Basic Electrical Engineering: AC Circuits

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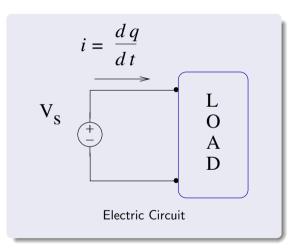
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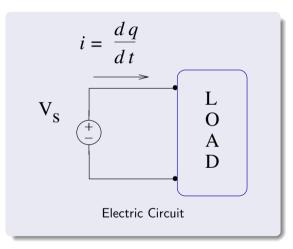


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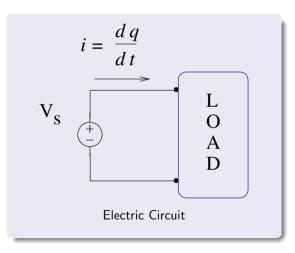




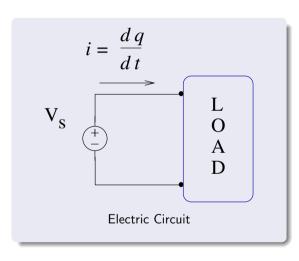
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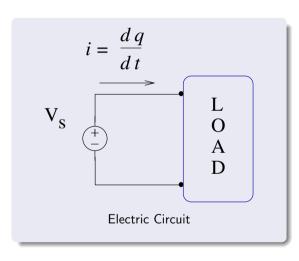


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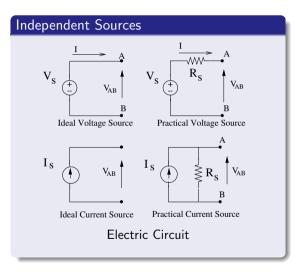
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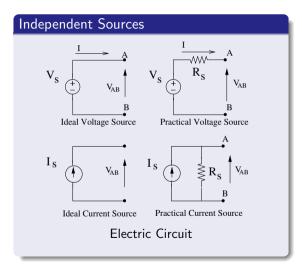


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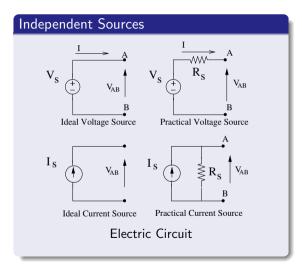




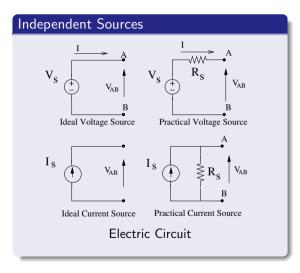
 Ideal Voltage Source can supply infinite current (Energy) at constant terminal voltage



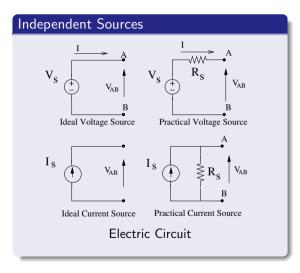
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Electric Power (Unit: Watt):
$$P = Voltage \cdot Current$$
 $V = \frac{P}{I}$ $V = \frac{Work}{time} \cdot \frac{time}{Q}$, $V = \frac{W}{Q}$



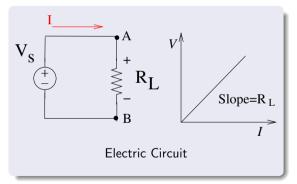
Electro Motive Force (Unit Volt) does not have unit of force, but is a driving force to produce electric current when connected to electrical load. Typically, it is the terminal voltage of a open circuited Battery ifference in Potentials between the two points. Its unit is *Volt*. The potential sources of e.m.f. are:

- The electrodes of dissimilar materials immersed in an electrolyte, as in primary and secondary cells, i.e. batteries.
- The relative movement of a conductor and a magnetic flux, as in electric generators; this source can, alternatively, be expressed as the variation of magnetic flux linked with a coil.
- The difference of temperature between junctions of dissimilar metals, as in thermo-junctions.



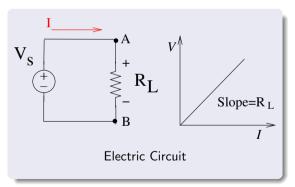
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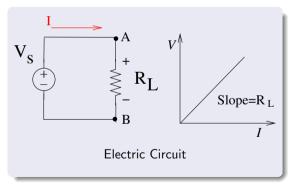
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- The resistance of a wire is expressed as $R = \frac{\rho \cdot length}{Area}$, where ρ is a constant for given conductor material (metal) and is called **Resistivity**, $(\Omega \cdot meters)$, temperature dependent

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A passive network of resistors may be available with following connections:

Series connections



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- Series connections
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- Delta connections
- Star connections



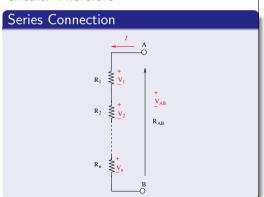
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- Delta connections
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- Q: Why equivalent resistance needs to be known to analyse the network?

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Let *I* the net current flowing through the circuit. Therefore

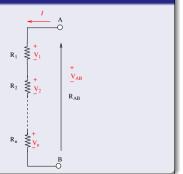


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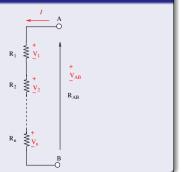
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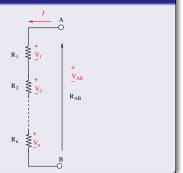
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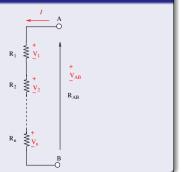
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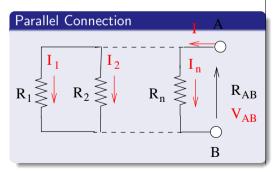
$$V_1 = I \cdot R_1 = \frac{R_1}{R_{AB}} \cdot V_{AB}$$

$$V_2 = \frac{R_2}{R_{AB}} \cdot V_{AB} \cdot \dots \cdot V_n = \frac{R_n}{R_{AB}} \cdot V_{AB}$$

Voltage Sharing Formula

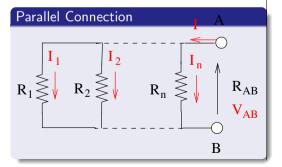


In Parallel connection, the Voltage across each resistor is same but current would be different. Let I the net current flowing through the circuit. Therefore, let equivalent resistance of the network is: R_{AB}



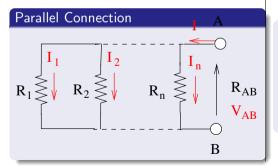
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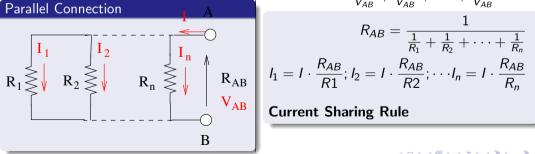
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$$V_{AB} = I_1 \cdot R_1 = I_2 \cdot R_2 = \dots = I_n \cdot R_n = I \cdot R_{AB}$$

$$R_{AB} = \frac{1}{\frac{I_1}{V_{AB}} + \frac{I_2}{V_{AB}} + \dots + \frac{I_n}{V_{AB}}}$$

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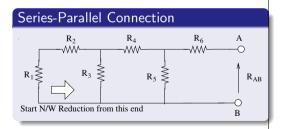
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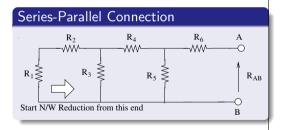
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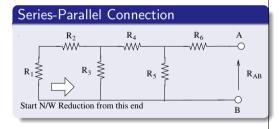
$$I_1 = I \cdot \frac{R_{AB}}{R_1}; I_2 = I \cdot \frac{R_{AB}}{R_2}; \dots I_n = I \cdot \frac{R_{AB}}{R_n}$$
Current Sharing Rule



