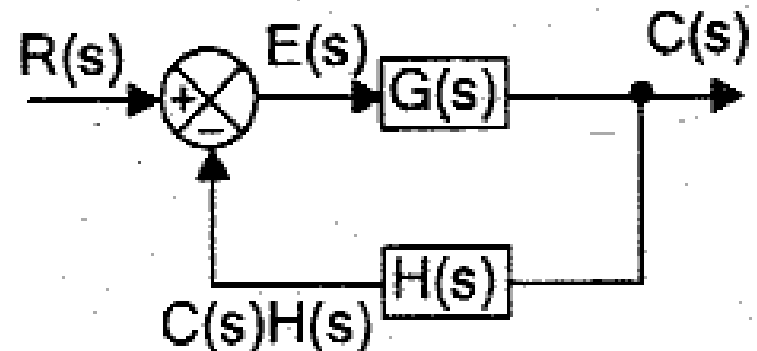


# Module 2

## Error analysis

# Steady state error

Let,  $R(s)$  = Input signal  
 $E(s)$  = Error signal  
 $C(s) H(s)$  = Feedback signal  
 $C(s)$  = Output signal or response



$$E(s) = R(s) - C(s) H(s)$$

$$E(s) = R(s) - [E(s) G(s)] H(s)$$

$$E(s) + E(s) G(s) H(s) = R(s)$$

$$E(s) [1 + G(s) H(s)] = R(s)$$

$$\therefore E(s) = \frac{R(s)}{1 + G(s) H(s)}$$

$$\therefore e(t) = \mathcal{L}^{-1}\{E(s)\} = \mathcal{L}^{-1}\left\{\frac{R(s)}{1 + G(s) H(s)}\right\}$$

Let,  $e_{ss}$  = steady state error.

The steady state error is defined as the value of  $e(t)$  when  $t$  tends to infinity.

$$\therefore e_{ss} = \lim_{t \rightarrow \infty} e(t)$$

The final value theorem of Laplace transform states that,

$$\text{If, } F(s) = \mathcal{L}\{f(t)\} \text{ then, } \lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} s F(s)$$

Using final value theorem,

$$\text{The steady state error, } e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s) = \lim_{s \rightarrow 0} \frac{s R(s)}{1 + G(s) H(s)}$$


---

# Static error constants

- Steady state error may be zero, constant or infinity depends on the type number and input signal
- **Type 0 and step input** – constant error
- **Type 1 and ramp input** (velocity signal) - constant error
- **Type 2 and parabolic input** (acceleration signal) - constant error

$$\lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)H(s)}$$

Positional error constant,  $K_p = \lim_{s \rightarrow 0} G(s) H(s)$

Velocity error constant,  $K_v = \lim_{s \rightarrow 0} s G(s) H(s)$

Acceleration error constant,  $K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s)$

The  $K_p$ ,  $K_v$  and  $K_a$  are in general called static error constants.

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## STEADY STATE ERROR WHEN THE INPUT IS UNIT STEP SIGNAL

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Steady state error,  $e_{ss} = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)H(s)}$

When the input is unit step,  $R(s) = 1/s$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} \frac{\frac{1}{s}}{1 + G(s)H(s)} = \lim_{s \rightarrow 0} \frac{1}{1 + G(s)H(s)} = \frac{1}{1 + \lim_{s \rightarrow 0} G(s)H(s)} = \frac{1}{1 + K_p}$$

$$\text{where, } K_p = \lim_{s \rightarrow 0} G(s)H(s)$$

The constant  $K_p$  is called *positional error constant*.

### Type-0 system

$$K_p = \lim_{s \rightarrow 0} G(s) H(s) = \lim_{s \rightarrow 0} K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{(s+p_1)(s+p_2)(s+p_3)\dots} = K \frac{z_1 \cdot z_2 \cdot z_3 \dots}{p_1 \cdot p_2 \cdot p_3 \dots} = \text{constant}$$

$$\therefore e_{ss} = \frac{1}{1+K_p} = \text{constant}$$

Hence in type-0 systems when the input is unit step there will be a constant steady state error.

