

Problem 5.59 Relate v_{out} to v_i in the circuit of Fig. P5.59. Assume $v_C = 0$ at $t = 0$.

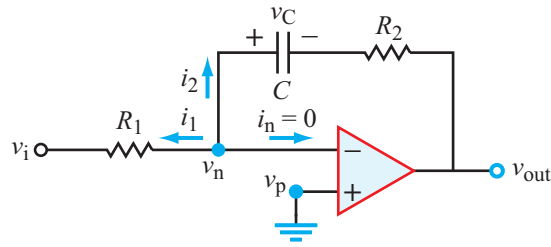


Figure P5.59: Circuit for Problem 5.59.

Solution:

$$i_1 = \frac{v_n - v_i}{R_1}$$

$$v_n - v_{\text{out}} = i_2 R_2 + \frac{1}{C} \int_0^t i_2 dt$$

But $v_n = v_p = 0$, and

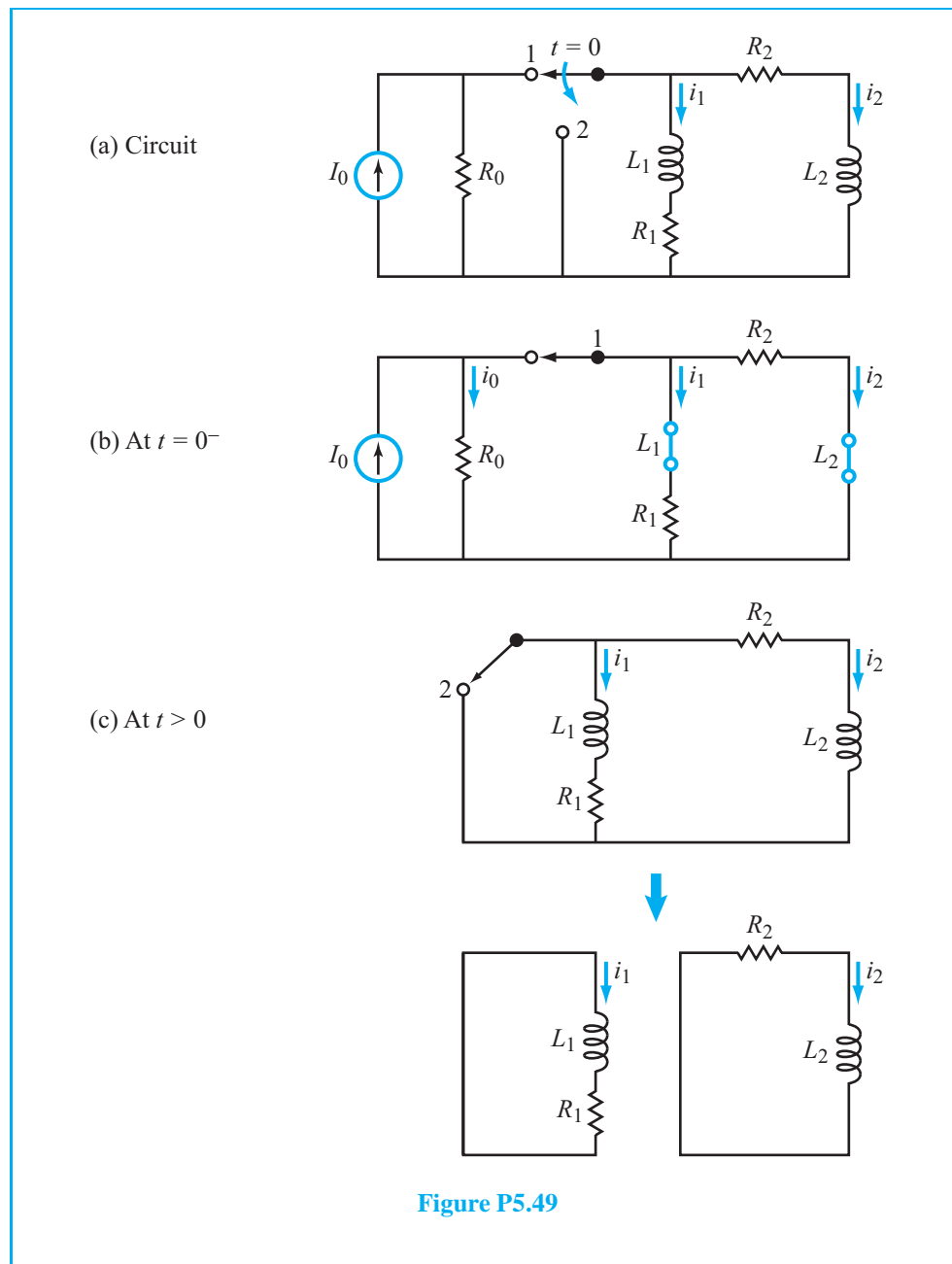
$$i_2 = -i_1 = \frac{v_i}{R_1},$$

which leads to

$$v_{\text{out}} = - \left(\frac{R_2}{R_1} v_i + \frac{1}{R_1 C} \int_0^t v_i dt \right).$$

