



**KALINGA INSTITUTE OF INDUSTRIAL TECHNOLOGY (KIIT)**

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## LABORATORY RECORD – AUTUMN 2020

MICROWAVE ENGINEERING LAB (EC 3015)

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ROLL NO: 1804373

Section: ETC-06

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<b>Experiment Number</b>	01
<b>Date of Experiment</b>	12/08/2020
<b>Date of Submission</b>	16/08/2020
<b>Name of the student</b>	Debagnik Kar
<b>Roll Number</b>	1804373
<b>Section</b>	ETC – 06

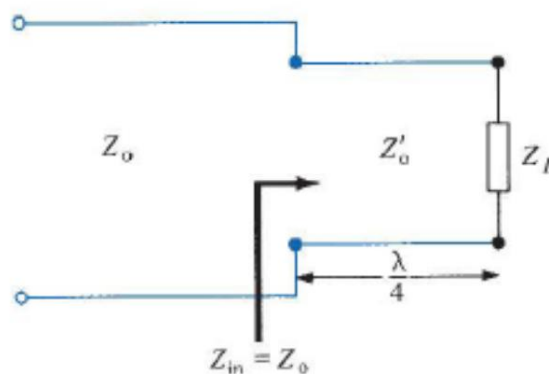
### **Aim of The Experiment :-**

To design a quarter wave transformer for matching a  $50\ \Omega$  microstrip line with a load of  $173\ \Omega$

### **Equipment / Software Required:-**

CST Studio Suite 2019 (Student Edition)

### **Theory:**



*Fig 1.1: Load matching using a quarter wave transformer*

When  $Z_0 \neq Z_L$ , the load is said to be mismatched and a reflected wave exist. So we use quarter wave transformer for impedance matching.

When  $l = \frac{\lambda}{4}$ ,

$$Z_{in} = Z_0 \left[ \frac{Z_L + \frac{jZ_0 \tan \pi}{2}}{Z_0 + \frac{jZ_L \tan \pi}{2}} \right] = \frac{Z_0^2}{Z_L}$$

(1)

A mismatched load can be properly matched to a line (with characteristic impedance  $Z_0$ ) by inserting prior to the transmission line  $\lambda/4$  long (with characteristic impedance  $Z'_0$ ) as depicted in Fig.1.

From (1),  $Z'_0$  is selected such that ( $Z_{in}=Z_0$ )

Therefore,

$$Z_0' = \sqrt{Z_0 Z_L} \quad (2)$$

Note: When microstrip line is used, then guided wavelength must be used ,i.e,

$$\lambda_g = \frac{\lambda_0}{\sqrt{\epsilon_{eff}}} \quad (3)$$

where,  $\lambda_g$ = guided wavelength.

When  $Z_0 \neq Z_L$ , the load is said to be mismatched and a reflected wave exist. So we use quarter wave transformer for impedance matching.

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$$Z_{in} = Z_0 \left[ \frac{Z_L + \frac{jZ_0 \tan \pi}{2}}{Z_0 + \frac{jZ_L \tan \pi}{2}} \right] = \frac{Z_0^2}{Z_L} \quad (1)$$

A mismatched load can be properly matched to a line (with characteristic impedance  $Z_0$  ) by inserting prior to the transmission line  $\lambda/4$  long (with characteristic impedance  $Z'_0$  ) as depicted in Fig.1.

From (1),  $Z'_0$  is selected such that ( $Z_{in}=Z_0$ )

Therefore,

$$Z_0' = \sqrt{(Z_0 Z_L)}$$

$$Z_0 = 50$$

$$Z_L = 173$$

$$Z_0' = \sqrt{(50 \times 173)}$$

$$Z_0' = 93.01 \, \Omega \quad (2)$$

Note: When microstrip line is used, then guided wavelength must be used ,i.e,

$$\lambda_g = \frac{\lambda_0}{\sqrt{\epsilon_{eff}}} \quad (3)$$

where,  $\lambda_g$ = guided wavelength.

