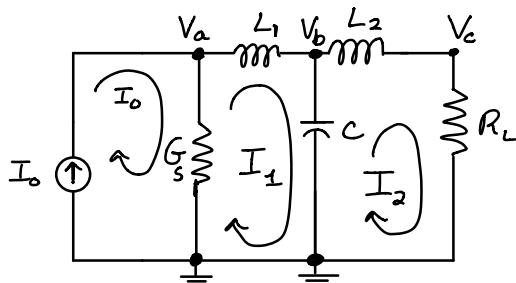


4.3.b) Replacing Voltage source and internal resistance to a current source w/ a paralleled resistance:



Starting w/ the mesh equations:

Path of I_o :

Not included because it is connected to Current source

Path of I_1 :

$$-L_1 \frac{dI_1}{dt} - \frac{1}{C} \int (I_1 - I_2) dt - \frac{1}{G_s} (I_1 - I_o) = 0$$

Path of I_2 :

$$-R_L I_2 - L_2 \frac{dI_2}{dt} - \frac{1}{C} \int (I_2 - I_1) dt = 0$$

\therefore Loop/mesh equations derived from KVL still hold true

Node equations:

Current in = '+', Current out = '-'

at Node a:

$$I_o - \frac{1}{L_1} \int (V_b - V_a) dt + (V_a - 0) G_s = 0 \quad * \quad G_s = \frac{1}{R_s}$$

$$I_o + \frac{1}{L_1} \int (V_a - V_b) dt + \frac{V_a}{R_s} = 0$$

at node b:

$$\frac{1}{L_1} \int (V_b - V_a) dt + C \frac{d}{dt} (V_b - 0) - \frac{1}{L_2} \int (V_b - V_c) dt = 0$$

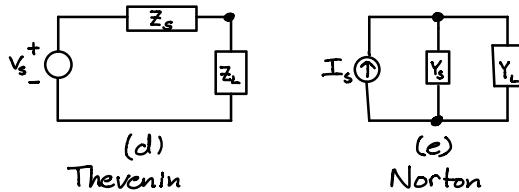
$$\frac{1}{L_1} \int (V_b - V_a) dt + C \frac{dV_b}{dt} + \frac{1}{L_2} \int (V_c - V_b) dt = 0$$

at node c:

$$+ \frac{(V_c - 0)}{R_L} - \frac{1}{L_2} \int (V_b - V_c) dt = 0$$

$$\frac{V_c}{R_L} + \frac{1}{L_2} \int (V_c - V_b) dt = 0$$

4.3c) Figure 4.3d and e are:



$$* Y_s \equiv Z_s^{-1} \rightarrow I_s = V_s Y_s$$

From Figure (d)

The current across load Z_L :

$$I_L = I_s = \frac{V_s}{Z_s + Z_L}$$

Voltage across load is given by voltage divider formula:

$$V_L = V_s \left(\frac{Z_L}{Z_L + Z_s} \right)$$

From Figure e:

Voltage delivered is given from:

$$V_s = I_s Z_{\text{tot}} = I_s (Z_s \parallel Z_L)$$

$$V_L = I_s \left(\frac{Z_L Z_s}{Z_L + Z_s} \right) * \text{due to both in parallel}$$

$$V_L = V_s \cancel{Y_s} \left(\frac{Z_L Z_s}{Z_L + Z_s} \right) * \text{using relationship: } I_s = V_s Y_s \quad \& \quad Z_s = \frac{1}{Y_s}$$

$$V_L = V_s \left(\frac{Z_L}{Z_L + Z_s} \right)$$

Current delivered to load is:

$$I_L = \frac{V_L}{Z_L} = \frac{V_s}{Z_L} \left(\frac{Z_L}{Z_L + Z_s} \right) = \frac{V_s}{Z_L + Z_s}$$

\therefore Current & Voltage formulas are equivalent

4.3.d) For fixed V_s & Z_s , where Z_s is a conjugate match: $Z_L = Z_s^*$

$$P = \frac{1}{2} R |I|^2 \text{ for complex circuits}$$

Power delivered to load depends on current delivered to load:

$$P_L = R_L |I_L|^2$$

Using figure (d) Thevenin equivalent:

$$I_L = I_s = \frac{V_s}{Z_L + Z_s} * Z = R + jX$$

$$I_L = \frac{V_s}{(R_L + R_s) + j(X_s + X_L)}$$

$$I_L^* = \frac{V_s}{(R_L + R_s) - j(X_s + X_L)}$$

$$P_L = \frac{R_L}{2} |I_L|^2 = \frac{R_L}{2} \left(\frac{V_s}{(R_L + R_s)^2 + (X_s + X_L)^2} \right)$$

