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Circuits and Transforms

VIBHAVASU

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Abstract—This manual provides a simple introduction to Transforms

1 Definitions

1. The unit step function is

$$u(t) = \begin{cases} 1 & t > 0 \\ \frac{1}{2} & t = 0 \\ 0 & t < 0 \end{cases}$$
 (1.1)

2. The Laplace transform of g(t) is defined as

$$G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$$
 (1.2)

2 Laplace Transform

1. In the circuit, the switch S is connected to position P for a long time so that the charge on the capacitor becomes $q_1 \mu C$. Then S is switched to position Q. After a long time, the charge on the capacitor is $q_2 \mu C$.

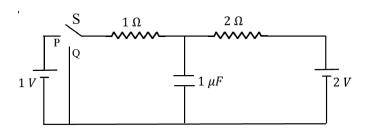


Fig. 2.1

2. Draw the circuit using latex-tikz. **Solution:**

3. Find q_1 .

Solution:

After infinite time with switch at P The capacitor is charged Applying KCL at X:

$$\frac{V_x - 1}{1} = \frac{2 - V_x}{2} \tag{2.1}$$

$$\implies V_x = \frac{4}{3} \text{ V} \tag{2.2}$$

$$q_1 = CV = 1 \ \mu\text{C} \tag{2.3}$$

4. Show that the Laplace transform of u(t) is $\frac{1}{s}$ and find the ROC.

Solution:

$$\mathcal{U}(s) = \int_0^\infty u(t)e^{-st}dt \qquad (2.4)$$

$$= \int_0^0 \frac{1}{2} e^{-st} dt + \int_0^\infty e^{-st} dt = \frac{1}{s}$$
 (2.5)

R.O.C:
$$Re(s) > 0$$
 (2.6)

5. Show that

$$e^{-at}u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \frac{1}{s+a}, \quad a > 0$$
 (2.7)

and find the ROC. Solution:

$$e^{-at}u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} \int_0^\infty u(t)e^{-(s+a)t}dt$$
 (2.8)

$$=\frac{1}{s+a}\tag{2.9}$$

R.O.C:
$$Re(s) > -a$$
 (2.10)

6. Now consider the following resistive circuit transformed from Fig. 2.1 where

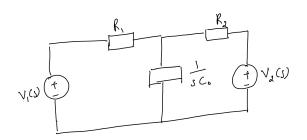


Fig. 2.3

$$u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_1(s)$$
 (2.11)

$$2u(t) \stackrel{\mathcal{L}}{\longleftrightarrow} V_2(s)$$
 (2.12)

Find the voltage across the capacitor $V_{C_0}(s)$. **Solution:** Applying KCL at X:

$$\frac{V_x - \frac{1}{s}}{R_1} + s(VC_0) = \frac{\frac{2}{s} - V_x}{R_2}$$
 (2.13)

$$V(s) = \frac{\frac{1}{R_1} + \frac{2}{R_2}}{s\left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right)}$$
 (2.14)

$$= \frac{2R_1 + R_2}{R_1 + R_2} \left(\frac{1}{s} - \frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right)$$
 (2.15)

7. Find $v_{C_0}(t)$. Plot using python.

$$v_{C_0}(t) = \frac{2R_1 + R_2}{R_1 + R_2} u(t) \left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} \right)$$
 (2.16)

$$v_{C_0}(t) = \frac{4}{3} \left(1 - e^{-(1.5 \times 10^6)t} \right) u(t)$$
 (2.17)

8. Verify your result using ngspice. **Solution:**

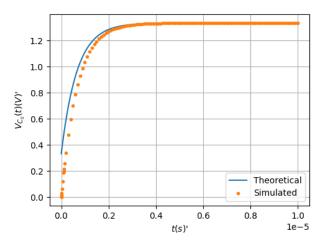


Fig. 2.4: $v_{C_0}(t)$ before the switch is flipped

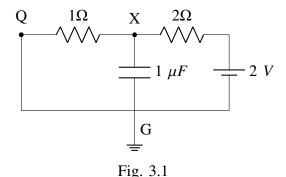
9. Obtain Fig. 2.3 using the equivalent differential equation.