### Type-1 system

$$K_p = \lim_{s \rightarrow 0} G(s) H(s) = \lim_{s \rightarrow 0} K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s(s+p_1)(s+p_2)(s+p_3)\dots} = \infty$$

$$\therefore e_{ss} = \frac{1}{1+K_p} = \frac{1}{1+\infty} = 0$$

In systems with type number 1 and above, for unit step input the value of  $K_p$  is infinity and so the steady state error is zero.

## STEADY STATE ERROR WHEN THE INPUT IS UNIT RAMP SIGNAL

$$\text{Steady state error, } e_{ss} = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)H(s)}$$

$$\text{When the input is unit ramp, } R(s) = \frac{1}{s^2}$$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} \frac{s \frac{1}{s^2}}{1 + G(s)H(s)} = \lim_{s \rightarrow 0} \frac{1}{s + sG(s)H(s)} = \frac{1}{\lim_{s \rightarrow 0} sG(s)H(s)} = \frac{1}{K_v}$$

$$\text{where, } K_v = \lim_{s \rightarrow 0} sG(s)H(s)$$

The constant  $K_v$  is called *velocity error constant*.



### Type-0 system

$$K_v = \lim_{s \rightarrow 0} sG(s)H(s) = \lim_{s \rightarrow 0} sK \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{(s+p_1)(s+p_2)(s+p_3)\dots} = 0$$

$$\therefore e_{ss} = 1/K_v = 1/0 = \infty$$

Hence in type-0 systems when the input is unit ramp, the steady state error is infinity.

### Type-1 system

$$K_v = \lim_{s \rightarrow 0} sG(s)H(s) = \lim_{s \rightarrow 0} sK \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s(s+p_1)(s+p_2)(s+p_3)\dots} = K \frac{z_1 \cdot z_2 \cdot z_3 \dots}{p_1 \cdot p_2 \cdot p_3 \dots} = \text{constant}$$

$$\therefore e_{ss} = 1/K_v = \text{constant}$$

Hence in type-1 systems when the input is unit ramp there will be a constant steady state error.

### Type-2 system

$$K_v = \lim_{s \rightarrow 0} sG(s)H(s) = \lim_{s \rightarrow 0} sK \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s^2(s+p_1)(s+p_2)(s+p_3)\dots} = \infty$$

$$\therefore e_{ss} = 1/K_v = 1/\infty = 0$$

In systems with type number 2 and above, for unit ramp input, the value of  $K_v$  is infinity so the steady state error is zero.

## STEADY STATE ERROR WHEN THE INPUT IS UNIT PARABOLIC SIGNAL

---

$$\text{Steady state error, } e_{ss} = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)H(s)}$$

$$\text{When the input is unit parabola, } R(s) = \frac{1}{s^3}$$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} \frac{s \frac{1}{s^3}}{1 + G(s)H(s)} = \lim_{s \rightarrow 0} \frac{1}{s^2 + s^2 G(s)H(s)} = \frac{1}{\lim_{s \rightarrow 0} s^2 G(s)H(s)} = \frac{1}{K_a}$$

$$\text{where, } K_a = \lim_{s \rightarrow 0} s^2 G(s)H(s)$$

The constant  $K_a$  is called *acceleration error constant*.

### Type-0 system

$$K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s) = \lim_{s \rightarrow 0} s^2 K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{(s+p_1)(s+p_2)(s+p_3)\dots} = 0$$

$$\therefore e_{ss} = \frac{1}{K_a} = \frac{1}{0} = \infty$$

Hence in type-0 systems for unit parabolic input, the steady state error is infinity.

### Type-1 system

$$K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s) = \lim_{s \rightarrow 0} s^2 K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s(s+p_1)(s+p_2)(s+p_3)\dots} = 0$$

$$\therefore e_{ss} = \frac{1}{K_a} = \frac{1}{0} = \infty$$

Hence in type-1 systems for unit parabolic input, the steady state error is infinity.

