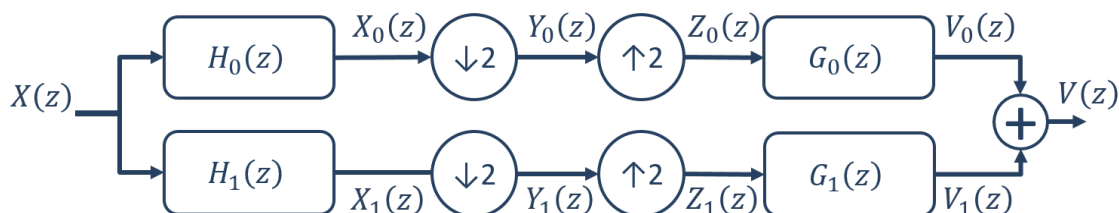


Question #1: (2 pts) How many hours did you spend on this homework?

Question #2: (12 pts) Consider the following 2-channel filter bank shown below.



Let the filters be defined by

$$H_0(z) = \frac{1}{\sqrt{2}} (1 + z^{-1}) \quad G_0(z) = \frac{1}{\sqrt{2}} (z^{+1} + 1)$$

$$H_1(z) = \frac{1}{\sqrt{2}} (1 - z^{-1}) \quad G_1(z) = \frac{1}{\sqrt{2}} (-z^{+1} + 1)$$

- Do the filters satisfy the alias canceling filter bank conditions? Show that they do or do not.
- Change $G_0(z)$ and $G_1(z)$ to be causal: $G_0(z) = \frac{1}{\sqrt{2}} (1 + z^{-1})$ and $G_1(z) = \frac{1}{\sqrt{2}} (-1 + z^{-1})$. How does the right-hand side of the alias canceling filter bank conditions change under this condition?
- Sketch all of the intermediate signals ($X_0(z)$, $X_1(z)$, $Y_0(z)$, $Y_1(z)$, $Z_0(z)$, $Z_1(z)$, $V_0(z)$, $V_1(z)$, $V(z)$) in the time domain for excitation $X(z) = 1 + z^{-1} - z^{-2} - z^{-3}$.

Question #3: (8 pts) Consider a 2-channel filter bank shown in Question #2.

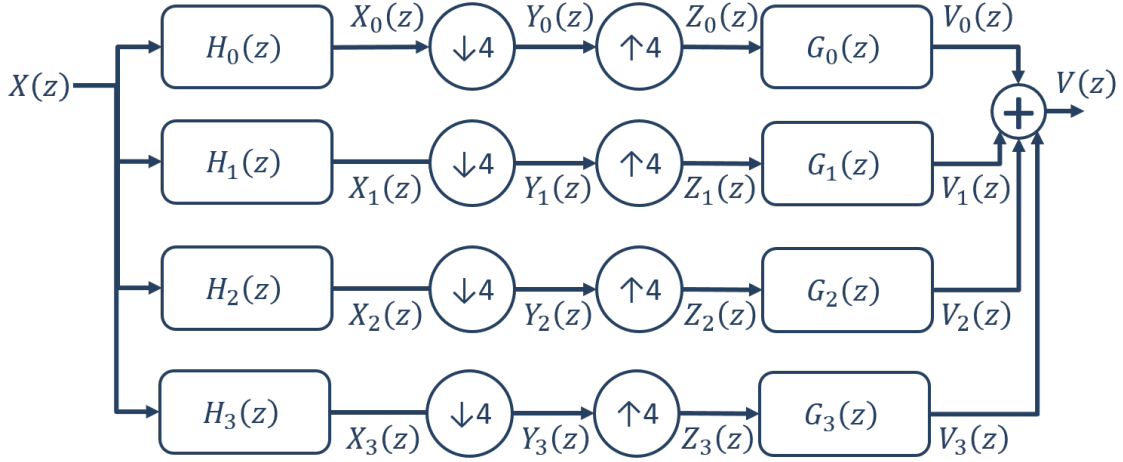
Let the filters be defined by

$$H_0(z) = \frac{1}{2} (1 + z^{-1} - z^{-2} - z^{-3}) \quad G_0(z) = \frac{1}{2} (-z^{+3} - z^{+2} + z^{+1} + 1)$$

$$H_1(z) = \frac{1}{2} (1 - z^{-1} - z^{-2} + z^{-3}) \quad G_1(z) = \frac{1}{2} (z^{+3} - z^{+2} - z^{+1} + 1)$$

- Do the filters satisfy the orthogonal filter bank conditions? Show that they do or do not.
- Sketch all of the intermediate signals ($X_0(z)$, $X_1(z)$, $Y_0(z)$, $Y_1(z)$, $Z_0(z)$, $Z_1(z)$, $V_0(z)$, $V_1(z)$, $V(z)$) in the time domain for excitation $X(z) = 1 + z^{-1} - z^{-2} - z^{-3}$.

Question #4: (8 pts) Consider the following 4-channel filter bank shown below.



