

# Circuits and Transforms

VIBHAVASU

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**Abstract**—This manual provides a simple introduction to Transforms

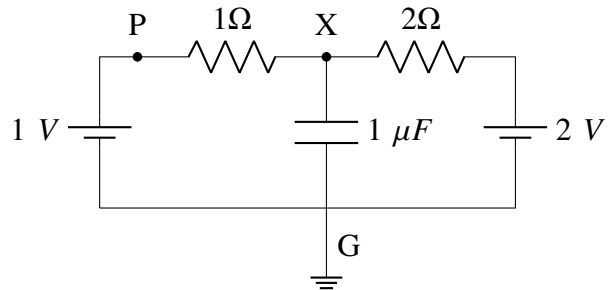


Fig. 2.2

## 1 DEFINITIONS

1. The unit step function is

$$u(t) = \begin{cases} 1 & t > 0 \\ \frac{1}{2} & t = 0 \\ 0 & t < 0 \end{cases} \quad (1.1)$$

2. The Laplace transform of  $g(t)$  is defined as

$$G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt \quad (1.2)$$

## 2 LAPLACE TRANSFORM

1. In the circuit, the switch S is connected to position P for a long time so that the charge on the capacitor becomes  $q_1 \mu C$ . Then S is switched to position Q. After a long time, the charge on the capacitor is  $q_2 \mu C$ .

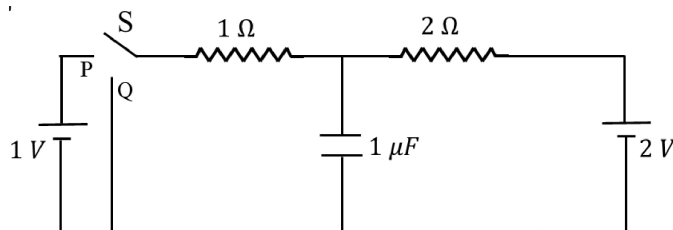


Fig. 2.1

2. Draw the circuit using latex-tikz.

**Solution:**

3. Find  $q_1$ .

**Solution:** After a long time, the capacitor starts to behave like an open switch, which means that no current will flow through the capacitor. Assume that the circuit is grounded at the negative terminals of the battery and current in the circuit is  $i$ . Applying KVL in the loop:

$$1 + i + 2i - 2 = 0 \quad (2.1)$$

$$\Rightarrow i = \frac{1}{3} A \quad (2.2)$$

$$\frac{q_1 \mu}{C} = 1 + \frac{1}{3} \quad (2.3)$$

$$\Rightarrow q_1 = \frac{4}{3} \quad (2.4)$$

4. Find  $q_1$ .

**Solution:**

After infinite time with switch at P

The capacitor is charged

Applying KCL at X:

$$\frac{V_x - 1}{1} = \frac{2 - V_x}{2} \quad (2.5)$$

$$\Rightarrow V_x = \frac{4}{3} V \quad (2.6)$$

$$q_1 = CV = 1 \mu C \quad (2.7)$$

5. Show that the Laplace transform of  $u(t)$  is  $\frac{1}{s}$  and find the ROC.

**Solution:**

$$\mathcal{U}(s) = \int_0^{\infty} u(t)e^{-st} dt \quad (2.8)$$

$$= \int_0^0 \frac{1}{2} e^{-st} dt + \int_0^{\infty} e^{-st} dt = \frac{1}{s} \quad (2.9)$$

$$\text{R.O.C: } \text{Re}(s) > 0 \quad (2.10)$$

6. Show that

$$e^{-at}u(t) \xleftrightarrow{\mathcal{L}} \frac{1}{s+a}, \quad a > 0 \quad (2.11)$$

and find the ROC.

