Statistical Computation

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Contents

1 Review

1.1 Complex Number

Definition 1.1. A complex number z consists of two component, real and imaginary:

$$z = x + \iota y$$

where $\iota = \sqrt{-1}$.

Property 1.1. If $z_1 = x_1 + \iota y_1, z_2 = x_2 + \iota y_2$ then

$$\begin{aligned} z_1 + z_2 &= (x_1 + x_2) + \iota(y_1 + y_2) \\ z_1 z_2 &= (x_1 x_2 - y_1 y_2) + \iota(x_1 y_2 + x_2 y_1) \end{aligned}$$

Property 1.2. $\exp(\iota\theta) = \cos(\theta) + \iota \sin(\theta)$.

Property 1.3. $z = x + \iota y = r \exp(\iota\theta)$ where $r = |z| = \sqrt{x^2 + y^2}, x = r\cos(\theta), y = r\sin(\theta)$.

Property 1.4. $\exp(\iota(\theta_1 + \theta_2)) = \cos(\theta_1 + \theta_2) + \iota \sin(\theta_1 + \theta_2)$.

2 Basics

2.1 Floating Point

2.1.1 Floating Point Representation

Definition 2.1. A ***floating point number*** is represented by three components: (S, F, E) where S is the sign of the number (± 1), F is a fraction (lying between 0 and 1), E is an exponent. S, F, E are all represented as binary digits (bits). The ***floating point representation*** of x, $\text{fl}(x)$ is

$$\text{fl}(x) = S \times F \times 2^E$$

Note. x and $\text{fl}(x)$ need not be the same, since $\text{fl}(x)$ is a binary approximation to x, and there are only a finite number of floating point numbers.

2.1.2 Round-Off Error

Mathematical operations introduce further approximation errors

$$f(\text{fl}(x)) = f(x + \varepsilon) \approx f(x) + \varepsilon f'(x)$$

and the goal is to make the round-off error $|f(x) - \text{fl}(f(\text{fl}(x)))|$ as small as possible.

2.1.3 Machine Epsilon and Other Constants

For a given real number x, we have

$$|\text{fl}(x) - x| \leqslant U|x| \text{ or } \text{fl}(x) = x(1 + u), |u| \leqslant U$$

where U is ***machine epsilon*** or ***machine unit***. U is machine dependent but very small. In R, $U = 2^{-52} = 2.220 \times 10^{-16}$.

Other machine dependent constants include:

1. The minimum and maximum positive floating point numbers: $x_{\min} = 2^{-1022} = 2.225 \times 10^{-308}$ and $x_{\max} = 2^{1024} - 1 = 1.798 \times 10^{308}$.
2. The maximum integer: $2147383647 = 2^{31} - 1$.

2.1.4 Overflow and Underflow Error

Definition 2.2. If the result of a floating point operation exceeds $x_{\max}$, then the value returned is `Inf`.

Note. `Inf` indicates an ***overflow error***.

Definition 2.3. If the result of a floating point operation is undefined then `NaN` is returned.

Definition 2.4. An ***underflow error*** occurs when the result of a floating point calculation is smaller (in absolute value) than $x_{\min}$.

Note. There are two possible outcomes: an error is reported or an exact 0 is returned. The latter outcome may cause problems in subsequent computations (e.g., division by 0).

Note. There are some ways to avoid overflow and underflow errors:

1. Use logarithmic scale: Changes multiplication/division into addition/subtraction, e.g., `lgamma`, `lfactorial`, `lchoose`.
2. Use series expansions (e.g., Taylor series).

Example 2.1. For x close to 0, $\dfrac{\exp(x)-1}{x} \approx 1$. Naive computation of $\dfrac{\exp(x)-1}{x}$ is problematic for x close to 0 due to possible round-off and underflow errors:

$$\frac{\mathrm{fl}(\exp(x)-1)}{\mathrm{fl}(x)} \neq \frac{\exp(x)-1}{x}$$

We solve the problem by using a series approximation, for $|x| \leqslant \varepsilon$,

$$\frac{\exp(x)-1}{x} = \frac{x + x^2/2 + x^3/6 + \cdots}{x} = 1 + \frac{x}{2} + \frac{x^2}{6} + \cdots$$

2.1.5 Catastrophic Cancellation

Suppose $z_1 = g_1(x_1, \cdots, x_n)$ and $z_2 = g_2(x_1, \cdots, x_n)$. We want to compute $y = z_1 - z_2$. What we actually compute is

$$y^* = \mathrm{fl}(\mathrm{fl}(z_1) - \mathrm{fl}(z_2))$$

where $\mathrm{fl}(z_1) = z_1(1+u_1)$ and $\mathrm{fl}(z_2) = z_2(1+u_2)$. We have

$$\mathrm{fl}(z_1) - \mathrm{fl}(z_2) = \underbrace{z_1 - z_2}_{y} + \underbrace{z_1 u_1 - z_2 u_2}_{\text{error}}$$

If z_1 and z_2 are large but $y = z_1 - z_2$ is small then the magnitude of the error may be larger than the magnitude of y - ***catastrophic cancellation***.

