Lecture 16: Filter Implementation... Again.

Foundations of Digital Signal Processing

Outline

- Reviewing different types of filters
- Designing the phase response
- Implementation of FIR Filters
- Implementation of IIR Filters

News

Homework #7

- Due <u>this Thursday</u>
- Submit via canvas

Coding Problem #4

- Due <u>this Thursday</u>
- Submit via canvas

In two weeks

Exam #2 (yay!)

News

Next week

- Guest lectures
- Tuesday: Resampling
- Thursday: Review for Exam #2

Practice Exam

Will come out early next week

Lecture 16: Filter Implementation... Again.

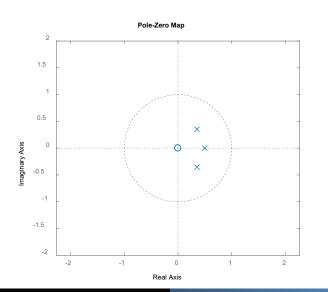
Foundations of Digital Signal Processing

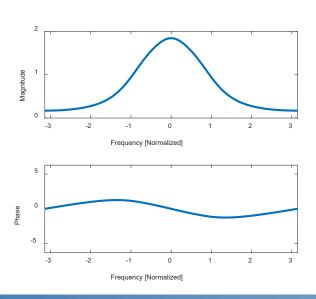
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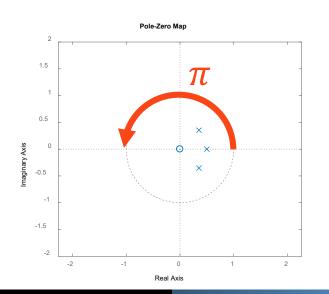
- Question: What types of filters are there?
 - Low pass
 - High pass
 - Band pass
 - All pass

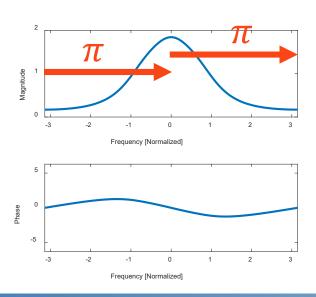
- Question: How do I turn a low pass filter into other filters?
 - Low pass (prototype filter) $H(\omega)$
 - High pass
 - Band pass
 - All pass



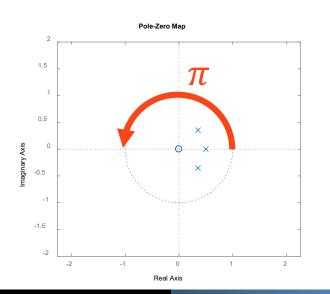


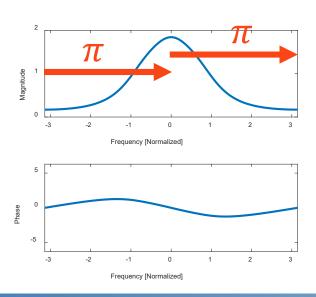
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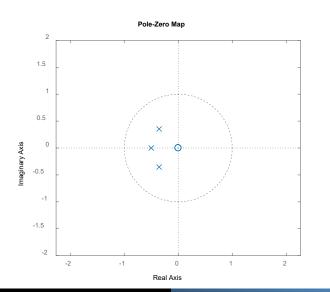


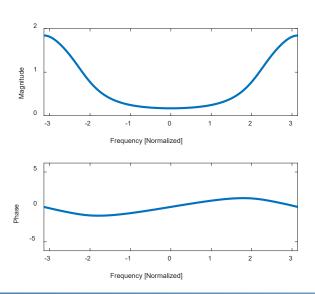
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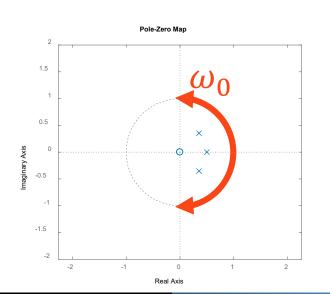
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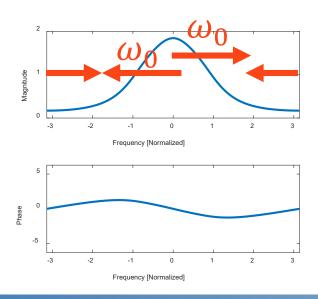




Question: How do I make a notch filter?

- Low pass (prototype filter) $H(\omega)$
- High pass $H(\omega \pi)$
- Band pass $H(\omega \omega_0) + H(\omega + \omega_0)$
- All pass



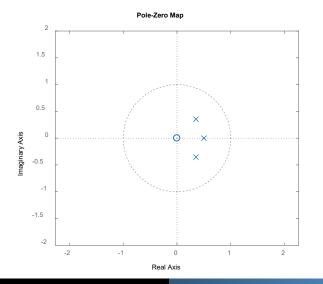


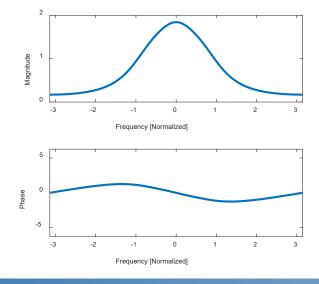
- Question: How do I make an all pass filter?
 - Low pass (prototype filter) $H(\omega)$
 - High pass $H(\omega \pi)$
 - Band pass $H(\omega \omega_0) + H(\omega + \omega_0)$

• All pass $H(\omega) = z^{-N} \frac{A(z^{-1})}{A(z)}$

Flip poles / zeros around the unit circle

Swap Poles and Zeros



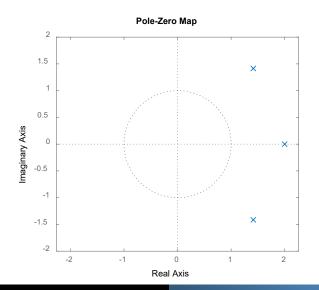


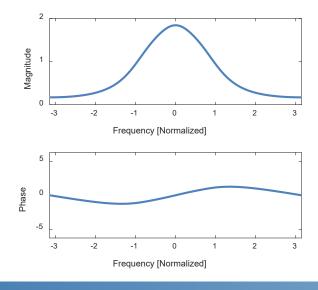
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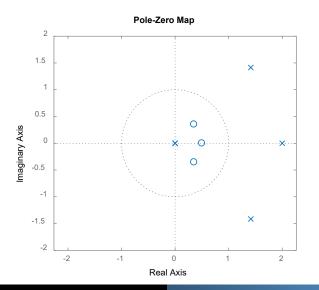


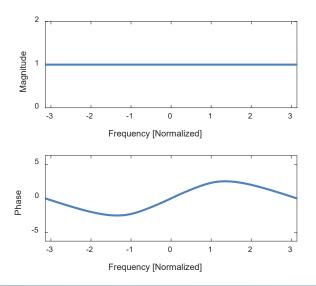
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Swap Poles and Zeros





- Question: What are these operations in time?
 - Low pass (prototype filter)

$$\diamond H(\omega)$$

High pass

$$\diamond H(\omega - \pi)$$

Band pass

$$\diamond H(\omega - \omega_0) + H(\omega + \omega_0)$$

All pass

$$\diamond H(\omega) = z^{-N} \frac{A(z^{-1})}{A(z)}$$

- Question: What are these operations in time?
 - Low pass (prototype filter)
 - $\diamond H(\omega)$
 - High pass
 - $\diamond H(\omega \pi)$
 - $\diamond h[n]\cos(\pi n)$
 - Band pass

$$\diamond H(\omega - \omega_0) + H(\omega + \omega_0)$$

- $\diamond 2 h[n] \cos(\omega_0 n)$
- All pass

$$\diamond H(\omega) = z^{-N} \frac{A(z^{-1})}{A(z)}$$

<u>Time-reversal</u>

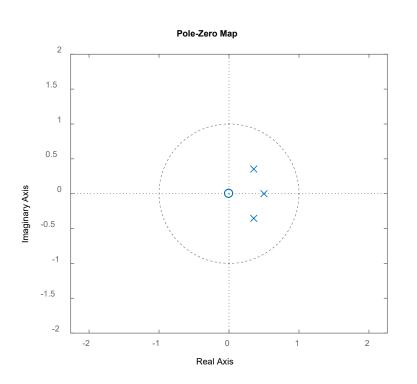
<u>De-convolution (inverse</u> convolution)

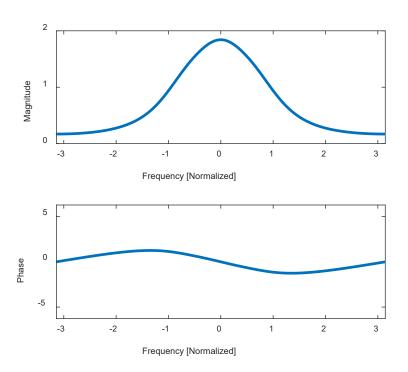
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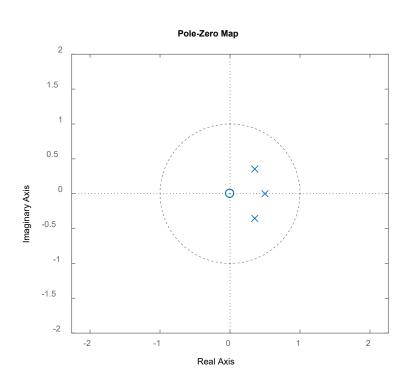
- Answer: Inverse
- $\blacksquare Y(\omega) = H(\omega)X(\omega)$
- $X(\omega) = \frac{Y(\omega)}{H(\omega)}$

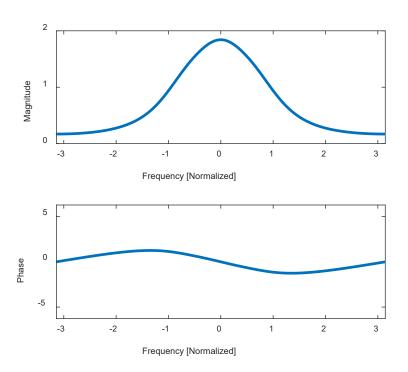
Question: If all <u>poles</u> are inside the unit circle, what does this mean?



