

# Control Systems

Deep\*

## CONTENTS

### 1 Feedback Voltage Amplifier: Shunt-Shunt

#### 1 FEEDBACK VOLTAGE AMPLIFIER: SHUNT-SHUNT

The circuit in Fig. 1.1 utilizes a voltage amplifier with gain  $\mu$  in a shunt-shunt feedback topology with the feedback network composed of resistor  $R_F$ . In order to be able to use the feedback equations you should first convert the signal source to its Norton Representation.

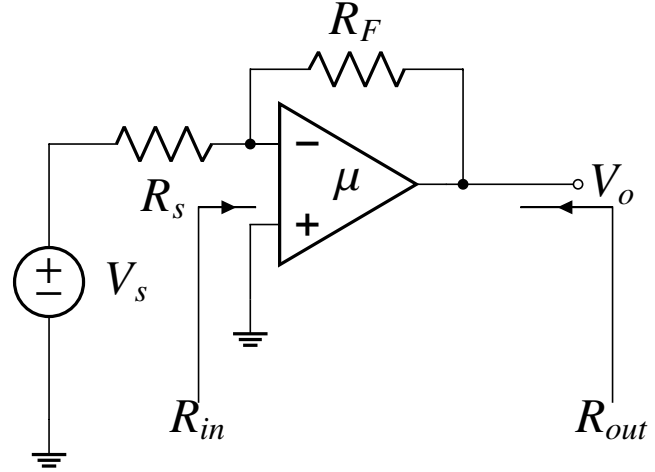


Fig. 1.1

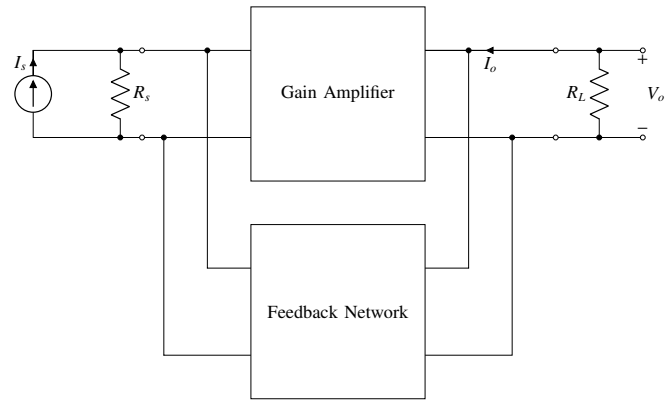


Fig. 1.2: Shunt Shunt Amplifier Block Diagram

- 1) If the loop gain is very large, what approximate closed loop voltage gain  $V_o/V_s$  is realized? If  $R_s = 1\text{k}\Omega$ , give the value of  $R_F$  that will result in  $V_o/V_s \simeq -10\text{V/V}$ .
- 2) If the amplifier  $\mu$  has a dc gain of  $10^3\text{ V/V}$ , an input resistance  $R_{id} = 100\text{k}\Omega$ , and an output resistance  $r_o = 1\text{k}\Omega$ , find the actual  $V_o/V_s$  realized. Also find  $R_{in}$  and  $R_{out}$ .
- 3) If the amplifier  $\mu$  has an upper 3-dB frequency of 1 kHz and a uniform -20-dB/decade gain rolloff, what is the 3-dB frequency of the gain  $|V_o/V_s|$ .

1. Fig. 1.1 shows the original circuit. Draw the Norton Representation, A-Circuit and H-Circuit.

**Solution:** See Fig. 1.3 for the Norton Representation, Fig. 1.4 and Fig. 1.5 for A-Circuit and Fig. 1.6 for the H-Circuit.

