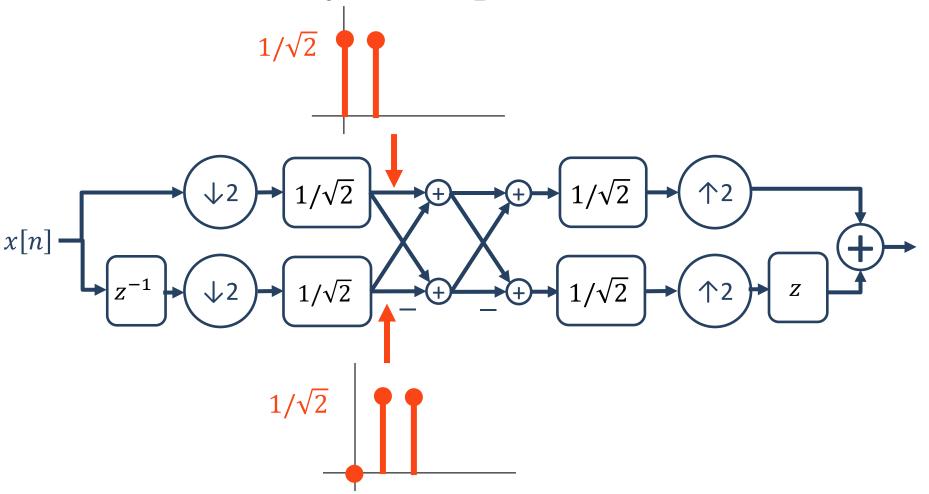


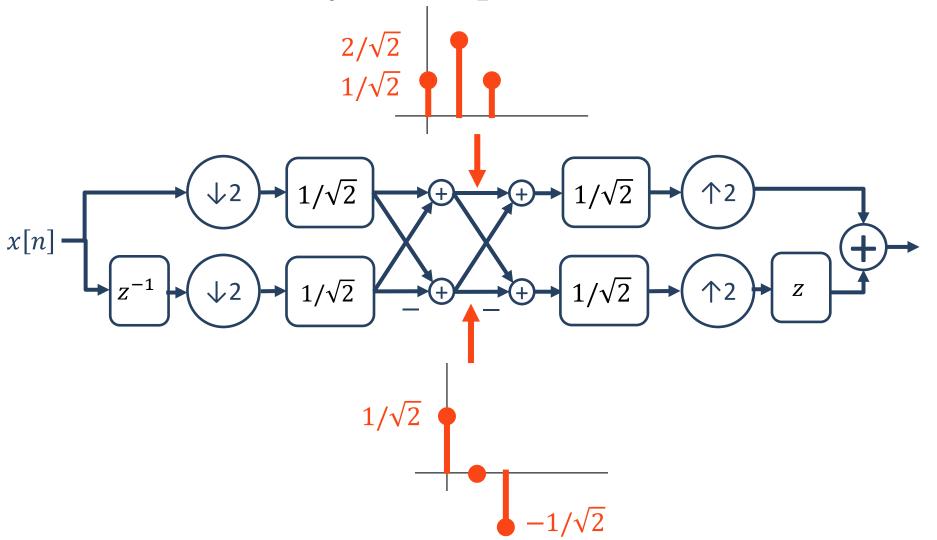


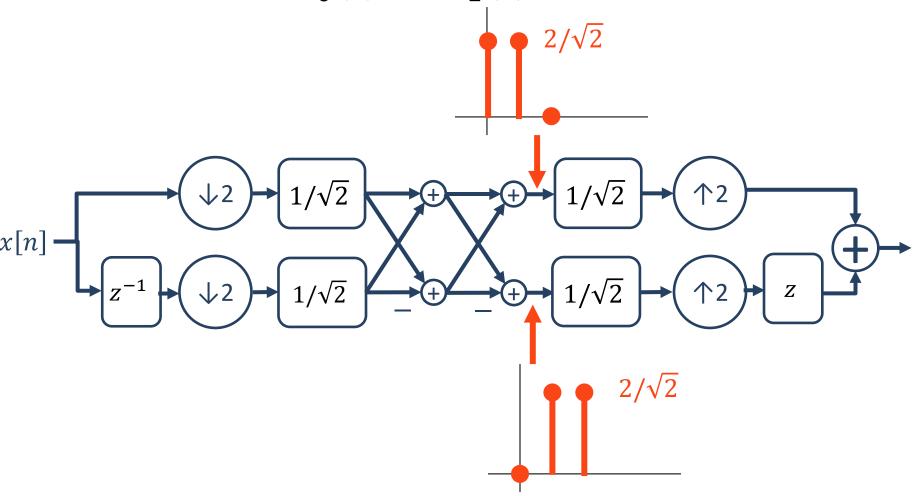