### Type-2 system

$$K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s) = \lim_{s \rightarrow 0} s^2 K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s^2(s+p_1)(s+p_2)(s+p_3)\dots} = K \frac{z_1 \cdot z_2 \cdot z_3 \dots}{p_1 \cdot p_2 \cdot p_3 \dots} = \text{constant}$$

$$\therefore e_{ss} = \frac{1}{K_a} = \text{constant}$$

Hence in type-2 system when the input is unit parabolic signal there will be a constant steady state error.

### Type-3 system

$$K_a = \lim_{s \rightarrow 0} s^2 G(s) H(s) = \lim_{s \rightarrow 0} s^2 K \frac{(s+z_1)(s+z_2)(s+z_3)\dots}{s^3 (s+p_1)(s+p_2)(s+p_3)\dots} = \infty$$

$$\therefore e_{ss} = \frac{1}{K_a} = \frac{1}{\infty} = 0$$

In systems with type number 3 and above for unit parabolic input the value of  $K_a$  is infinity and so the steady state error is zero.

**TABLE-2.2 : Static Error Constant for  
Various Type Number of Systems**

Error Constant	Type number of system			
	0	1	2	3
$K_p$	constant	$\infty$	$\infty$	$\infty$
$K_v$	0	constant	$\infty$	$\infty$
$K_a$	0	0	constant	$\infty$

**TABLE-2.3 : Steady State Error for  
Various Types of Inputs**

Input Signal	Type number of system			
	0	1	2	3
Unit Step	$\frac{1}{1+K_p}$	0	0	0
Unit Ramp	$\infty$	$\frac{1}{K_v}$	0	0
Unit Parabolic	$\infty$	$\infty$	$\frac{1}{K_a}$	0

# Generalized error coefficients

- Static error coefficients does not show variations of error with time and input should be standard input
- Generalised error coefficients give steady state error at any instants of time and for any type of input.

$$E(s) = \frac{R(s)}{1 + G(s) H(s)} = \frac{1}{1 + G(s) H(s)} R(s) = F(s) R(s)$$

where,  $F(s) = \frac{1}{1 + G(s) H(s)}$

Let,  $e(t) = \mathcal{L}^{-1}\{E(s)\}$  (error signal in time domain)

$f(t) = \mathcal{L}^{-1}\{F(s)\}$

$r(t) = \mathcal{L}^{-1}\{R(s)\}$  (input signal in time domain)

$$\text{i.e., } \mathcal{L}\{f(t) * r(t)\} = F(s) R(s)$$

where  $*$  is the symbol for convolution operation

$$\therefore \mathcal{L}^{-1}\{F(s) R(s)\} = f(t) * r(t)$$

$$f(t) * r(t) = \int_{-\infty}^{+\infty} f(T) r(t-T) dT ;$$

$$\therefore e(t) = \int_{-\infty}^{+\infty} f(T) r(t-T) dT$$

Using Taylor's series expansion the signal  $r(t-T)$  can be expressed as,

$$r(t-T) = r(t) - T \dot{r}(t) + \frac{T^2}{2!} \ddot{r}(t) - \frac{T^3}{3!} \dddot{r}(t) + \dots + (-1)^n \frac{T^n}{n!} r^{(n)}(t) \dots$$

where,  $\dot{r}(t) = 1^{\text{st}}$  derivative of  $r(t)$

$\ddot{r}(t) = 2^{\text{nd}}$  derivative of  $r(t)$

$\vdots$

$r^{(n)}(t) = n^{\text{th}}$  derivative of  $r(t)$



On substituting the Taylor's series expansion of  $r(t - T)$ , the error  $e(t)$  can be written as,