* $Z_L = R_L + jX_L$, minimizing each individually:

$$\frac{\partial P_L}{\partial X_L} = \frac{V_s R_L}{2} \frac{\partial}{\partial X_L} \left(\frac{1}{(R_L + R_s)^2 + (X_s + X_L)^2} \right) = 0$$

$$= \frac{V_s R_L}{2} \left(\frac{-2(X_s + X_L)}{[(R_L + R_s)^2 + (X_s + X_L)^2]^2} \right) = 0$$

$$X_L = -X_s$$

Minimizing R_L

$$\frac{\partial P_L}{\partial R_L} = \left(\frac{V_s}{2} \right) \frac{\partial}{\partial R_L} \left(\frac{R_L}{(R_L + R_s)^2 + (X_s + X_L)^2} \right) = 0$$

$$\frac{V_s}{2} \left(\frac{1}{(R_L + R_s)^2 + (X_s + X_L)^2} + \frac{R_L (-1)(2(R_L + R_s))}{[(R_L + R_s)^2 + (X_s + X_L)^2]^2} \right) = 0$$

$$1 - \frac{2R_L(R_L + R_s)}{(R_L + R_s)^2 + (X_s + X_L)^2} = 0$$

$$R_L(R_L + R_s) = [(R_L + R_s)^2 + (X_s + X_L)^2] \leq * \min X_L \text{ is when } X_L = -X_s$$

$$R_L(R_L + R_s) = (R_L + R_s)^2 \frac{1}{2}$$

$$2R_L = R_L + R_s \\ \therefore R_L = R_s$$

\therefore for P_L to be at max value :

$$Z_L = R_s - jX_s = Z_s^*$$

4.6c) Figure 3.17



From Faradays Law

$$\oint \vec{E} \cdot d\vec{l} = -\frac{\partial \Phi_B}{\partial t}$$

$$\oint \vec{E} \cdot d\vec{l} = -\frac{\partial}{\partial t} \int \vec{B} \cdot d\vec{A}$$

$$\int_A^B \vec{E} \cdot d\vec{l} + \int_B^C \vec{E} \cdot d\vec{l} + \int_C^D \vec{E} \cdot d\vec{l} + \int_D^A \vec{E} \cdot d\vec{l} = -\frac{\partial}{\partial t} \int \vec{B} \cdot d\vec{A}$$

in region between coaxial cable :

if considering external inductance independent of time :

$$\vec{B} = \vec{B}_0 e^{j\omega t} \rightarrow -\frac{\partial \vec{B}}{\partial t} = -j\omega \vec{B}$$

$$-\frac{\partial \vec{B}}{\partial t} = -j\omega \vec{B} = j\omega L_e \cdot I \quad * \text{external inductance per unit length due to } B\text{-field}$$

Left hand integrals can be related to Potential difference across points :

$$V_{AB} + V_{BC} + V_{CD} + V_{DA} = j\omega L_e I \Delta z$$

Finding V_{CD} :

$$E_z = I (Z_{int})_{outer} \quad * (Z_{int})_{outer} = \text{internal impedance of outer conductor}$$

where internal impedance of a conductor is given as

$$\frac{Z_{in}}{\Delta z} = \left(\frac{R_s}{2\pi f_0} + \frac{j\omega L_s}{2\pi f_0} \right) \frac{\Omega}{m}$$

at path $C \rightarrow D$, $r_o = b$

$$(Z_{int})_{outer} = \left(\frac{R_{s_b}}{2\pi b} + \frac{j\omega L_{s_b}}{2\pi b} \right) \frac{\Omega}{m}$$

$\therefore V_{CD} = -I (Z_{int})_{outer} \Delta z \quad * \text{KVL says going w/ path of current, gives neg. Voltage}$

Similarly one can obtain an expression of impedance for path between $A \rightarrow B$

$$V_{AB} = I (Z_{int})_{inner} \Delta z = -I \left(\frac{R_s}{2\pi a} + \frac{j\omega L_s}{2\pi a} \right) \Delta z$$

Substituting into equation:

$$V_{DA} - I \left(\frac{R_{sb} + j\omega L_{ib}}{2\pi b} \right) \Delta z + V_{BC} - I \left(\frac{R_{sa} + j\omega L_{ia}}{2\pi a} \right) \Delta z = (-j\omega L_e I) \Delta z$$

$$I \left(\frac{R_{sb} + R_{sa}}{2\pi b} \right) \Delta z + I \left(\frac{j\omega L_{ib} + j\omega L_{ia}}{2\pi a} + j\omega L_e \right) \Delta z = V_{DA} + V_{BC}$$

$$V_{DA} + V_{BC} = (I R_{eq} + I L_{eq}) \Delta z \quad * \text{Req & Reg hold same formula as circuit diagram}$$

* Inductors & Resistors add in series

where Δz is dependent on length of path, as $\Delta z \rightarrow 0$

$$V_{DA} + V_{BC} = 0, * \text{circuit diagram matches voltage gains/drops}$$

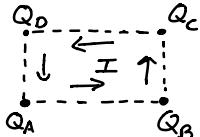
Considering paths $B \rightarrow C$ & $D \rightarrow A$

- Coaxial cylinder conductors act as capacitors and as such:

$$|V_{BC}| = \frac{Q_{BC}}{C}$$

$$|V_{DA}| = \frac{Q_{DA}}{C}$$

due to conservation of charge:



$$Q_{DA} = Q_{BC} * \text{current along path is continuous}$$

$$\therefore V_{DA} = -V_{BC} \Rightarrow V_{DA} = V_{CB}$$

\therefore Circuit equivalent of $\oint \vec{E} \cdot d\vec{l} = -\frac{\partial}{\partial t} \int B \cdot dA$