2. Refer Table 2 for the parameters.
3. Write all the feedback equations based on all the Figs. using KCL/KVL.

**Solution:** The equations are as follows:

$$T = \frac{V_o}{I_s} = \frac{G}{1 + GH} \quad (3.1)$$

$$H = \frac{I_f}{V_o} = -\frac{1}{R_F} \quad (3.2)$$

$$I_i = I_s - I_f = \frac{I_s}{1 + GH} \quad (3.3)$$

$$R_{if} = \frac{V_i}{I_s} = \frac{V_i}{(1 + GH)I_i} = \frac{R_i}{1 + GH} \quad (3.4)$$

$$R_{of} = \frac{R_o}{1 + GH} \quad (3.5)$$

$$R_{in} = \frac{1}{\frac{1}{R_{if}} - \frac{1}{R_s}}, R_{out} = \frac{1}{\frac{1}{R_{of}} - \frac{1}{R_L}} \quad (3.6)$$

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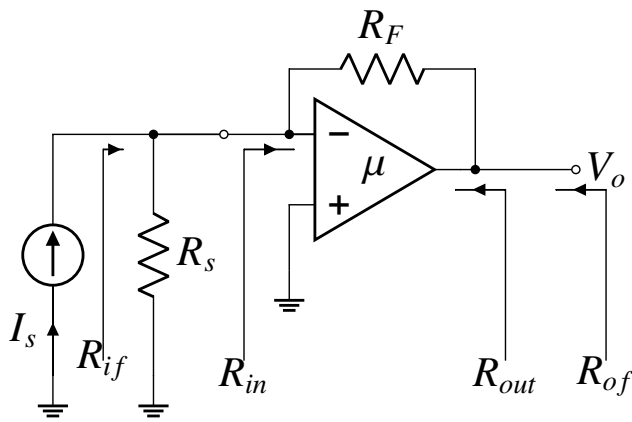