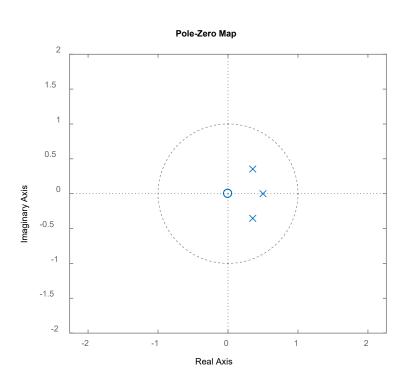


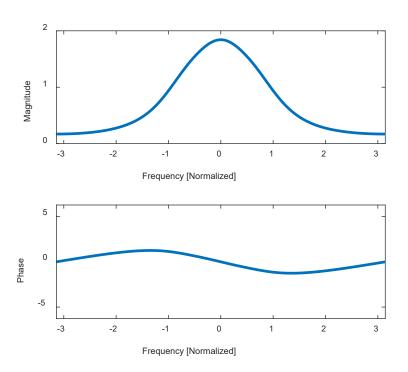
Question: If all <u>poles</u> are inside the unit circle, what does this mean? System is stable



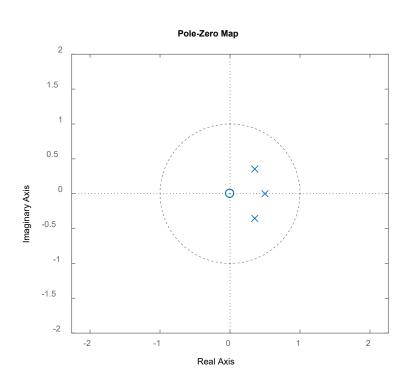


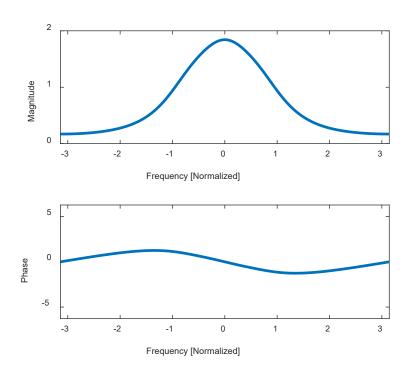
Question: If all <u>zeros</u> are inside the unit circle, what does this mean?



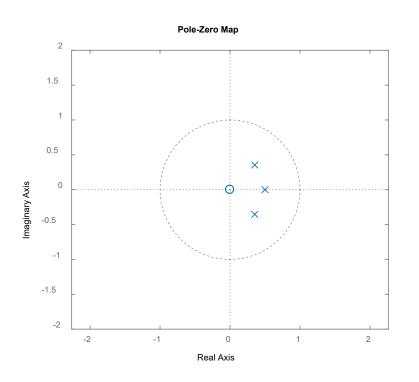


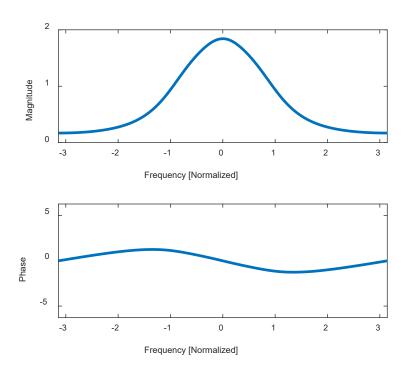
Question: If all <u>zeros</u> are inside the unit circle, what does this mean? System inverse is stable



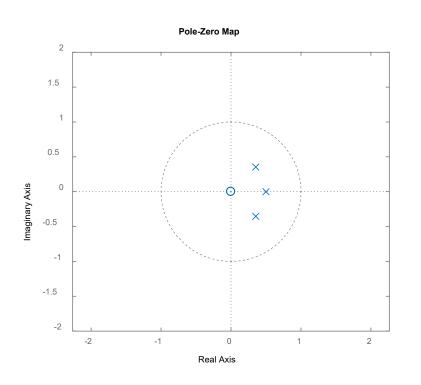


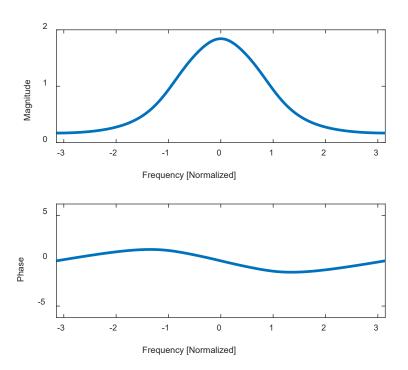
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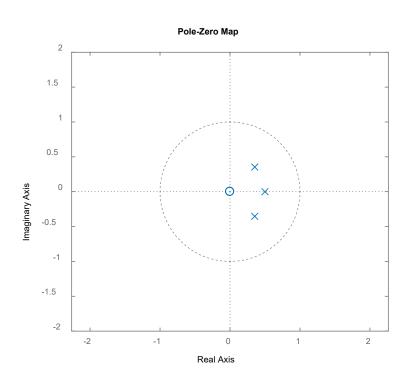


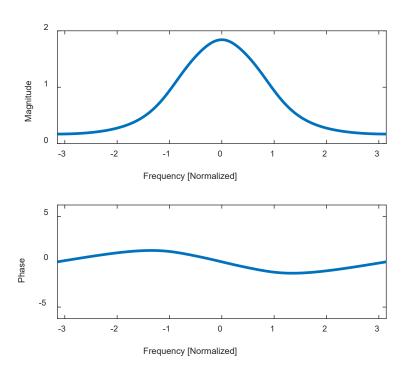
Question: If all poles and zeros are inside the unit circle, what does this mean? System and its inverse are stable



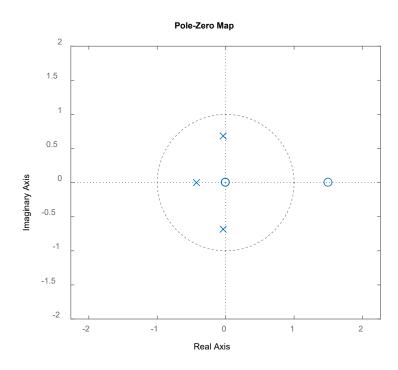


- Question: If all poles and zeros are inside the unit circle, what does this mean?
- We call these types of filters minimum phase filters.

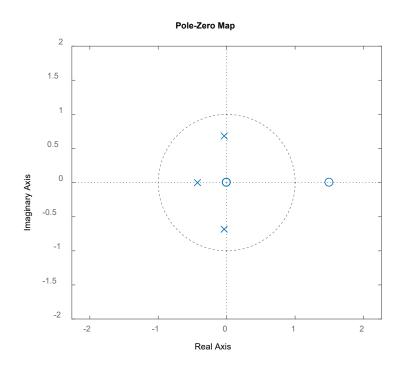




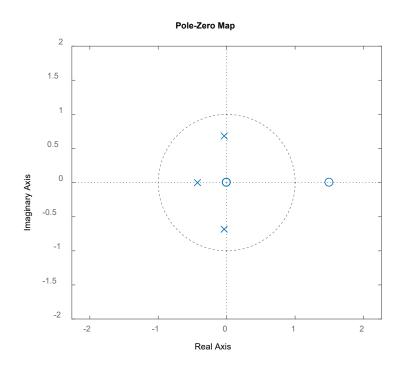
- Example: Is the filter defined by the following pole-zero plot...
 - A low pass, high pass, or band pass filter?
 - Stable?
 - Have a Stable Inverse?