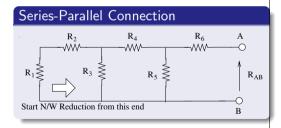
In **Series-Parallel connection**, the resistors are seen in series and parallel combination as **Ladder Network**. let equivalent resistance of the network is: R_{AB}



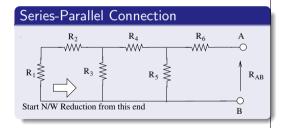
Start circuit reduction from the opposite end of the terminals A-B, (Ladder-end) and go-on solving the ladder in alternate series and parallel connection of resistors until the last node terminals



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- Resistor R_1 and R_2 are in series and their equivalence is $R_{12} = R_1 + R_2$

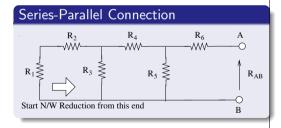


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- ② Resistor R_1 and R_2 are in series and their equivalence is $R_{12} = R_1 + R_2$
- The resistor R_{12} will then come in parallel with R_3 Therefore: $R_{123} = \frac{R_{12} \cdot R_3}{R_{12} + R_2}$



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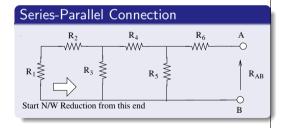
$$R_{1234} = R_{123} + R_4$$



- Start circuit reduction from the opposite end of the terminals A-B, (Ladder-end) and go-on solving the ladder in alternate series and parallel connection of resistors until the last node terminals
- **2** Resistor R_1 and R_2 are in series and their equivalence is $R_{12} = R_1 + R_2$
- **3** The resistor R_{12} will then come in parallel with R_3 Therefore: $R_{123} = \frac{R_{12} \cdot R_3}{R_{12} + R_2}$

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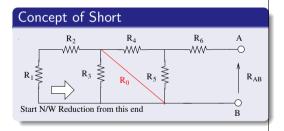


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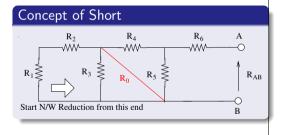
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$$R_{AB} = R_{12345} + R_6$$

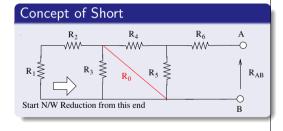




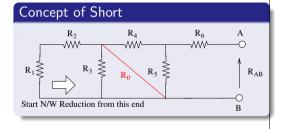
Now let us see what happens when a shorting link is introduced in the circuit



• The Resistors R_1 and R_2 are in series and their equivalence is $R_{12} = R_1 + R_2$



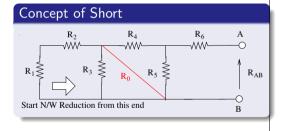
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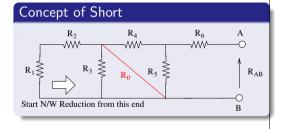
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- **1** This leads to $R_{1230} = \frac{R_{123} \cdot R_0}{R_{123} + R_0} = 0$



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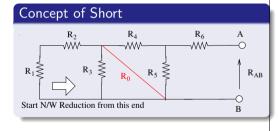
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- $R_{AB} = R_{12345} + R_6$ This is less than the previous resistor



To find Equivalent Resistance: Example

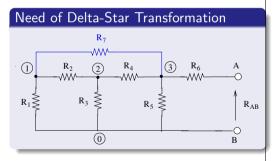
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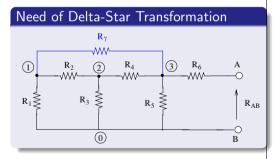
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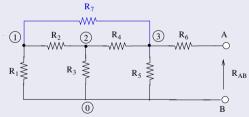
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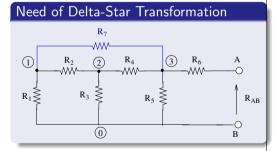
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Need of Delta-Star Transformation

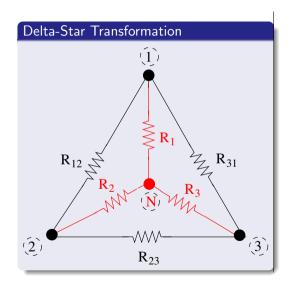


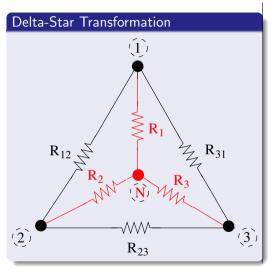
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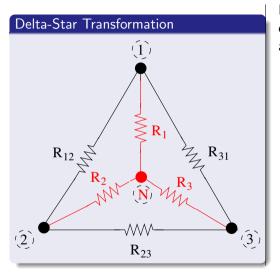


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- Therefore, the series parallel technique of network reduction fails.
- We can use Delta-StarTransformation or Star-DeltaTransformation
- Here, resistor R₇ forms Delta (nodes 1,2 and 3)with resistor R₂ and R₄. Then, converting this into equivalent Delta can solve the problem.OR Resistor R₂,R₃ and R₄ forms a Star(nodes 1,3 and 0), transform



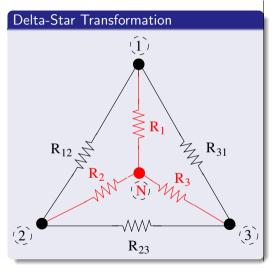


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$$R_1 + R_2 = \frac{R_{12} \cdot (R_{23} + R_{31})}{(R_{12} + R_{23} + R_{31})} \tag{1}$$

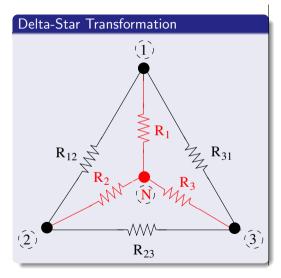


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$$R_2 + R_3 = \frac{R_{23} \cdot (R_{31} + R_{12})}{(R_{12} + R_{23} + R_{31})}$$
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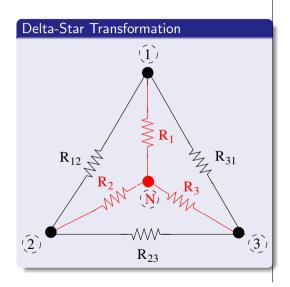
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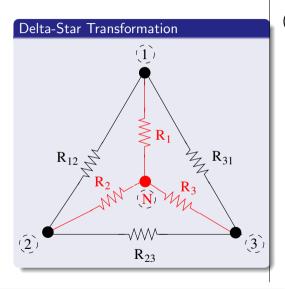
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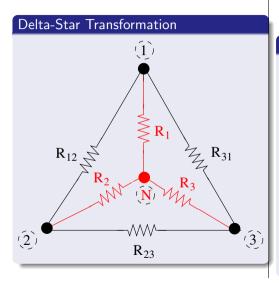
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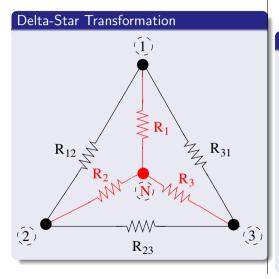
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Delta-Star Transformation

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$$R_1 = \frac{R_{12} \cdot R_{31}}{(R_{12} + R_{23} + R_{31})} \tag{4}$$

Similarly, we can perform (2)-(3)+(1) and (3)-(1)+(2), we write:

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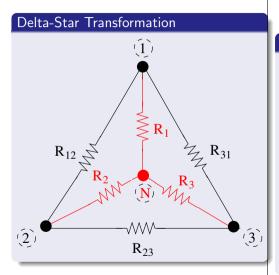
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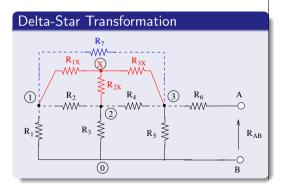
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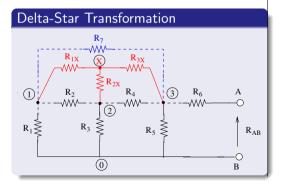


Now we shall solve this problem by existing Delta: $(R_2, R_4 \& R_7)$ connections to Star: $(R_{1X}, R_{2X} \& R_{3X})$ Network by Delta-Star Transformation.