Fig. 1.3

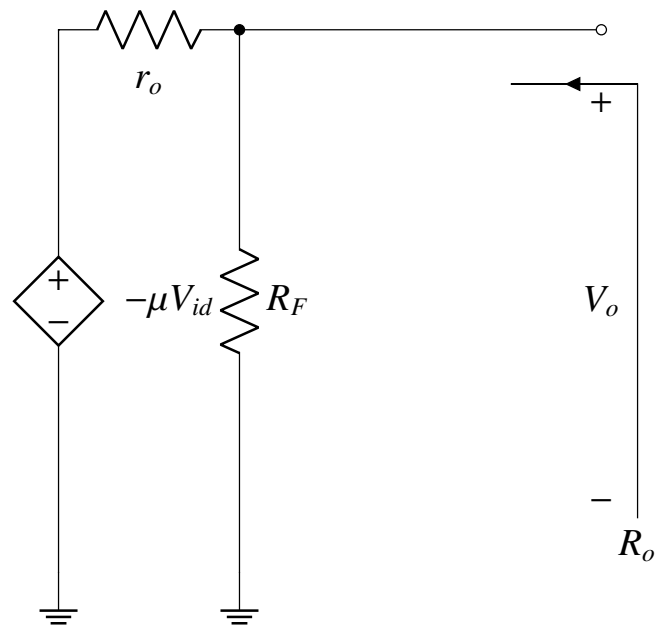


Fig. 1.5

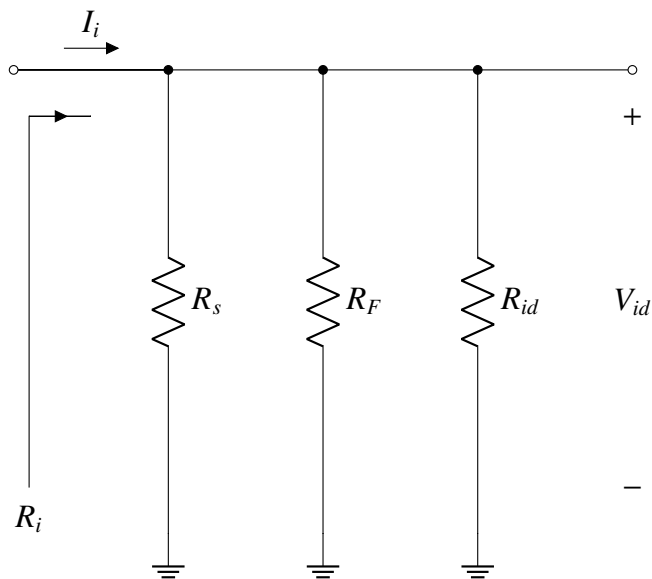


Fig. 1.4

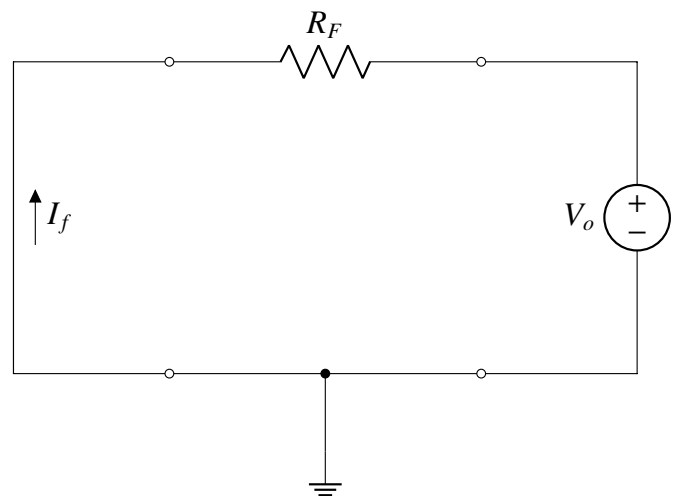


Fig. 1.6

4. If the loop gain is very large, what approximate closed-loop voltage gain  $V_o/V_s$  is realized? Also if  $R_s = 1 \text{ k}\Omega$ , give the value of  $R_F$  that will result in  $V_o/V_s \approx -10 \text{ V/V}$ .

**Solution:** If the loop gain  $GH$  is very large then the closed loop gain is,

$$T = \frac{V_o}{I_s} = \frac{G}{1 + GH} \quad (4.1)$$

$$\because GH \gg 1 \Rightarrow T \approx \frac{1}{H} \quad (4.2)$$

From equation 3.2 and 4.2 we get,

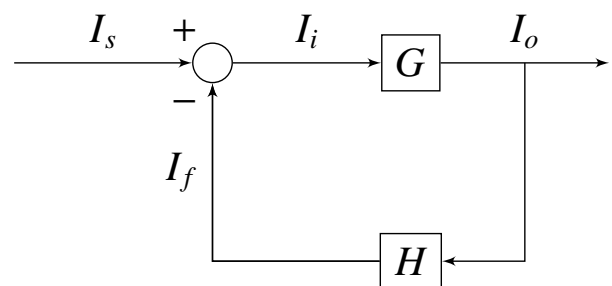


Fig. 1.7: Block Diagram

$$T \approx -R_F \quad (4.3)$$

$$\Rightarrow \frac{V_o R_s}{V_s} \approx -R_F \quad (4.4)$$

$$\Rightarrow R_F = 10 \text{ k}\Omega \quad (4.5)$$

Parameters	Description
$G$	Open Loop Gain
$H$	Feedback Factor
$T$	Closed Loop Gain
$V_0$	Output Voltage
$V_s$	Signal Source Voltage
$V_{id}$	Input Voltage of Opamp
$I_s$	Signal Source Current
$I_f$	Feedback Current
$R_i$	Total Input Resistance
$R_{out}$	Total Output Resistance
$R_{id}$	Input resistance of Opamp
$r_o$	Output resistance of Opamp
$R_i$	Input resistance of Open Loop
$R_o$	Output resistance of Open Loop
$R_{if}$	Input resistance of Feedback
$R_{of}$	Output resistance of Feedback
$R_s$	Resistance of Current Source
$V_f$	Voltage across $R_s$
$V_{in}$	Voltage at -ve terminal of opamp
$f$	Closed loop 3-dB freq.

