

# Computer Simulation of Systems/Circuits

## CENG 215 Lecture Notes

### 1 States and State Variables (Quick Primer)

A *state* is the minimal set of variables whose values at time  $t_0$  together with the input for  $t \geq t_0$  determine the future of the system.

Energy-storing elements define natural state choices:

- Capacitor:  $v_C(t)$  with stored energy  $W_C = \frac{1}{2}Cv_C^2$ ,
- Inductor:  $i_L(t)$  with stored energy  $W_L = \frac{1}{2}Li_L^2$ .

Hence in lumped RLC networks, a convenient state vector is  $x = \begin{bmatrix} v_C \\ i_L \end{bmatrix}$  (for a single  $C$  and  $L$ ).

#### General Series RLC (output $v_C$ )

For the series  $R\{L\{C$  with source  $v_s(t)$ :

$$v_s = v_R + v_L + v_C = R i + L \dot{i} + v_C, \quad i = C \dot{v}_C.$$

A *second-order ODE* for  $v_C$  follows:

$$LC \ddot{v}_C + RC \dot{v}_C + v_C = v_s(t). \quad (1)$$

A state-space form (one of many) is obtained by choosing  $x_1 = v_C$  and  $x_2 = i$ :

$$\dot{x}_1 = \frac{1}{C}x_2, \quad L\dot{x}_2 = -R x_2 - x_1 + v_s(t),$$

or in matrix form  $\dot{x} = Ax + Bu$ ,  $y = Cx$  with  $u = v_s$ ,  $y = v_C$ .

### 2 Series RC as a First-Order Special Case

Setting  $L = 0$  in (1) yields the familiar RC model:

$$RC \dot{v}_C + v_C = v_s(t) \iff \dot{v}_C = -\frac{1}{\tau}v_C + \frac{1}{\tau}v_s(t), \quad \tau = RC. \quad (2)$$

Here the single state is  $x(t) = v_C(t)$  and the state equation is

$$\dot{x}(t) = f(t, x(t), u(t)) = -\frac{1}{\tau}x(t) + \frac{1}{\tau}u(t), \quad u(t) = v_s(t).$$

### 3 Euler's (Forward) Method for Simulation

Given  $\dot{x}(t) = f(t, x, u)$  and a time grid  $t_k = t_0 + k h$  with step  $h > 0$ :

$$x_{k+1} = x_k + h f(t_k, x_k, u_k) \quad (\text{Forward Euler}). \quad (3)$$

Stability example (RC): for  $u$  bounded and  $\dot{x} = -\frac{1}{\tau}x$  (homogeneous),

$$x_{k+1} = \left(1 - \frac{h}{\tau}\right)x_k \Rightarrow \text{stability requires } \left|1 - \frac{h}{\tau}\right| < 1 \iff 0 < h < 2\tau.$$

Smaller  $h$  improves accuracy and damping of numerical oscillations; for stiff or oscillatory second order RLC, consider smaller  $h$  or implicit/Runge-Kutta methods.

## 4 Driven RC: From Model to Simulation

We simulate (2) for three common inputs:

$$(i) \text{ Step: } u(t) = A u(t), \quad (ii) \text{ Ramp: } u(t) = A t u(t), \quad (iii) \text{ Sinusoid: } u(t) = A \sin(\omega t).$$

### Analytic reference (closed-form solutions)

With  $x(0) = x_0$ :

$$\begin{aligned} \text{Step: } x(t) &= A + (x_0 - A)e^{-t/\tau}. \\ \text{Ramp: } x(t) &= A(t - \tau) + (x_0 + A\tau)e^{-t/\tau}. \\ \text{Sinusoid (steady state): } x_{ss}(t) &= \frac{A}{\sqrt{1 + (\omega\tau)^2}} \sin(\omega t - \tan^{-1}(\omega\tau)). \end{aligned}$$