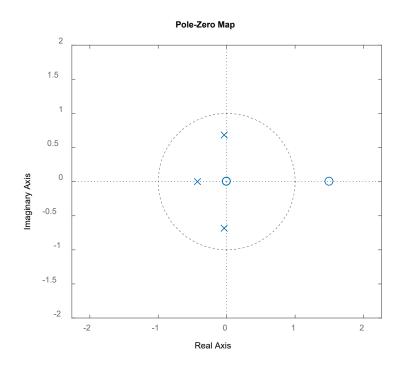
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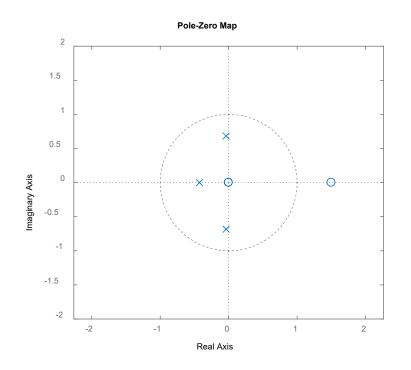
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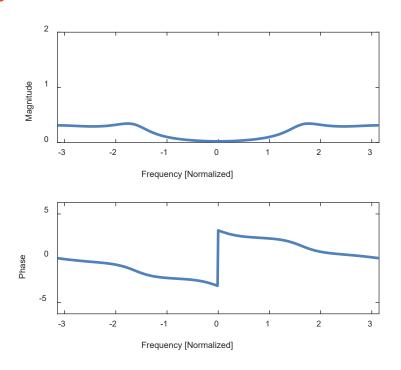


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Question: Are there other types of filters?

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Question: What is the frequency-domain phase if the impulse response is <u>even</u>?

$$h[n] = h[-n]$$

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$$h[n] = h[-n]$$

Result: There is no phase.

$$H(\omega) = H^*(-\omega)$$

$$H(\omega) = |H(\omega)|e^{j\angle H(\omega)} = |H(\omega)|$$

Question: What is the frequency-domain phase if the <u>even</u> impulse response is shifted?

$$h[n] = h[-n]$$
$$h[n - n_0]$$

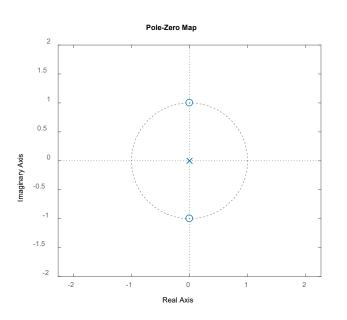
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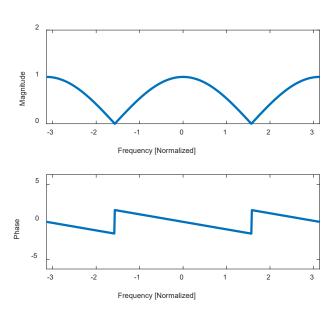
$$h[n] = h[-n]$$
$$h[n - n_0]$$

Result: There is a linear phase.

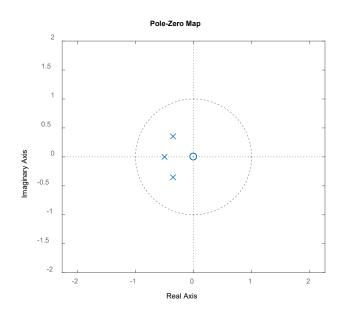
$$H(\omega) = |H(\omega)|e^{-j\omega n_0}$$

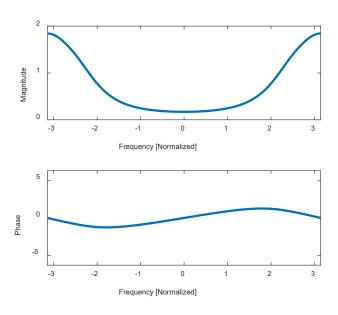
- Linear phase -> Delay
 - $H(\omega) = |H(\omega)|e^{-j\omega n_0}$ <- slope = delay



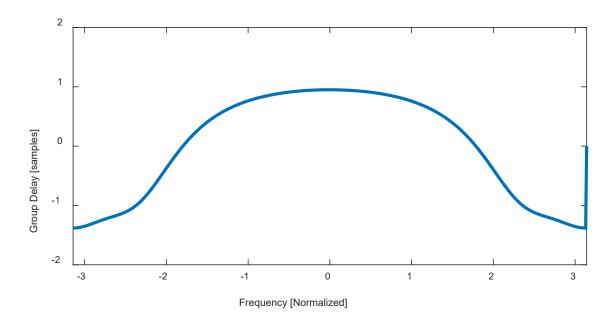


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- More generally
 - $\frac{d \angle H(\omega)}{d\omega}$ = group delay (delay at each frequency)





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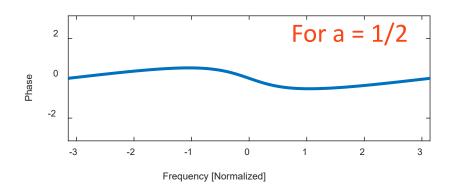
$$\angle H(\omega) = \angle (1) - \angle \left(1 - e^{-j\omega}\right)$$

$$\angle H(\omega) = 0 - \angle \left(1 - a(\cos(\omega) - j\sin(\omega))\right)$$

$$\angle H(\omega) = -\arctan\left(\frac{a\sin(\omega)}{1 - a\cos(\omega)}\right)$$

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Example: Compute the phase response of

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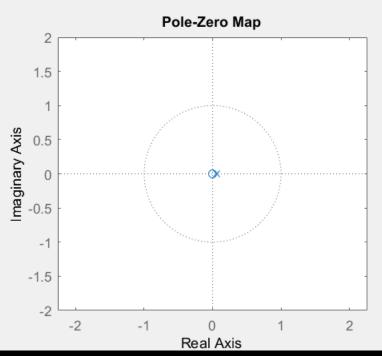
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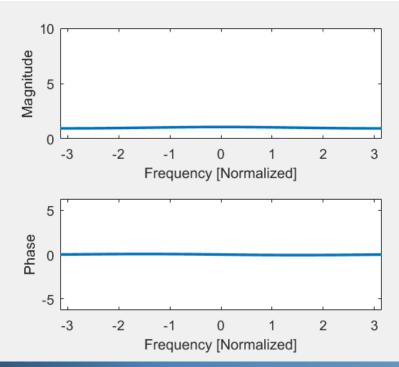
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$$\angle H(\omega) = \angle \left[\frac{1}{(1 - ae^{-j\omega})} \right] + \angle \left[\frac{1}{(1 - be^{-j\omega})} \right]$$

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Question: What happens as a, b gets large?

Example: Compute the phase response of

$$H(\omega) = \frac{1}{(1 - ae^{-j\omega})(1 - be^{-j\omega})}$$

$$\angle H(\omega) = -\operatorname{atan}\left(\frac{a\sin(\omega)}{1 - a\cos(\omega)}\right) - \operatorname{atan}\left(\frac{b\sin(\omega)}{1 - b\cos(\omega)}\right) \to 2\omega$$

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Question: Assume I have 46 poles. What is the minimum possible delay? What is the maximum possible delay?

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Answer:

- Minimum delay = 0
- Maximum delay = 46

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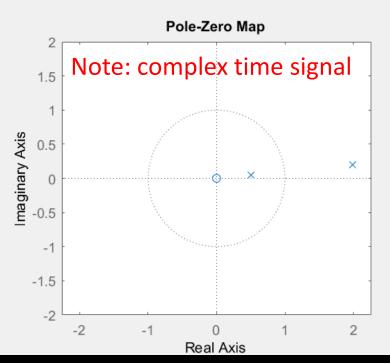
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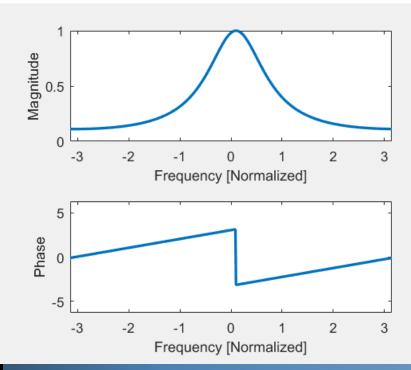
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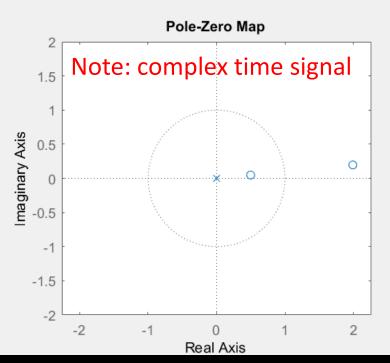
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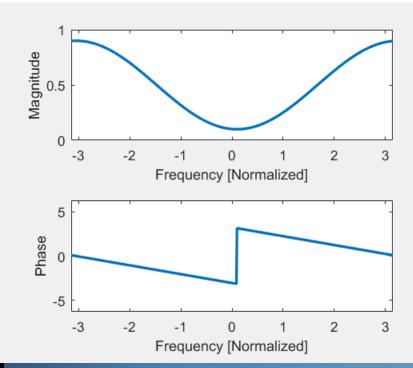
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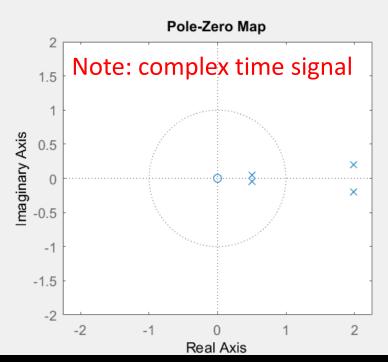