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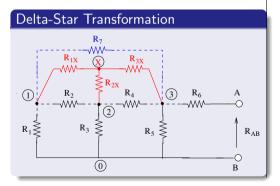


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$$R1X = \frac{R_2 \cdot R_7}{R_2 + R_4 + R_7}$$

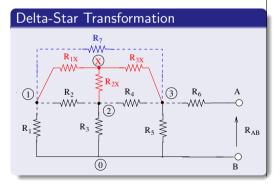
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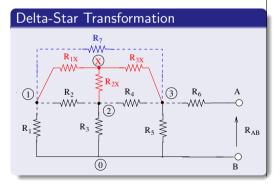


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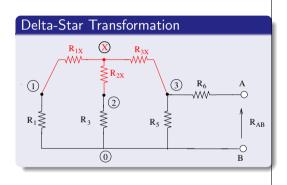
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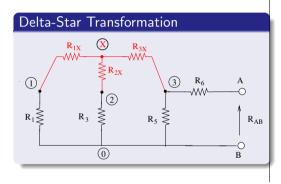


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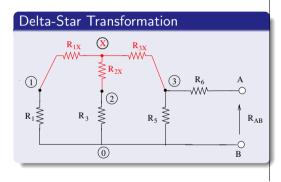
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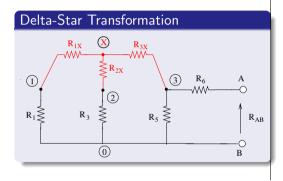


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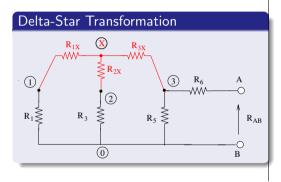
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 $R_{11X} = R_1 + R_{1X}$, $R_{23X} = R_3 + R_{2X}$



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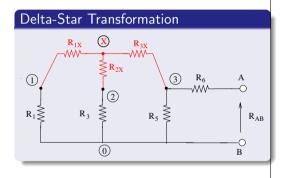


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$$\bullet$$
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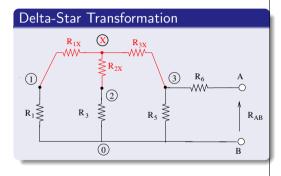
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$$R_{1123X} = \frac{R_{11X} \cdot R_{23X}}{R_{11X} + R_{23X}}$$

$$\bullet$$
 $R_{11233X} = R_{1123X} + R_{3X}$

$$P_{112335X} = \frac{R_{11233X} \cdot R_5}{R_{11233X} + R_5}$$



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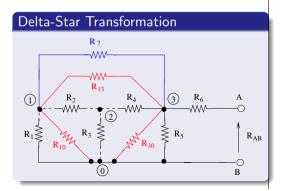
$$\bullet$$
 $R_{11233X} = R_{1123X} + R_{3X}$

•
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•
$$R_{AB} = R_{112335X} + R_6$$

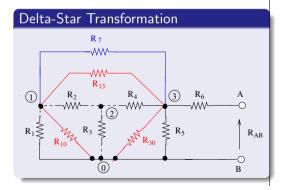
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Delta-Star Transformation R 7 -**√**//\/ R₁₂ R_{AB} By using equations: (7), (8) and (9)

Star-Delta Transformation

$$R_{10} = \frac{R_2 \cdot R_3 + R_3 \cdot R_4 + R_2 \cdot R_4}{R_4}$$

Now we shall solve this problem by existing Star: $(R_2,R_3 \& R_4)$ connections to new Delta: $(R_{10},R_{30} \& R_{13})$ Network by Star-Delta Transformation.

Delta-Star Transformation R_{AB}

By using equations: (7), (8) and (9)

Star-Delta Transformation

$$R_{10} = \frac{R_2 \cdot R_3 + R_3 \cdot R_4 + R_2 \cdot R_4}{R_4} \quad (10)$$

Similarily,

$$R_{30} = \frac{R_2 \cdot R_3 + R_3 \cdot R_4 + R_2 \cdot R_4}{R_2}$$

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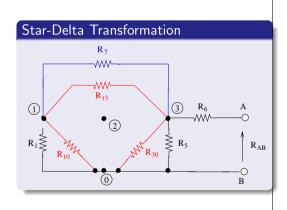
Star-Delta Transformation

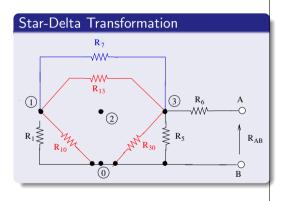
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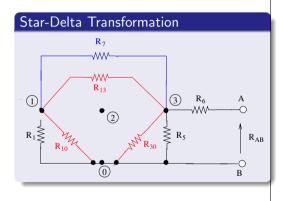
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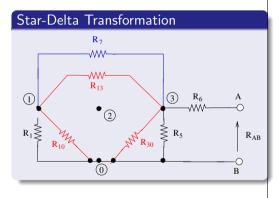


Now, we have to analyse new network and neglect the resistors $(R_2, R_3 \& R_4)$. The node (2) will vanish now in the new network.



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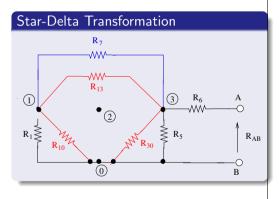
$$R_{137} = \frac{R_{13} \cdot R_7}{R_{13} + R_7}$$



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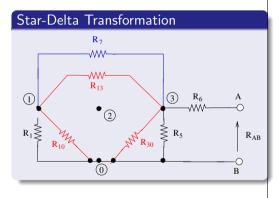


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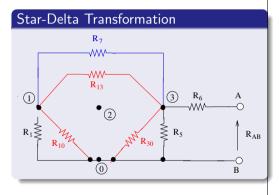
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$$R_{305} = \frac{R_{30} \cdot R_5}{R_{30} + R_5}$$

$$R_{137101} = R_{137} + R_{101}$$



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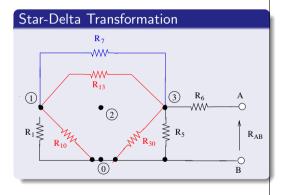
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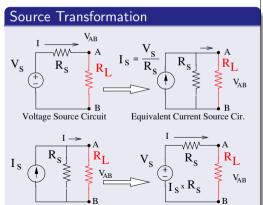
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$$\bullet$$
 $R_{AB} = R_{137101305} + R_6$



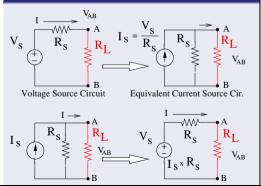
Voltage to Current Source Transformation: Let us consider a primary circuit as Voltage source with voltage V_s and having a series resistor having value R_s .

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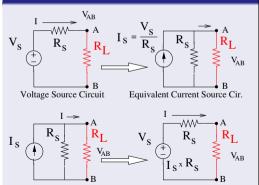
Source Transformation



Let us assume that this voltage source circuit can be transformed to an equivalent current source circuit with a parallel resistor of value R_s . Due to equivalence, the terminal current in both the circuit should be equal.

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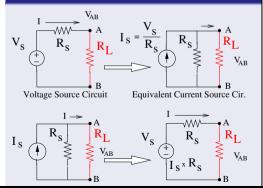


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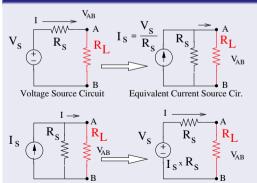
Let an external load resistor R_L be connected at the terminal A-B of both the circuits. If we equate the current produced by both the circuits in the load resistor, we get:

$$\frac{V_s}{R_s + R_L} = I_s \cdot \frac{R_s}{R_s + R_L}$$



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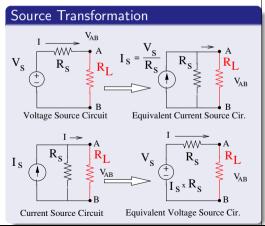
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Current to Voltage Source Transformation: Let us consider a primary circuit as Current source with value I_s and having a parallel resistor with value R_s .