**Solution:**

$$e^{-at}u(t) \xleftrightarrow{\mathcal{L}} \int_0^{\infty} u(t)e^{-(s+a)t} dt \quad (2.12)$$

$$= \frac{1}{s+a} \quad (2.13)$$

$$\text{R.O.C: } \text{Re}(s) > -a \quad (2.14)$$

7. Now consider the following resistive circuit transformed from Fig. 2.1 where

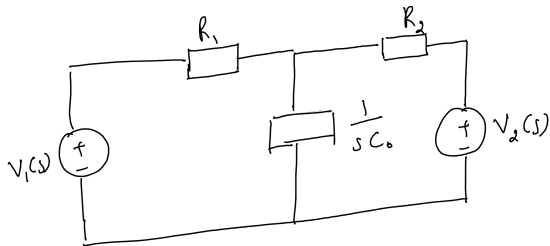


Fig. 2.3

$$u(t) \xleftrightarrow{\mathcal{L}} V_1(s) \quad (2.15)$$

$$2u(t) \xleftrightarrow{\mathcal{L}} V_2(s) \quad (2.16)$$

Find the voltage across the capacitor  $V_{C_0}(s)$ .

**Solution:** Applying KCL at X:

$$\frac{V_x - \frac{1}{s}}{R_1} + s(V_{C_0}) = \frac{\frac{2}{s} - V_x}{R_2} \quad (2.17)$$

$$V(s) = \frac{\frac{1}{R_1} + \frac{2}{R_2}}{s\left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right)} \quad (2.18)$$

$$= \frac{2R_1 + R_2}{R_1 + R_2} \left( \frac{1}{s} - \frac{1}{\frac{1}{C_0}\left(\frac{1}{R_1} + \frac{1}{R_2}\right) + s} \right) \quad (2.19)$$

$$= \frac{4}{3} \left( \frac{1}{s} - \frac{1}{\frac{3}{2C_0} + s} \right) \quad (2.20)$$

8. Find  $v_{C_0}(t)$ . Plot using python.

$$v_{C_0}(t) = \frac{2R_1 + R_2}{R_1 + R_2} u(t) \left( 1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} \right) \quad (2.21)$$

$$v_{C_0}(t) = \frac{4}{3} \left( 1 - e^{-(1.5 \times 10^6)t} \right) u(t) \quad (2.22)$$

9. Verify your result using ngspice.

**Solution:**

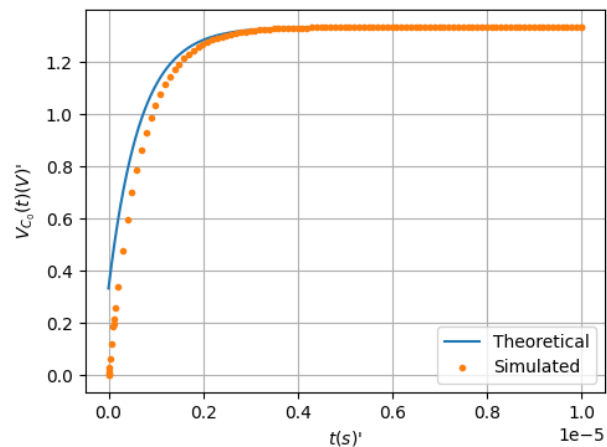


Fig. 2.4:  $v_{C_0}(t)$  before the switch is flipped

10. Obtain Fig. 2.3 using the equivalent differential equation.

### 3 INITIAL CONDITIONS

1. Find  $q_2$  in Fig. 2.1.

**Solution:** The circuit at steady state when the

switch is at Q:

At steady state: Capacitor is charged

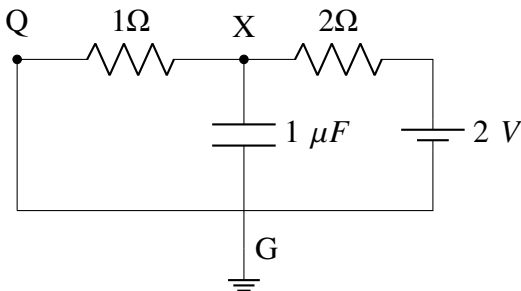


Fig. 3.1

Applying KCL at X.

$$\frac{V-0}{1} + \frac{V-2}{2} = 0 \quad (3.1)$$

$$\Rightarrow V = \frac{2}{3} \text{ V} \quad q_2 = \frac{2}{3} \mu\text{C} \quad (3.2)$$

2. Draw the equivalent  $s$ -domain resistive circuit when S is switched to position Q. Use variables  $R_1, R_2, C_0$  for the passive elements. Use latex-tikz.

