

Computer Simulation of Systems/Circuits

CENG 215 Lecture Notes

1 States and State Variables (Quick Primer)

A *state* is the minimal set of variables whose values at time t_0 together with the input for $t \geq t_0$ determine the future of the system.

Energy-storing elements define natural state choices:

- Capacitor: $v_C(t)$ with stored energy $W_C = \frac{1}{2}Cv_C^2$,
- Inductor: $i_L(t)$ with stored energy $W_L = \frac{1}{2}Li_L^2$.

Hence in lumped RLC networks, a convenient state vector is $x = \begin{bmatrix} v_C \\ i_L \end{bmatrix}$ (for a single C and L).

General Series RLC (output v_C)

For the series $R\{L\{C$ with source $v_s(t)$:

$$v_s = v_R + v_L + v_C = Ri + L\dot{i} + v_C, \quad i = C\dot{v}_C.$$

A *second-order* ODE for v_C follows:

$$LC\ddot{v}_C + RC\dot{v}_C + v_C = v_s(t). \quad (1)$$

A state-space form (one of many) is obtained by choosing $x_1 = v_C$ and $x_2 = \dot{v}_C$:

$$\dot{x}_1 = x_2, \quad L\dot{x}_2 = -Rx_2 - x_1 + v_s(t),$$

or in matrix form $\dot{x} = Ax + Bu$, $y = Cx$ with $u = v_s$, $y = v_C$.

2 Series RC as a First-Order Special Case

Setting $L = 0$ in (1) yields the familiar RC model:

$$RC\dot{v}_C + v_C = v_s(t) \iff \dot{v}_C = -\frac{1}{\tau}v_C + \frac{1}{\tau}v_s(t), \quad \tau = RC. \quad (2)$$

Here the single state is $x(t) = v_C(t)$ and the state equation is

$$\dot{x}(t) = f(t, x(t), u(t)) = -\frac{1}{\tau}x(t) + \frac{1}{\tau}u(t), \quad u(t) = v_s(t).$$

3 Euler's (Forward) Method for Simulation

Given $\dot{x}(t) = f(t, x, u)$ and a time grid $t_k = t_0 + kh$ with step $h > 0$:

$$x_{k+1} = x_k + hf(t_k, x_k, u_k) \quad (\text{Forward Euler}). \quad (3)$$

Stability example (RC): for u bounded and $\dot{x} = -\frac{1}{\tau}x$ (homogeneous),

$$x_{k+1} = \left(1 - \frac{h}{\tau}\right)x_k \implies \text{stability requires } |1 - \frac{h}{\tau}| < 1 \iff 0 < h < 2\tau.$$

Smaller h improves accuracy and damping of numerical oscillations; for stiff or oscillatory second order RLC, consider smaller h or implicit/Runge-Kutta methods.

4 Driven RC: From Model to Simulation

We simulate (2) for three common inputs:

(i) Step: $u(t) = A u(t)$, (ii) Ramp: $u(t) = A t u(t)$, (iii) Sinusoid: $u(t) = A \sin(\omega t)$.

Analytic reference (closed-form solutions)

With $x(0) = x_0$:

Step: $x(t) = A + (x_0 - A)e^{-t/\tau}$.

Ramp: $x(t) = A(t - \tau) + (x_0 + A\tau)e^{-t/\tau}$.

Sinusoid (steady state): $x_{ss}(t) = \frac{A}{\sqrt{1 + (\omega\tau)^2}} \sin(\omega t - \tan^{-1}(\omega\tau))$.

4.1 Pseudo-code (Forward Euler)

Given R, C, tau=R*C, A, w, T_end, step h, initial x0

Define input u(t):

STEP: u(t) = A

RAMP: u(t) = A * t

SINE: u(t) = A * sin(w * t)

Initialize:

t = 0

x = x0

Save (t, x)

Loop while t < T_end:

u = input(t)

f = -(1/tau) * x + (1/tau) * u

x = x + h * f # Forward Euler update

t = t + h

Save (t, x)

Optionally compute analytic x_ref(t) for comparison.

Plot u(t), x(t), (and x_ref(t) if computed).