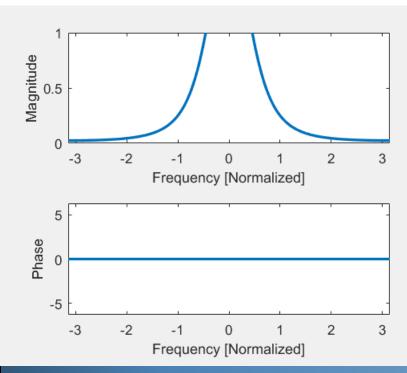
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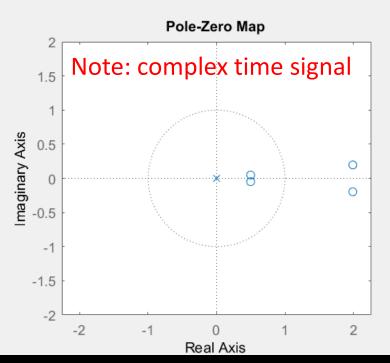


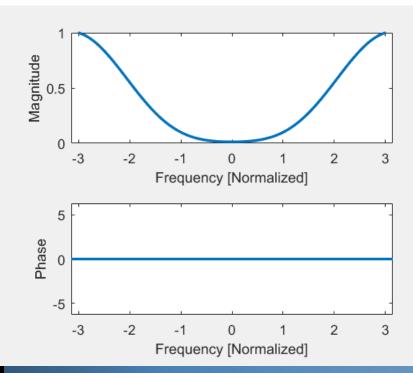
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Question: So why do I care about phase?

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- Implementation of FIR Filters
- Implementation of IIR Filters

Question: What do we mean by implementation? Why do we care?

- Two ways to look at FIR filters
 - Convolution perspective

$$y[n] = \sum_{m=0}^{M-1} h[m]x[n-m]$$

Filtering perspective

$$Y(Z) = X(z) \sum_{m=0}^{M-1} h[m] z^{-m}$$

FIR Direct Form

$$y[n] = \sum_{m=0}^{M-1} h[m]x[n-m]$$

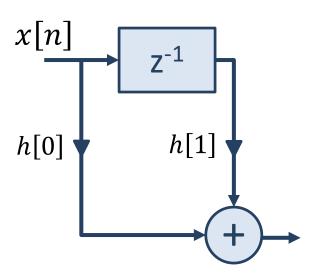
Question: What are the defining characteristics of the pole-zero plot?

FIR Direct Form

$$y[n] = \sum_{m=0}^{M-1} h[m]x[n-m]$$

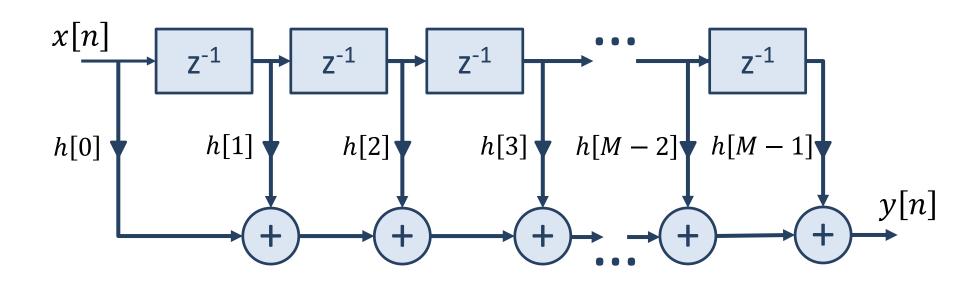
FIR Direct Form (M = 1)