**Solution:**

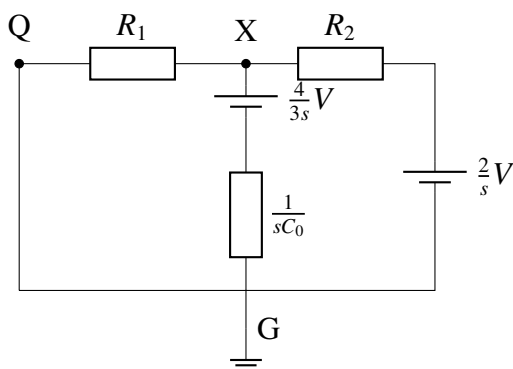


Fig. 3.2

Applying KCL at node X in Fig. 3.2

$$\frac{V-0}{R_1} + \frac{V-\frac{2}{s}}{R_2} + sC_0 \left( V - \frac{4}{3s} \right) = 0 \quad (3.3)$$

$$\Rightarrow V_{C_0}(s) = \frac{\frac{2}{sR_2} + \frac{4C_0}{3}}{\frac{1}{R_1} + \frac{2}{R_2} + sC_0} \quad (3.4)$$

$$V_{C_0}(s) = \frac{4}{3} \left( \frac{1}{\frac{1}{C_0} \left( \frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right) + \frac{2}{R_2 \left( \frac{1}{R_1} + \frac{1}{R_2} \right)} \left( \frac{1}{s} - \frac{1}{\frac{1}{C_0} \left( \frac{1}{R_1} + \frac{1}{R_2} \right) + s} \right) \quad (3.5)$$

4.  $v_{C_0}(t) = ?$  Plot using python. Taking an inverse Laplace Transform,

$$v_{C_0}(t) = \frac{4}{3} e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} u(t) + \frac{2}{R_2 \left(\frac{1}{R_1} + \frac{1}{R_2}\right)} \left( 1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} \right) u(t) \quad (3.6)$$

Substituting values gives

$$v_{C_0}(t) = \frac{2}{3} \left( 1 + e^{-(1.5 \times 10^6)t} \right) u(t) \quad (3.7)$$

The python code used to plot Fig. 3.3

<https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e3.4.py>

5. Verify your result using ngspice.

**Solution:** The following ngspice script simu-

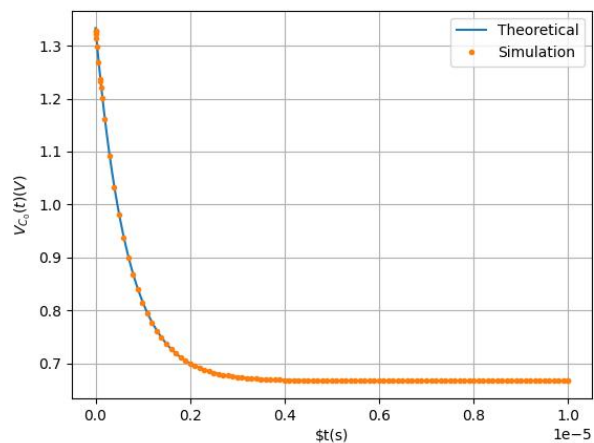


Fig. 3.3:  $v_{C_0}(t)$  after the switch is flipped

lates the given circuit

<https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e3.cir>

3.  $V_{C_0}(s) = ?$  **Solution:**

6. Find  $v_{C_0}(0-)$ ,  $v_{C_0}(0+)$  and  $v_{C_0}(\infty)$ .

**Solution:**

$$v_{C_0}(0-) = \lim_{t \rightarrow 0-} v_{C_0}(t) = \frac{q_1}{C} = \frac{4}{3} \text{ V} \quad (3.8)$$

$$v_{C_0}(0+) = \lim_{t \rightarrow 0+} v_{C_0}(t) = \frac{4}{3} \text{ V} \quad (3.9)$$

$$v_{C_0}(\infty) = \lim_{t \rightarrow \infty} v_{C_0}(t) = \frac{2}{3} \text{ V} \quad (3.10)$$