Problem 5.49 After having been in position 1 for a long time, the switch in the circuit of Fig. P5.49 was moved to position 2 at $t = 0$. Determine $i_1(t)$ and $i_2(t)$ for $t \geq 0$ given that $I_0 = 6 \text{ mA}$, $R_0 = 12 \Omega$, $R_1 = 10 \Omega$, $R_2 = 40 \Omega$, $L_1 = 1 \text{ H}$, and $L_2 = 2 \text{ H}$.

Solution:



At $t = 0^-$ (Fig. P5.49(b)), I_0 will flow through the three branches such that

$$i_0 R_0 = i_1 R_1 = i_2 R_2,$$

and $I_0 = i_0 + i_1 + i_2$. Hence,

$$i_1(0^-) = \frac{R_0 R_2 I_0}{R_0 R_1 + R_0 R_2 + R_1 R_2} = 2.88 \quad (\text{mA}),$$

$$i_2(0^-) = \frac{R_0 R_1 I_0}{R_0 R_1 + R_0 R_2 + R_1 R_2} = 0.72 \quad (\text{mA}).$$

At $t > 0$, we have two independent RL circuits sharing a common short circuit.

$R_1 L_1$ Circuit

$$i_1(0) = i_1(0^-) = 2.88 \quad (\text{mA})$$

$$i_1(\infty) = 0$$

$$\tau_1 = \frac{L_1}{R_1} = \frac{1}{10} = 0.1 \text{ s}$$

$$i_1(t) = 2.88e^{-10t} \quad (\text{mA}), \quad \text{for } t \geq 0.$$

$R_2 L_2$ Circuit

$$i_2(0) = i_2(0^-) = 0.72 \quad (\text{mA})$$

$$i_2(\infty) = 0$$

$$\tau_2 = \frac{L_2}{R_2} = \frac{2}{40} = 0.05 \text{ s}$$

$$i_2(t) = 0.72e^{-20t} \quad (\text{mA}), \quad \text{for } t \geq 0.$$

CHAPTER 7

Section 7-1: Sinusoidal Signals

Problem 7.1 Express the sinusoidal waveform

$$v(t) = -4 \sin(8\pi \times 10^3 t - 45^\circ) \text{ V}$$

in standard cosine form and then determine its amplitude, frequency, period, and phase angle.

Solution:

(a)

$$\begin{aligned} v(t) &= -4 \sin(8\pi \times 10^3 t - 45^\circ) \\ &= 4 \cos(8\pi \times 10^3 t - 45^\circ + 90^\circ) \quad [-\sin x = \cos(x + 90^\circ)] \\ &= 4 \cos(8\pi \times 10^3 t + 45^\circ) \quad (\text{V}). \end{aligned}$$

(b)

$$\text{amplitude} = 4 \text{ V}$$

(c)

$$f = 4 \times 10^3 \text{ Hz} = 4 \text{ kHz}$$

(d)

$$T = \frac{1}{f} = \frac{1}{4 \times 10^3} = 0.25 \text{ ms}$$

(e)

$$\phi = 45^\circ$$

Problem 5.62 The two-stage op-amp circuit in Fig. P5.62 is driven by an input step voltage given by $v_i(t) = 10u(t)$ mV. If $V_{cc} = 10$ V for both op amps and the two capacitors had no charge prior to $t = 0$, determine and plot:

- (a) $v_{out1}(t)$ for $t \geq 0$
- (b) $v_{out2}(t)$ for $t \geq 0$

Solution:

(a)

$$v_{out1}(t) = -\frac{1}{R_1 C_1} \int_0^t v_i dt$$

$$R_1 C_1 = 5 \times 10^3 \times 4 \times 10^{-6} = 0.02.$$

Hence,

$$v_{out1}(t) = -50 \int_0^t 10 \times 10^{-3} dt = -0.5t \quad (\text{V}), \quad \text{for } t \geq 0.$$

(b) For the second stage:

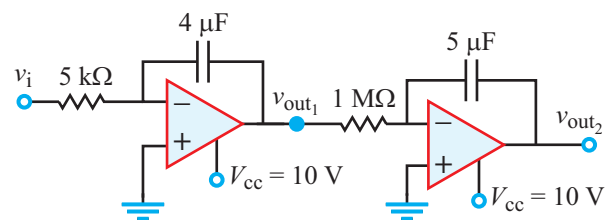
$$v_{out2}(t) = -\frac{1}{R_2 C_2} \int_0^t v_{out1}(t) dt$$

$$R_2 C_2 = 1 \times 10^6 \times 5 \times 10^{-6} = 5$$

$$v_{out2}(t) = -\frac{1}{5} \int_0^t (-0.5t) dt = 0.1 \frac{t^2}{2} = 0.05t^2 \quad (\text{V}), \quad \text{for } t \geq 0.$$

Plots of $v_i(t)$, $v_{out1}(t)$, and $v_{out2}(t)$ are shown below. We note that $v_{out1}(t)$ reaches saturation at $-V_{cc} = -10$ V after 20 s, and $v_{out2}(t)$ reaches saturation at $V_{cc} = +10$ V at $t = 14.14$ s.

(a)



(b)

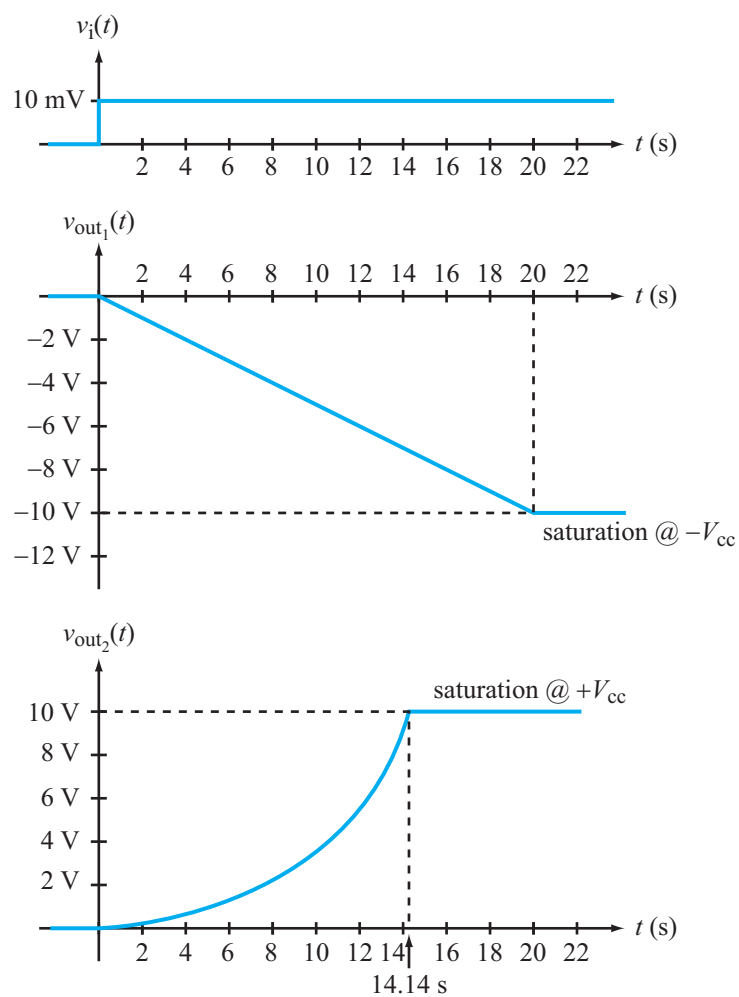


Figure P5.62

Problem 7.11 Express the following complex numbers in rectangular form:

(a) $\mathbf{z}_1 = 2e^{j\pi/6}$

(b) $\mathbf{z}_2 = -3e^{-j\pi/4}$

(c) $\mathbf{z}_3 = \sqrt{3} e^{-j3\pi/4}$

(d) $\mathbf{z}_4 = -j^3$

(e) $\mathbf{z}_5 = -j^{-4}$

(f) $\mathbf{z}_6 = (2 + j)^2$

(g) $\mathbf{z}_7 = (3 - j2)^3$

Solution:

(a) $\mathbf{z}_1 = 2e^{j\pi/6} = 2e^{j30^\circ} = 2\cos 30^\circ + j2\sin 30^\circ = 1.73 + j1.$

(b) $\mathbf{z}_2 = -3e^{-j\pi/4} = -3e^{-j45^\circ} = -3[\cos(-45^\circ) + j\sin(-45^\circ)] = -2.12 + j2.12.$

(c) $\mathbf{z}_3 = \sqrt{3} e^{-j3\pi/4} = \sqrt{3} e^{-j135^\circ} = \sqrt{3} [\cos 135^\circ - j\sin 135^\circ] = -1.22 - j1.22.$