$$y[n] = h[0] + h[1]x[n-1]$$

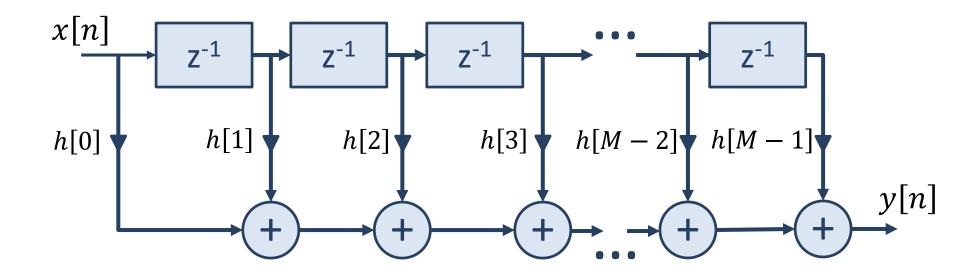


FIR Direct Form

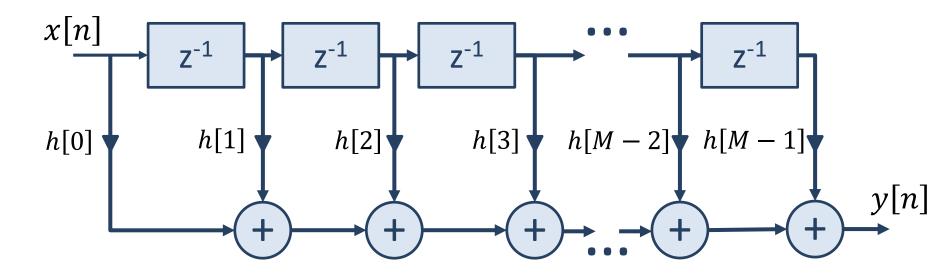
$$y[n] = \sum_{m=0}^{M-1} h[m]x[n-m]$$



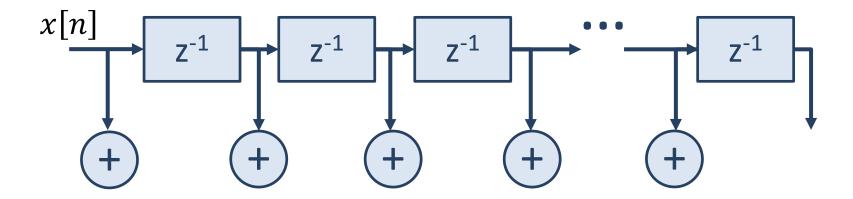
FIR Direct Form



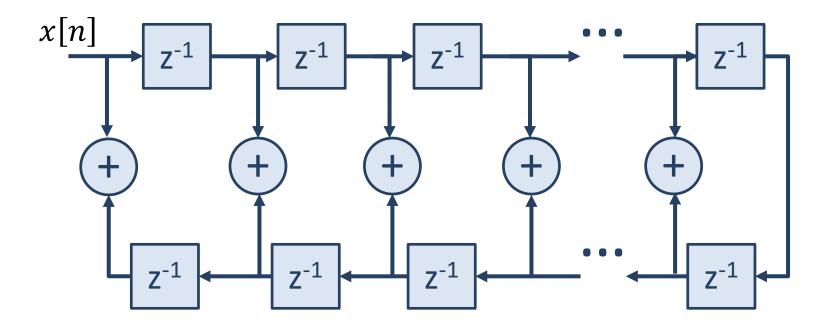
FIR Direct Form



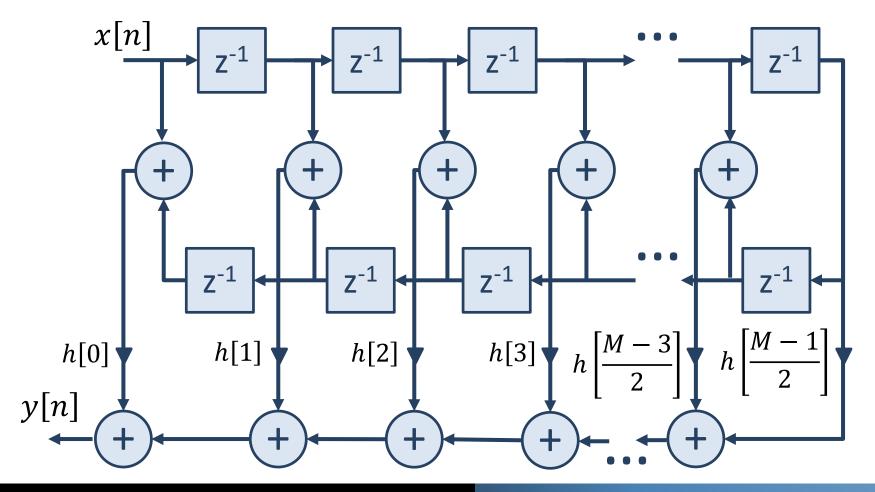
FIR Direct Form



FIR Direct Form

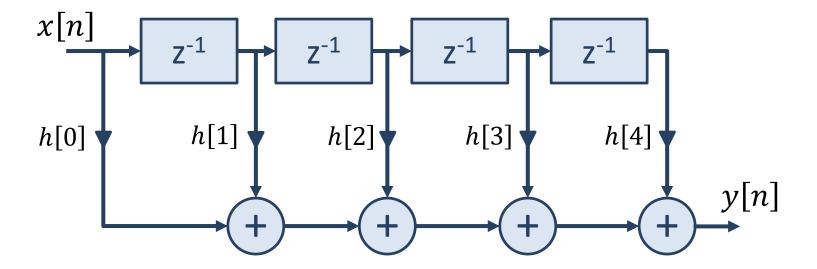


■ FIR Direct Form



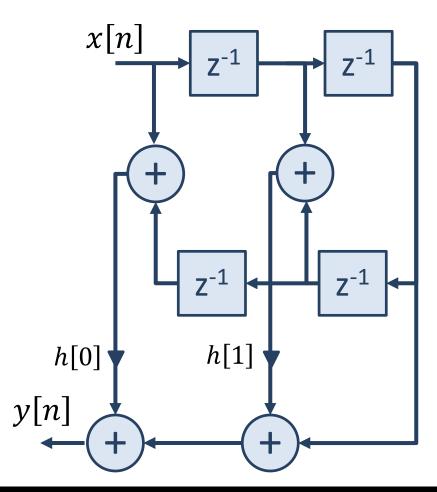
FIR Direct Form

Non-symmetric impulse response (4 multiplications)



FIR Direct Form

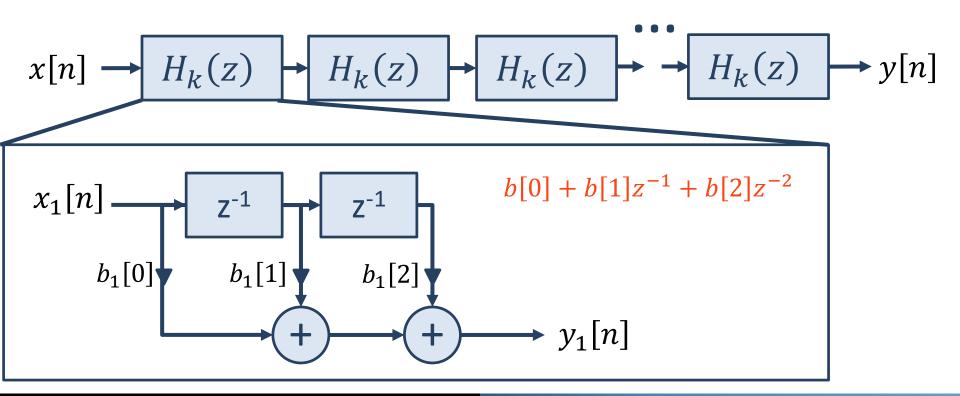
Symmetric impulse response (2 multiplications)



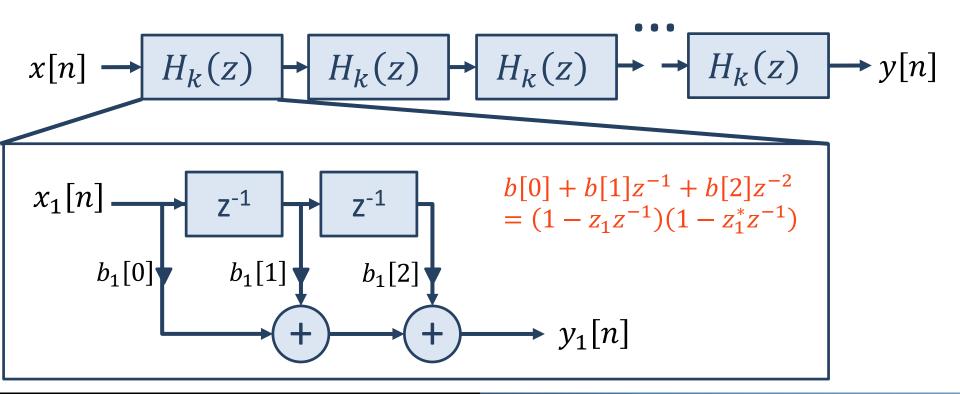
$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$

$$x[n] \longrightarrow H_k(z) \longrightarrow H_k(z) \longrightarrow H_k(z) \longrightarrow y[n]$$

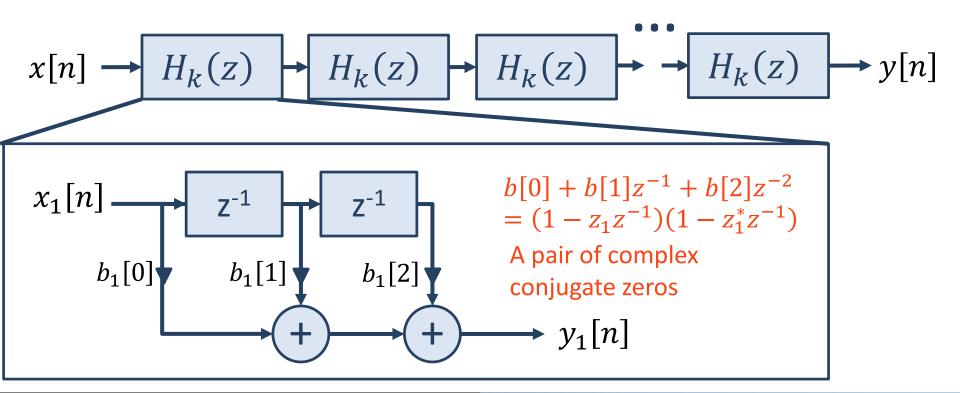
$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



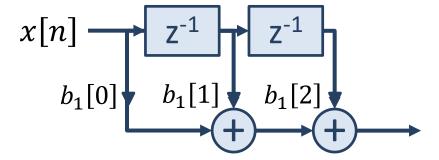
$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



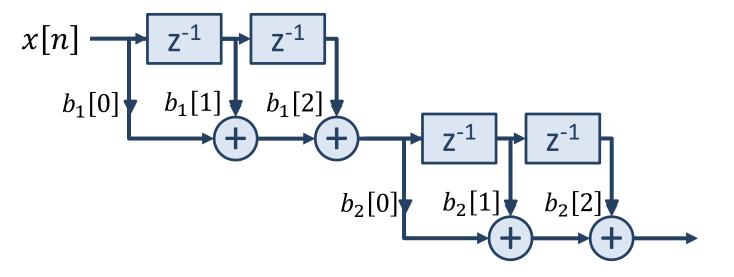
$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