3 Initial Conditions

1. Find q_2 in Fig. 2.1.

Solution: The circuit at steady state when the switch is at Q:

At steady state: Capacitor is charged



Applying KCL at X.

$$\frac{V-0}{1} + \frac{V-2}{2} = 0 \tag{3.1}$$

$$\implies V = \frac{2}{3} \operatorname{V} q_2 = \frac{2}{3} \mu C \tag{3.2}$$

2. Draw the equivalent *s*-domain resistive circuit when S is switched to position Q. Use variables R_1, R_2, C_0 for the passive elements. Use latex-tikz.

Solution:

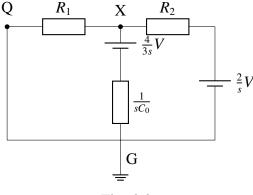


Fig. 3.2

3. $V_{C_0}(s) = ?$ Solution: Applying KCL at node X in Fig. 3.2

$$\frac{V-0}{R_1} + \frac{V-\frac{2}{s}}{R_2} + sC_0\left(V - \frac{4}{3s}\right) = 0$$
 (3.3)

$$\implies V_{C_0}(s) = \frac{\frac{2}{sR_2} + \frac{4C_0}{3}}{\frac{1}{R_1} + \frac{2}{R_2} + sC_0}$$
 (3.4)

$$V_{C_0}(s) = \frac{4}{3} \left(\frac{1}{\frac{1}{R_1} \left(\frac{1}{R_1} + \frac{1}{R_2}\right) + s}\right)$$

$$+\frac{2}{R_2\left(\frac{1}{R_1}+\frac{1}{R_2}\right)}\left(\frac{1}{s}-\frac{1}{\frac{1}{C_0}\left(\frac{1}{R_1}+\frac{1}{R_2}\right)+s}\right) \quad (3.5)$$

4. $v_{C_0}(t) = ?$ Plot using python. Taking an inverse Laplace Transform,

$$v_{C_0}(t) = \frac{4}{3}e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}u(t) + \frac{2}{R_2\left(\frac{1}{R_1} + \frac{1}{R_2}\right)}\left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}\right)u(t)$$
 (3.6)

Substituting values gives

$$v_{C_0}(t) = \frac{2}{3} \left(1 + e^{-\left(1.5 \times 10^6\right)t} \right) u(t) \tag{3.7}$$

The python code used to plot Fig. 3.3

https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e3.4.py

Verify your result using ngspice.
 Solution: The following ngspice script simulates the given circuit

https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e3.cir

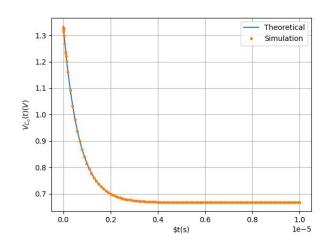


Fig. 3.3: $v_{C_0}(t)$ after the switch is flipped

6. Find $v_{C_0}(0-), v_{C_0}(0+)$ and $v_{C_0}(\infty)$. **Solution:**

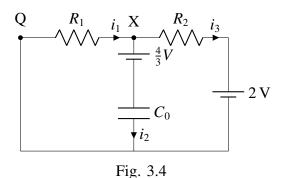
$$v_{C_0}(0-) = \lim_{t \to 0-} v_{C_0}(t) = \frac{q_1}{C} = \frac{4}{3} \text{ V}$$
 (3.8)

$$v_{C_0}(0+) = \lim_{t \to 0+} v_{C_0}(t) = \frac{4}{3} V$$
 (3.9)

$$v_{C_0}(\infty) = \lim_{t \to \infty} v_{C_0}(t) = \frac{2}{3} V$$
 (3.10)

7. Obtain the Fig. in problem 3.2 using the equivalent differential equation.

Solution: The equivalent circuit in the *t*-domain is shown below.



From KCL and KVL,

$$i_1 = i_2 + i_3 \tag{3.11}$$

$$i_1 R_1 + \frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt = 0$$
 (3.12)

$$\frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt - i_3 R_2 - 2 = 0$$
 (3.13)

(3.14)

Taking Laplace Transforms on both sides and using the properties of Laplace Transforms,

Let
$$i(t) \stackrel{\mathcal{L}}{\longleftrightarrow} I(s)$$
 (3.15)

$$\implies I_1 = I_2 + I_3 \tag{3.16}$$

$$I_1R_1 + \frac{4}{3} + \frac{1}{sC_0}I_2 = 0$$
 (3.17)

$$\frac{4}{3} + \frac{1}{sC_0}I_2 - I_3R_2 - 2 = 0 \tag{3.18}$$

4 BILINEAR TRANSFORM

1. In Fig. 2.1, consider the case when S is switched to Q right in the beginning. Formulate the differential equation.

Solution: The equivalent circuit in the *t*-domain is shown below.