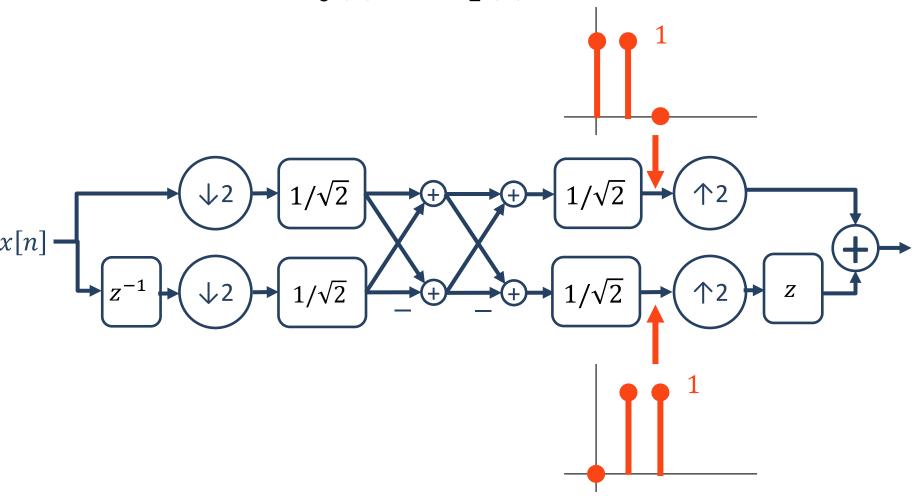


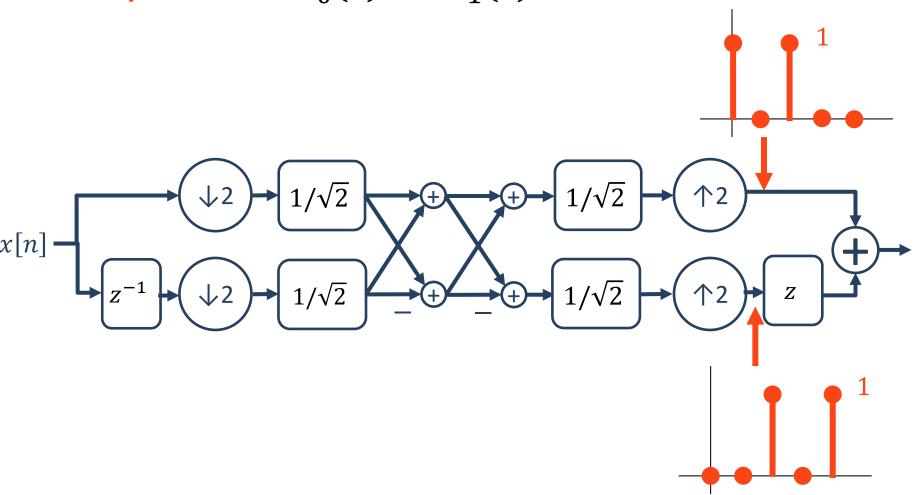


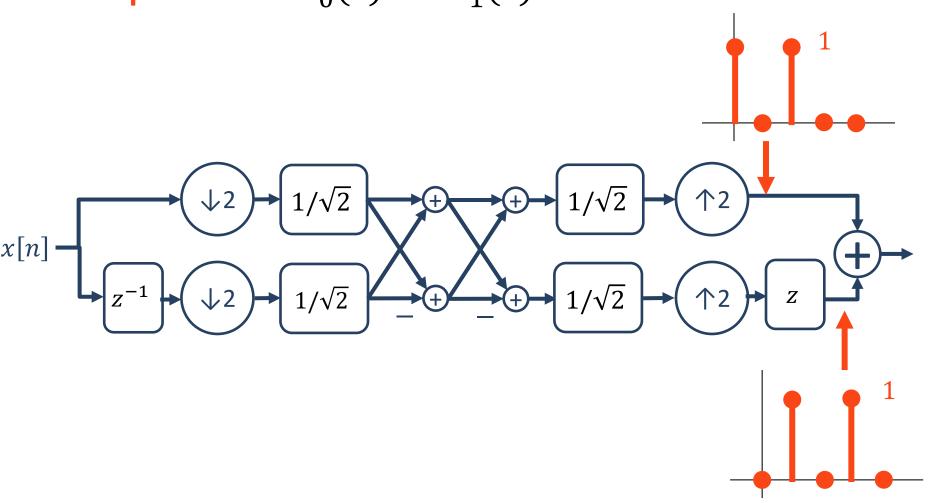


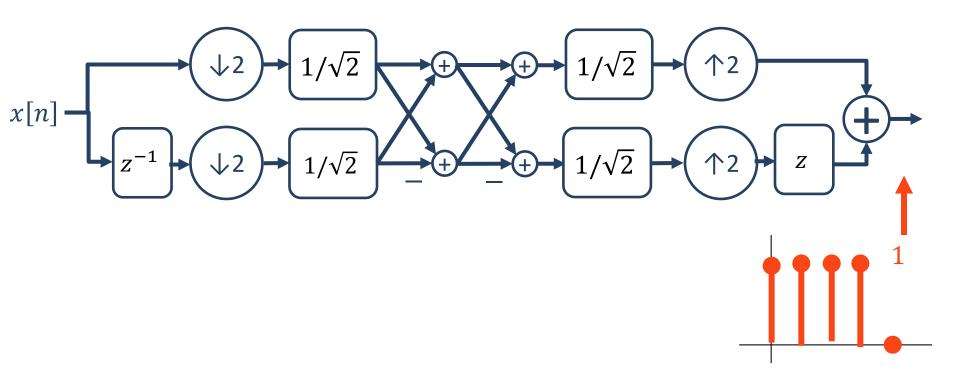