$$e(t) = \int_0^t f(T) \left[ r(t) - T \dot{r}(t) + \frac{T^2}{2!} \ddot{r}(t) - \frac{T^3}{3!} \dddot{r}(t) + \dots + (-1)^n \frac{T^n}{n!} r^{(n)}(t) \dots \right] dT$$

$$e(t) = \int_0^t f(T) r(t) dT - \int_0^t f(T) T \dot{r}(t) dT + \int_0^t f(T) \frac{T^2}{2!} \ddot{r}(t) dT \\ - \int_0^t f(T) \frac{T^3}{3!} \dddot{r}(t) dT + \dots + \int_0^t f(T) (-1)^n \frac{T^n}{n!} r^{(n)}(t) dT \dots \infty$$

$$e(t) = r(t) \int_0^t f(T) dT - \dot{r}(t) \int_0^t T f(T) dt + \frac{\ddot{r}(t)}{2!} \int_0^t T^2 f(T) dt \\ - \frac{\dddot{r}(t)}{3!} \int_0^t T^3 f(T) dt + \dots + (-1)^n \frac{r^{(n)}(t)}{n!} \int_0^t T^n f(T) dt \dots$$

$$\text{Let, } C_0 = + \int_0^t f(T) dT \qquad C_3 = - \int_0^t T^3 f(T) dT$$

$$C_1 = - \int_0^t T f(T) dT$$

⋮

$$C_2 = + \int_0^t T^2 f(T) dT$$

$$C_n = (-1)^n \int_0^t T^n f(T) dT$$

$$\begin{aligned}
 e(t) &= r(t) C_0 + \dot{r}(t) C_1 + \ddot{r}(t) \frac{C_2}{2!} + \ddot{\ddot{r}}(t) \frac{C_3}{3!} + \dots + r^{(n)}(t) \frac{C_n}{n!} + \dots \\
 &= C_0 r(t) + C_1 \dot{r}(t) + \frac{C_2}{2!} \ddot{r}(t) + \frac{C_3}{3!} \ddot{\ddot{r}}(t) + \dots + \frac{C_n}{n!} r^{(n)}(t) \dots
 \end{aligned}$$

The coefficients  $C_0, C_1, C_2, \dots, C_n$  are called the generalized error coefficients or dynamic error coefficients.

The steady state error  $e_{ss}$  is obtained by taking limit  $t \rightarrow \infty$  on  $e(t)$ .

$$\begin{aligned}
 \therefore \text{Steady state error, } e_{ss} &= \lim_{t \rightarrow \infty} \left[ r(t) C_0 + \dot{r}(t) C_1 + \ddot{r}(t) \frac{C_2}{2!} + \ddot{\ddot{r}}(t) \frac{C_3}{3!} + \dots + r^{(n)}(t) \frac{C_n}{n!} + \dots \right] \\
 &= C_0 r(t) + C_1 \dot{r}(t) + \frac{C_2}{2!} \ddot{r}(t) + \frac{C_3}{3!} \ddot{\ddot{r}}(t) + \dots + \frac{C_n}{n!} r^{(n)}(t) \dots \quad \dots(2.66)
 \end{aligned}$$

The generalized error coefficient is given by,

$$C_n = (-1)^n \int_0^t T^n f(T) dT; \quad \text{where } F(s) = \frac{1}{1 + G(s) H(s)}$$

We know that  $\mathcal{L}\{f(T)\} = F(s)$ , hence by the definition of Laplace transform,

$$F(s) = \int_0^t f(T) e^{-sT} dT$$

On taking  $\lim_{s \rightarrow 0}$  on both sides of equation (2.67) we get,

$$\begin{aligned} \lim_{s \rightarrow 0} F(s) &= \lim_{s \rightarrow 0} \int_0^t f(T) e^{-sT} dT \\ &= \int_0^t f(T) \lim_{s \rightarrow 0} e^{-sT} dT = \int_0^t f(T) dT = C_0 \end{aligned}$$

$$\therefore \boxed{C_0 = \lim_{s \rightarrow 0} F(s)}$$

$$F(s) = \int_0^t f(T) e^{-sT} dT$$

On differentiating equation

$$\begin{aligned} \frac{d}{ds} F(s) &= \frac{d}{ds} \int_0^t f(T) e^{-sT} dT \\ &= \int_0^t f(T) \frac{d}{ds} (e^{-sT}) dT = \int_0^t f(T) (-T) e^{-sT} dT \\ &= - \int_0^t T f(T) e^{-sT} dT \end{aligned}$$