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Source Transformation Voltage Source Circuit Equivalent Current Source Cir. Current Source Circuit Equivalent Voltage Source Cir.

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Current to Voltage Source Transformation: Let us consider a primary circuit as Current source with value I_s and having a parallel resistor with value R_s .

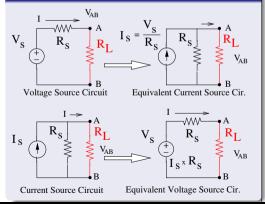
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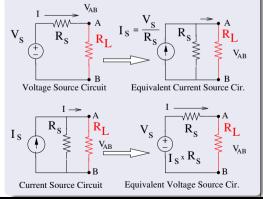
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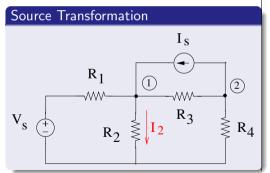
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$$I_{s} \cdot \frac{R_{s} \cdot R_{L}}{R_{s} + R_{L}} = V_{s} \cdot \frac{R_{L}}{R_{s} + R_{L}}$$
$$V_{s} = I_{s} \cdot R_{s}$$

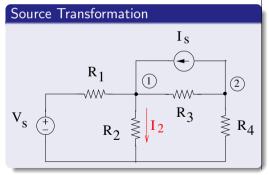


Example: Let us consider a circuit as shown below. The objective is to find current through the resistor R_2 to be evaluated by Source Transformation Technique.

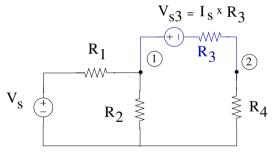
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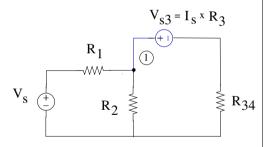
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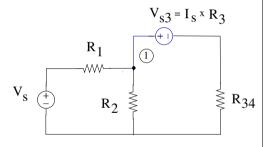
Now let us transform the current source I_s along with parallel resistor R_3 to an equivalent voltage source V_{53} and a series resistor R_3 . We can write $V_{53} = I_s \cdot R_3$.



Now solve $R_{34} = R_3 + R_4$, as series combination.

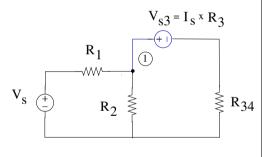


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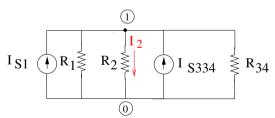
Now transform the voltage source V_s3 alongwith series resistor R_{34} , to a current source $I_{s334} = \frac{V_s3}{R_{34}}$ with a parrallel resistor R_{34} .

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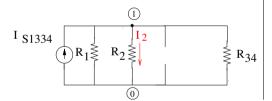
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Now, transform voltage source V_s into an equivalent current source I_{s1} with a parallel resistor R_1 .

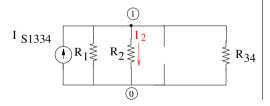


Since, both current sources are connected between same pair of nodes and they are in the same direction, we can get resultant current as $I_{51334} = I_{51} + I_{5334}$.

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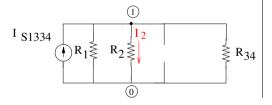
Since, both current sources are connected between same pair of nodes and they are in the same direction, we can get resultant current as $I_{S1334} = I_{s1} + I_{s334}$.



Calculating total equivalent parallel resistor across R_2 as: $R134 = \frac{R_1 \cdot R_{34}}{R_1 + R_{34}}$

Circuit Analysis by Source Transformation

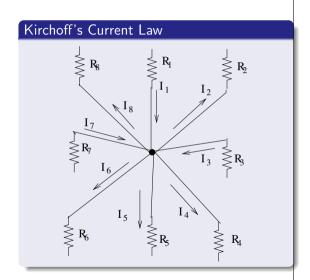
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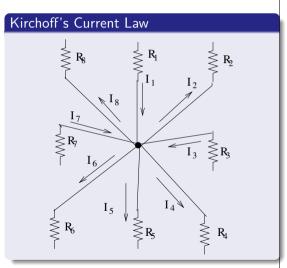


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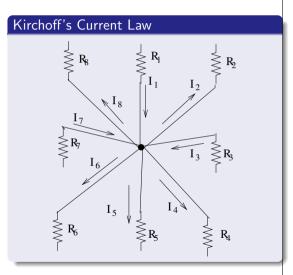
Finally, the current I_2 can be computed from following equation:

$$I_2 = I_{S1334} \cdot \frac{R_{134}}{R_2 + R_{134}}$$





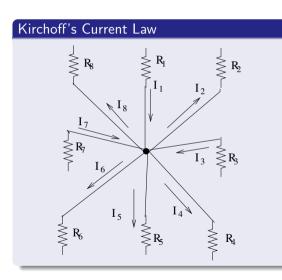
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$$I_1 + I_3 + I_7 - I_2 - I_4 - I_5 - I_6 - I_8 = 0$$

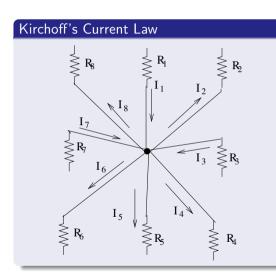


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Where, b=branches, n=no. of braches meeting at a junction/node. If appropriate sign-convention for the current is followed.



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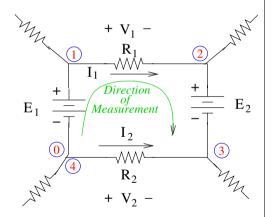
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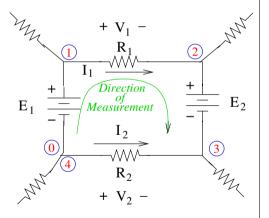
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Sign-Convention Entering current is positive (+ve) and leaving current is negative (-ve).

Kirchoff's Voltage Law

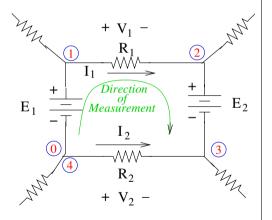


Kirchoff's Voltage Law



In a linear network, Summation of all the voltages included in a circuit-loop/mesh is zero with appropriate sign-convention.

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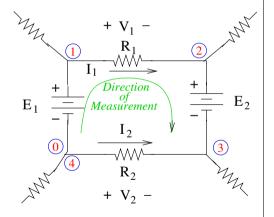


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Kirchoff's Voltage Law



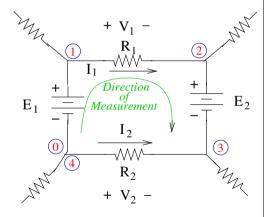
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Sign-Convention In the direction of measurement, if a voltage is meaured from -ve polarity to +ve polarity, considered as (+ve) and vice-versa (as negative (-ve)).



This is an application of **KCL** and **Ohm's Law**. Objective is to compute the unknown node voltages.

• Identify prominent nodes in the circuit. One node is identified as groumd node or reference node.

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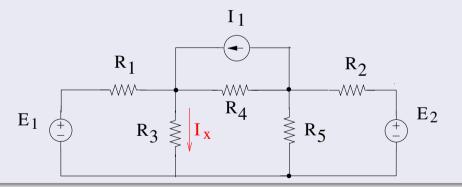
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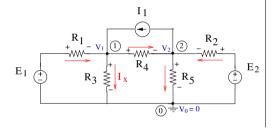
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Example

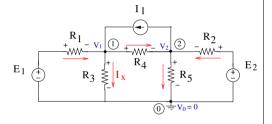
Example: To find the current I_x as shown in the figure below by Nodal Analysis.