4.2 Python code (NumPy + Matplotlib)

The script below simulates all three inputs, compares with analytic formulas, and saves figures:

- rc_step.png, rc_ramp.png, rc_sine.png

Listing 1: rc-sim.py Euler simulation of driven RC with step, ramp, and sinusoid

```
1 import numpy as np
2 import matplotlib.pyplot as plt
3
4 # ----- Parameters (edit freely) -----
5 R = 1_000.0                      # Ohms
6 C = 1e-4                        # Farads
7 A = 5.0                         # Volts
8 w = 50.0                        # rad/s (for sinusoid)
9 tau = R*C
10 x0 = 0.0                        # initial capacitor voltage (V)
11 T_end = 5*tau                  # total simulation time
12 h = 0.002*tau                 # step size (keep 0 < h < 2*tau for Euler stability)
13 # -----
14
15 # Time grid
```

```

16 N = int(np.ceil(T_end/h)) + 1
17 t = np.linspace(0.0, N*h, N)
18 t = np.minimum(t, T_end)
19
20 def euler_sim(u_func):
21     x = np.zeros_like(t)
22     x[0] = x0
23     for k in range(len(t)-1):
24         u = u_func(t[k])
25         f = -(1.0/tau)*x[k] + (1.0/tau)*u
26         x[k+1] = x[k] + h*f
27     return x
28
29 # Inputs
30 u_step = lambda tt: A
31 u_ramp = lambda tt: A*tt
32 u_sine = lambda tt: A*np.sin(w*tt)
33
34 # Simulations
35 x_step = euler_sim(u_step)
36 x_ramp = euler_sim(u_ramp)
37 x_sine = euler_sim(u_sine)
38
39 # Analytic references
40 x_step_ref = A + (x0 - A)*np.exp(-t/tau)
41 x_ramp_ref = A*(t - tau) + (x0 + A*tau)*np.exp(-t/tau)
42 # For sinusoid, plot steady state reference only:
43 x_sine_ss = (A/np.sqrt(1+(w*tau)**2))*np.sin(w*t - np.arctan(w*tau))
44
45 # ----- Plot helpers -----
46 def plot_and_save(t, u, x_num, x_ref, title, fname, show_ref=True):
47     plt.figure(figsize=(7.0,3.6))
48     plt.plot(t, u, label='Input u(t)')
49     plt.plot(t, x_num, label='Numerical x(t) (Euler)', linewidth=2)
50     if show_ref and x_ref is not None:
51         plt.plot(t, x_ref, '--', label='Analytic reference')
52     plt.xlabel('t (s)')
53     plt.ylabel('Voltage (V)')
54     plt.grid(True, alpha=0.3)
55     plt.title(title)
56     plt.legend()
57     plt.tight_layout()
58     plt.savefig(fname, dpi=160)
59     plt.close()
60
61 # Save figures
62 plot_and_save(t, A*np.ones_like(t), x_step, x_step_ref,
63 f'RC Step Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_step.png')
64
65 plot_and_save(t, A*t, x_ramp, x_ramp_ref,
66 f'RC Ramp Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_ramp.png')
67
68 plot_and_save(t, A*np.sin(w*t), x_sine, x_sine_ss,
69 f'RC Sinusoid Response (tau={tau:.3e}s, h={h:.3e}s)', 'rc_sine.png')
70 print("Saved: rc_step.png, rc_ramp.png, rc_sine.png")

```

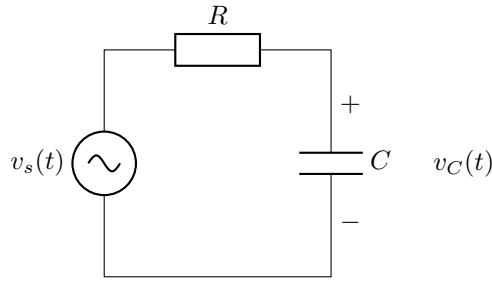


Figure 1: Series RC with capacitor voltage $v_C(t)$ as output.

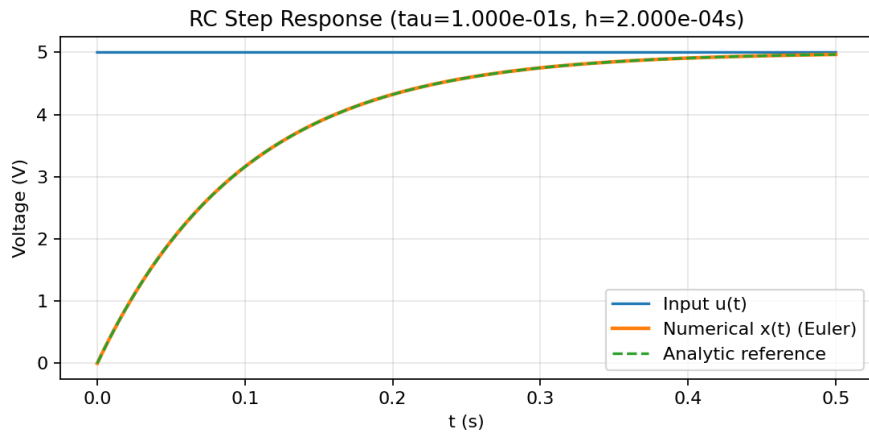


Figure 2: Euler vs Analytic: Step input.