Let the filters be defined by

$$H_0(\omega) = G_0(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi/4 - 2\pi k) - u(\omega - \pi/4 - 2\pi k)$$

$$H_1(\omega) = G_1(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi/2 - 2\pi k) - u(\omega + \pi/4 - 2\pi k) + u(\omega - \pi/4 - 2\pi k) - u(\omega - \pi/2 - 2\pi k)$$

$$H_2(\omega) = G_2(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + 3\pi/4 - 2\pi k) - u(\omega + \pi/2 - 2\pi k) + u(\omega - \pi/2 - 2\pi k) - u(\omega - 3\pi/4 - 2\pi k)$$

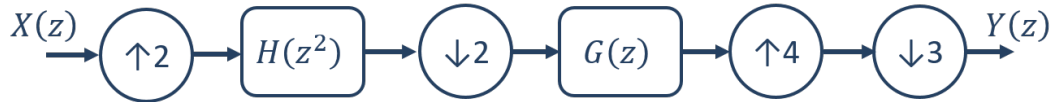
$$H_3(\omega) = G_3(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi - 2\pi k) - u(\omega + 3\pi/4 - 2\pi k) + u(\omega - 3\pi/4 - 2\pi k) - u(\omega - \pi - 2\pi k)$$

Sketch all of the intermediate signals ($X_m(z)$, $Y_m(z)$, $Z_m(z)$, $V_m(z)$, $V(z)$ for $0 \leq m \leq 3$) in the frequency domain for excitation $X(z) = 1$.

Question #5: (8 pts) *Noble Properties*

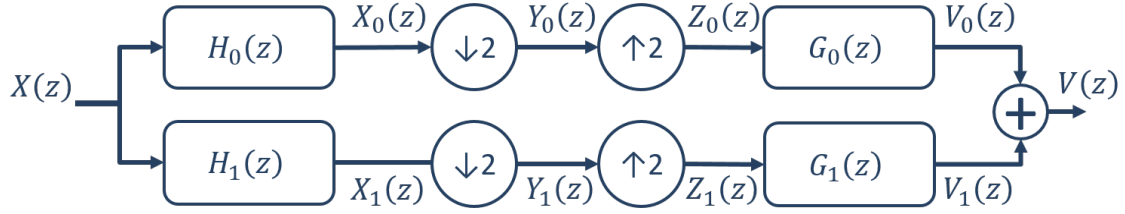
One of the useful aspects of Noble properties is that they can help simplify complex expressions with downsampling and upsampling.

- (a) Use the Noble properties for upsampling and downsampling to simplify the following block diagram. Represent the results as a block diagram. It should only have one downsampling operation, one upsampling operation, and one filter operation



- (b) Let $H(\omega) = G(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi/2 - 2\pi k) - u(\omega - \pi/2 - 2\pi k)$. Plot the magnitude response of the resulting filter in your block diagram.

Question #6: (8 pts) Consider the following 2-channel bank.



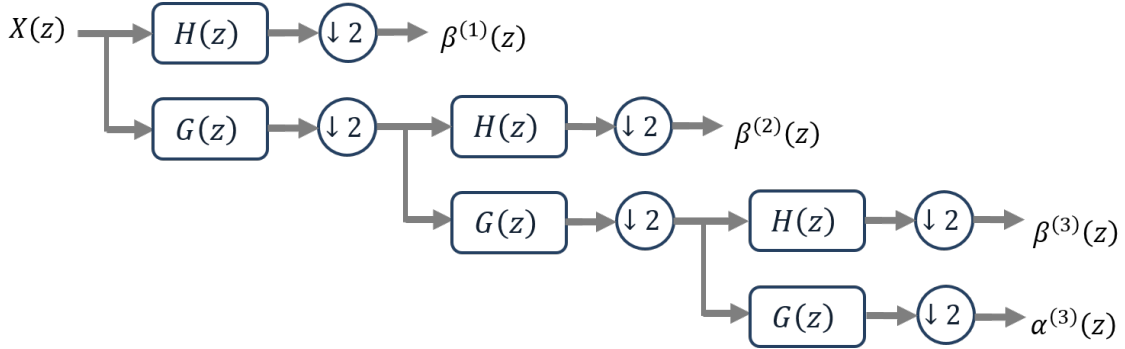
Let the filters be defined by

$$H_0(z) = \frac{1}{2} (1 + z^{-1} - z^{-2} - z^{-3}) \quad H_1(z) = \frac{1}{2} (-z^{+1} - z^{+2} + z^{+1} + 1)$$

$$G_0(z) = \frac{1}{2} (1 - z^{-1} - z^{-2} + z^{-3}) \quad G_1(z) = \frac{1}{2} (z^{+3} - z^{+2} - z^{+1} + 1)$$

Determine the equivalent polyphase filter implementation for this filter bank. Stretch the impulse response of the filter coefficients for each polyphase filter.

Question #7: (12 pts) Consider the following wavelet bank.



Let the high pass filter $H(z)$ and low pass filter $G(z)$ be defined by:

$$G(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi/2 - 2\pi k) - u(\omega - \pi/2 - 2\pi k)$$

$$H(\omega) = \sum_{k=-\infty}^{\infty} u(\omega + \pi/2 - \pi - 2\pi k) - u(\omega - \pi/2 - \pi - 2\pi k)$$

Sketch the magnitude responses of $|\beta^{(1)}(\omega)|, |\beta^{(2)}(\omega)|, |\beta^{(3)}(\omega)|, |\alpha^{(3)}(\omega)|$ when $X(\omega) = 1$