Example

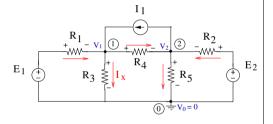


Example



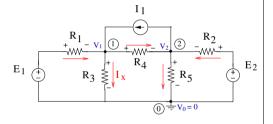
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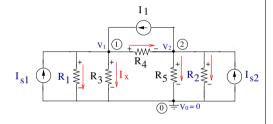
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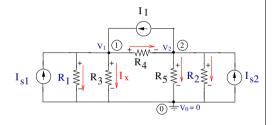
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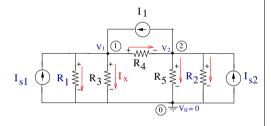
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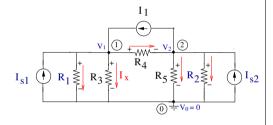
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Example



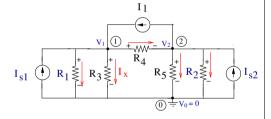
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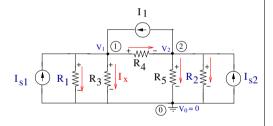
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Example



Write KCL equations for each independent node voltages. Form simultaneous linear equations in terms of node voltages as unknown quantities.

Example



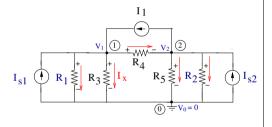
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Applying KCL at Node-1 will results in equation (13) and KCL at Node-2 will results in equation (14).

$$I_{s1} - \frac{V_1}{R_1} - \frac{V_1}{R_3} - \frac{V_1 - V_2}{R_4} + I_1 = 0 \quad (13)$$

$$I_{s2} - \frac{V_2}{R_2} - \frac{V_2}{R_5} + \frac{V_1 - V_2}{R_4} - I_1 = 0$$
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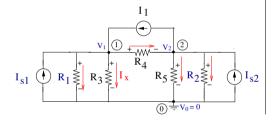
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Please note that here, only unknown quantitites are node voltages V_1 and V_2 .

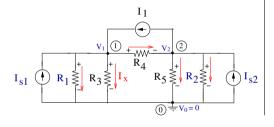


Example



• Solving equations (13) and (14), simultaneously, we can find the unknown node voltages V_1 and V_2 .

Example



- Solving equations (13) and (14), simultaneously, we can find the unknown node voltages V_1 and V_2 .
- **2** Finally we can find current through any branch of the network. However, we can find current $I_x = \frac{V_1}{R_3}$ since it is specifically asked in the exercise.

Mesh Analysis

KVL and **Ohm's Law**: Steps to be followd for generalised circuit Evaluation. Objective is to frame a procedure to compute the unkmown mesh current for complete analysis of the given circuit.

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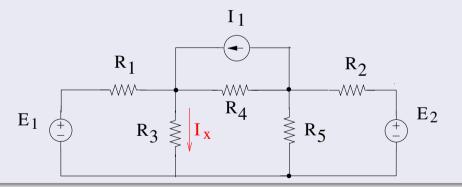
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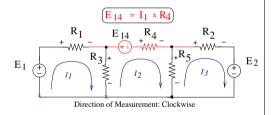
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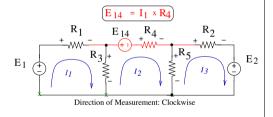
Example

Example: To find the current I_x as shown in the figure below by Mesh Analysis.

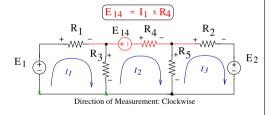




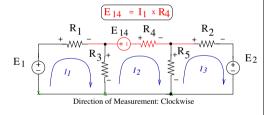
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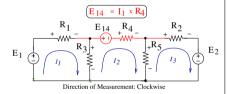
• The circuit has three independent meshes/windows (1-3).



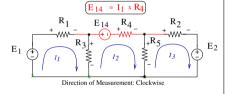
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- We then transform the current source I_1 to the equivalent voltage source $E_{14} = I_1 \cdot R_4$
- Now mark the polarity voltage drop across each resistor depending on the direction of the mesh current

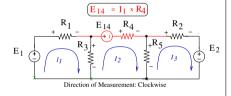


Example



 Now, start writting KVL equation for each mesh by assuming direction ofs, the resulting equations for Mesh-1 to Mesh-3 are given below, sequentially.

Example



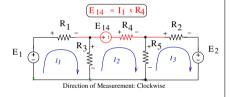
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$$E_1 - I_1 \cdot R_1 - (I_1 - I_2) \cdot R_3 = 0$$

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Now, by solving above three equations, simultaneously, the three unknown mesh currents can be computed. The current I_x can be computed by solving $(I_1 - I_2)$.

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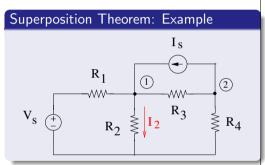
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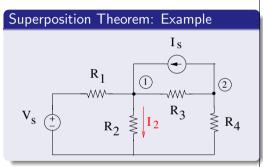


Example-1: To find current I_2 in the circuit below



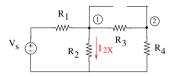
$$I_2 = I_{2X} + I_{2Y}$$

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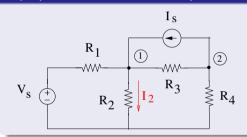
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Considering Voltage source V_s only



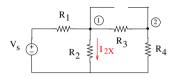
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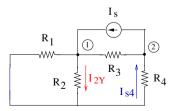


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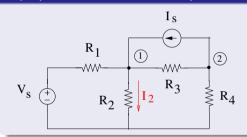


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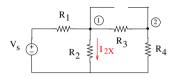
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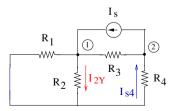


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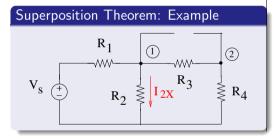
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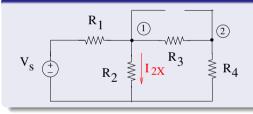


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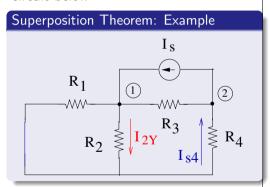
Superposition Theorem: Example



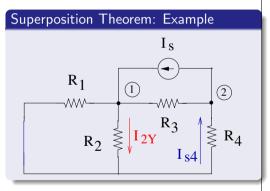
Considering Current V_s only Compute I_{2X} :

$$R_{234} = rac{R_2 \cdot (R_3 + R_4)}{R_2 + R_3 + R_4}$$
 $I_{2X} = rac{1}{R_2} \cdot V_s \cdot rac{R_{234}}{R_1 + R_{234}}$

Circuit-2: To compute current I_{2Y} in the circuit below



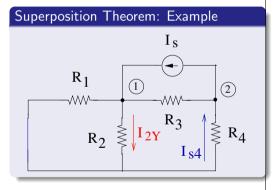
Circuit-2: To compute current I_{2Y} in the circuit below



Considering Current l_s only Compute l_{2X} :

$$R_{12} = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

Circuit-2: To compute current I_{2Y} in the circuit below



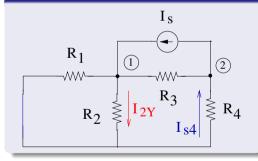
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Superposition Theorem: Example



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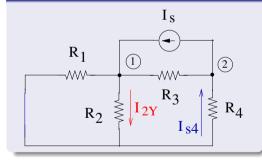
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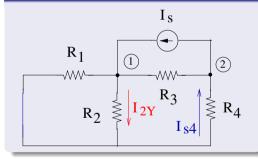
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Finally, By superposition theorem,

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Superposition Theorem: Example



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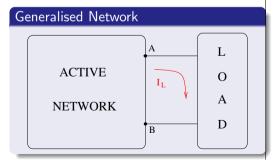
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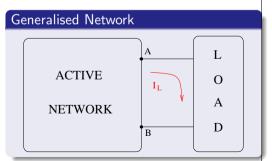
Generalised Active Electrical Network

An active network consists of one or more energy sources. A load branch may be passive circuit branch or active branch. Both Thevenin and Norton Theorem are applied to simplify and evaluate the given network.

