Control Engineering (SC42095)

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Lecture outline

- Analysis of discrete-time systems
 - Stability (definitions, tests, Lyapunov method)
 - Controllability and reachability
 - Observability and detectability
 - Simulation

Stability: basic notions

Defined with respect to change in initial conditions.

The solution $x^0(k)$ of

$$x(k+1) = f(x(k), k)$$

is stable if for a given $\epsilon > 0$, there exists a $\delta(\epsilon, k_0) > 0$

such that all solutions x(k) with $||x(k_0) - x^0(k_0)|| < \delta$

are such that $||x(k) - x^{0}(k)|| < \epsilon$ for $k \ge k_0$.

Definition - Asymptotic stability

The solution $x^0(k)$ is asymptotically stable

if it is stable and δ can be chosen such that

$$||x(k_0) - x^0(k_0)|| < \delta$$
 implies that

$$||x(k) - x^{0}(k)|| \to 0 \text{ when } k \to \infty.$$

Linear discrete-time systems

$$x(k+1) = \Phi x(k), \quad x(0) = a$$

Stability of *one* solution implies stability of all solutions, i.e., stability of the system.

A linear discrete-time system is asymptotically stable if and only if all eigenvalues of Φ are strictly inside the unit disc $(|\lambda_i| < 1)$.

(Solutions $x(k) = \Phi^k x(0)$ are combinations of terms λ_i^k or $p_i(k)\lambda_i^k$.)

BIBO stability

A linear time-invariant system is BIBO stable if a bounded input gives a bounded output for all initial values.

Asymptotic stability is the strongest concept.

Asymptotically stable \Rightarrow BIBO stable,

but

stable \Rightarrow BIBO stable.

Stability tests

- Direct numerical or algebraic computation of $\lambda(\Phi)$ in MATLAB: eig(Phi).
- Methods based on properties of the characteristic polynomial (Jury's criterion).

$$A(z) = a_0 z^n + a_1 z^{n-1} + \ldots + a_n = 0$$

- Root-locus method, Nyquist criterion, Bode plot.
- Lyapunov method.

Lyapunov theory

$$x(k+1) = f(x(k)), \quad f(0) = 0$$

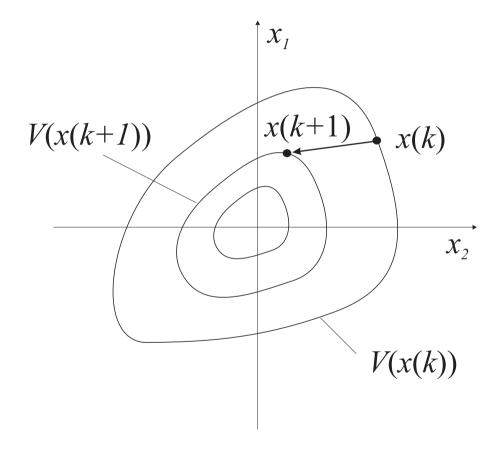
Monotonic convergence ||x(k+1)|| < ||x(k)|| – too strong condition for stability. Another measure: Lyapunov function:

- V(x) is continuous in x and V(0) = 0
- \bullet V(x) is positive definite
- \bullet $\Delta V(x) = V(f(x)) V(x)$ is negative definite

Existence of a Lyapunov function implies asymptotic stability for the solution x=0

Geometric interpretation

$$x(k+1) = f(x(k)), \quad f(0) = 0$$



Linear system

$$x(k+1) = \Phi x(k), \qquad V(x) = x^T P x, \qquad P > 0$$

$$\Delta V(k) = V(\Phi x) - V(x) = x^T \Phi^T P \Phi x - x^T P x$$

$$= x^T (\Phi^T P \Phi - P) x$$

V is a Lyapunov function iff there exists a P>0 that satisfies the $Lyapunov\ equation$

$$\Phi^T P \Phi - P = -Q, \qquad Q > 0$$

Controllability and reachability

Controllability: Possible to find a control sequence such that the origin can be reached from any initial state in finite time.