2.2 Sparse Matrices

Definition 2.5. We say an $n \times n$ matrix is ***sparse*** if it has $k \times n$ non-zero elements where $k \ll n$.

Note 1. An $n \times n$ matrix needs at least n non-zero elements to be invertible.

Note 2. Sparse matrices are useful because we need only store non-zero elements and their row and column indices; multiplication by and addition to 0 are free operations.

2.3 Application: Computation of Probability Distributions

Question: Suppose X_i are independent discrete r.v.s. taking values $0, \cdots, l$ with

$$P(X_i = x) = p(x), x = 0, \cdots, l$$

Define $S = X_1 + \cdots + X_n$ and find the probability distribution of S.

2.3.1 Brute Force Approach

Start with $n = 2$ and proceed inductively:

$$p_2(x) := P(X_1 + X_2 = x) = \sum_{y=0}^{x} P(X_1 = y, X_2 = x - y)$$

$$p_3(x) := P(X_1 + X_2 + X_3 = x) = \sum_{y=0}^{x} P(X_1 + X_2 = y, X_3 = x - y)$$

$$\vdots$$

$p_k(x)$ requires $x+1$ multiplications and to evaluate $p_k(x)$ for $x = 0, \cdots, kl$, we need

$$N(k) = \sum_{x=0}^{kl}(x+1) \approx \frac{(kl)^2}{2} \text{ multiplications}$$

Thus the total number of multiplications is

$$\sum_{k=2}^{n} N(k) \approx \frac{n^3 l^2}{6} = O(n^3 l^2)$$

2.3.2 Probability Generating Function

Definition 2.6. If X is a discrete r.v. taking values $0, 1, \cdots$, then its ***probability generating function*** is

$$\phi(t) = \mathbb{E}[t^X] = \sum_{x=0}^{\infty} P(X = x)t^x$$

Note. If X takes values $0, \cdots, l$, then $P(X = x)$ can be recovered from evaluating $\phi(t)$ at $l+1$ distinct (non-zero) points $t_0, \cdots, t_l$.

If $\phi(t) = \mathbb{E}[t^{X_i}]$, then the probability generating function of S is

$$\mathbb{E}[t^S] = \mathbb{E}[t^{X_1+\cdots+X_n}] = [\phi(t)]^n$$

Thus we can recover $P(S = x)$ for $x = 0, \cdots, nl$ by evaluating $[\phi(t)]^n$ at $t_0, \cdots, t_{nl}$. We have $nl+1$ linear equations in $nl+1$ unknowns, and solving typically requires $O(n^3 l^3)$ operations, which is slower than the brute force approach.

2.3.3 Discrete Fourier Transform (DFT)

A choice for $t_0, \cdots, t_{nl}$ are complex exponentials

$$t_j = \exp\left(-2\pi\iota \frac{j}{nl+1}\right), j = 0, \cdots, nl$$

where $\iota = \sqrt{-1}$. Since $p(x) = 0$ for $x = l+1, \cdots, nl$, we have

$$\phi(t_j) = \sum_{x=0}^{l} p(x) \exp\left(-2\pi\iota \frac{jx}{nl+1}\right) = \sum_{x=0}^{nl} p(x) \exp\left(-2\pi\iota \frac{jx}{nl+1}\right)$$

$\phi(t_0), \cdots, \phi(t_{nl})$ is the ***discrete Fourier transform*** (DFT) of $p(0), \cdots, p(nl)$, and thus, the DFT of $P(S = 0), \cdots, P(S = nl)$ is $[\phi(t_0)]^n, \cdots, [\phi(t_{nl})]^n$. Hence, given $\phi(t_0), \cdots, \phi(t_{nl})$, we can compute the probability distribution of S using the inverse DFT:

$$P(S = x) = \frac{1}{nl+1} \sum_{j=0}^{nl} [\phi(t_j)]^n \exp\left(2\pi\iota \frac{jx}{nl+1}\right), x = 0, \cdots, nl$$

Naive computation of $P(S = x)$ using DFT requires $O(n^3 l^2)$ multiplications; but with divide-and-conquer algorithm, we can reduce the number of multiplications by a factor of n.