### 4.1 Pseudo-code (Forward Euler)

```
Given R, C, tau=R*C, A, w, T_end, step h, initial x0
Define input u(t):
    STEP: u(t) = A
    RAMP: u(t) = A * t
    SINE: u(t) = A * sin(w * t)

Initialize:
t = 0
x = x0
Save (t, x)

Loop while t < T_end:
    u = input(t)
    f = -(1/tau) * x + (1/tau) * u
    x = x + h * f           # Forward Euler update
    t = t + h
    Save (t, x)

Optionally compute analytic x_ref(t) for comparison.
Plot u(t), x(t), (and x_ref(t) if computed).
```

### 4.2 Python code (NumPy + Matplotlib)

The script below simulates all three inputs, compares with analytic formulas, and saves figures:

- rc\_step.png, rc\_ramp.png, rc\_sine.png

Listing 1: rc\_sim.py Euler simulation of driven RC with step, ramp, and sinusoid

```
1 import numpy as np
2 import matplotlib.pyplot as plt
3
4 # ----- Parameters (edit freely) -----
5 R = 1_000.0          # Ohms
6 C = 1e-4             # Farads
7 A = 5.0              # Volts
8 w = 50.0             # rad/s (for sinusoid)
9 tau = R*C
10 x0 = 0.0            # initial capacitor voltage (V)
11 T_end = 5*tau        # total simulation time
12 h = 0.002*tau        # step size (keep 0 < h < 2*tau for Euler stability)
13 # -----
14
15 # Time grid
```

```

16 N = int(np.ceil(T_end/h)) + 1
17 t = np.linspace(0.0, N*h, N)
18 t = np.minimum(t, T_end)
19
20 def euler_sim(u_func):
21     x = np.zeros_like(t)
22     x[0] = x0
23     for k in range(len(t)-1):
24         u = u_func(t[k])
25         f = -(1.0/tau)*x[k] + (1.0/tau)*u
26         x[k+1] = x[k] + h*f
27     return x
28
29 # Inputs
30 u_step = lambda tt: A
31 u_ramp = lambda tt: A*tt
32 u_sine = lambda tt: A*np.sin(w*tt)
33
34 # Simulations
35 x_step = euler_sim(u_step)
36 x_ramp = euler_sim(u_ramp)
37 x_sine = euler_sim(u_sine)
38
39 # Analytic references
40 x_step_ref = A + (x0 - A)*np.exp(-t/tau)
41 x_ramp_ref = A*(t - tau) + (x0 + A*tau)*np.exp(-t/tau)
42 # For sinusoid, plot steady state reference only:
43 x_sine_ss = (A/np.sqrt(1+(w*tau)**2))*np.sin(w*t - np.arctan(w*tau))
44
45 # ----- Plot helpers -----
46 def plot_and_save(t, u, x_num, x_ref, title, fname, show_ref=True):
47     plt.figure(figsize=(7.0,3.6))
48     plt.plot(t, u, label='Input u(t)')
49     plt.plot(t, x_num, label='Numerical x(t) (Euler)', linewidth=2)
50     if show_ref and x_ref is not None:
51         plt.plot(t, x_ref, '--', label='Analytic reference')
52     plt.xlabel('t (s)')
53     plt.ylabel('Voltage (V)')
54     plt.grid(True, alpha=0.3)
55     plt.title(title)
56     plt.legend()
57     plt.tight_layout()
58     plt.savefig(fname, dpi=160)
59     plt.close()
60
61 # Save figures
62 plot_and_save(t, A*np.ones_like(t), x_step, x_step_ref,
63 f'RC Step Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_step.png')
64
65 plot_and_save(t, A*t, x_ramp, x_ramp_ref,
66 f'RC Ramp Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_ramp.png')
67
68 plot_and_save(t, A*np.sin(w*t), x_sine, x_sine_ss,
69 f'RC Sinusoid Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_sine.png')
70 print("Saved: rc_step.png, rc_ramp.png, rc_sine.png")

```

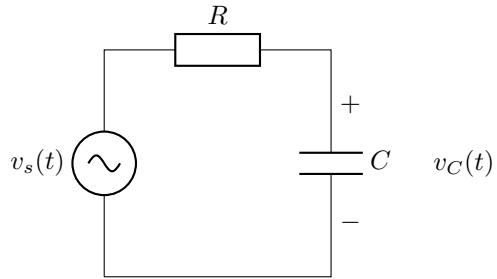


Figure 1: Series RC with capacitor voltage  $v_C(t)$  as output.