Reachability: Possible to find a control sequence such that an arbitrary state can be reached from any initial state in finite time.

With continuous-time systems, controllability implies reachability.

Not for discrete-time systems...

Controllability and reachability (cont'd)

$$x(k+1) = \Phi x(k) + \Gamma u(k), \quad x(0)$$
 given
$$y(k) = Cx(k)$$

At time n (n = order of the system):

$$x(n) = \Phi^{n} x(0) + \Phi^{n-1} \Gamma u(0) + \dots + \Gamma u(n-1)$$

$$x(n) = \Phi^{n} x(0) + W_{c} U$$

$$W_c = \begin{pmatrix} \Gamma & \Phi \Gamma & \dots & \Phi^{n-1} \Gamma \end{pmatrix}, \quad U = \begin{pmatrix} u^T (n-1) & \dots & u^T (0) \end{pmatrix}^T$$

System is reachable iff W_c has rank n W_c is called the controllability matrix

Reachability and coordinate transformation

$$\tilde{W}_c = \begin{pmatrix} \tilde{\Gamma} & \tilde{\Phi}\tilde{\Gamma} & \dots & \tilde{\Phi}^{n-1}\tilde{\Gamma} \end{pmatrix}
= \begin{pmatrix} T\Gamma & T\Phi T^{-1}T\Gamma & \dots & T\Phi^{n-1}T^{-1}T\Gamma \end{pmatrix}
= TW_c$$

⇒ reachability is not influenced by change in coordinates

Controllable canonical form

If W_c is nonsingular and Φ has the characteristic polynomial:

$$\det[\lambda I - \Phi] = \lambda^n + a_1 \lambda^{n-1} + \dots + a_n = 0$$

then there exists a transformation $T = \tilde{W}_c W_c^{-1}$ such that

$$z(k+1) = \begin{pmatrix} -a_1 - a_2 \dots -a_{n-1} - a_n \\ 1 & 0 & \dots & 0 & 0 \\ 0 & 1 & \dots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & 1 & 0 \end{pmatrix} z(k) + \begin{pmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} u(k)$$

$$y(k) = (b_1 \dots b_n) z(k)$$

Observability and detectability

The system

$$x(k+1) = \Phi x(k) + \Gamma u(k)$$
$$y(k) = Cx(k)$$

is observable if there is a finite k such that the knowledge of $u(0), \ldots, u(k-1)$ and $y(0), \ldots, y(k-1)$ is sufficient to determine the initial state x(0) of the system.

Observability and detectability (cont'd)

Assume u(k) = 0, iterate the system equation:

$$y(0) = Cx(0)$$

$$y(1) = Cx(1) = C\Phi x(0)$$

$$\vdots$$

$$y(n-1) = C\Phi^{n-1}x(0)$$

$$\begin{pmatrix} y(0) \\ y(1) \\ \vdots \\ y(n-1) \end{pmatrix} = \begin{pmatrix} C \\ C\Phi \\ \vdots \\ C\Phi^{n-1} \end{pmatrix} x(0)$$

Observability and detectability (cont'd)

The system is observable iff the observability matrix

$$W_o = \begin{pmatrix} C \\ C\Phi \\ \vdots \\ C\Phi^{n-1} \end{pmatrix}$$

has rank n.

A system is detectable if the unobservable states decay to the origin (the corresponding eigenvalues are stable).