In R, if `x` is a vector of length n we can compute its DFT with `fft(x)` and the inverse DFT with `fft(tx, inv=T) / length(x)`:

```
probs = # The vector for P(X=x)
dft = fft(probs)
dft.s = dtf^n # S=X1+...+Xn
idft.s = fft(dft.s, inv=T) / length(probs)
Re(idft.s) # Real component of idft.s, or P(S=x)
```

Note. `fft` is the fast Fourier transform, which is an efficient algorithm for computing the DFT when the length of the sequence is a product of small primes.

2.4 Application: Image Processing

Question: We observe an image denoted by $x(i,j).i = 1, \cdots, m, j = 1, \cdots, n$, where (i,j) denotes a pixel location. We want:

1. Denoising: Think of $\{x(i,j)\}$ as a image corrupted by noise

$$x(i,j) = \underbrace{s(i,j)}_{\text{True}} + \underbrace{\varepsilon(i,j)}_{\text{Noise}}$$

2. Compression: Approximate $x(i,j)$ by $x^*(i,j)$ where

$$x^*(i,j) = \sum_{k=1}^{p} \beta_k \phi_k(i,j)$$

where $p \ll m \times n$ and $\phi_1, \cdots, \phi_p$ are known functions.

2.4.1 Transformation

Define X to be the $m \times n$ matrix whose elements are $x(i,j)$. Define orthogonal matrices H_1 $(m \times m)$ and H_2 $(n \times n)$ and define $\widehat{X} = H_1 X H_2$, which has the same dimensions as X. Since for orthogonal matrix H, $H^{-1} = H^T$ and so $X = H_1^T \widehat{X} H_2^T$. Assume the noisy image model $X = S + E$, if H_1 and H_2 are chosen appropriately,

$$\widehat{X} = \underbrace{H_1 S H_2}_{\text{Sparse}} + \underbrace{H_1 E H_2}_{\approx 0}$$

Therefore,

1. Denoising: Given $\widehat{X}$, find a transformation $\widehat{X} \mapsto T(\widehat{X})$ and define the denoised image

$$X_{\text{dn}} = H_1^T T(\widehat{X}) H_2^T$$

where we assume the smallest elements of $\widehat{X}$ are due to noise and set these equal to 0

$$T(\widehat{X})(i,j) = 0, |\widehat{X}(i,j)| \leqslant \text{Threshold}$$

2. Compression: The same idea is used for compression: for some T,

$$X_{\text{c}} = H_1^T T(\widehat{X}) H_2^T$$

Note. T is usually defined more deterministically. The form of T depends on the amount of compression and the type of image.

2.4.2 Hadamard Matrices and Walsh-Hadamard Transform

Definition 2.7. A ***Hadamard matrix*** is an $n \times n$ matrix whose elements are all ± 1 with orthogonal rows s.t. $HH^T = nI$.

Note 1. $H^{-1} = \dfrac{H^T}{n}$.

Note 2. Hadamard matrices only exist if $n = 1, n = 2$, or n is a multiple of 4.

Note 3. We focus on the case where $n = 2^k$ since it is simple to construct and we can write the Hadamard matrix as a product of sparse matrices. We start with the trivial 1×1 Hadamard matrix $H_1 = 1$, and then define $H_2, H_4, H_8, \cdots$ recursively:

$$H_2 = \begin{pmatrix} 1 & 1 \\ 1 & -1 \end{pmatrix}$$
$$H_{2^k} = \begin{pmatrix} H_{2^{k-1}} & H_{2^{k-1}} \\ H_{2^{k-1}} & -H_{2^{k-1}} \end{pmatrix}$$

for $k = 2, 3, \cdots$.

Note 4. H_2 is symmetric and so H_{2^k} is symmetric and thus $H_{2^k}^{-1} = \dfrac{H_{2^k}}{2^k}$.

Definition 2.8. Given arbitrary matrices A and B, the ***Kronecker product*** $A \otimes B$ is

$$A \otimes B = \begin{pmatrix} a_{11}B & a_{12}B & \cdots & a_{1n}B \\ a_{21}B & a_{22}B & \cdots & a_{2n}B \\ \vdots & \vdots & \ddots & \vdots \\ a_{m1}B & a_{m2}B & \cdots & a_{mn}B \end{pmatrix}$$

for an $m \times n$ matrix A.