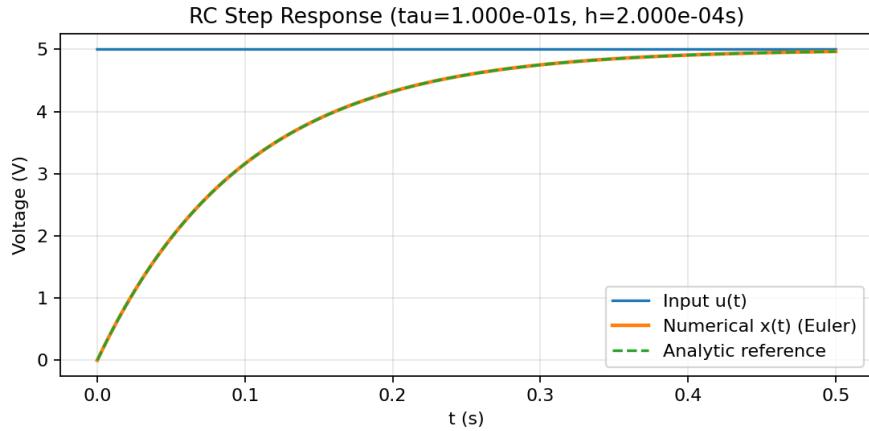


Figure 2: Euler vs Analytic: Step input.

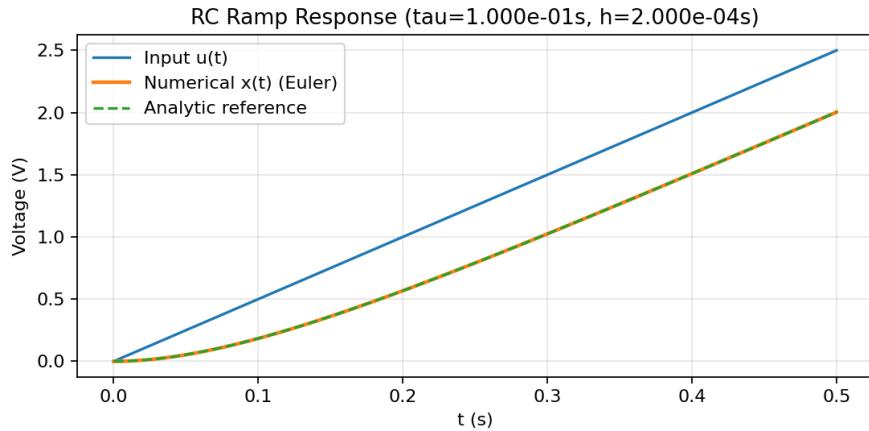


Figure 3: Euler vs Analytic: Ramp input.

#### 4.3 Circuit schematic (for the driven RC)

### 5 Simulation Plots

#### Step, Ramp, and Sinusoid Results

### 6 Practical Guidance

- Step size ( $h$ ): For RC with  $\tau = RC$ , keep  $0 < h < 2\tau$  for forward Euler stability; usually  $h \leq 0.05\tau$  gives good accuracy.
- Inputs: If  $u(t)$  is discontinuous (step), smaller  $h$  around the jump captures the corner more accurately.