Lecture 26: Filter Banks to Wavelets

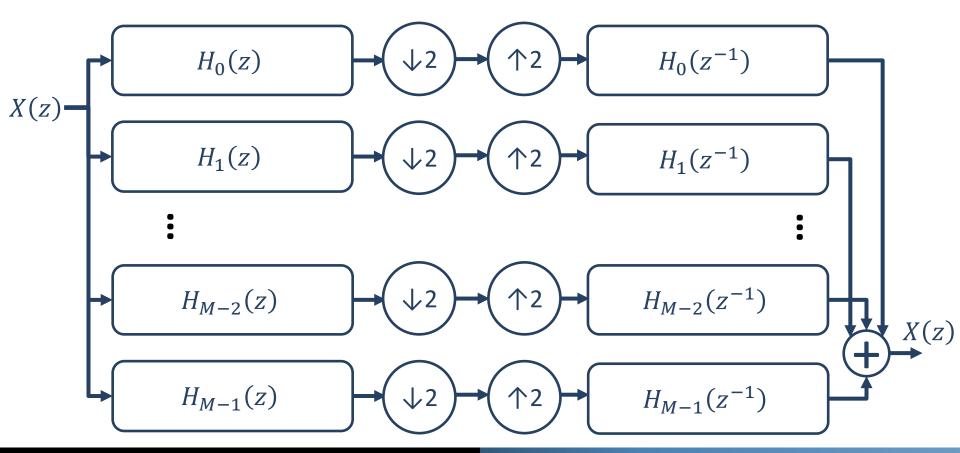
Foundations of Digital Signal Processing