On taking  $\lim_{s \rightarrow 0}$  on both sides of equation

$$\lim_{s \rightarrow 0} \frac{d}{ds} F(s) = \lim_{s \rightarrow 0} - \int_0^t T f(T) e^{-sT} dT$$

$$= - \int_0^t T f(T) \lim_{s \rightarrow 0} e^{-sT} dT = - \int_0^t T f(T) dT = C_1$$

$$\therefore \boxed{C_1 = \lim_{s \rightarrow 0} \frac{d}{ds} F(s)}$$

On differentiating equation (2.68) on both sides with respect to  $s$  we get,

$$\frac{d}{ds} \left[ \frac{d}{ds} (F(s)) \right] = \frac{d}{ds} \left[ - \int_0^t T f(T) e^{-sT} dT \right]$$

$$\frac{d^2}{ds^2} F(s) = \left[ - \int_0^t T f(T) \frac{d}{ds} (e^{-sT}) dT \right] = - \int_0^t T f(T) (-T) e^{-sT} dT$$

$$\frac{d^2 (F(s))}{ds^2} = \int_0^t T^2 f(T) e^{-sT} dT$$

Applying the limit  $s \rightarrow 0$  on both sides of the equation (2.71) we get,

$$\begin{aligned}\lim_{s \rightarrow 0} \frac{d^2}{ds^2} F(s) &= \lim_{s \rightarrow 0} \int_0^t T^2 f(T) e^{-sT} dT \\ &= \int_0^t T^2 f(T) \lim_{s \rightarrow 0} e^{-sT} dT = \int_0^t T^2 f(T) dT = C_2\end{aligned}$$

$$\therefore C_2 = \lim_{s \rightarrow 0} \frac{d^2}{ds^2} F(s)$$

Similarly it can be shown that,

$$C_n = \lim_{s \rightarrow 0} \frac{d^n}{ds^n} F(s)$$



# Relation between static and dynamic error coefficient

$$C_0 = \frac{1}{1 + K_p}$$

$$C_1 = \frac{1}{K_v}$$

$$C_2 = \frac{1}{K_a}$$

Proof

$$C_0 = \lim_{s \rightarrow 0} F(s) = \lim_{s \rightarrow 0} \frac{1}{1 + G(s) H(s)} = \frac{1}{1 + \lim_{s \rightarrow 0} G(s) H(s)} = \frac{1}{1 + K_p}$$

---

For a unity feedback control system the open loop transfer function,  $G(s) = \frac{10(s+2)}{s^2(s+1)}$ . Find

a) the position, velocity and acceleration error constants,

b) the steady state error when the input is  $R(s)$ , where  $R(s) = \frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3}$

## To find static error constants

For a unity feedback system,  $H(s)=1$

$$\text{Position error constant, } K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} G(s) = \lim_{s \rightarrow 0} \frac{10(s+2)}{s^2(s+1)} = \infty$$

$$\text{Velocity error constant, } K_v = \lim_{s \rightarrow 0} s G(s)H(s) = \lim_{s \rightarrow 0} s G(s) = \lim_{s \rightarrow 0} s \frac{10(s+2)}{s^2(s+1)} = \infty$$

$$\begin{aligned} \text{Acceleration error constant, } K_a &= \lim_{s \rightarrow 0} s^2 G(s)H(s) = \lim_{s \rightarrow 0} s^2 G(s) \\ &= \lim_{s \rightarrow 0} s^2 \frac{10(s+2)}{s^2(s+1)} = \frac{10 \times 2}{1} = 20 \end{aligned}$$

**b) To find steady state error**

**Method-I**

Steady state error for non-standard input is obtained using generalized error series, given below.