**Substrate:** FR4 (Lossless) ( $\epsilon_r = 4.3$ )

Width of the substrate is 50 mm and the length is 100 mm

$h = 1.6$  mm

$t = 0.2$  mm

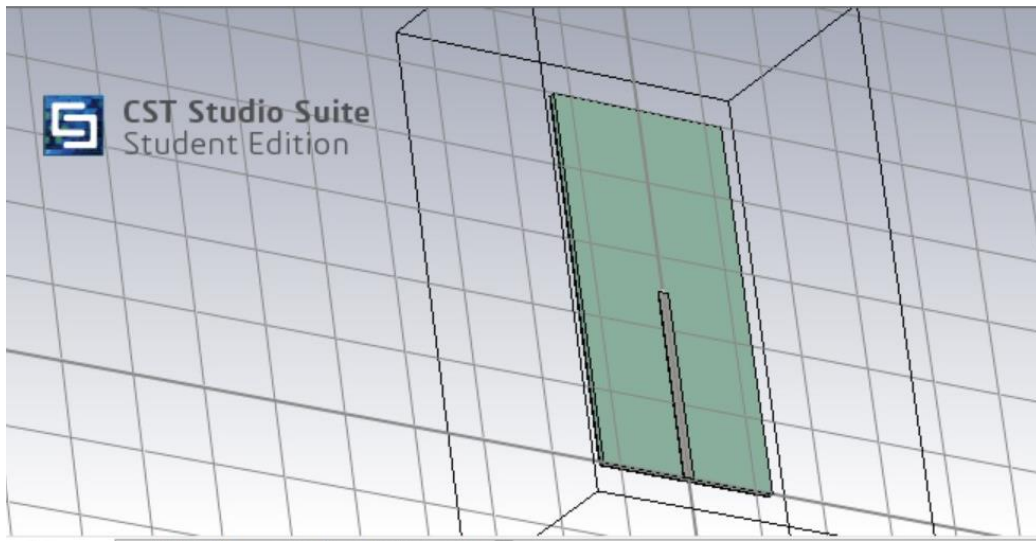
$W = 2.93$  mm (determined using Analysis and synthesis of transmission lines)