Outline

- DFT Filter Banks [without downsampling]
- DFT Filter Bank [with downsampling]
- General Two-Channel Filter Banks
- Two-Channel Filter Bank Perfect Reconstruction (aliasing canceling)
- Two-Channel Filter Bank Perfect Reconstruction (orthogonal filter bank)
- Polyphase Filters
- Multi-Channel Filter Bank Perfect Reconstruction
- Wavelets

Multi-Channel Filter Banks

Question: Can we generalize perfect reconstruction?

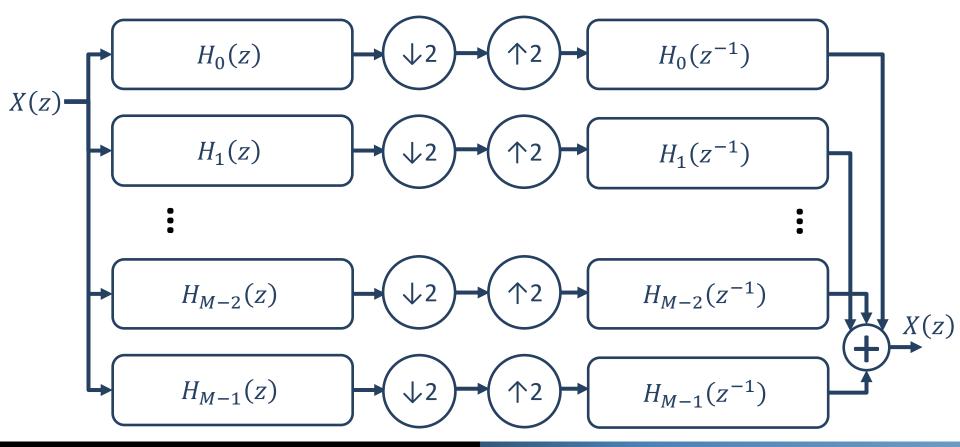


Multi-Channel Filter Banks

Question: Can we generalize perfect reconstruction?

•
$$G_m(z)G_m(z^{-1}) + G_m(-z)G_m(-z^{-1}) = 2$$
 for all m

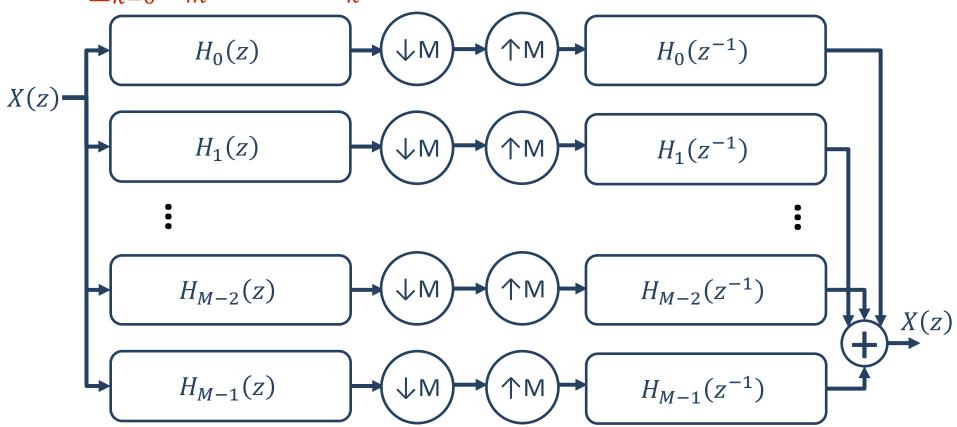
•
$$G_m(z)G_k(z^{-1}) + G_m(-z)G_k(-z^{-1}) = 0$$
 for all m, k



Multi-Channel Filter Banks

Question: Can we generalize perfect reconstruction?