The error signal,  $e(t) = r(t)C_0 + \dot{r}(t)C_1 + \ddot{r}(t)\frac{C_2}{2!} + \dots + \ddot{r}(t)\frac{C_n}{n!} + \dots$

Given that,  $R(s) = \frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3}$

Input signal in time domain,  $r(t) = \mathcal{L}^{-1}\{R(s)\} = \mathcal{L}^{-1}\left\{\frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3}\right\}$

$$= 3 - 2t + \frac{1}{3} \frac{t^2}{2!} = 3 - 2t + \frac{t^2}{6}$$

$$\therefore \dot{r}(t) = \frac{d}{dt}r(t) = -2 + \frac{1}{6}2t = -2 + \frac{t}{3}$$

$$\ddot{r}(t) = \frac{d^2}{dt^2}r(t) = \frac{d}{dt}\dot{r}(t) = \frac{1}{3}$$

$$\dddot{r}(t) = \frac{d^3}{dt^3}r(t) = \frac{d}{dt}\ddot{r}(t) = 0$$

$$C_0 = \lim_{s \rightarrow 0} F(s); \quad C_1 = \lim_{s \rightarrow 0} \frac{d}{ds} F(s); \quad C_2 = \lim_{s \rightarrow 0} \frac{d^2}{ds^2} F(s)$$

$$C_0 = \lim_{s \rightarrow 0} F(s); \quad C_1 = \lim_{s \rightarrow 0} \frac{d}{ds} F(s); \quad C_2 = \lim_{s \rightarrow 0} \frac{d^2}{ds^2} F(s)$$

$$F(s) = \frac{1}{1+G(s)H(s)} = \frac{1}{1+G(s)} = \frac{1}{1 + \frac{10(s+2)}{s^2(s+1)}} = \frac{s^2(s+1)}{s^2(s+1) + 10(s+2)} = \frac{s^3 + s^2}{s^3 + s^2 + 10s + 20}$$

$$C_0 = \lim_{s \rightarrow 0} F(s) = \lim_{s \rightarrow 0} \left[ \frac{s^3 + s^2}{s^3 + s^2 + 10s + 20} \right] = 0$$

$$C_1 = \lim_{s \rightarrow 0} \frac{d}{ds} F(s) = \lim_{s \rightarrow 0} \frac{d}{ds} \left[ \frac{s^3 + s^2}{s^3 + s^2 + 10s + 20} \right]$$

$$= \lim_{s \rightarrow 0} \left[ \frac{(s^3 + s^2 + 10s + 20)(3s^2 + 2s) - (s^3 + s^2)(3s^2 + 2s + 10)}{(s^3 + s^2 + 10s + 20)^2} \right]$$

$$= \lim_{s \rightarrow 0} \left[ \frac{3s^5 + 2s^4 + 3s^4 + 2s^3 + 30s^3 + 20s^2 + 60s^2 + 40s - 3s^5 - 2s^4 - 10s^3 - 3s^4 - 2s^3 - 10s^2}{(s^3 + s^2 + 10s + 20)^2} \right]$$

$$= \lim_{s \rightarrow 0} \left[ \frac{20s^3 + 70s^2 + 40s}{(s^3 + s^2 + 10s + 20)^2} \right] = 0$$

$$C_2 = \lim_{s \rightarrow 0} \frac{d^2}{ds^2} F(s) = \lim_{s \rightarrow 0} \frac{d}{ds} \left[ \frac{d}{ds} F(s) \right] = \lim_{s \rightarrow 0} \frac{d}{ds} \left[ \frac{20s^3 + 70s^2 + 40s}{(s^3 + s^2 + 10s + 20)^2} \right]$$