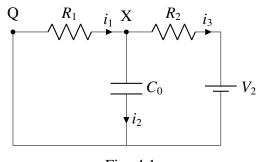


Fig. 4.1

Applying KCL and KVL,

$$i_1 = i_2 + i_3 \tag{4.1}$$

$$i_1 R_1 + \frac{1}{C_0} \int_0^t i_2 \, dt = 0 \tag{4.2}$$

$$i_3R_2 + 2 - \frac{1}{C_0} \int_0^t i_2 dt = 0$$
 (4.3)

Differentiating the above equations,

$$\frac{\mathrm{d}i_1}{\mathrm{d}t} = \frac{\mathrm{d}i_2}{\mathrm{d}t} + \frac{\mathrm{d}i_3}{\mathrm{d}t} \tag{4.4}$$

$$R_1 \frac{\mathrm{d}i_1}{\mathrm{d}t} + \frac{i_2}{C_0} = 0 \tag{4.5}$$

$$R_2 \frac{\mathrm{d}i_3}{\mathrm{d}t} - \frac{i_2}{C_0} = 0 \tag{4.6}$$

Using (4.4) and (4.6) in (4.5),

$$R_1 \left(\frac{\mathrm{d}i_2}{\mathrm{d}t} + \frac{\mathrm{d}i_3}{\mathrm{d}t} \right) + \frac{i_2}{C_0} = 0$$
 (4.7)

$$R_1 \frac{\mathrm{d}i_2}{\mathrm{d}t} + \left(1 + \frac{R_1}{R_2}\right) \frac{i_2}{C_0} = 0$$
 (4.8)

$$\frac{\mathrm{d}i_2}{\mathrm{d}t} + \left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{i_2}{C_0} = 0\tag{4.9}$$

$$\frac{di_2}{dt} + \frac{i_2}{\tau} = 0 {(4.10)}$$

where $\tau = \frac{C_0 R_1 R_2}{R_1 + R_2}$ is the RC time constant of the circuit. Note that $i_2(0) = \frac{V_2}{R_2}$ A and $i_2 = C_0 \frac{dV}{dt}$, where V is the voltage of the capacitor. Hence, integrating (4.10),

$$C_0 \frac{\mathrm{d}V}{\mathrm{d}t} - \frac{V_2}{R_2} + \frac{C_0 V}{\tau} = 0 \tag{4.11}$$

$$\implies \frac{\mathrm{d}V}{\mathrm{d}t} + \frac{V}{\tau} = \frac{V_2}{C_0 R_2} \tag{4.12}$$

2. Find H(s) considering the outur voltage at the capacitor.

Solution:

$$H(s) = \frac{V_{C_0}(s)}{V_2(s)} \tag{4.13}$$

In the s domain:

$$\frac{V_{C_0}}{R_1} + \frac{V_{C_0}}{\frac{1}{sC_0}} + \frac{V_{C_0} - V_2}{R_2} = 0 {(4.14)}$$

$$H(s)\left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right) = \frac{1}{R_2}$$
 (4.15)

$$H(s) = \frac{\frac{1}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2} + sC_0} = \frac{1}{3 + sC_0}$$
 (4.16)

3. Plot H(s). What kind of filter is it?

It is a Low-Pass filter

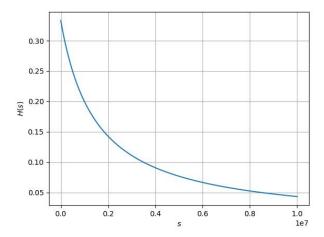


Fig. 4.2: Plot of H(s).

4. Using trapezoidal rule for integration, formulate the difference equation by considering

$$y(n) = y(t)|_{t=n}$$
 (4.17)

Solution:

- 5. Find H(z).
- 6. How can you obtain H(z) from H(s)? **Solution:** Apply a Bilinear Transform

$$s \to \frac{2}{T} \frac{z-1}{z+1} \tag{4.18}$$