Property 2.1. Assume below that any matrix sums, products or inverses are well-defined.

1. $A \otimes (B + C) = (A \otimes B) + (A \otimes C)$.
2. $(B + C) \otimes A = (B \otimes A) + (C \otimes A)$.
3. $A \otimes (B \otimes C) = (A \otimes B) \otimes C$.
4. $(A \otimes B)(C \otimes D) = (AC) \otimes (BD)$.
5. $(A \otimes B)^T = A^T \otimes B^T$.
6. $(A \otimes B)^{-1} = A^{-1} \otimes B^{-1}$.

Note. For Hadamard matrices, $H_{2^k} = H_2 \otimes H_{2^{k-1}}$. We rewrite it as $H_{2^k} = (H_2 I_2) \otimes (I_{2^{k-1}} H_{2^{k-1}})$ and using the property, we have

$$H_{2^k} = (H_2 \otimes I_{2^{k-1}})(I_2 \otimes H_{2^{k-1}})$$

Repeating the process with $H_{2^{k-1}}, H_{2^{k-2}}, \cdots$, we get

$$H_{2^k} = \underbrace{(H_2 \otimes I_{2^{k-1}})(I_2 \otimes H_2 \otimes I_{2^{k-2}})(I_4 \otimes H_2 \otimes I_{2^{k-3}}) \cdots (I_{2^{k-1}} \otimes H_2)}_{k = \log_2(n) \text{ terms}}$$

Definition 2.9. Given an $n \times n$ Hadamard matrix H and a vector $\mathbf{x}$ of length n, we define its ***Walsh-Hadamard transform*** by $\hat{\mathbf{x}} = H\mathbf{x}$.

Note 1. Given the W-H transform, we can recover $\mathbf{x}$

$$\mathbf{x} = \frac{1}{n} H^T \hat{\mathbf{x}}$$

Note 2. If $n = 2^k$, since $H = H^T$, then

$$\mathbf{x} = \frac{1}{n} H \hat{\mathbf{x}}$$

2.5 Application: Denoising

Question: Suppose we observe $\mathbf{x} = (x_1, \cdots, x_n)^T$ where we assume that

$$\mathbf{x} = \mathbf{s} + \mathbf{e} = \text{Signal} + \text{Noise}$$

We want to recover or estimate the signal $\mathbf{s}$.

2.5.1 Assumption

Assume $\mathbf{s}$ is structured so that its W-H transform $\widehat{\mathbf{s}} = H\mathbf{s}$ contains mostly 0s

$$\widehat{\mathbf{x}} = H\mathbf{x} = \underbrace{H\mathbf{s}}_{\text{Sparse}} + \underbrace{H\mathbf{e}}_{\text{Relatively small}}$$

2.5.2 Thresholding

We shrink smaller components of $\widehat{\mathbf{x}}$ towards 0, and then estimate $\mathbf{s}$ by the inverse W-H transform of the thresholded $\widehat{\mathbf{x}}$. Thresholded W-H transform $\widehat{\mathbf{x}}_s$ is an estimate of the W-H transform of $\mathbf{s}$, and thus we can estimate $\mathbf{s}$ by the inverse W-H transform

$$\widetilde{\mathbf{s}} = \frac{1}{n} H^T \widehat{\mathbf{x}}_s$$

Define thresholds $\lambda_1, \cdots, \lambda_n \geqslant 0$. The ***hard thresholding*** is to modify $\widehat{\mathbf{x}}$ as follows:

$$\widehat{\mathbf{x}}_s = \begin{pmatrix} \widehat{x}_1 I(|\widehat{x}_1| \geqslant \lambda_1) \\ \vdots \\ \widehat{x}_n I(|\widehat{x}_n| \geqslant \lambda_n) \end{pmatrix}$$

The ***soft thresholding*** is to modify $\widehat{\mathbf{x}}$ as follows:

$$\widehat{\mathbf{x}}_s = \begin{pmatrix} \text{sgn}(\widehat{x}_1)(|\widehat{x}_1| - \lambda_1)_+ \\ \vdots \\ \text{sgn}(\widehat{x}_n)(|\widehat{x}_n| - \lambda_n)_+ \end{pmatrix}$$

where $\text{sgn}(y)$ is the sign of y, and y_+ equals y if $y > 0$ and 0 if $y \leqslant 0$.