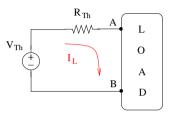


Generalised Active Electrical Network

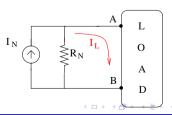
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Thevenin's Equivalent Circuit



Norton's Equivalent Circuit



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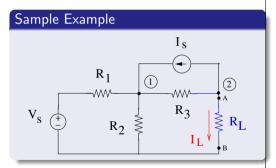
Thevenin's and Norton's Theorem

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- **1** The Norton's equivalent Current source I_N means the short-circuit current appearing at A and B terminals of the active network (AN), when load branch is removed from the original AN.



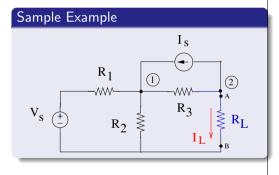
Thevenin's and Norton's Theorem

Circuit-1: To compute current I_L in the circuit below

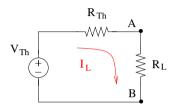


Thevenin's and Norton's Theorem

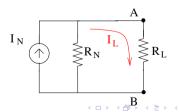
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Thevenin's Equivalent Circuit

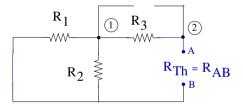


Norton's Equivalent Circuit



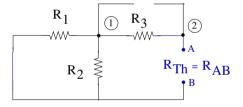
Sample Example:1

To find R_{Th} , Thevenin's Resistance



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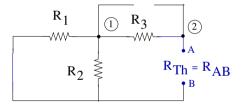
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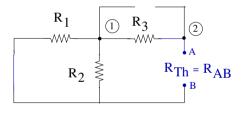


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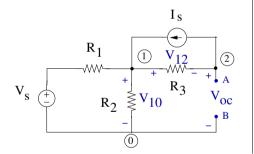
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This is same as Norton's Resistor R_N

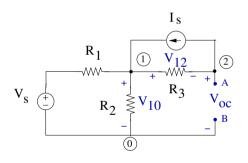
Sample Example:1

To find V_{Th} , Thevenin's Voltage Source



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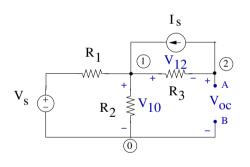
By Kirchoff's Voltage Law

$$V_{Th} = V_{oc} = V_{10} - V_{12}$$

= $V_s \cdot \frac{R_2}{R_1 + R_2} - I_s \cdot R_3$

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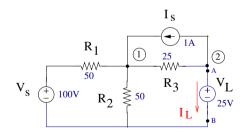
Finally, the load current I_L by Thevenin's Theorem is given as:

$$I_L = rac{V_{Th}}{R_{Th} + R_L}$$



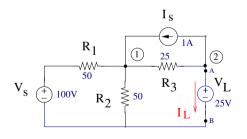
Sample Example:2

Now, let us take an example when the load branch is consisting of an Independent source.



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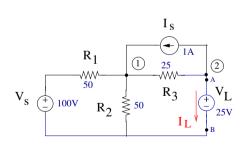
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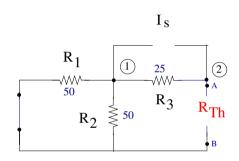
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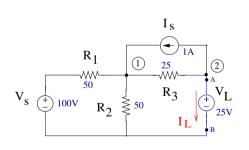


$$R_{Th} = R_{AB} = R_3 + \frac{R_1 \cdot R_2}{R_1 + R_2} = 25 + 25 = 50$$

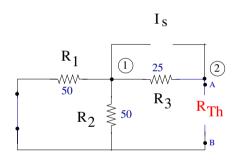


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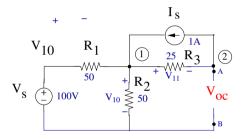


$$R_{Th} = R_{AB} = R_3 + \frac{R_1 \cdot R_2}{R_1 + R_2} = 25 + 25 = 50$$

This is same as Norton's Resistor $R_N = 990$

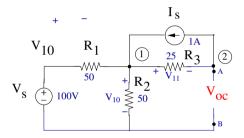
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Now, let us calculate Thevenin's equivalent source V_{Th}



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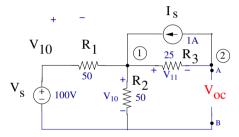
$$V_{Th} = V_{oc} = V_{10} - V_{12}$$

= $50 - 25 = 25V$



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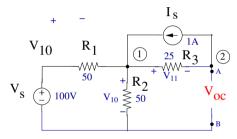
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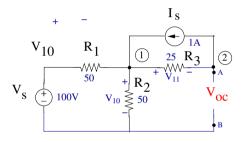
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Solving Thevenin's Equivalent Circuit

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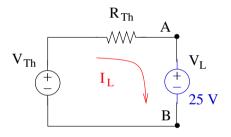
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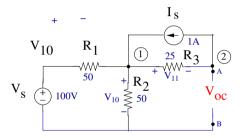
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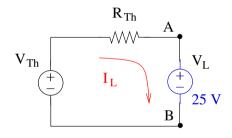
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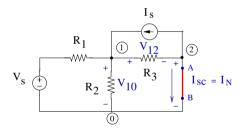
Solving Thevenin's Equivalent Circuit



$$I_L = \frac{V_{Th} - V_L}{R_{Th}}$$
$$= \frac{75}{50} = 1.5A$$

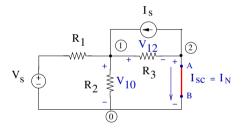
Sample Example-1

To find I_N , Norton's Current Source



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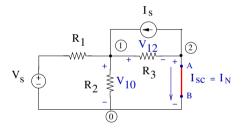


By Superposition theorem

$$I_N = I_N^{(1)} + I_N^{(2)}$$

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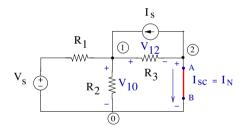


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$$I_N = I_N^{(1)} + I_N^{(2)}$$
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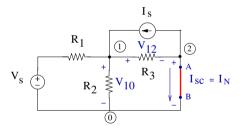
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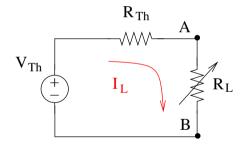
$$I_{N}^{(2)} = I_{s} \cdot \frac{R_{3}}{R_{3} + \frac{R_{1} \cdot R_{2}}{R_{1} + R_{2}}}$$

Finally, the load current I_L by Norton's Theorem is given as:

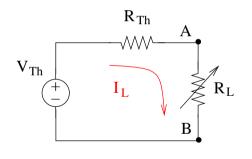
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Maximum Power in a load resistor



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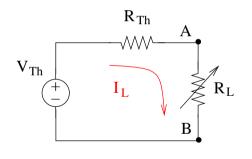
If any active network is replaced by Thevenin's eequivalent circuit, then there should be unique value of R_L for which maximum power transfer would takes place from the active network to the load Resistor. Therefore, the power transferred to the load resistor is given as $I_L^2 \cdot R_L$.