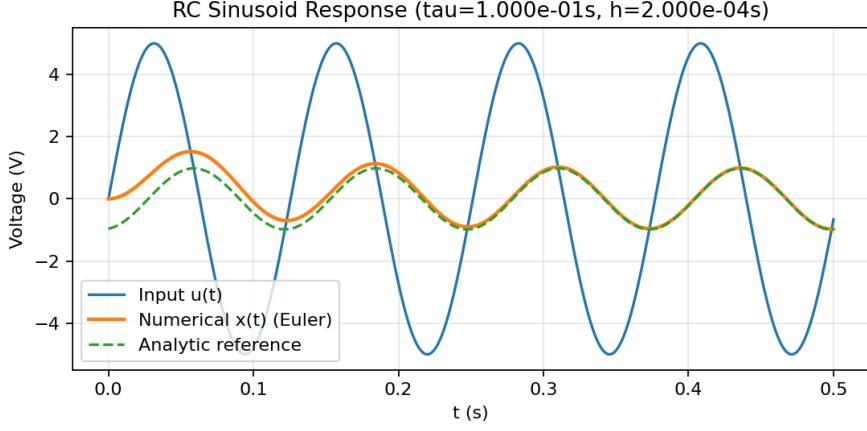


Figure 4: Euler vs Analytic (steady state reference shown): Sinusoid input.

- RLC systems: Oscillatory or stiff cases may require smaller  $h$  or higher-order/implicit schemes (e.g., RK4, Trapezoidal, BDF).
- Verification: Always compare with a known analytic solution (when available) to calibrate  $h$  and validate your simulator.

## 7 Nonlinear RC Simulation with a Quadratic i–v Device

In the previous sections, we considered a linear resistor in series with a capacitor and derived the state equation

$$\dot{x}(t) = -\frac{1}{\tau}x(t) + \frac{1}{\tau}v_s(t), \quad x(t) = v_C(t).$$

We now replace the resistor by a nonlinear two-terminal device whose current–voltage characteristic is

$$i = 0.01 v^2,$$

where  $v$  is the voltage across the nonlinear element. We keep the series connection with the capacitor, and the source  $v_s(t)$  remains the input.

### Nonlinear Device Model and State Equation

Let  $v_C(t)$  be the capacitor voltage (state) and  $i(t)$  the series current. The circuit is

$$v_s(t) = v_{NL}(t) + v_C(t),$$

where  $v_{NL}$  is the voltage across the nonlinear device. The device law is

$$i(t) = 0.01 v_{NL}^2(t),$$

and the capacitor current relation is

$$i(t) = C \dot{v}_C(t).$$

Because the elements are in series, the same current flows through both, so

$$C \dot{v}_C(t) = 0.01 v_{NL}^2(t).$$

Using  $v_{NL}(t) = v_s(t) - v_C(t)$ , we obtain the scalar nonlinear state equation

$$\dot{v}_C(t) = \frac{0.01}{C} (v_s(t) - v_C(t))^2.$$

Defining  $x(t) = v_C(t)$  gives

$$\dot{x}(t) = f(t, x) = \frac{0.01}{C} (v_s(t) - x(t))^2.$$

For a step input  $v_s(t) = A u(t)$ , the state equation becomes

$$\dot{x}(t) = \frac{0.01}{C} (A - x(t))^2, \quad x(0) = x_0.$$

There is no simple closed-form exponential solution as in the linear RC case; instead we will integrate this nonlinear equation numerically.

## Forward Euler Update for the Nonlinear RC

Given a time grid  $t_k = kh$  with step size  $h > 0$ , and state samples  $x_k \approx x(t_k)$ , Forward Euler applied to

$$\dot{x}(t) = \frac{0.01}{C} (v_s(t) - x(t))^2$$

yields

$$x_{k+1} = x_k + h \frac{0.01}{C} (v_s(t_k) - x_k)^2.$$

For a step input  $v_s(t) = A u(t)$  (so  $v_s(t_k) = A$  for  $t_k > 0$ ),

$$x_{k+1} = x_k + h \frac{0.01}{C} (A - x_k)^2.$$

As in the linear case,  $h$  must be chosen sufficiently small to obtain a stable and accurate numerical solution. Because the right-hand side grows with  $(A-x)^2$ , too large a step size can cause numerical blowup.