$$= \lim_{s \rightarrow 0} \left[ \frac{(s^3 + s^2 + 10s + 20)^2 (60s^2 + 140s + 40) - (20s^3 + 70s^2 + 40s) 2 \times (s^3 + s^2 + 10s + 20) (3s^2 + 2s + 10)}{(s^3 + s^2 + 10s + 20)^4} \right] = \frac{20^2 \times 40}{20^4} = \frac{1}{10}$$

$$\text{Error signal, } e(t) = r(t)C_0 + \dot{r}(t)C_1 + \ddot{r}(t)\frac{C_2}{2!} = \left(3 - 2t + \frac{t^2}{6}\right) \times 0 + \left(-2 + \frac{t}{3}\right) \times 0 + \frac{1}{3} \times \frac{1}{10} \times \frac{1}{2!} = \frac{1}{60}$$

$$\text{Steady state error, } e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{t \rightarrow \infty} \frac{1}{60} = \frac{1}{60}$$



## Method - II

The error signal in s-domain,  $E(s) = \frac{R(s)}{1 + G(s)H(s)}$

$$\text{Given that, } R(s) = \frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3} ; \quad G(s) = \frac{10(s+2)}{s^2(s+1)} ; \quad H(s) = 1$$

$$\therefore E(s) = \frac{\frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3}}{1 + \frac{10(s+2)}{s^2(s+1)}} = \frac{\frac{3}{s} - \frac{2}{s^2} + \frac{1}{3s^3}}{\frac{s^2(s+1) + 10(s+2)}{s^2(s+1)}}$$

$$= \frac{3}{s} \left[ \frac{s^2(s+1)}{s^2(s+1) + 10(s+2)} \right] - \frac{2}{s^2} \left[ \frac{s^2(s+1)}{s^2(s+1) + 10(s+2)} \right] + \frac{1}{3s^3} \left[ \frac{s^2(s+1)}{s^2(s+1) + 10(s+2)} \right]$$

The steady state error  $e_{ss}$  can be obtained from final value theorem.

$$\text{Steady state error, } e_{ss} = \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s)$$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} s \left\{ \frac{3}{s} \left[ \frac{s^2(s+1)}{s^2(s+1)+10(s+2)} \right] - \frac{2}{s^2} \left[ \frac{s^2(s+1)}{s^2(s+1)+10(s+2)} \right] + \frac{1}{3s^3} \left[ \frac{s^2(s+1)}{s^2(s+1)+10(s+2)} \right] \right\}$$

$$= \lim_{s \rightarrow 0} \left\{ \frac{3s^2(s+1)}{s^2(s+1)+10(s+2)} - \frac{2s(s+1)}{s^2(s+1)+10(s+2)} + \frac{(s+1)}{3s^2(s+1)+30(s+2)} \right\} = 0 - 0 + \frac{1}{60}$$

$$= \frac{1}{60}$$

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For servomechanisms with open loop transfer function given below explain what type of input signal give rise to a constant steady state error and calculate their values.

$$\text{a) } G(s) = \frac{20(s+2)}{s(s+1)(s+3)}; \quad \text{b) } G(s) = \frac{10}{(s+2)(s+3)}; \quad \text{c) } G(s) = \frac{10}{s^2(s+1)(s+2)}$$

$$a) \quad G(s) = \frac{20(s+2)}{s(s+1)(s+3)}$$

Let us assume unity feedback system,  $\therefore H(s)=1$

The open loop system has a pole at origin. Hence it is a type-1 system. In systems with type number-1, the velocity (ramp) input will give a constant steady state error.

The steady state error with unit velocity input,  $e_{ss} = \frac{1}{K_v}$

Velocity error constant,  $K_v = \lim_{s \rightarrow 0} s G(s) H(s) = \lim_{s \rightarrow 0} s G(s)$

$$= \lim_{s \rightarrow 0} s \frac{20(s+2)}{s(s+1)(s+3)} = \frac{20 \times 2}{1 \times 3} = \frac{40}{3}$$

Steady state error,  $e_{ss} = \frac{1}{K_v} = \frac{3}{40} = 0.075$

$$b) G(s) = \frac{10}{(s+2)(s+3)}$$

Let us assume unity feedback system,  $\therefore H(s)=1$ .