$\epsilon_{eff} = 3.204$

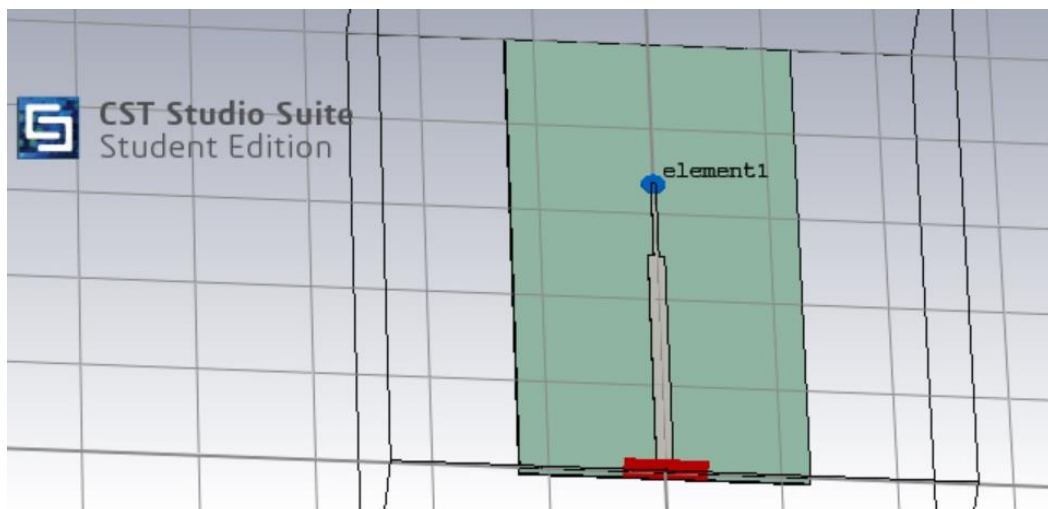
$Z'_0 = 93.01 \Omega$  length = 17 mm

Therefore, width of the quarter wave line is 0.87mm

**Design:**

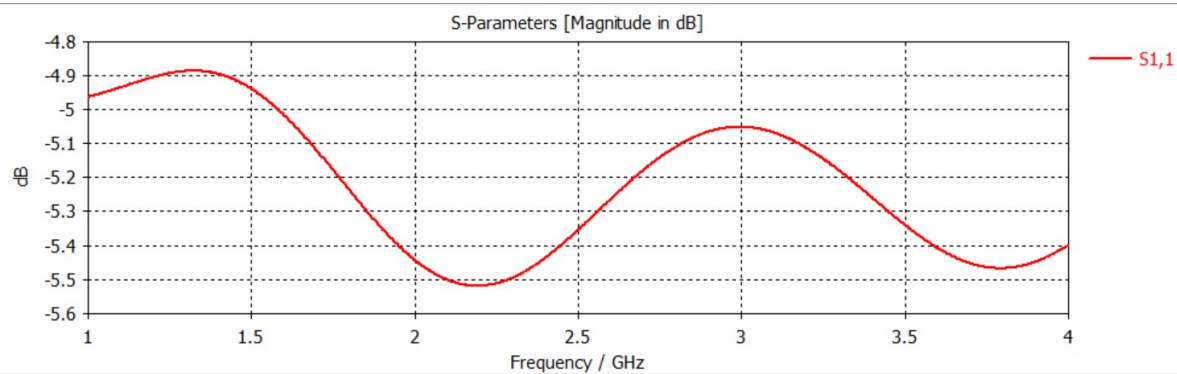


*Fig 1.2: Design of microstrip line terminated with the desired load.*

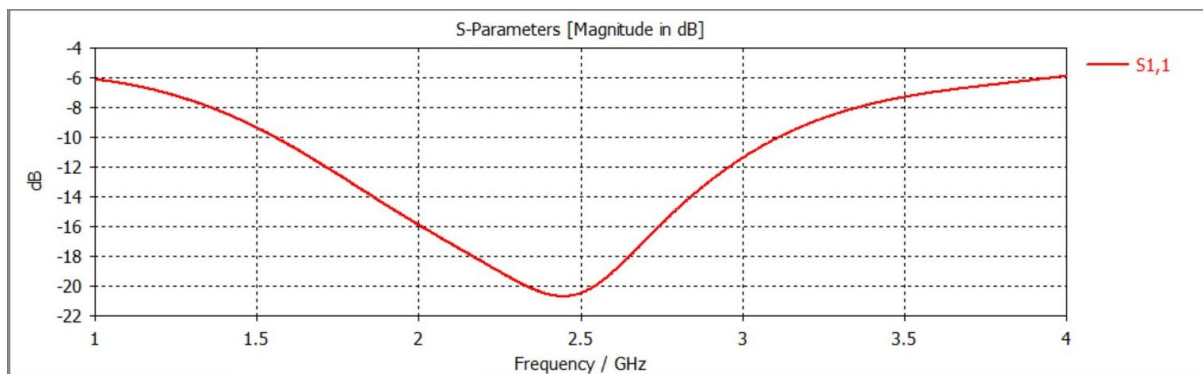


*Fig 1.3: Design of microstrip line terminated with quarter wave line and desired load*

### **Output/Graph:-**



*Fig 1.4: Result of the design of the microstrip line*



*Fig 1.5: Result of the design of the microstrip line terminated by quarter wave line.*

### **Observation of the experiment:**

- For fig 4, No resonance is observed around 2.4 GHz which implies impedance mismatch.
- For fig 5, an impedance is achieved at 2.4 GHz by using a quarter wave transformer.

### **Conclusion:-**

The designing of a quarter wave transformer for matching a  $50\ \Omega$  microstrip line with a load of  $173\ \Omega$  is successfully achieved.

<b>Experiment Number</b>	02
<b>Date of Experiment</b>	19/08/2020
<b>Date of Submission</b>	24/08/2020
<b>Name of the student</b>	Debagnik Kar
<b>Roll Number</b>	1804373
<b>Section</b>	ETC - 06

### **Aim of The Experiment :-**

To design a wire dipole antenna operating at 373 MHz and to find the directive gain and half power beam width from the radiation pattern.