Typically we set $\lambda_1 = 0$, and use knowledge of the problem to decide $\lambda_2, \cdots, \lambda_n$; or take $\lambda_2 = \cdots = \lambda_n$ and choose the common value based on tools such as half normal plots.

2.5.3 The Fast W-H Transform

A Hadamard matrix H consists of ± 1 so computation of $H\mathbf{x}$ involves only additions and subtractions, but naive computation involves $n(n-1) = O(n^2)$ additions and subtractions, which is less than ideal if n is very large. We can write H as a product of sparse matrices to reduce complexity.

Example 2.2 ($n = 2^3 = 8$)**.** The 8×8 Hadamard matrix is

$$H_8 = \begin{pmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & -1 & 1 & -1 & 1 & -1 & 1 & -1 \\ 1 & 1 & -1 & -1 & 1 & 1 & -1 & -1 \\ 1 & -1 & -1 & 1 & 1 & -1 & -1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ 1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 \\ 1 & 1 & -1 & -1 & -1 & -1 & 1 & 1 \\ 1 & -1 & -1 & 1 & -1 & 1 & 1 & -1 \end{pmatrix}$$

Naive computation of $H_8\mathbf{x}$ needs 56 additions and subtractions. But if $H_8 = A^3$ where

$$A = \begin{pmatrix} 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 1 & -1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & -1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & -1 \end{pmatrix}$$

Computation of $AAA\mathbf{x}$ needs $3 \times 8 = 24$ additions and subtractions.

2.5.4 R code for FWHT

The function `fwht` below computes the W-H transform of data in a vector `x`.

```
fwht = function(x) {
  h=1
  len = length(x)
  while (h < len) {
    for (i in seq(1, len, by=h*2)) {
      for (j in seq(i, i+h-1)) {
        a = x[j]
        b = x[j+h]
        x[j] = a + b
        x[j+h] = a - b
      }
    }
  h = 2 * h
  }
  x
}
```

We can compute the inverse W-H transform using `fwht` by dividing the output by the length of the vector.

2.6 Fast Fourier Transform (FFT)

Definition 2.10 (Discrete Fourier Transform)**.** Suppose we have data $x_0, \cdots, x_{n-1}$, and define $\widehat{x}_0, \cdots, \widehat{x}_{n-1}$ by

$$\widehat{x}_j = \sum_{t=0}^{n-1} \exp\left(-2\pi\iota \frac{j}{n} t\right) x_t$$

where $\iota = \sqrt{-1}$.

Property 2.2 (Inverse DFT)**.** Given DFT, recover the original sequence by

$$x_t = \frac{1}{n} \sum_{j=0}^{n-1} \exp\left(2\pi\iota \frac{j}{n} t\right) \widehat{x}_j$$

Proof. For complex numbers z,

$$\sum_{j=0}^{n-1} z^j = \begin{cases} n, & z = 1 \\ \dfrac{1-z^n}{1-z}, & \text{otherwise} \end{cases}$$

Thus if $z = \exp\left(\dfrac{2\pi\iota t}{n}\right)$ for an integer t. we have

$$\sum_{j=0}^{n-1} z^j = \sum_{j=0}^{n-1} \exp\left(2\pi\iota \frac{t}{n} j\right) = \frac{1-\exp(2\pi\iota t)}{1-\exp(2\pi\iota t/n)} = 0$$

since $\exp(2\pi\iota t) = \cos(2\pi t) + \iota \sin(2\pi t) = 1$. Hence,

$$\begin{aligned} \frac{1}{n}\sum_{j=0}^{n-1} \exp\left(2\pi\iota \frac{j}{n} t\right) \widehat{x}_j &= \frac{1}{n}\sum_{j=0}^{n-1}\sum_{s=0}^{n-1} \exp\left(2\pi\iota \frac{t-s}{n} j\right) x_s \\ &= \frac{1}{n}\sum_{s=0}^{n-1} x_s \sum_{j=0}^{n-1} \exp\left(2\pi\iota \frac{t-s}{n} j\right) \\ &= x_t \end{aligned}$$

since

$$\sum_{j=0}^{n-1} \exp\left(2\pi\iota \frac{t-s}{n} j\right) = \begin{cases} n, & s = t \\ 0, & s \neq t \end{cases}$$

□

Definition 2.11 (Matrix Formulation of DFT)**.** Define $\mathbf{x} = (x_0, \cdots, x_{n-1})^T$ and $\widehat{\mathbf{x}} = (\widehat{x}_0, \cdots, \widehat{x}_{n-1})^T$. Then

$$\widehat{\mathbf{x}} = F\mathbf{x}$$

where F is an $n \times n$ matrix whose jth row and kth column is

$$f_{jk} = \exp\left(-2\pi\iota \frac{(j-1)(k-1)}{n}\right)$$

The elements of F^{-1} are

$$\overline{f}_{jk} = \frac{1}{n}\exp\left(2\pi\iota \frac{(j-1)(k-1)}{n}\right)$$

Note 1. Using the matrix form directly, we need $O(n^2)$ additions and multiplications to compute the DFT (and its inverse).