Observability - Example

$$x(k+1) = \begin{pmatrix} 1.1 & -0.3 \\ 1 & 0 \end{pmatrix} x(k)$$
$$y(k) = (1 & -0.5)x(k)$$

Observability matrix

$$W_o = \begin{pmatrix} C \\ C\Phi \end{pmatrix} = \begin{pmatrix} 1 & -0.5 \\ 0.6 & -0.3 \end{pmatrix}$$

The rank of W_o is 1

The Kalman decomposition

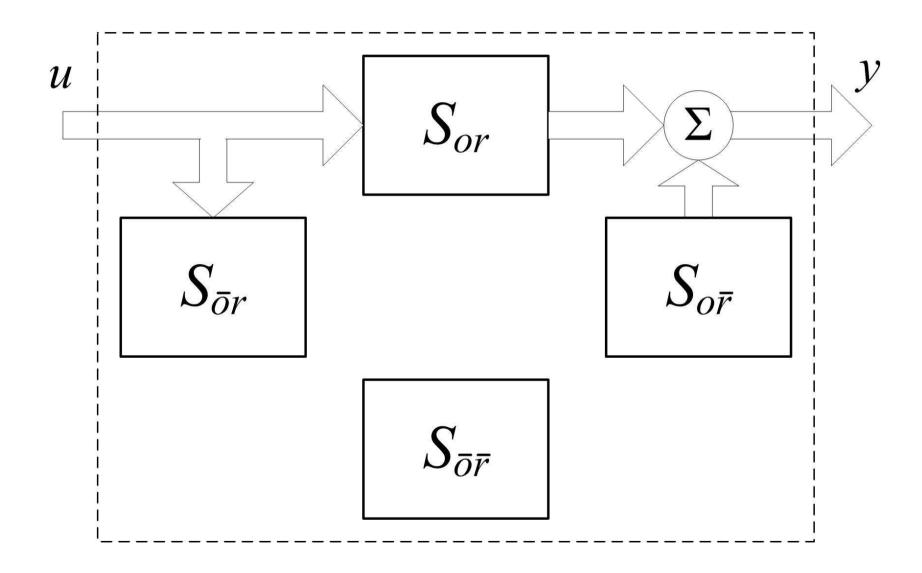
It is possible introduce coordinates that lead to this partitioning

$$x(k+1) = \begin{pmatrix} \Phi_{11} & \Phi_{12} & 0 & 0 \\ 0 & \Phi_{22} & 0 & 0 \\ \Phi_{31} & \Phi_{32} & \Phi_{33} & \Phi_{34} \\ 0 & \Phi_{42} & 0 & \Phi_{44} \end{pmatrix} x(k) + \begin{pmatrix} \Gamma_1 \\ 0 \\ \Gamma_3 \\ 0 \end{pmatrix} u(k)$$
$$y(k) = \begin{pmatrix} C_1 & C_2 & 0 & 0 \end{pmatrix} x(k)$$

Pulse-transfer operator:

$$H(q) = C_1 (qI - \Phi_{11})^{-1} \Gamma_1$$

The Kalman decomposition



Sampling -- reachability and observability

- \bullet Φ and Γ depend on h.
- To get reachable discrete-time system it is necessary that the continuous-time system is controllable.
- May happen that reachability is lost for some h
- Sampled-data system may be unobservable even if the continuoustime system is observable (hidden oscillations).

Hidden oscillations

Inter-sample ripple

Two cases:

- Oscillations seen in continuous-time output of an open-loop / closed-loop system.
 Loss of observability purely by sampling.
- 2. Oscillations seen in the control input, but not in the sampled output. Loss of observability by feedback, cancellation of poorly damped zeros by the controller.