(d) $\mathbf{z}_4 = -j^3 = -j \cdot j^2 = j.$

(e) $\mathbf{z}_5 = -j^{-4} = \frac{-1}{j^4} = -1.$

(f)

$$\begin{aligned}\mathbf{z}_6 &= (2 + j)^2 = \left[\sqrt{2^2 + 1^2} e^{j\tan^{-1}(1/2)} \right]^2 = \left[\sqrt{5} e^{j26.565^\circ} \right]^2 \\ &= 5e^{j53.13^\circ} \\ &= 5\cos 53.13^\circ + j5\sin 53.13^\circ = 3 + j4.\end{aligned}$$

(g)

$$\begin{aligned}\mathbf{z}_7 &= (3 - j2)^3 = \left[\sqrt{3^2 + 2^2} e^{-j\tan^{-1}(2/3)} \right]^3 = \left[\sqrt{13} e^{-j33.69^\circ} \right]^3 \\ &= 46.87e^{-j101.1^\circ} \\ &= 46.87(\cos 101.1^\circ - j\sin 101.1^\circ) \\ &= -9 - j46.\end{aligned}$$

Problem 7.29 The circuit in Fig. P7.29 is in the phasor domain. Determine the following:

- (a) The equivalent input impedance \mathbf{Z} at terminals (a, b) .
- (b) The phasor current \mathbf{I} , given that $\mathbf{V}_s = 25 \angle 45^\circ \text{ V}$.

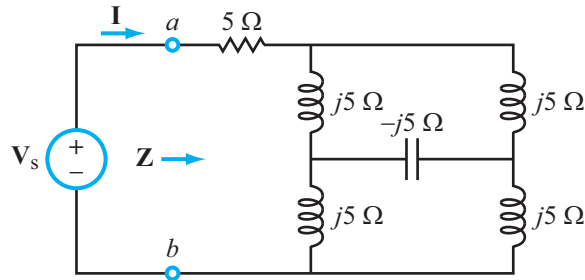
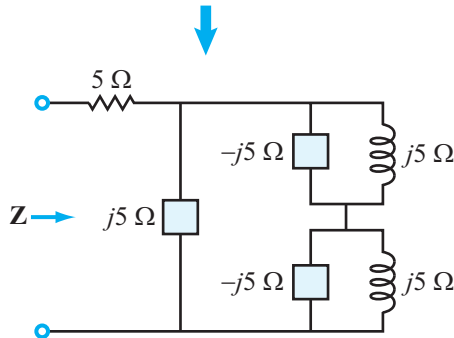
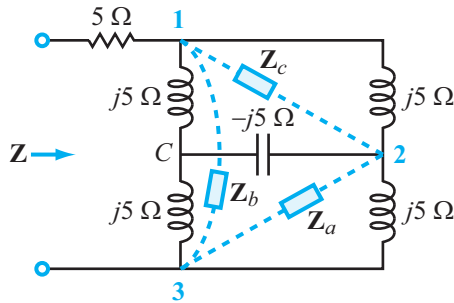


Figure P7.29: Circuit for Problem 7.29.

Solution:

(a)



$$\begin{aligned} \mathbf{Z}_a &= \frac{\mathbf{Z}_1 \mathbf{Z}_2 + \mathbf{Z}_2 \mathbf{Z}_3 + \mathbf{Z}_1 \mathbf{Z}_3}{\mathbf{Z}_1} \\ &= \frac{(j5)(-j5) + (-j5)(j5) + (j5)^2}{j5} = -j5 \, \Omega \\ \mathbf{Z}_b &= \frac{\mathbf{Z}_1 \mathbf{Z}_2 + \mathbf{Z}_2 \mathbf{Z}_3 + \mathbf{Z}_1 \mathbf{Z}_3}{\mathbf{Z}_2} = j5 \, \Omega \\ \mathbf{Z}_c &= \mathbf{Z}_a = -j5 \, \Omega. \\ \mathbf{Z} &= 5 + j5 \parallel 2(j5 \parallel -j5) \end{aligned}$$

$$\begin{aligned}
&= 5 + j5 \parallel 2 \left(\frac{25}{j5 - j5} \right) \\
&= 5 + j5 \parallel \infty \\
&= (5 + j5) = 5\sqrt{2} e^{j45^\circ} \quad (\Omega).
\end{aligned}$$

(b)

$$\mathbf{I} = \frac{\mathbf{V}_s}{\mathbf{Z}} = \frac{25e^{j45^\circ}}{5\sqrt{2} e^{j45^\circ}} = \frac{5}{\sqrt{2}} = 3.54 \text{ V}.$$