Note 2. We can write F as a product of sparse matrices, but unlike the W-H transform, factorization of the DFT matrix is more complicated.

2.6.1 FFT Derivation

Assume n is a product of prime numbers $n_1, \cdots, n_k : n = n_1 \times \cdots \times n_k$.

2.6.1.1 Case I: Even Number and Product of Small Prime Numbers

Assume n is even, then

$$\begin{aligned}\widehat{x}_j &= \sum_{t=0}^{n/2-1} \exp\left(-2\pi\iota\frac{j}{n}2t\right) x_{2t} + \sum_{t=0}^{n/2-1} \exp\left(-2\pi\iota\frac{j}{n}(2t+1)\right) x_{2t+1} \\ &= \underbrace{\sum_{t=0}^{n/2-1} \exp\left(-2\pi\iota\frac{j}{n/2}t\right) x_{2t}}_{\text{DFT of } x_0, x_2, \cdots} + \exp\left(-2\pi\iota\frac{j}{n}\right) \underbrace{\sum_{t=0}^{n/2-1} \exp\left(-2\pi\iota\frac{j}{n/2}t\right) x_{2t+1}}_{\text{DFT of } x_1, x_3, \cdots}\end{aligned}$$

Hence, the DFT of $x_0, \cdots, x_{n-1}$ is a linear combination of the DFT of the even and odd indices. Our rearrangement into DFT of odd and even indices can be written in matrix form as

$$\widehat{\mathbf{x}} = \underbrace{\begin{pmatrix} I & \Omega \\ I & -\Omega \end{pmatrix}}_{\text{Sparse}} \underbrace{\begin{pmatrix} F_{n/2} & 0 \\ 0 & F_{n/2} \end{pmatrix}}_{\text{Sparser}} P\mathbf{x}$$

where Ω is a diagonal matrix (sparse) and P is a permutation matrix (sparse), i.e., if n is even, we can write F as a product of two sparse matrices and a matrix that is sparser than F ($n^2/2$ 0s).

If $n/2$ is divisible by a prime number n', we can perform a similar decomposition of $F_{n/2}$ and F is now the product of sparser matrices. When $n_1, \cdots, n_k$ are small then we need $O(n\ln(n))$ additions and multiplications.

2.6.1.2 Case II: Prime Number with Zero-Padding

Definition 2.12 (Zero Padding)**.** Add 0s to the end of the sequence so that the length of the ***zero padded*** sequence is a product of small prime numbers:

$$x_0, \cdots, x_{n-1}, \underbrace{0, \cdots, 0}_{m}$$

with $n + m = n_1 \times \cdots \times n_k$ where $n_1, \cdots, n_k$ are small primes.

Note 1. The function `nextn` is useful for zero-padding.

Note 2. Adding 0s to a sequence changes the nature of the sequence - creating a large discontinuity, which is reflected in the DFT.

2.6.2 Analysis of DFT Approach

For the application in computation of probability distributions with DFT approach, we take $m \geqslant nl = 1$ where m is a product of small prime numbers, and follow the steps:

1. Define $\widehat{p}_i(0), \cdots, \widehat{p}_i(m-1)$ to be the DFT of $p_i(0), \cdots, p_i(m-1)$ for $i = 1, \cdots, n$.
2. Define
$$\widehat{p}_s(k) = \prod_{i=1}^{n} \widehat{p}_i(k), k = 0, \cdots, m-1$$
3. Inverse DFT: $P(S=0), \cdots, P(S=m-1)$ is the inverse DFT of $\widehat{p}_s(0), \cdots, \widehat{p}_s(m-1)$.

The number of multiplications at each step is:

1. DFT: $n \times O(m\ln(m)) = O(nm\ln(m))$.
2. Product of DFTs: $O(nm)$.

3. Inverse DFT: $O(m \ln(m))$.

The total number of multiplications is $O(nm \ln(m))$ and thus if $m \approx nl$, the number of multiplications is $O(n^2 l \ln(nl))$ versus $O(n^3 l^2)$ for the brute force algorithm.