7. Obtain the Fig. in problem 3.2 using the equivalent differential equation.

**Solution:** The equivalent circuit in the  $t$ -domain is shown below.

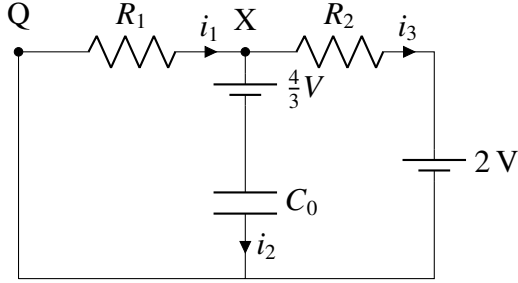


Fig. 3.4

From KCL and KVL,

$$i_1 = i_2 + i_3 \quad (3.11)$$

$$i_1 R_1 + \frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt = 0 \quad (3.12)$$

$$\frac{4}{3} + \frac{1}{C_0} \int_0^t i_2 dt - i_3 R_2 - 2 = 0 \quad (3.13)$$

Taking Laplace Transforms on both sides and using the properties of Laplace Transforms,

$$\text{Let } i(t) \xrightarrow{\mathcal{L}} I(s) \quad (3.14)$$

$$\Rightarrow I_1 = I_2 + I_3 \quad (3.15)$$

$$I_1 R_1 + \frac{4}{3} + \frac{1}{s C_0} I_2 = 0 \quad (3.16)$$

$$\frac{4}{3} + \frac{1}{s C_0} I_2 - I_3 R_2 - 2 = 0 \quad (3.17)$$

#### 4 BILINEAR TRANSFORM

1. In Fig. 2.1, consider the case when  $S$  is switched to  $Q$  right in the beginning. Formulate the differential equation.

**Solution:** The equivalent circuit in the  $t$ -domain is shown below.

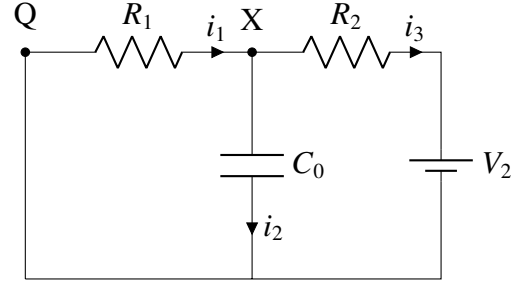


Fig. 4.1

Applying KCL and KVL,

$$i_1 = i_2 + i_3 \quad (4.1)$$

$$i_1 R_1 + \frac{1}{C_0} \int_0^t i_2 dt = 0 \quad (4.2)$$

$$i_3 R_2 + 2 - \frac{1}{C_0} \int_0^t i_2 dt = 0 \quad (4.3)$$

Differentiating the above equations,

$$\frac{di_1}{dt} = \frac{di_2}{dt} + \frac{di_3}{dt} \quad (4.4)$$

$$R_1 \frac{di_1}{dt} + \frac{i_2}{C_0} = 0 \quad (4.5)$$

$$R_2 \frac{di_3}{dt} - \frac{i_2}{C_0} = 0 \quad (4.6)$$

Using (4.4) and (4.6) in (4.5),

$$R_1 \left( \frac{di_2}{dt} + \frac{di_3}{dt} \right) + \frac{i_2}{C_0} = 0 \quad (4.7)$$

$$R_1 \frac{di_2}{dt} + \left( 1 + \frac{R_1}{R_2} \right) \frac{i_2}{C_0} = 0 \quad (4.8)$$

$$\frac{di_2}{dt} + \left( \frac{1}{R_1} + \frac{1}{R_2} \right) \frac{i_2}{C_0} = 0 \quad (4.9)$$

$$\frac{di_2}{dt} + \frac{i_2}{\tau} = 0 \quad (4.10)$$

where  $\tau = \frac{C_0 R_1 R_2}{R_1 + R_2}$  is the RC time constant

$i_2(0) = \frac{V_2}{R_2}$  and  $i_2 = C_0 \frac{dV}{dt}$ , where  $V = V_{C_0}$  is the voltage of the capacitor.

$$C_0 \frac{dV}{dt} - \frac{V_2}{R_2} + \frac{C_0 V}{\tau} = 0 \quad (4.11)$$

$$\Rightarrow \frac{dV}{dt} + \frac{V}{\tau} = \frac{V_2}{C_0 R_2} \quad (4.12)$$

2. Find  $H(s)$  considering the output voltage at the

capacitor.