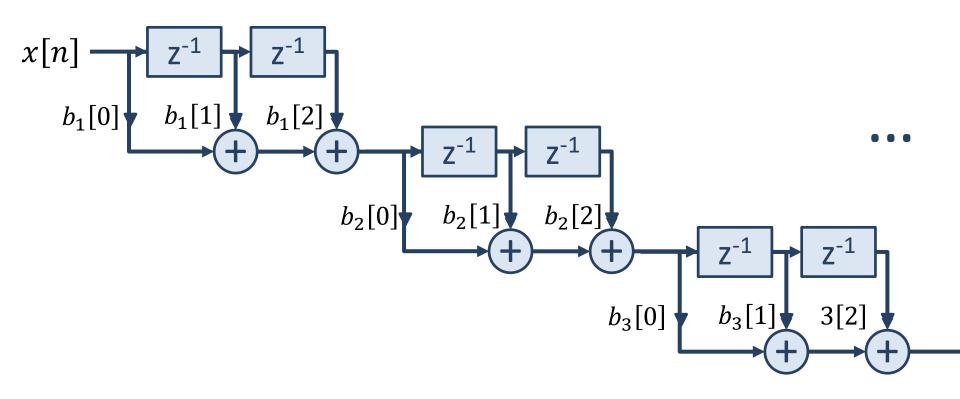
FIR Cascade Form



FIR Cascade Form

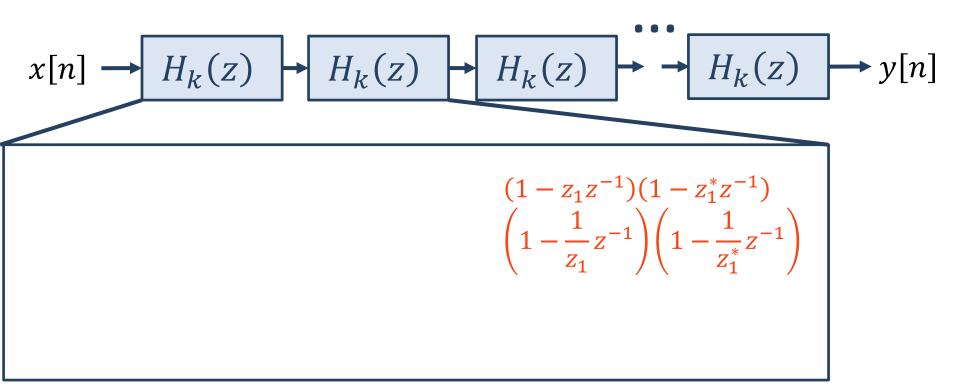


FIR Cascade Form



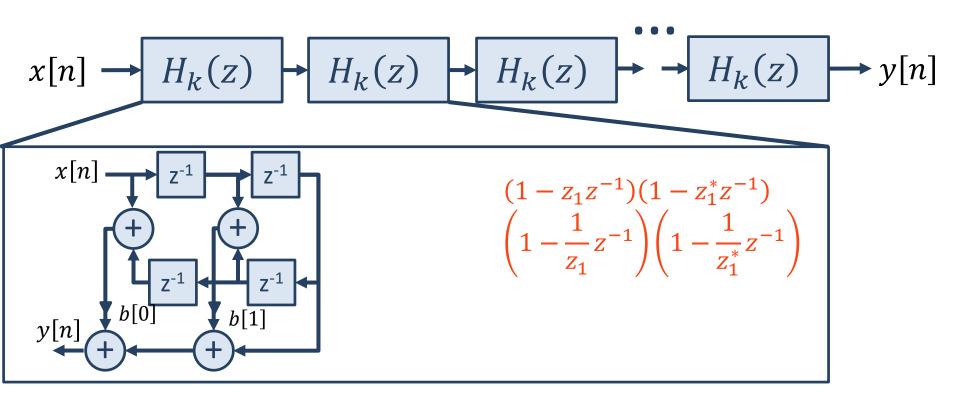
FIR Cascade Form (if the impulse response is symmetric)

$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m]z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$

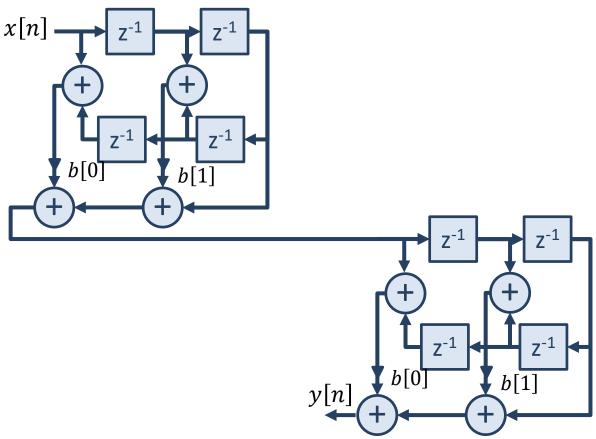


FIR Cascade Form (if the impulse response is symmetric)

$$Y[z] = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



FIR Cascade Form



$$H[k] = \sum_{n=0}^{N-1} h[n]e^{-j\frac{2\pi}{N}kn} \quad , \qquad h[n] = \frac{1}{N} \sum_{k=0}^{N-1} H[k]e^{j\frac{2\pi}{N}kn}$$

$$H[k] = \sum_{n=0}^{N-1} h[n]e^{-j\frac{2\pi}{N}kn} \quad , \qquad h[n] = \frac{1}{N} \sum_{k=0}^{N-1} H[k]e^{j\frac{2\pi}{N}kn}$$

$$H(z) = \sum_{n=0}^{N-1} h[n] z^{-n} = \sum_{n=0}^{N-1} \left(\frac{1}{N} \sum_{k=0}^{N-1} H[k] e^{j\frac{2\pi}{N}kn} \right) z^{-n}$$

$$= \sum_{n=0}^{N-1} \left(\frac{1}{N} \sum_{k=0}^{N-1} H[k] e^{j\frac{2\pi}{N}kn} \right) z^{-n}$$

$$= \sum_{k=0}^{N-1} \left(H[k] \frac{1}{N} \sum_{n=0}^{N-1} e^{j\frac{2\pi}{N}kn} z^{-n} \right)$$

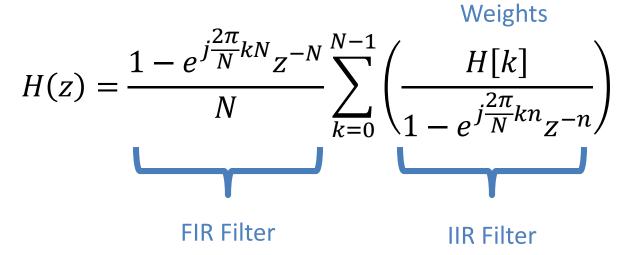
$$H[k] = \sum_{n=0}^{N-1} h[n]e^{-j\frac{2\pi}{N}kn} \quad , \qquad h[n] = \frac{1}{N} \sum_{k=0}^{N-1} H[k]e^{j\frac{2\pi}{N}kn}$$

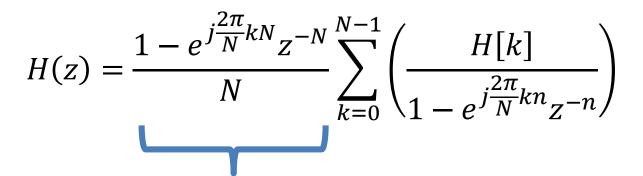
$$H(z) = \sum_{n=0}^{N-1} h[n] z^{-n} = \sum_{n=0}^{N-1} \left(\frac{1}{N} \sum_{k=0}^{N-1} H[k] e^{j\frac{2\pi}{N}kn} \right) z^{-n}$$

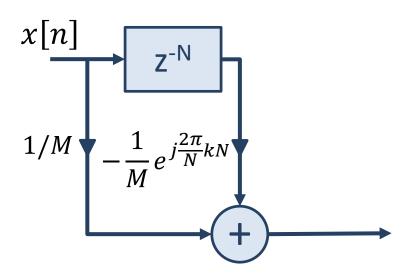
$$= \sum_{k=0}^{N-1} \left(H[k] \frac{1}{N} \frac{1 - e^{j\frac{2\pi}{N}kN} z^{-N}}{1 - e^{j\frac{2\pi}{N}kn} z^{-n}} \right)$$

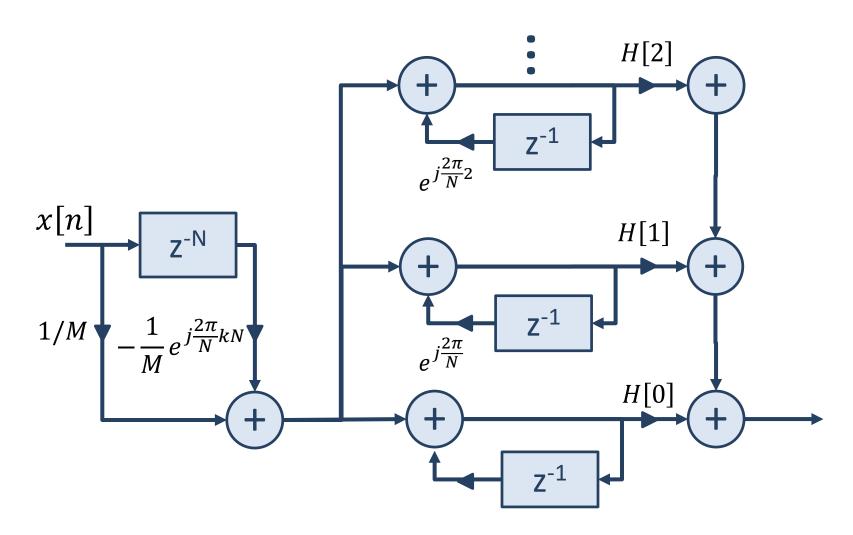
$$= \frac{1 - e^{j\frac{2\pi}{N}kN} z^{-N}}{N} \sum_{k=0}^{N-1} \left(\frac{H[k]}{1 - e^{j\frac{2\pi}{N}kn} z^{-n}} \right)$$

$$H(z) = \frac{1 - e^{j\frac{2\pi}{N}kN}z^{-N}}{N} \sum_{k=0}^{N-1} \left(\frac{H[k]}{1 - e^{j\frac{2\pi}{N}kn}z^{-n}}\right)$$

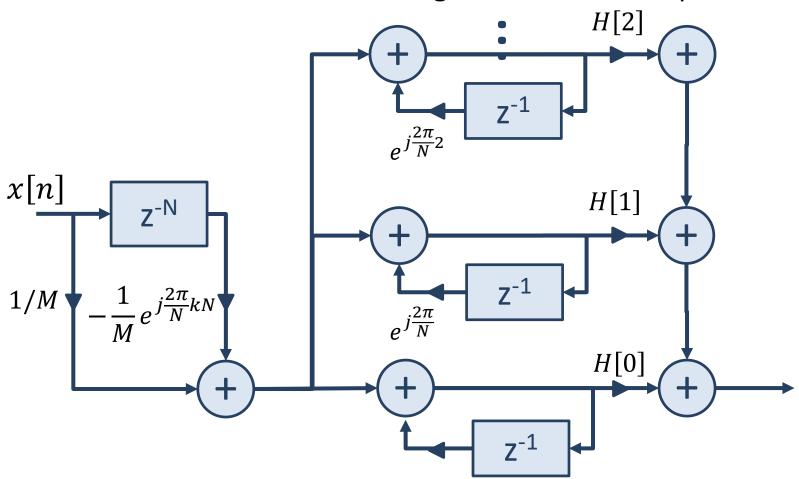








Frequency Sampling Form



Lecture 16: Filter Implementation... Again.