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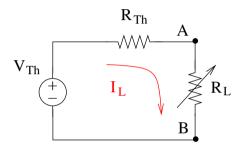
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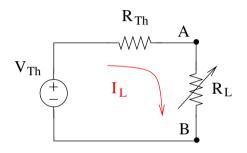
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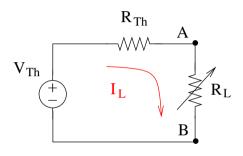
$$= \frac{V_{Th}^{2} \cdot R_{L}}{(R_{Th} + R_{L})^{2}}$$

$$\frac{dP_{L}}{dR_{L}} = V_{Th}^{2} \cdot \left(\frac{1}{R_{Th} + R_{L})^{2}} + \frac{-2 \cdot R_{L}}{(R_{Th} + R_{L})^{3}}\right)$$

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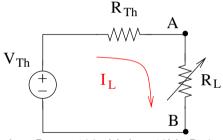
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The leads to the Condition for Maximum Power Transfer

$$R_L = R_{Th}$$

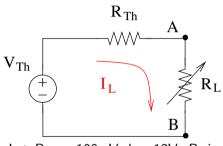


Maximum Power in a load resistor by Theven's Theorem

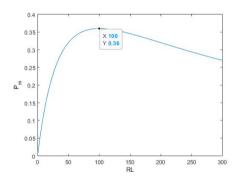


Let $R_{Th} = 100$, $V_T h = 12V$, R_L is varying from 0 to 300. The power consumed in R_L is plotted against the load Resistor.

Maximum Power in a load resistor by Theven's Theorem



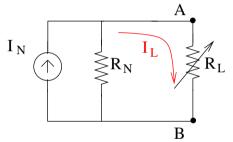
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This clearely verifies the condition of maximum power transfer.

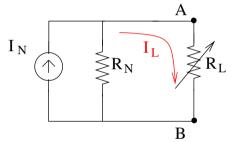


Maximum Power in a load resistor by Norton's Theorem

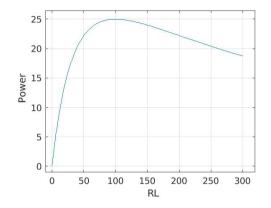


Let $R_{100} = 100$, $I_N = 1$ A, R_L is varying from 0 to 300. The power consumed in R_L is plotted against the load Resistor.

Maximum Power in a load resistor by Norton's Theorem



Let $R_{100} = 100$, $I_N = 1$ A, R_L is varying from 0 to 300. The power consumed in R_L is plotted against the load Resistor.



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DC Transients in R-L and R-C Circuit

Initial and Final Conditions

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- It also tells us the behaviour of the elements at the instant of switching.
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- **3** Sometimes we may use conditions at $t = \infty$; these are called final conditions. Final conditions are normally described as the values of the state variable after long time from the instance of switching and therefore describes the new network equilibrium condition.

Initial and Final Conditions

 \bullet At the reference time, t = 0, one or more switches operate.

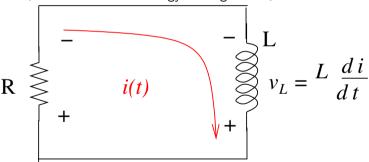
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- Initial conditions in an electric circuit depend on the state of the network prior to $t=0^+$, and the network structure at $t=0^+$, after switching.

The **source-free** response is called as natural response.

Consider a simple R-L circuit as shown below. The network has no excitation, but assumed that there is some initial current I_0 present in the network prior to t=0 where the analysis starts, as inductor is an energy storing device,



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$$i(t) = I_0 \cdot e^{\frac{-R}{L} \cdot t}$$



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At
$$t = \tau$$

$$i(\tau) = I_0 \cdot e^{\frac{-\tau}{\tau}}$$
$$= I_0 \cdot 0.368$$

$$i(t) = I_0 \cdot e^{\frac{-R}{L} \cdot t}$$
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$$i(\tau) = I_0 \cdot e^{\frac{-\tau}{\tau}}$$
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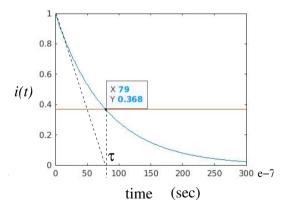
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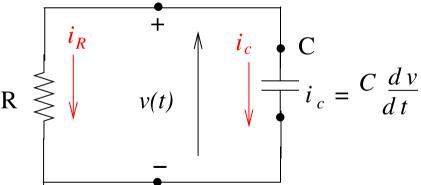
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Consider a simple R-C circuit as shown below. The network has no excitation, but assume that there is some initial voltage V_0 present across the capacitor prior to t=0, where the analysis starts, as the capacitor is an energy storing device.



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Natural Response in R-C Circuit

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Then let us define $\tau=R\cdot C$, Therefore, here, $\tau=7.957\mu$ sec. The symbol τ is called as **Time-Constant** of R-C Circuit. Following is the plot for v(t) at above parameters.

$$v(t) = V_0 \cdot e^{\frac{-t}{R \cdot C}}$$
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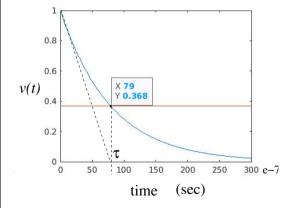
Where,
$$au=R\cdot C$$
 ; At $t= au$
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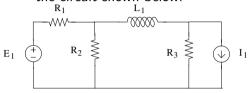
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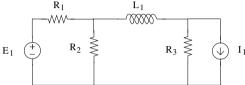
Draw a tangent to v(t) curve at t=0, the line cuts the time-axis at $t = \tau$.



Example-1 Find the time constant of the circuit shown below.

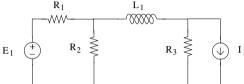


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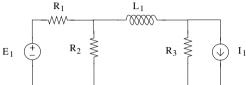
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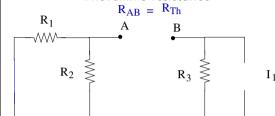
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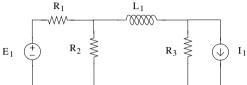


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Passive equivalent circuit to find
Theyenin's resistance

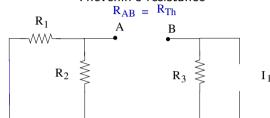


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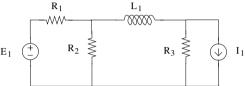
Passive equivalent circuit to find Thevenin's resistance



$$R_{Th} = R_{AB} = R_3 + \frac{R_1 \cdot R_2}{R_1 + R_2}$$

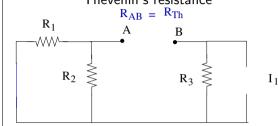


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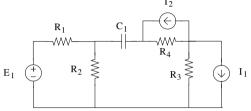


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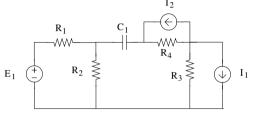
Therefore time-constant of the given circuit is given as: $\tau = \frac{L_1}{R_{Th}}$.



Example-2 Find the time constant of the circuit shown below.

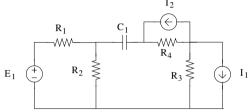


Example-2 Find the time constant of the circuit shown below.



The above example can be solved by simplifying the above circuit to get equivalent series R-C Circuit.

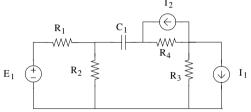
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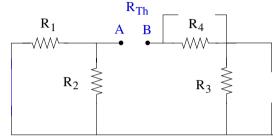


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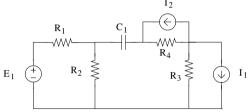


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Passive equivalent circuit to find Thevenin's resistance

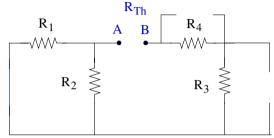


Example-2 Find the time constant of the circuit shown below.



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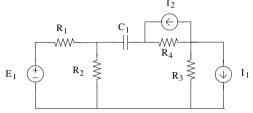
Passive equivalent circuit to find Thevenin's resistance



$$R_{Th} = R_{AB} = R_4 + R_3 + \frac{R_1 \cdot R_2}{R_1 + R_2}$$

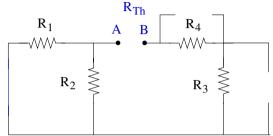


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Passive equivalent circuit to find Thevenin's resistance



$$R_{Th} = R_{AB} = R_4 + R_3 + \frac{R_1 \cdot R_2}{R_1 + R_2}$$

Therefore time-constant of the given circuit is given as: $\tau = R_{Th} \cdot C_1$.