TABLE 2

5. If the amplifier  $\mu$  has a dc gain of  $10^3$  V/V, an input resistance  $R_{id} = 100$  k $\Omega$ , and an output resistance  $r_o = 1$  k $\Omega$ , find the actual  $V_o/V_s$  realized. Also find  $R_{in}$  and  $R_{out}$ .

**Solution:** To find  $V_o/V_s$ ,  $R_{in}$  and  $R_{out}$  first find the other necessary parameters. From Fig. 1.4 we get,

$$R_i = R_{id} || R_F || R_s \quad (5.1)$$

$$R_i = 100k || 10k || 1k = 0.90k\Omega \quad (5.2)$$

$$V_{id} = I_i R_i \quad (5.3)$$

From Fig. 1.5 we get,

$$R_o = r_o || R_F \quad (5.4)$$

$$\Rightarrow R_o = 1k || 10k = 0.91k\Omega \quad (5.5)$$

$$V_o = -\mu V_{id} \frac{R_F}{r_o + R_F} \quad (5.6)$$

From equation 5.3 and 5.6 we get the open-loop gain as,

$$G = \frac{V_o}{I_i} = -\mu R_i \frac{R_F}{r_o + R_F} \quad (5.7)$$

$$\Rightarrow G = -(1000)(0.90) \frac{10}{11} = -819.00k\Omega \quad (5.8)$$

From equation 3.2 and 5.8 we get closed loop gain T as,

$$T = \frac{G}{1 + GH} = \frac{-819}{82.9} = -9.88k\Omega \quad (5.9)$$

From equation 3.1 we know,

$$T = \frac{V_o}{I_s} \quad (5.10)$$

$$\Rightarrow T = \frac{V_o R_s}{V_s} \quad (5.11)$$

$$\Rightarrow \frac{V_o}{V_s} = \frac{T}{R_s} \quad (5.12)$$

$$\Rightarrow \frac{V_o}{V_s} = \frac{-9.88}{1} = -9.88V/V \quad (5.13)$$

From equation 3.4 and 3.6 we know,

$$R_{if} = \frac{R_i}{1 + GH} = \frac{0.90}{82.9} \quad (5.14)$$

$$\Rightarrow R_{if} = 10.87\Omega \quad (5.15)$$

$$R_{in} = \frac{1}{\frac{1}{R_{if}} - \frac{1}{R_s}} \quad (5.16)$$

$$\Rightarrow R_{in} = \frac{1}{\frac{1}{10.87} - \frac{1}{1000}} = 10.99\Omega \quad (5.17)$$

Because  $R_L$  is not there in the circuit so we take its value as  $\infty$ , so from equation 3.5 and 3.6 we know,

$$R_{of} = \frac{R_o}{1 + GH} = \frac{0.91}{82.9} \quad (5.18)$$

$$\Rightarrow R_{of} = 10.97\Omega \quad (5.19)$$

$$R_{out} = \frac{1}{\frac{1}{R_{of}} - \frac{1}{R_L}} \quad (5.20)$$

$$\Rightarrow R_{out} = \frac{1}{\frac{1}{10.97} - \frac{1}{\infty}} = 10.97\Omega \quad (5.21)$$

Verify the above calculations using the following Python code.

```
codes/ee18btech11011/ee18btech11011_cal.
```

ipynb

6. If the amplifier  $\mu$  has an upper 3-dB frequency of 1 kHz and a uniform -20-dB/decade gain rolloff, what is the 3-dB frequency of the gain  $|V_o/V_s|$ .