**Solution:**

$$H(s) = \frac{V_{C_0}(s)}{V_2(s)} \quad (4.13)$$

In the  $s$  domain:

$$\frac{V_{C_0}}{R_1} + \frac{V_{C_0}}{\frac{1}{sC_0}} + \frac{V_{C_0} - V_2}{R_2} = 0 \quad (4.14)$$

$$H(s) \left( \frac{1}{R_1} + \frac{1}{R_2} + sC_0 \right) = \frac{1}{R_2} \quad (4.15)$$

$$H(s) = \frac{\frac{1}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2} + sC_0} = \frac{1}{3 + sC_0} \quad (4.16)$$

3. Plot  $H(s)$ . What kind of filter is it?

<https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e4.3.py>

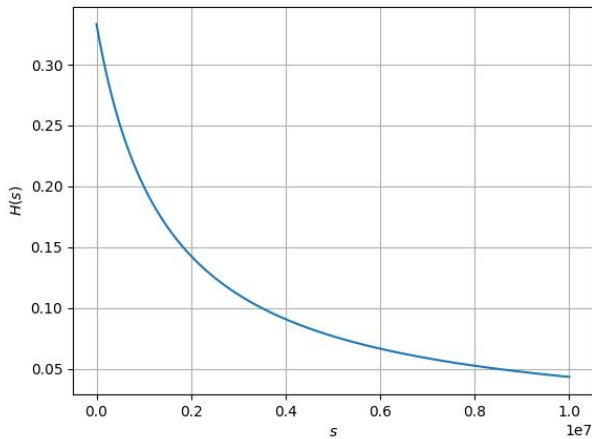


Fig. 4.2: Plot of  $H(s)$ .

It is a Low-Pass filter

4. Using trapezoidal rule for integration, formulate the difference equation by considering

$$y(n) = y(t)|_{t=n} \quad (4.17)$$

**Solution:**

$$v_{in} = v_2(n) = 2u(n) \text{ and } v_{out} = v(n)$$

Integrating (4.12) from  $n$  to  $n+1$

and using (2.16):

$$\begin{aligned} v(n+1) - v(n) + \frac{v(n+1) + v(n)}{2\tau} \\ = \frac{2}{C_0 R_2} \frac{(u(n+1) + u(n))}{2} \end{aligned} \quad (4.18)$$

$$\begin{aligned} v(n+1)(2\tau + 1) + v(n)(1 - 2\tau) \\ = \frac{2}{C_0 R_2} \tau (u(n+1) + u(n)) \end{aligned} \quad (4.19)$$

$$\begin{aligned} v(n)(2\tau + 1) + v(n-1)(1 - 2\tau) \\ = \frac{2}{C_0 R_2} \tau (u(n) + u(n-1)) \end{aligned} \quad (4.20)$$

5. Find  $H(z)$ .

**Solution:**

$$H(z) = \frac{V(z)}{V_2(z)} \quad (4.21)$$

We know that

$$v_2(n) = 2u(n) \implies V_2(z) = \frac{2}{1 - z^{-1}} \quad (4.22)$$

Applying  $\mathcal{Z}$ -transform on (4.20)

$$V(z)(2\tau + 1) + z^{-1}V(z)(1 - 2\tau) = \frac{2\tau(1 + z^{-1})}{C_0 R_2(1 - z^{-1})} \quad (4.23)$$

$$V(z)(2\tau + 1 - z^{-1}(2\tau - 1)) = \frac{\tau V_2(z)(1 + z^{-1})}{C_0 R_2} \quad (4.24)$$

$$H(z) = \frac{V(z)}{V_2(z)} = \frac{\tau(1 + z^{-1})}{C_0 R_2(2\tau + 1 - z^{-1}(2\tau - 1))} \quad (4.25)$$

$$H(z) = \frac{\tau(1 + z^{-1})}{R_2 C_0 (2\tau + 1 - z^{-1}(2\tau - 1))} \quad (4.26)$$

$$= \frac{\tau(z + 1)}{R_2 C_0 (2\tau(z - 1) + z + 1))} \quad (4.27)$$

$$= \frac{1}{2C_0 \left( 2\frac{z-1}{z+1} + \frac{1}{\tau} \right)} \quad (4.28)$$