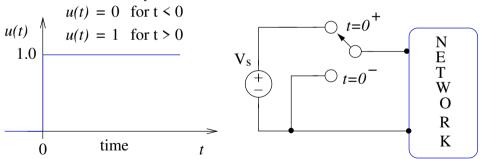
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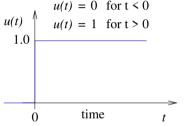
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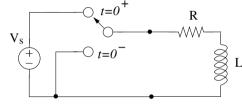
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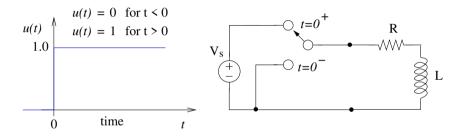
Let us analyse the time domain response of R-L Circuit described in following figure:

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The assumption is the initial inductor current prior to t=0 is zero. Therefore. $i(0)^- = i(0)^+ = 0$.



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- **3** Finally add the two solutions to get final expression in time-domain as $i(t) = i_{(natural)} + i_{(forced)}$.



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The Forced Response

Since the forcing function is constant for t > 0, the particular solution is assumed be constant as $i_t = I$. Therefore, substituting this in eq:(26),

$$L \cdot \frac{dI}{dt} + R \cdot I = V_s$$

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Since I is constant, $\frac{dI}{dt} = 0$, $I = \frac{V_s}{R} = I_f$ is called as Particular Solution.



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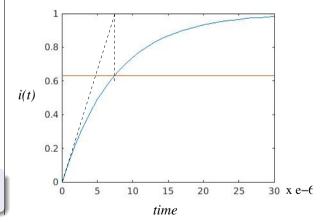
Now, substituting initial condition as $i0^- = i0^+ = 0$

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Therefore, $A = -\frac{V_s}{R}$, hence, substituting this in eq:(24), we get final solution as:

$$i(t) = \frac{V_s}{R} \cdot \left(1 - e^{\frac{-R}{L} \cdot t}\right)$$

Let
$$V_s=1$$
 V, R=1 Ω , L=79.57 μ H $i(t)=\left(1-e^{rac{-1}{79.57\mu}}
ight)$



The Complete Response with non-zero initial condition

$$i(t) = i_n + i_f = A \cdot e^{\frac{-t}{\tau}} + \frac{V_s}{R}$$
 (24)

The Complete Response with nonzero initial condition

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Now, substituting initial condition as $i0^- = i0^+ = I_0$

$$i(0) = A + \frac{V_s}{R} = I_0$$

The Complete Response with non-zero initial condition

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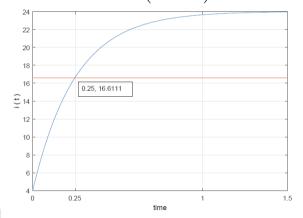
Now, substituting initial condition as $i0^- = i0^+ = I_0$

$$i(0) = A + \frac{V_s}{R} = I_0$$

Therefore, $A = -\frac{V_s}{R} + I_0$, hence, substituting this in eq:(24), we get final solution as:

$$i(t) = \frac{V_s}{R} \cdot \left(1 - e^{\frac{-R}{L} \cdot t}\right) + I_0 \cdot e^{\frac{-R}{L} \cdot t}$$

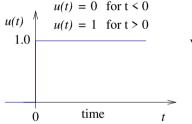
Let
$$V_s = 40$$
V, $I_{0_{\frac{t}{t}}} = 4$ A, R=2 Ω , L=0.5H $i(t) = 4 \frac{e^{\frac{-t}{0.25}}}{e^{2.25}} + 20\left(1 - e^{\frac{-t}{0.25}}\right)$.

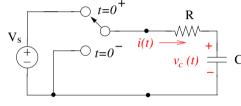


Transient Response of R-L Circuit

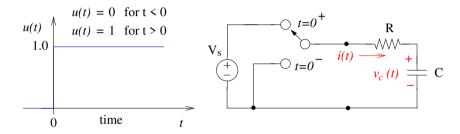
Let us analyse the time domain response of R-C Circuit described in following figure:

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The assumption is the initial capacitor voltage prior to t=0 is zero. Therefore.

$$v_c(0)^- = v_c(0)^+ = 0.$$



At t=0 and onwards, $i = C \cdot \frac{dv_c}{dt}$. The voltage $v_c(t)$ can be solved by following differential equation obtained by applying KVL:

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$$R \cdot i(t) + v_c(t) = V_s \tag{25}$$

$$R \cdot C \cdot \frac{dv_c}{dt} + v_c = V_s \tag{26}$$

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This is a non-homogeneous differential equation with constant coefficients.

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$$v_n = A \cdot e^{\frac{-t}{R \cdot C}} = A \cdot e^{\frac{-t}{\tau}}$$

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The Forced Response

(27) Since the forcing function is constant for t > 0, the particular solution is assumed be constant as $v_c t = V_c$. Therefore, subdifferenstituting this in eq:(26),

$$R \cdot C \cdot \frac{dV_c}{dt} + R \cdot V_c = V_s$$

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Since the forcing function is constant for t > 0, the particular solution is assumed be constant as $v_c t = V_c$. Therefore, substituting this in eq:(26),

$$R \cdot C \cdot \frac{dV_c}{dt} + R \cdot V_c = V_s$$

Since V_c is constant, $\frac{dV_c}{dt} = 0$, $V_c = V_s = v_f$ is called as Particular Solution.



The Complete Response with zero Initial Condition

$$v_c(t) = v_n + v_f = A \cdot e^{\frac{-t}{\tau}} + V_s$$
 (28)

The Complete Response with zero Initial Condition

$$v_c(t) = v_n + v_f = A \cdot e^{\frac{-t}{\tau}} + V_s$$
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Now, substituting initial condition as $v_c 0^- = v_c 0^+ = 0$

$$v_c(0) = A + V_s = 0$$

The Complete Response with zero Initial Condition

$$v_c(t) = v_n + v_f = A \cdot e^{\frac{-t}{\tau}} + V_s$$
 (28)

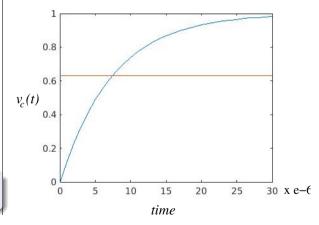
Now, substituting initial condition as $v_c 0^- = v_c 0^+ = 0$

$$v_c(0) = A + V_s = 0$$

Therefore, $A = -V_s$, hence, substituting this in eq:(29), we get final solution as:

$$v_c(t) = V_s \cdot \left(1 - e^{\frac{-t}{R \cdot C}}\right)$$

Let
$$V_s=1$$
 V, R=1 Ω , C=79.57 μ H $v_c(t)=\left(1-e^{rac{-1}{79.57\mu}}
ight)$



The Complete Response with Nonzero Initial Condition

$$v_c(t) = v_n + v_f = A \cdot e^{\frac{-t}{\tau}} + V_s$$
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The Complete Response with Nonzero Initial Condition

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$$v_c(0) = A + V_s = V_0$$

The Complete Response with Nonzero Initial Condition

$$v_c(t) = v_n + v_f = A \cdot e^{\frac{-t}{\tau}} + V_s$$
 (29)

Now, substituting initial condition as $v_c 0^- = v_c 0^+ = V_0$

$$v_c(0) = A + V_s = V_0$$

Therefore, $A = V_0 - V_s$, hence, substituting this in eq:(29), we get final solution as:

$$v_c(t) = V_s \cdot \left(1 - e^{\frac{-t}{R \cdot C}}\right) + V_0 \cdot e^{\frac{-t}{R \cdot C}}$$

Let
$$V_s = 100 \text{V}, V_0 = 12 \text{V R} = 10 \ \Omega,$$
 $C = 100 \mu \text{H}$ $v_c(t) = 100 \left(1 - e^{\frac{-1}{1e - 3}}\right) + 12 \ e^{\frac{-1}{1e - 3}}$