Foundations of Digital Signal Processing

Outline

- Reviewing different types of filters
- Designing the phase response
- Implementation of FIR Filters
- Implementation of IIR Filters

- Two ways to look at IIR filters with only recursive components
 - Convolution perspective

$$y[n] + \sum_{m=1}^{M-1} g[m]y[n-m] = x[n]$$

Filtering perspective

$$Y(z) = \frac{X(z)}{1 + \sum_{m=Q}^{M-1} g[m]z^{-m}}$$

- Two ways to look at IIR filters with only recursive components
 - Convolution perspective

$$y[n] = x[n] - \sum_{m=1}^{M-1} g[m]y[n-m]$$

Filtering perspective

$$Y(z) = \frac{X(z)}{1 + \sum_{m=Q}^{M-1} g[m]z^{-m}}$$

IIR Direct Form

$$Y(Z) = \frac{X(z)}{1 + \sum_{m=Q}^{M-1} g[m] z^{-m}}$$

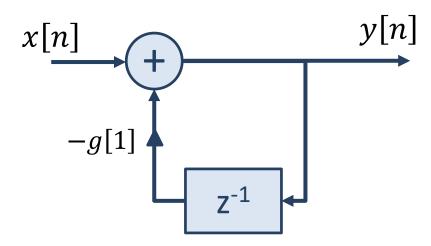
Question: What are the defining characteristics of the pole-zero plot?

IIR Direct Form

$$y[n] = x[n] - \sum_{m=1}^{M-1} g[m]y[n-m]$$

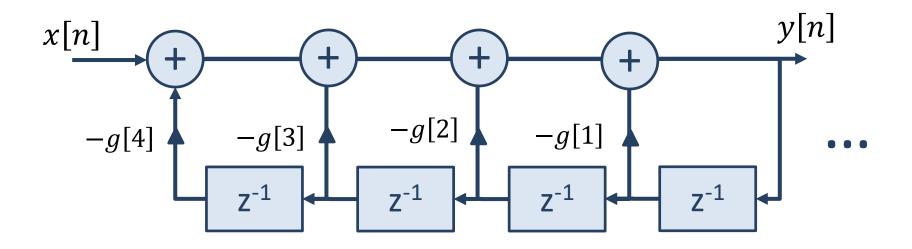
IIR Direct Form (M = 1)

$$y[n] = x[n] - g[1]y[n-1]$$



IIR Direct Form

$$y[n] = x[n] - \sum_{m=1}^{M-1} g[m]y[n-m]$$



- Two ways to look at general IIR filters
 - Convolution perspective

$$y[n] + \sum_{m=1}^{M-1} g[m]y[n-m] = \sum_{k=1}^{M-1} h[k]x[n-k]$$

Filtering perspective

$$\frac{Y(z)}{X(z)} = \frac{\sum_{m=Q}^{M-1} h[m]z^{-m}}{1 + \sum_{m=Q}^{M-1} g[m]z^{-m}}$$

- Two ways to look at general IIR filters
 - Convolution perspective

$$y[n] = \sum_{k=1}^{M-1} h[k]x[n-k] - \sum_{m=1}^{M-1} g[m]y[n-m]$$

Filtering perspective

$$\frac{Y(z)}{X(z)} = \frac{\sum_{m=Q}^{M-1} h[m]z^{-m}}{1 + \sum_{m=Q}^{M-1} g[m]z^{-m}}$$

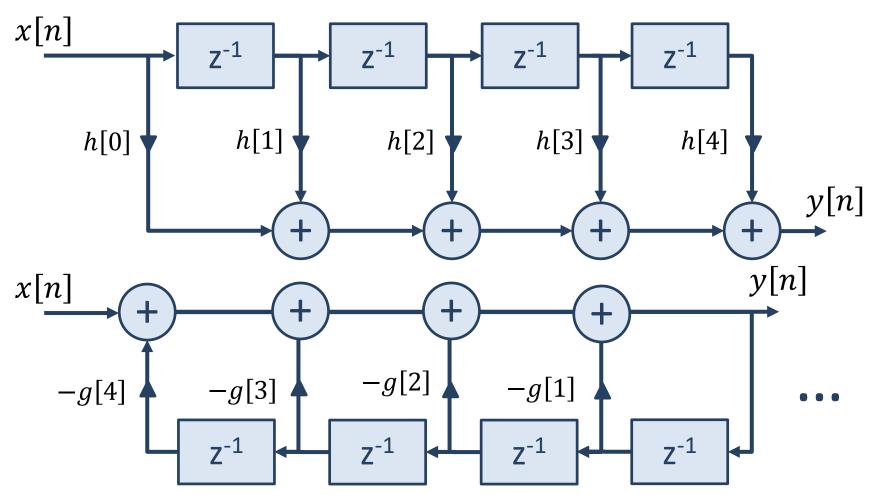
- Two ways to look at IIR filters with only recursive components
 - Convolution perspective

$$y[n] = x[n] - \sum_{m=1}^{M-1} g[m]y[n-m]$$

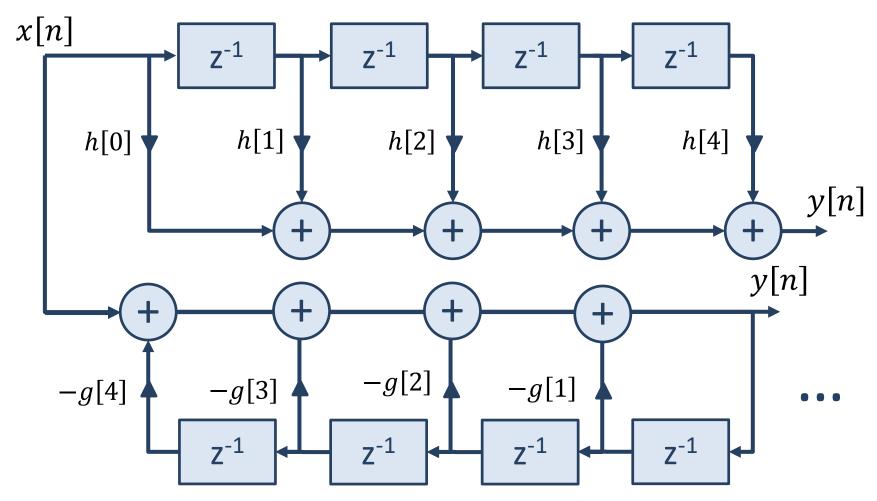
Filtering perspective

$$\frac{Y(z)}{X(z)} = \frac{\sum_{m=Q}^{M-1} h[m]z^{-m}}{1 + \sum_{m=Q}^{M-1} g[m]z^{-m}}$$