**Solution:** To find the 3-dB frequency i.e.,  $\omega_{3dB}$  we need to look at the Fig.6.8.

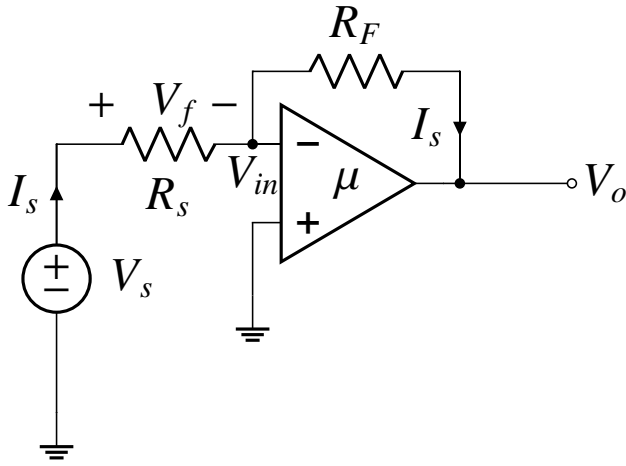


Fig. 6.8

The open loop gain  $G$  is given as follows in terms of frequency:

$$G = \frac{\mu}{1 + \frac{jf}{f_c}} \quad (6.1)$$

From Fig.6.8 we can say that:

$$V_{in} = V_s - V_f \quad (6.2)$$

$$V_o = -GV_{in} \quad (6.3)$$

$$\frac{V_f}{R_s} = \frac{V_{in} - V_o}{R_F} \quad (6.4)$$

From equation 6.3 and 6.4 we get:

$$\frac{V_f}{R_s} = \frac{-\frac{V_o}{G} - V_o}{R_F} \quad (6.5)$$

$$\Rightarrow \frac{V_f}{V_o} = -\frac{(1 + G)(R_s)}{G(R_F)} = -H \quad (6.6)$$

$$\because G \gg 1 \Rightarrow H = \frac{R_s}{R_F} \quad (6.7)$$

Now from equation 6.2, 6.3 and 6.6 we get:

$$-\frac{V_o}{G} = V_s + HV_o \quad (6.8)$$

$$\Rightarrow \frac{V_o}{V_s} = -\frac{G}{1 + GH} \quad (6.9)$$

Now, for "f" to be 3-dB frequency given condition should match i.e.,:

$$\left| \frac{V_o}{V_s} \right| = \frac{1}{\sqrt{2}} \quad (6.10)$$

$$\Rightarrow \left| -\frac{G}{1 + GH} \right| = \frac{1}{\sqrt{2}} \quad (6.11)$$

$$\Rightarrow \frac{\frac{\mu}{1 + \frac{jf}{f_c}}}{1 + \frac{(R_s)}{(R_F)} \frac{\mu}{1 + \frac{jf}{f_c}}} = \frac{1}{\sqrt{2}} \quad (6.12)$$

Parameters	Values
$R_s$	1k $\Omega$
$R_F$	10k $\Omega$
$\mu$	1000
$f_c$	1kHz

TABLE 6

Now putting the appropriate values as given in Table 6 we get:

$$\frac{\frac{1000}{1 + \frac{jf}{1000}}}{1 + \frac{(1)}{(10)} \frac{1000}{1 + \frac{jf}{1000}}} = \frac{1}{\sqrt{2}} \quad (6.13)$$

$$\frac{f^2}{10^{12}} + \frac{101^2}{10^6} = 2 \quad (6.14)$$

$$f \approx 1.41 \text{ MHz} \quad (6.15)$$

7. Using ngspice verify the Closed-Loop Transfer function or  $V_o/V_s$ .

**Solution:** From 5.13 we know that:

$$\frac{V_o}{V_s} = -9.88 \text{ V/V} \quad (7.1)$$

So, to verify this use the following spice file.

spice/ee18btech11011/ee18btech11011.net

and finally to get the result use the following python code.

spice/ee18btech11011/ee18btech11011\_spice.py

Result:

```
figs/ee18btech11011/  
ee18btech11011_spice_result.eps
```

```
figs/ee18btech11011/  
ee18btech11011_spice_result.pdf
```

Following are the instructions to run the spice file.

```
spice/ee18btech11011/README.md
```