### **Software Required:-**

CST Studio Suite 2019 (Student Edition)

### **Theory**

The length of the Dipole Antenna is given by the formula:

$$L = \frac{\lambda}{2} = \frac{c}{2f}$$

f = 373 MHz

The length of the dipole is 0.402 m = 402 mm

For the 402 mm length, we have got , 0.373 GHz which is far from our desired result, so we cant consider it as our result.

We know, the input impedance =  $(73 \pm 40j)$

To cancel the complex part we have to use a shorter length.

I am taking 360 mm to get a value which is nearly equal to the desired result 0.373 GHz

## Design:-

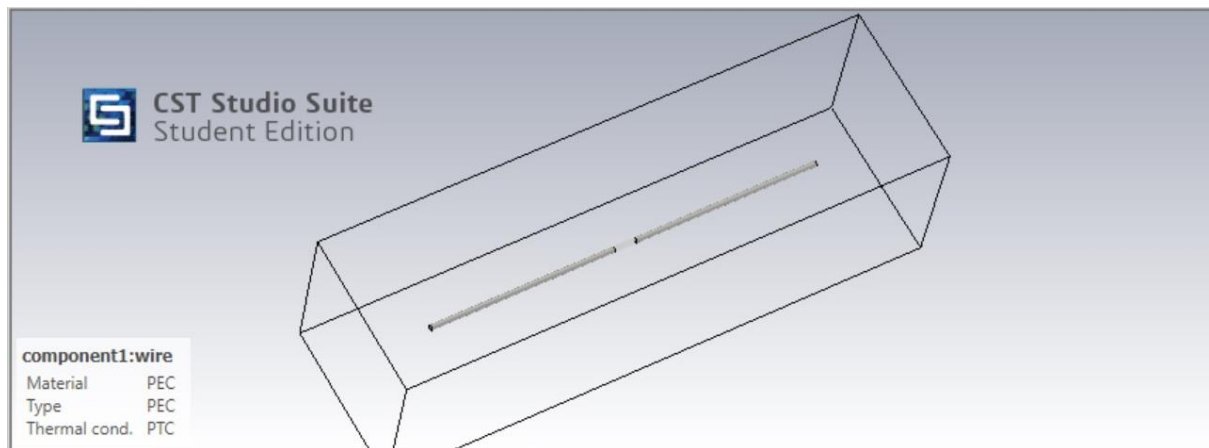


Fig 2.1: Design of wire dipole antenna

## Observation:-

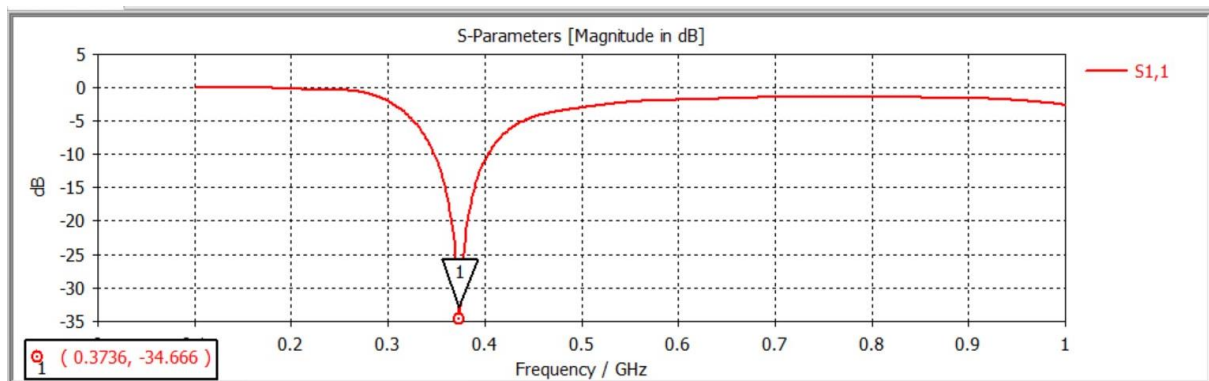


Fig 2.2: S11 Charecteristics

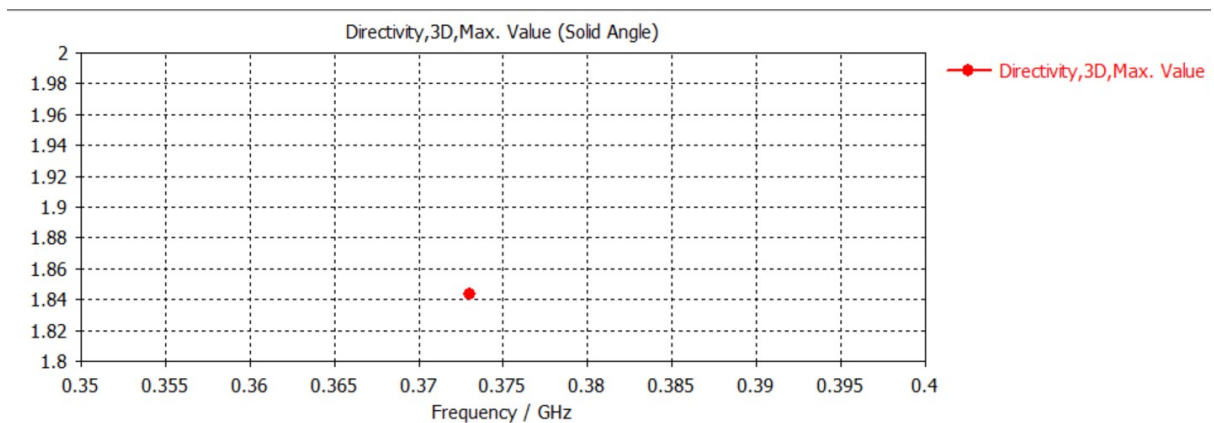


Fig 2.3: Directive Gain



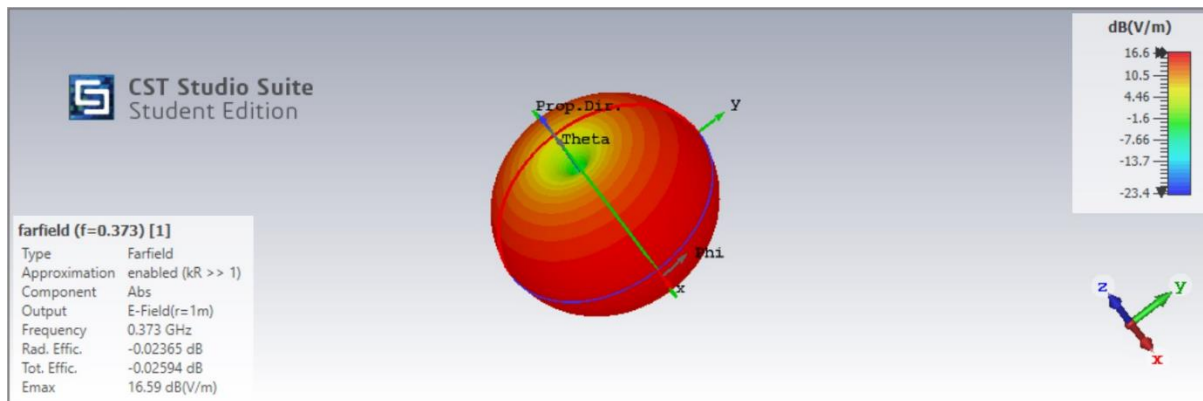


Fig 2.4: Radiation Pattern in X-Z and X-Y plane

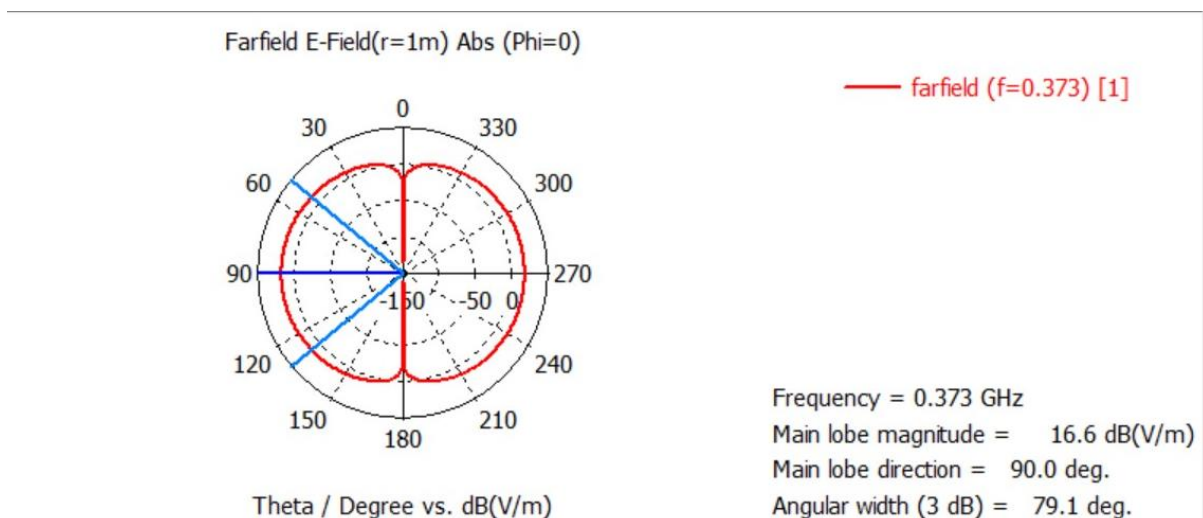