But  $\tau = \frac{C_0 R_1 R_2}{R_1 + R_2} = \frac{2C_0}{3}$  and  $R_2 = 2$

$$\Rightarrow H(z) = \frac{1}{2C_0 \left( 2 \frac{z-1}{z+1} + \frac{3}{2C_0} \right)} \quad (4.29)$$

$$H(z) = \frac{1}{3 + 4C_0 \frac{z-1}{z+1}} \quad (4.30)$$

R.O.C:  $|z| < 1$

6. How can you obtain  $H(z)$  from  $H(s)$ ?

**Solution:**

Apply a Bilinear Transform

$$s \rightarrow \frac{2}{T} \frac{z-1}{z+1} \quad (4.31)$$

$$H(z) = \frac{1}{3 + C_0 \frac{2}{T} \frac{z-1}{z+1}} \quad (4.32)$$

Putting  $T = \frac{1}{2}$  will give (4.30)

7. Find  $v(n)$ . Verify using ngspice and differential equation

**Solution:**

$$V(z) = H(z)V_2(z) \quad (4.33)$$

$$= \frac{1}{3 + \frac{2}{T} C_0 \frac{1-z^{-1}}{1+z^{-1}}} \frac{2}{1-z^{-1}} \quad (4.34)$$

$$= \frac{2}{1-z^{-1}} \frac{\frac{2}{T} C_0 (1+z^{-1})}{3 \left[ 3(1+z^{-1}) + \frac{2}{T} C_0 (1-z^{-1}) \right]} \quad (4.35)$$

$$= \frac{2}{3(1-z^{-1})} - \frac{2 \frac{\tau}{T}}{3 \left[ (1+z^{-1}) + 2 \frac{\tau}{T} (1-z^{-1}) \right]} \quad (4.36)$$

$$= \frac{2}{3(1-z^{-1})} - \frac{2\tau}{3 \left[ 2\tau + 1 - z^{-1}(2\tau - 1) \right]} \quad (4.37)$$

$$= \frac{2}{3(1-z^{-1})} - \frac{2 \frac{\tau}{T}}{2 \frac{\tau}{T} + 1} \frac{1}{3 \left[ 1 - z^{-1} \left[ \frac{2 \frac{\tau}{T} - 1}{2 \frac{\tau}{T} + 1} \right] \right]} \quad (4.38)$$

$$= \frac{2}{3(1-z^{-1})} - \frac{2 \frac{\tau}{T}}{2 \frac{\tau}{T} + 1} \frac{1}{3 \left[ 1 - z^{-1} \left[ \frac{2\tau - T}{2\tau + T} \right] \right]} \quad (4.39)$$

Taking Inverse  $\mathcal{Z}$  transform on both sides

$$v(n) = \left[ \frac{2}{3} - \frac{2\tau}{3(2\tau + T)} \left( \frac{2\tau + T}{2\tau - T} \right)^n \right] u(n) \quad (4.40)$$

Python code used to plot can be found at:

<https://github.com/DarkWake9/EE3900/blob/main/cktsig/codes/e4.7.py>

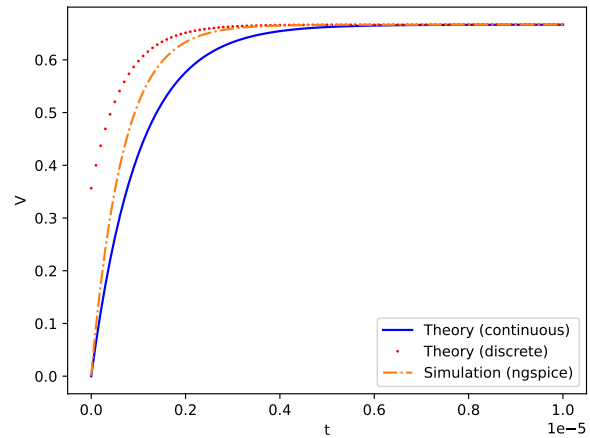


Fig. 4.3