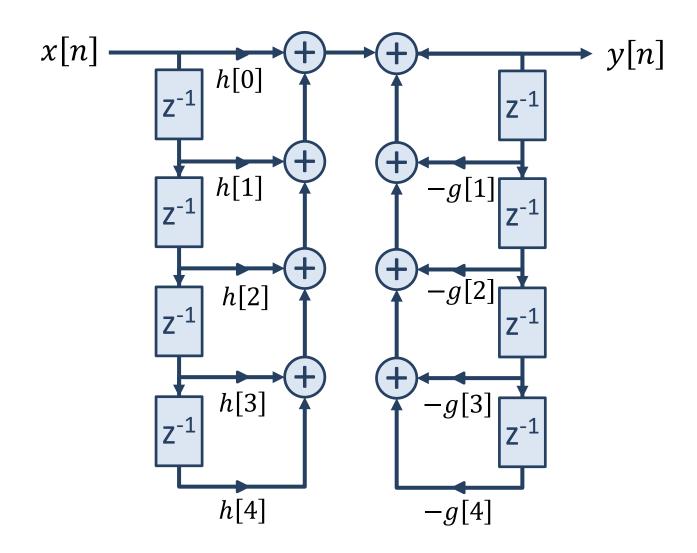
IIR Direct Form



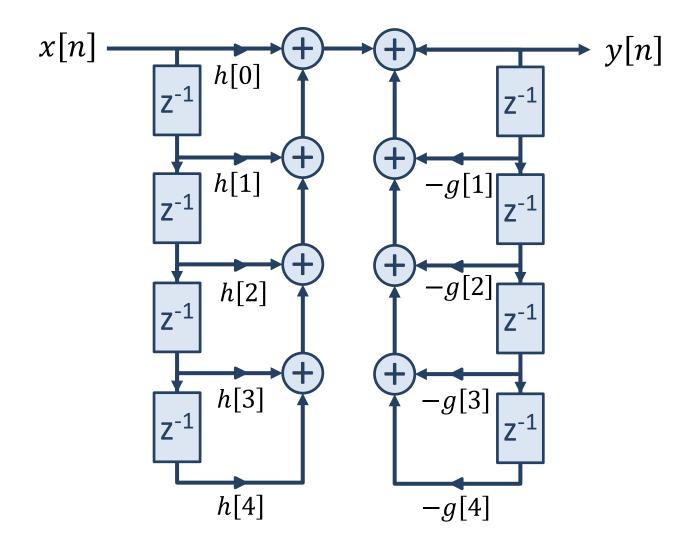
IIR Direct Form



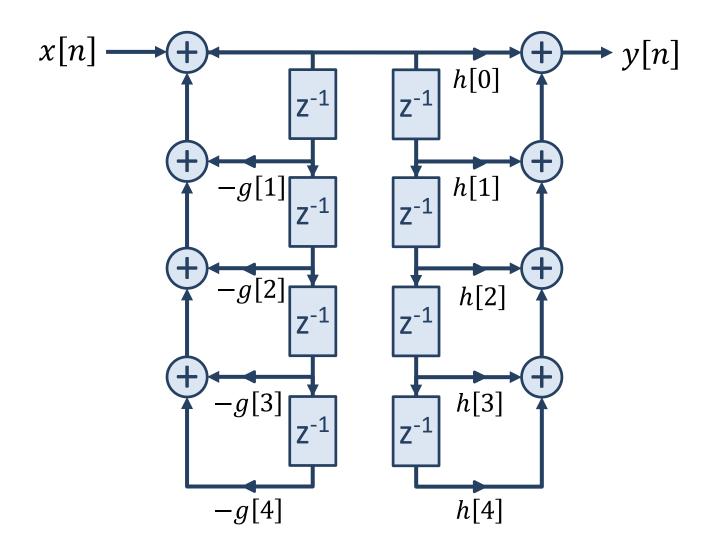
IIR Direct Form



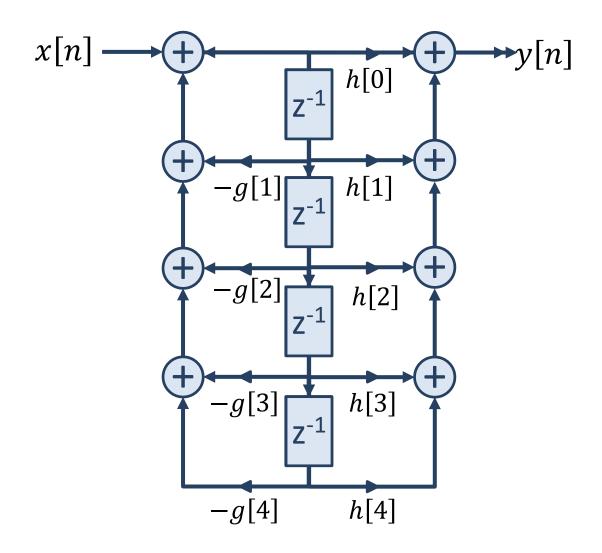
IIR Direct Form I



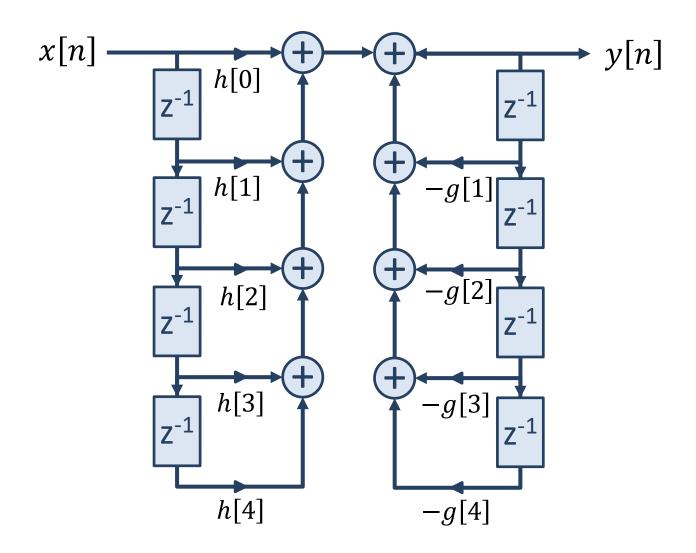
IIR Direct Form II



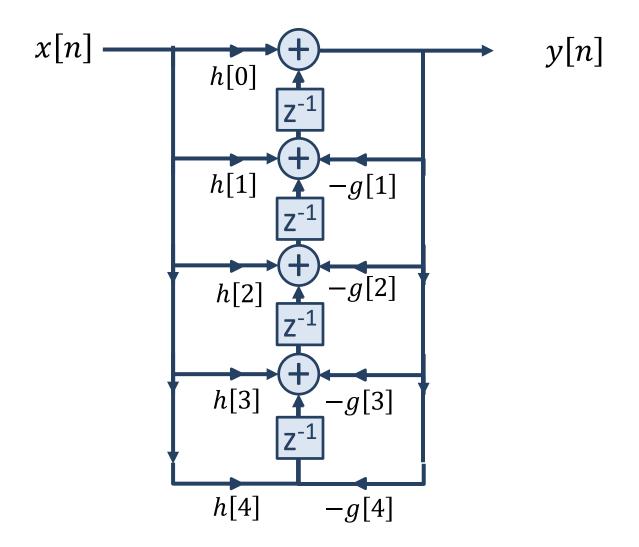
IIR Direct Form II



IIR Direct Form I

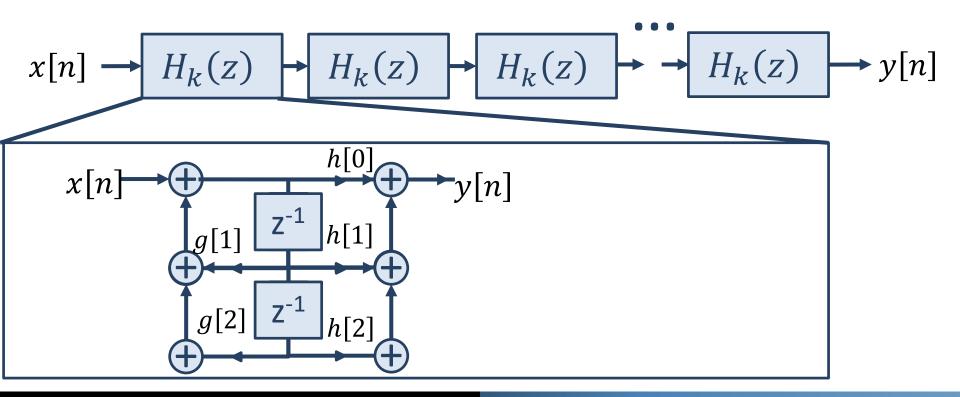


Transposed IIR Direct Form II



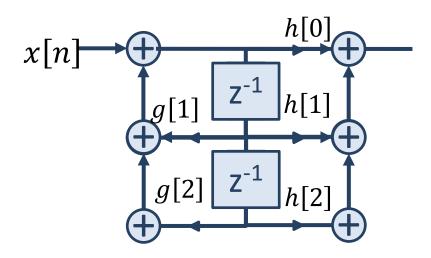
IIR Cascade Form

$$Y(z) = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



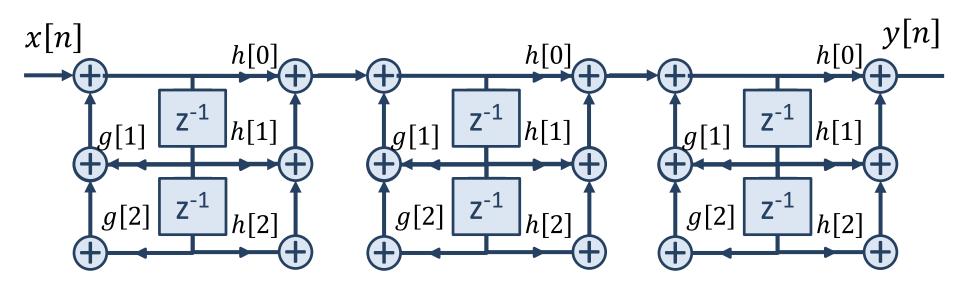
IIR Cascade Form

$$Y(z) = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



IIR Cascade Form

$$Y(z) = X(z) \sum_{m=0}^{M-1} h[m] z^{-m} = X(z) \prod_{k=1}^{K} H_k(z)$$



Parallel Form