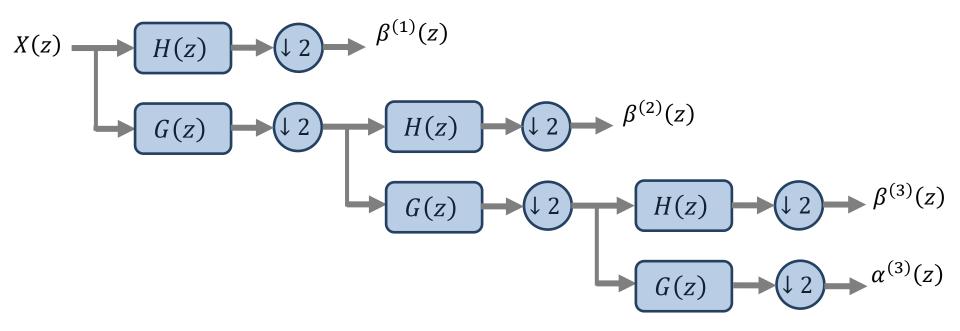


Lecture 26: Filter Banks to Wavelets

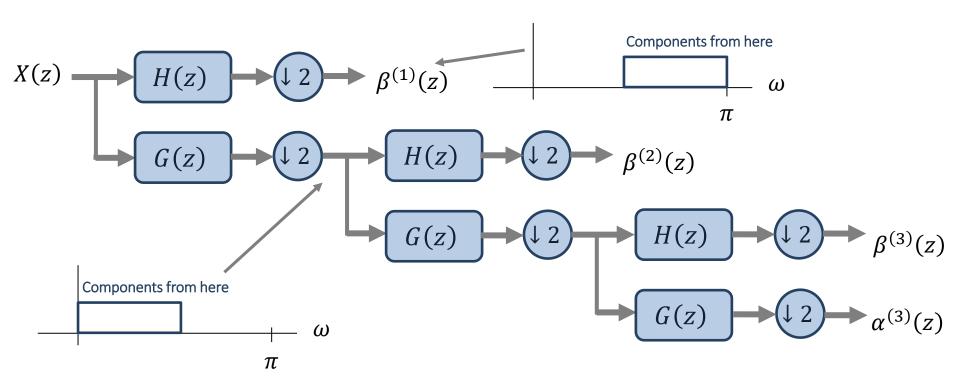
Foundations of Digital Signal Processing