The open loop system has no pole at origin. Hence it is a type-0 system. In systems with type number-0, the step input will give a constant steady state error.

The steady state error with unit step input,  $e_{ss} = \frac{1}{1+K_p}$

$$\text{Position error constant, } K_p = \lim_{s \rightarrow 0} G(s)H(s) = \lim_{s \rightarrow 0} G(s) = \lim_{s \rightarrow 0} \frac{10}{(s+2)(s+3)} = \frac{10}{2 \times 3} = \frac{5}{3}$$

$$\text{Steady state error, } e_{ss} = \frac{1}{1+K_p} = \frac{1}{1+\frac{5}{3}} = \frac{3}{3+5} = \frac{3}{8} = 0.375$$

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c)  $G(s) = \frac{10}{s^2(s+1)(s+2)}$

Let us assume unity feedback system,  $\therefore H(s)=1$ .

The open loop system has two poles at origin. Hence it is a type-2 system. In systems with type number-2, the acceleration (parabolic) input will give a constant steady state error.

The steady state error with unit acceleration input,  $e_{ss} = \frac{1}{K_a}$

$$\text{Acceleration error constant, } K_a = \lim_{s \rightarrow 0} s^2 G(s)H(s) = \lim_{s \rightarrow 0} s^2 G(s) = \lim_{s \rightarrow 0} s^2 \frac{10}{s^2(s+1)(s+2)} = \frac{10}{1 \times 2} = 5$$

$$\text{Steady state error, } e_{ss} = \frac{1}{K_a} = \frac{1}{5} = 0.2$$

A unity feedback system has the forward transfer function  $G(s) = \frac{K_1(2s+1)}{s(5s+1)(1+s)^2}$ . When the input  $r(t) = 1+6t$ ,

determine the minimum value of  $K_1$  so that the steady error is less than 0.1.

Given that, input  $r(t) = 1 + 6t$

On taking laplace transform of  $r(t)$  we get  $R(s)$ .

$$\therefore R(s) = \mathcal{L}\{r(t)\} = \mathcal{L}\{1 + 6t\} = \frac{1}{s} + \frac{6}{s^2}$$

The error signal in s-domain  $E(s)$  is given by,

$$\begin{aligned}\therefore E(s) &= \frac{R(s)}{1 + G(s)H(s)} = \frac{\frac{1}{s} + \frac{6}{s^2}}{1 + \frac{K_1(2s+1)}{s(5s+1)(1+s)^2}} = \frac{\frac{1}{s} + \frac{6}{s^2}}{\frac{s(5s+1)(1+s)^2 + K_1(2s+1)}{s(5s+1)(1+s)^2}} \\ &= \frac{1}{s} \left[ \frac{s(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} \right] + \frac{6}{s^2} \left[ \frac{s(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} \right]\end{aligned}$$



The steady state error  $e_{ss}$  can be obtained from final value theorem.

$$\begin{aligned} e_{ss} &= \lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s) \\ &= \lim_{s \rightarrow 0} s \left\{ \frac{1}{s} \left[ \frac{s(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} \right] + \frac{6}{s^2} \left[ \frac{s(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} \right] \right\} \\ &= \lim_{s \rightarrow 0} \left\{ \frac{s(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} + \frac{6(5s+1)(1+s)^2}{s(5s+1)(1+s)^2 + K_1(2s+1)} \right\} = 0 + \frac{6}{K_1} = \frac{6}{K_1} \end{aligned}$$

Given that,  $e_{ss} < 0.1$        $\therefore 0.1 = \frac{6}{K_1}$       or       $K_1 = \frac{6}{0.1} = 60$