Fig 2.5: Plane radiation pattern and half powered beam width

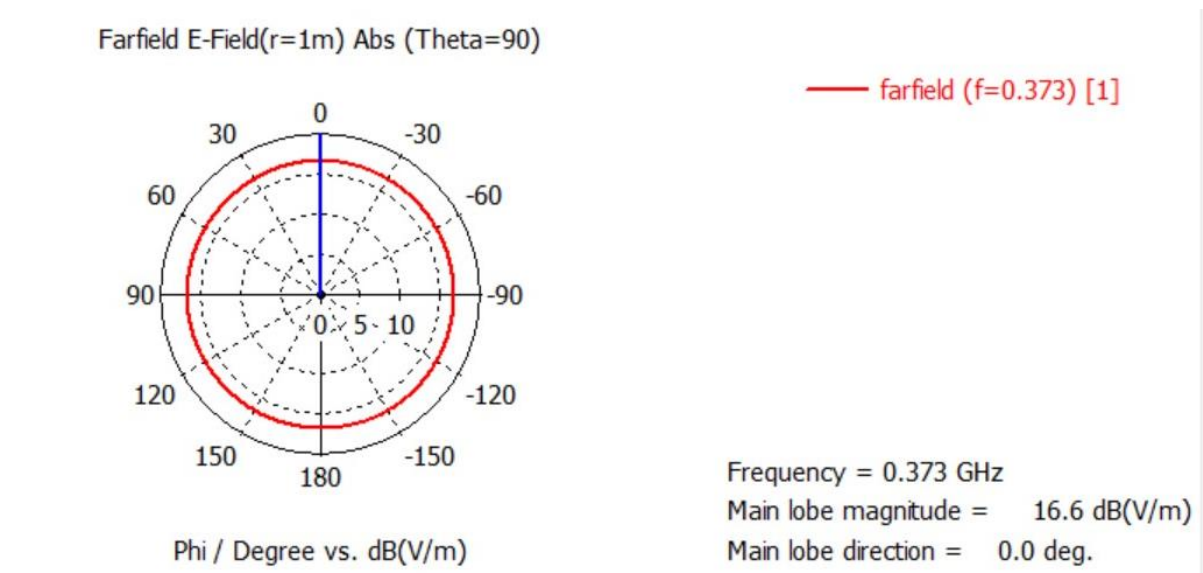


Fig 2.6: H-plane omni-directional Radiation Pattern

**Inference of the experiment:**

From this experiment, we learnt how to design a wire dipole antenna using CST Studio Suite. We also got to learn about the concept of radiation pattern after performing the experiment

**Conclusion:-**

A successful design of a wire dipole antenna operating at 373 MHz is successfully simulated on a virtual platform.