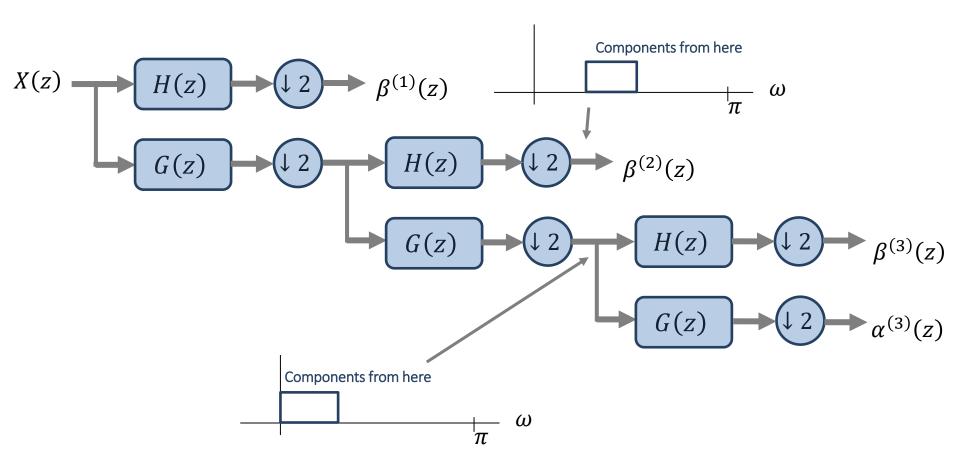
Outline

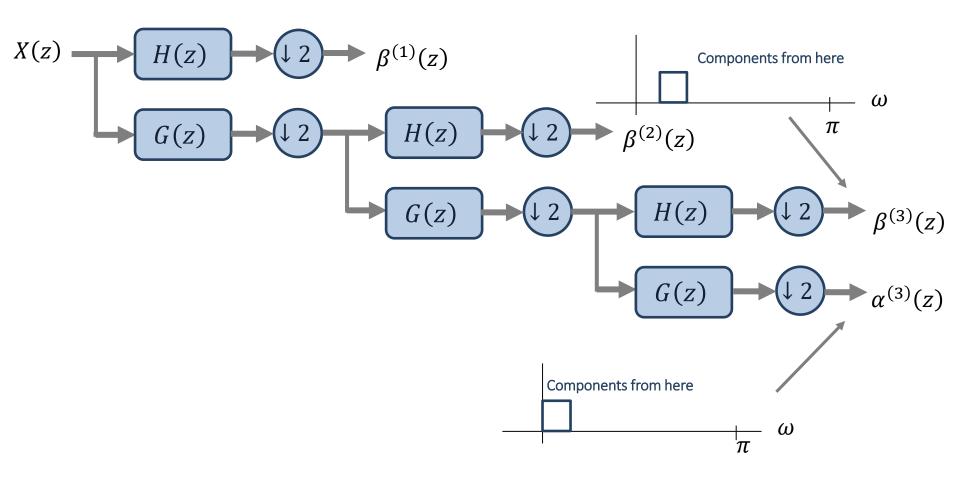
- DFT Filter Banks [without downsampling]
- DFT Filter Bank [with downsampling]
- General Two-Channel Filter Banks
- Two-Channel Filter Bank Perfect Reconstruction (aliasing canceling)
- Two-Channel Filter Bank Perfect Reconstruction (orthogonal filter bank)
- Polyphase Filters
- Multi-Channel Filter Bank Perfect Reconstruction
- Wavelets