## Python Simulation Code (Step Input)

The script below simulates the nonlinear RC circuit with step input  $v_s(t) = A u(t)$  using Forward Euler, and plots both the input voltage and the capacitor voltage.

Listing 2: Euler simulation of nonlinear RC with quadratic i-v device

```

1      import numpy as np
2      import matplotlib.pyplot as plt
3
4      # ----- Parameters (edit freely) -----
5      C = 1e-4          # Farads
6      A = 5.0           # Step amplitude (Volts)
7      x0 = 0.0           # Initial capacitor voltage (V)
8      Tend = 0.3         # Total simulation time (s)
9      h = 1e-5           # Time step (s) - choose small for stability
10     # -----
11
12     # Time grid
13     N = int(np.ceil(Tend / h)) + 1
14     t = np.linspace(0.0, Tend, N)
15
16     def vsource(tt):
17         # Step input v_s(t) = A u(t)
18         return A
19
20     def f(tval, xval):
21         v = vsource(tval)
22         return 0.01 / C * (v - xval)**2
23
24     # Forward Euler simulation
25     x = np.zeros_like(t)
26     x[0] = x0
27     for k in range(len(t) - 1):
28         x[k+1] = x[k] + h * f(t[k], x[k])
29
30     # Build input waveform for plotting
31     u = np.array([vsouce(tt) for tt in t])
32
33     # Plot results

```

```

34     plt.figure(figsize=(7.0, 3.6))
35     plt.plot(t, u, label="Input v_s(t)")
36     plt.plot(t, x, label="Capacitor voltage v_C(t)", linewidth=2)
37     plt.xlabel("t (s)")
38     plt.ylabel("Voltage (V)")
39     plt.grid(True, alpha=0.3)
40     plt.title("Nonlinear RC: step response with i = 0.01 v^2")
41     plt.legend()
42     plt.tight_layout()
43     plt.show()

```

## Remarks

- The right{hand side is always nonnegative for a step with  $A > 0$ , so  $x(t)$  is monotonically increasing as long as  $x(t) < A$ .
- Unlike the linear RC, the approach to the input level is not exponential; the dynamics slow down as  $x(t) \rightarrow A$  because the current becomes small.
- The same code structure can be reused for other nonlinear devices by simply changing the function  $f(tval, xval)$  to implement the desired  $i\{v$  relation.

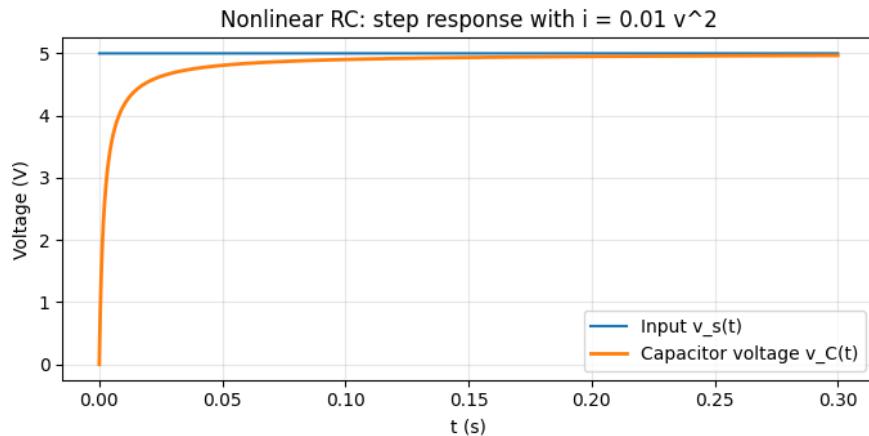


Figure 5: Simulation results: Non-linear device in series with capacitor