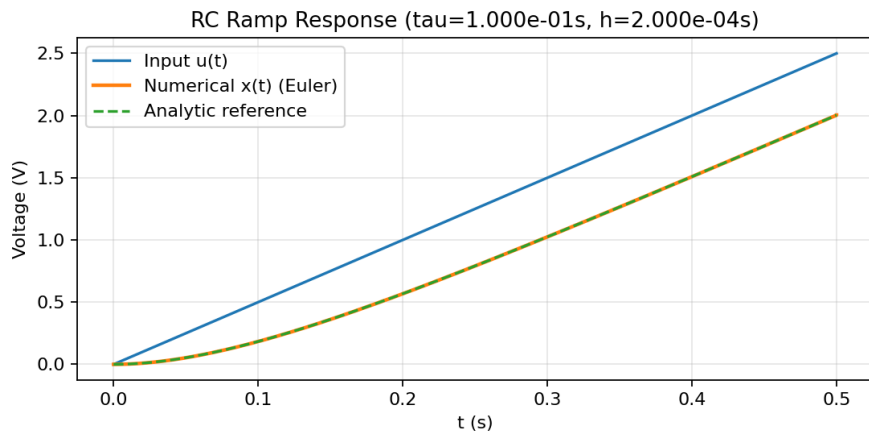


Figure 3: Euler vs Analytic: Ramp input.

4.3 Circuit schematic (for the driven RC)

5 Simulation Plots

Step, Ramp, and Sinusoid Results

6 Practical Guidance

- Step size (h): For RC with $\tau = RC$, keep $0 < h < 2\tau$ for forward Euler stability; usually $h \leq 0.05\tau$ gives good accuracy.
- Inputs: If $u(t)$ is discontinuous (step), smaller h around the jump captures the corner more accurately.

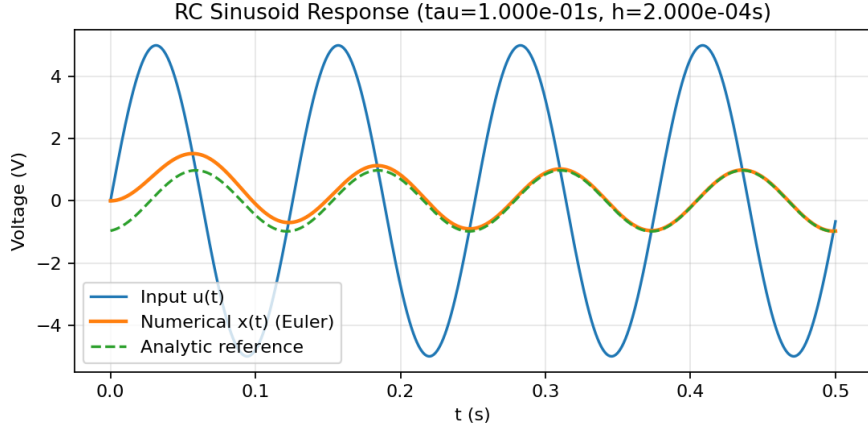


Figure 4: Euler vs Analytic (steady state reference shown): Sinusoid input.

- RLC systems: Oscillatory or stiff cases may require smaller h or higher-order/implicit schemes (e.g., RK4, Trapezoidal, BDF).
- Verification: Always compare with a known analytic solution (when available) to calibrate h and validate your simulator.

7 Nonlinear RC Simulation with a Quadratic i–v Device

In the previous sections, we considered a linear resistor in series with a capacitor and derived the state equation

$$\dot{x}(t) = -\frac{1}{\tau}x(t) + \frac{1}{\tau}v_s(t), \quad x(t) = v_C(t).$$

We now replace the resistor by a nonlinear two--terminal device whose current--voltage characteristic is

$$i = 0.01 v^2,$$

where v is the voltage across the nonlinear element. We keep the series connection with the capacitor, and the source $v_s(t)$ remains the input.

Nonlinear Device Model and State Equation

Let $v_C(t)$ be the capacitor voltage (state) and $i(t)$ the series current. The circuit is

$$v_s(t) = v_{\text{NL}}(t) + v_C(t),$$

where v_{NL} is the voltage across the nonlinear device. The device law is

$$i(t) = 0.01 v_{\text{NL}}^2(t),$$

and the capacitor current relation is

$$i(t) = C \dot{v}_C(t).$$

Because the elements are in series, the same current flows through both, so

$$C \dot{v}_C(t) = 0.01 v_{\text{NL}}^2(t).$$

Using $v_{\text{NL}}(t) = v_s(t) - v_C(t)$, we obtain the scalar nonlinear state equation

$$\dot{v}_C(t) = \frac{0.01}{C} (v_s(t) - v_C(t))^2.$$

Defining $x(t) = v_C(t)$ gives

$$\dot{x}(t) = f(t, x) = \frac{0.01}{C} (v_s(t) - x(t))^2.$$

For a step input $v_s(t) = Au(t)$, the state equation becomes

$$\dot{x}(t) = \frac{0.01}{C} (A - x(t))^2, \quad x(0) = x_0.$$

There is no simple closed-form exponential solution as in the linear RC case; instead we will integrate this nonlinear equation numerically.