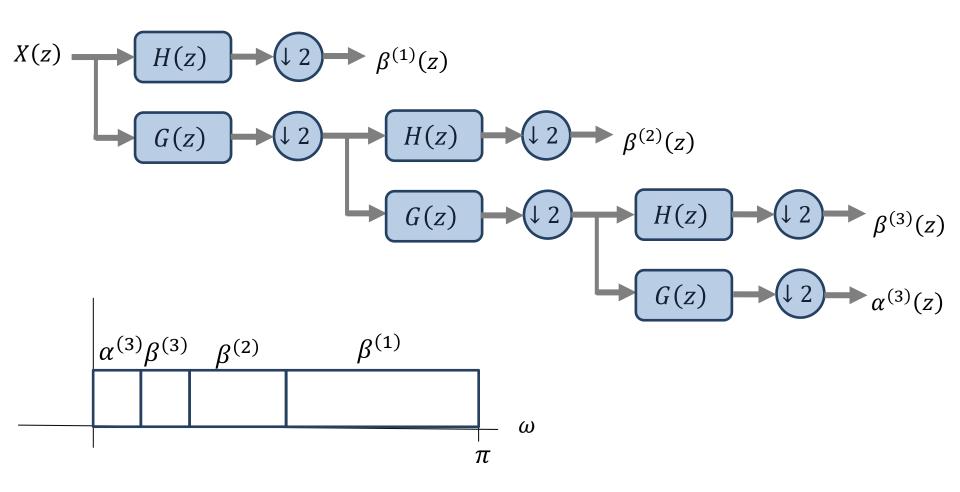


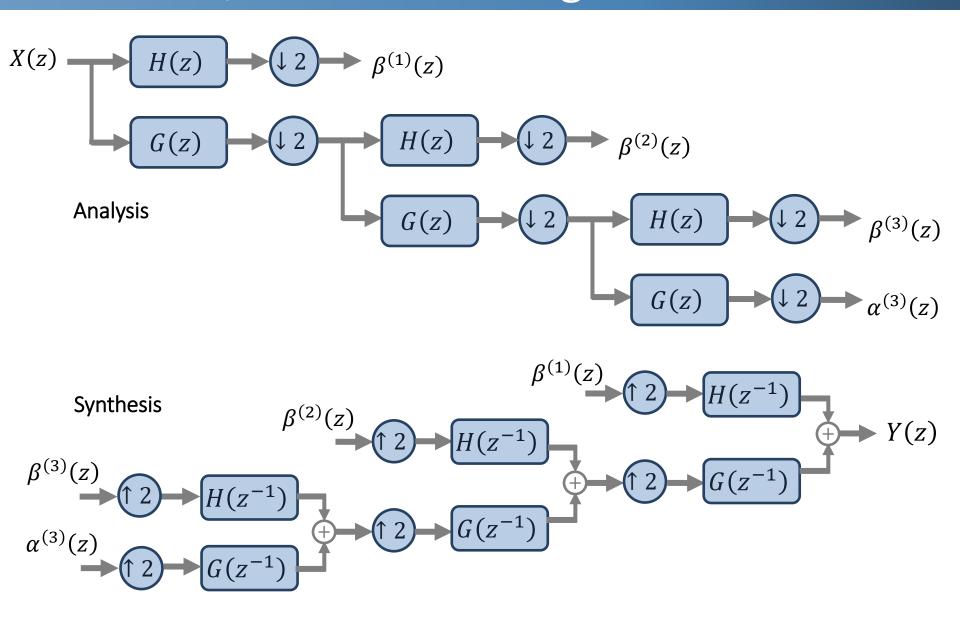
- **Assume** H^* is a half-band high pass filter
- **Assume** G^* is a half-band low pass filter







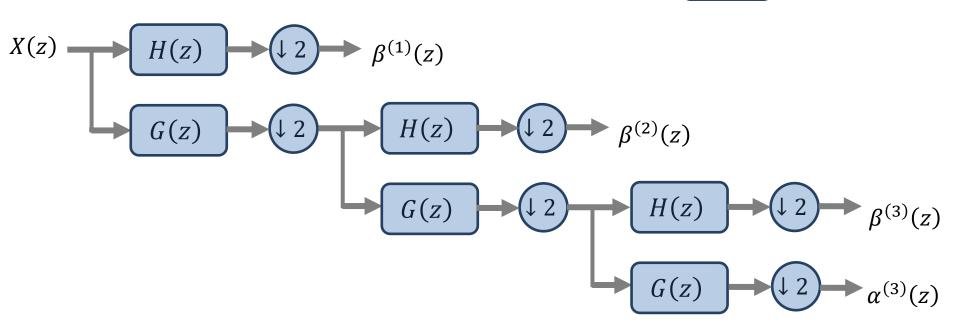




$$x[n] \xrightarrow{\downarrow 2} G(z) \xrightarrow{} y[n]$$

$$=$$

$$x[n] \xrightarrow{\downarrow G(z^2)} \downarrow 2 \xrightarrow{} y[n]$$



$$x[n] \xrightarrow{\downarrow 2} G(z) \xrightarrow{} y[n]$$

$$=$$

$$x[n] \xrightarrow{\downarrow G(z^2)} \downarrow 2 \xrightarrow{} y[n]$$