Hidden oscillation in open-loop system

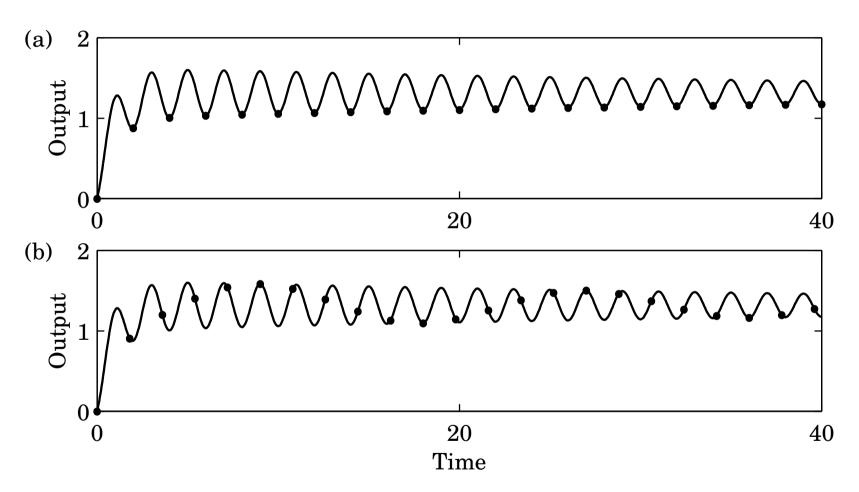
$$G(s) = \frac{1}{s+1} + \frac{\pi}{(s+0.02)^2 + \pi^2}$$

Sample with $h = 2, (\omega_s = \pi)$

$$H(z) = \frac{1 - a}{z - a} + \frac{0.0125}{z - \alpha}$$

where $a = e^{-2}$ and $\alpha = e^{-0.04}$

Hidden oscillation in open-loop system



a)
$$h = 2$$
, b) $h = 1.8$

Controller-induced hidden oscillation

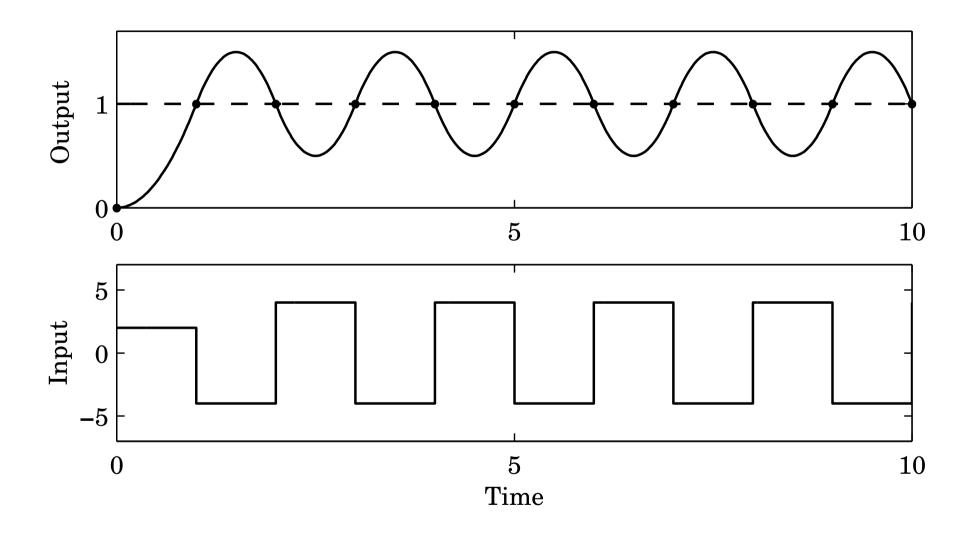
Consider a double integrator with the controller:

$$u(k) = \frac{2q}{q+1}u_c(k) - \frac{2(2q-1)}{q+1}y(k)$$

Closed-loop system

$$y(k) = \frac{q(q+1)}{(q+1)(q^2 - 2q + 1 - (-2q+1))} u_c(k)$$
$$= \frac{q(q+1)}{q^2(q+1)} u_c(k) = u_c(k-1)$$

Controller-induced hidden oscillation



How to detect hidden oscillations?

ullet Modified z-transform or solve system equation for times between sampling points

 Simulation. Check continuous-time output and control signal

Importance of simulation

- Simpler, cheaper, safer to experiment with a model.
- If the system does not exist.
- Investigate influence of parameter variations.

Simulation can never replace a field test.

Beware of numerical aspects, choice of integration method!

Summary

- Stability
- Controllability and reachability
- Observability and detectability
- Simulation