Forward Euler Update for the Nonlinear RC

Given a time grid $t_k = kh$ with step size $h > 0$, and state samples $x_k \approx x(t_k)$, Forward Euler applied to

$$\dot{x}(t) = \frac{0.01}{C} (v_s(t) - x(t))^2$$

yields

$$x_{k+1} = x_k + h \frac{0.01}{C} (v_s(t_k) - x_k)^2.$$

For a step input $v_s(t) = Au(t)$ (so $v_s(t_k) = A$ for $t_k > 0$),

$$x_{k+1} = x_k + h \frac{0.01}{C} (A - x_k)^2.$$

As in the linear case, h must be chosen sufficiently small to obtain a stable and accurate numerical solution. Because the right-hand side grows with $(A-x)^2$, too large a step size can cause numerical blowup.

Python Simulation Code (Step Input)

The script below simulates the nonlinear RC circuit with step input $v_s(t) = Au(t)$ using Forward Euler, and plots both the input voltage and the capacitor voltage.

Listing 2: Euler simulation of nonlinear RC with quadratic i-v device

```

1  import numpy as np
2  import matplotlib.pyplot as plt
3
4  # ----- Parameters (edit freely) -----
5  C = 1e-4          # Farads
6  A = 5.0           # Step amplitude (Volts)
7  x0 = 0.0          # Initial capacitor voltage (V)
8  Tend = 0.3        # Total simulation time (s)
9  h = 1e-5          # Time step (s) - choose small for stability
10 # -----
11
12 # Time grid
13 N = int(np.ceil(Tend / h)) + 1
14 t = np.linspace(0.0, Tend, N)
15
16 def vsource(tt):
17     # Step input v_s(t) = A u(t)
18     return A
19
20 def f(tval, xval):
21     v = vsource(tval)
22     return 0.01 / C * (v - xval)**2
23
24 # Forward Euler simulation
25 x = np.zeros_like(t)
26 x[0] = x0
27 for k in range(len(t) - 1):
28     x[k+1] = x[k] + h * f(t[k], x[k])
29
30 # Build input waveform for plotting
31 u = np.array([vsource(tt) for tt in t])
32
33 # Plot results

```

```

34 plt.figure(figsize=(7.0, 3.6))
35 plt.plot(t, u, label="Input v_s(t)")
36 plt.plot(t, x, label="Capacitor voltage v_C(t)", linewidth=2)
37 plt.xlabel("t (s)")
38 plt.ylabel("Voltage (V)")
39 plt.grid(True, alpha=0.3)
40 plt.title("Nonlinear RC: step response with i = 0.01 v^2")
41 plt.legend()
42 plt.tight_layout()
43 plt.show()

```

Remarks

- The right hand side is always nonnegative for a step with $A > 0$, so $x(t)$ is monotonically increasing as long as $x(t) < A$.
- Unlike the linear RC, the approach to the input level is not exponential; the dynamics slow down as $x(t) \rightarrow A$ because the current becomes small.
- The same code structure can be reused for other nonlinear devices by simply changing the function $f(tval, xval)$ to implement the desired $i\{v$ relation.

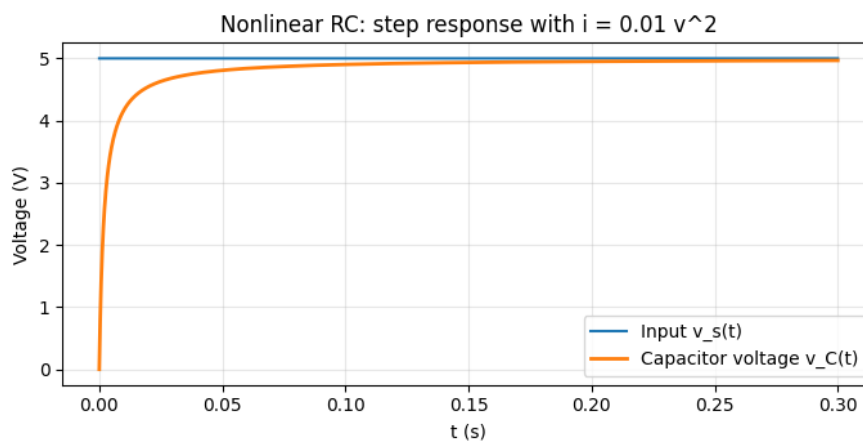


Figure 5: Simulation results: Non-linear device in series with capacitor