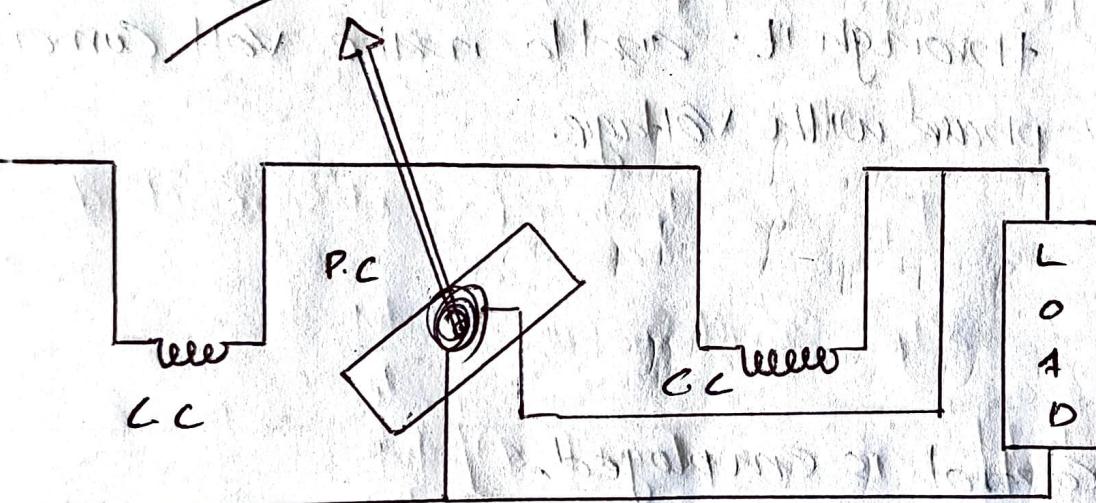


Electrodynamometer Wattmeter

An electrodynamometer-type wattmeter is similar in design and construction with the analog electrodynamometer-type ammeter and voltmeter.

Construction:

Scale:



fixed Coil:

- * It has 2 coils connected in different ways to the same set of which power is to be measured.
- * Fixed coils are connected in series.
- * pressure coils are connected in parallel with the Supply.
- * The magnetic field is produced by these fixed coils.
- * These coils are divided into 2 sections so as to provide uniform magnetic field.

Moving Coil

- * Connected across the load. Carries current proportional to load current.
- * Also called as pressure coils.
- * Mounted on a pivoted spindle.
- * A high value non-inductive fixed ~~so~~ resistance is connected in series so as to limit the current through it and to make current as in phase with voltage.

Control

- * Spring control is employed.

Damping

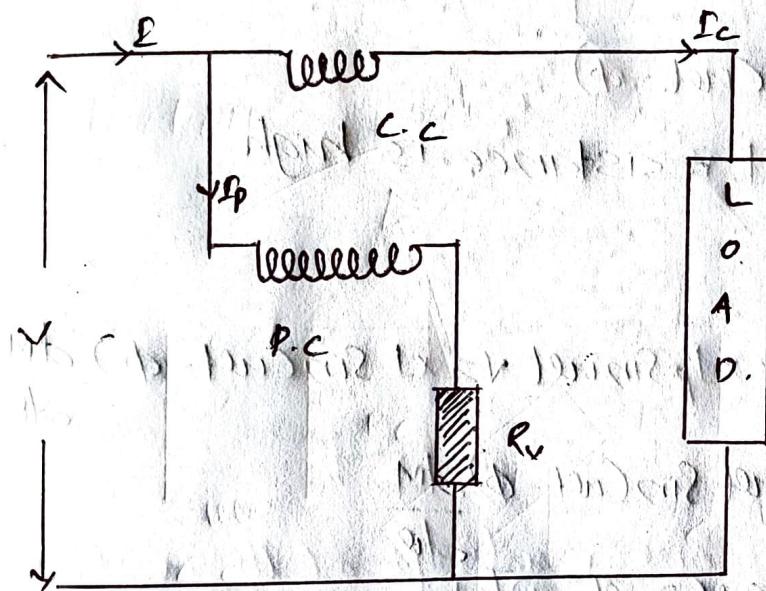
Air friction damping is used.

Shielding System

- * The operating field produced by the fixed coils are comparatively lower in electrodynamometer type instruments as compared to other instruments.
- * So it is essential to shield the instrument from effects of stray magnetic field.

So enclosure is made of alloys with high permeability to restrict penetration of external stray magnetic field into the instrument.

Working:



V → Voltage.

I → Current

V.C → Current Coil

P.G → Pressure Coil.

V_p → Voltage coil & first

I_C → Current coil first

R_s → External resistance with pressure coil.

R_p → Pressure coil resistance.

M → Mutual inductance b/w current coil and pressure coil.

θ → angle of deflection of the moving system

w → Angular frequency ϕ → phase angle, lag of current w.r.t voltage.

The instantaneous torque of electrodynamometer

w/m is,

$$T_i = V_p i C \frac{d\phi}{d\theta}$$

Instantaneous value of voltage across the pressure coil circuit is

$$V_p = \sqrt{2} V_{\text{sin} \omega t}$$

If the pressure coil resistance is assumed to be high.

The current up in the pressure coil thus, can be assumed to be in phase with the voltage V_p , and instantaneous value

$$i_p = V_p / R_p = \sqrt{2} V_p / R_p \sin \text{net} = \sqrt{2} I_p \sin \text{net}$$

where $I_p = V_p / R_p$ is the rms value of current in pressure coil.

$$i_c = \sqrt{2} \times I \sin (\text{net} - \phi)$$

Since pressure coil resistance is high

Instantaneous torque.

$$\tau_i = \sqrt{2} \times I_p \sin \text{net} \sqrt{2} \times I \sin (\text{net} - \phi) \frac{dM}{d\theta}$$

$$= 2 I_p L \sin \text{net} \sin (\text{net} - \phi) \frac{dM}{d\theta}$$

$$= I_p L [\cos \phi - \cos (2 \text{net} - \phi)] \frac{dM}{d\theta}$$

$$T_d = \frac{1}{4} T \int_0^T \tau_i d\theta = \frac{1}{2} T \int_0^T I_p L (\cos \phi - \cos (2 \text{net} - \phi)) \frac{dM}{d\theta} d\theta$$

$$\frac{I_p L}{2} [\text{net} \cos \phi] \frac{dM}{d\theta}, \quad I_p L \cos \phi \frac{dM}{d\theta} = \frac{V}{R_p} I \cos \phi \frac{dM}{d\theta}$$

Thus,

$$T_d = \frac{V I \cos \phi}{R_p} \frac{dM}{d\theta}$$

With a spring constant k , the controlling torque is provided by the spring for a final steady-state deflection of θ is given by

$$T_c = k \theta$$

* Under Steady State Condition, the avg. deflecting torque will be balanced by the controlling torque provided by the spring.

* Thus, at balanced condition $T_c = T_d$.

* Steady state deflection θ is thus found to be an indication of the power P to be measured.

$$PD = \frac{VI \cos\phi}{R_p} \frac{dm}{d\theta}$$

$$\theta = \frac{VI \cos\phi}{k_p} \frac{dm}{d\theta} \quad \theta = k_1 \left[\frac{dm}{d\theta} \right] P.$$

Blondel's Theorem

If a network is supplied through n conductors, the total power is measured by summing the reading of n w/m.

By taking one line as common, we can measure the power by using $(n-1)$ w/m.

3φ Power Measurement

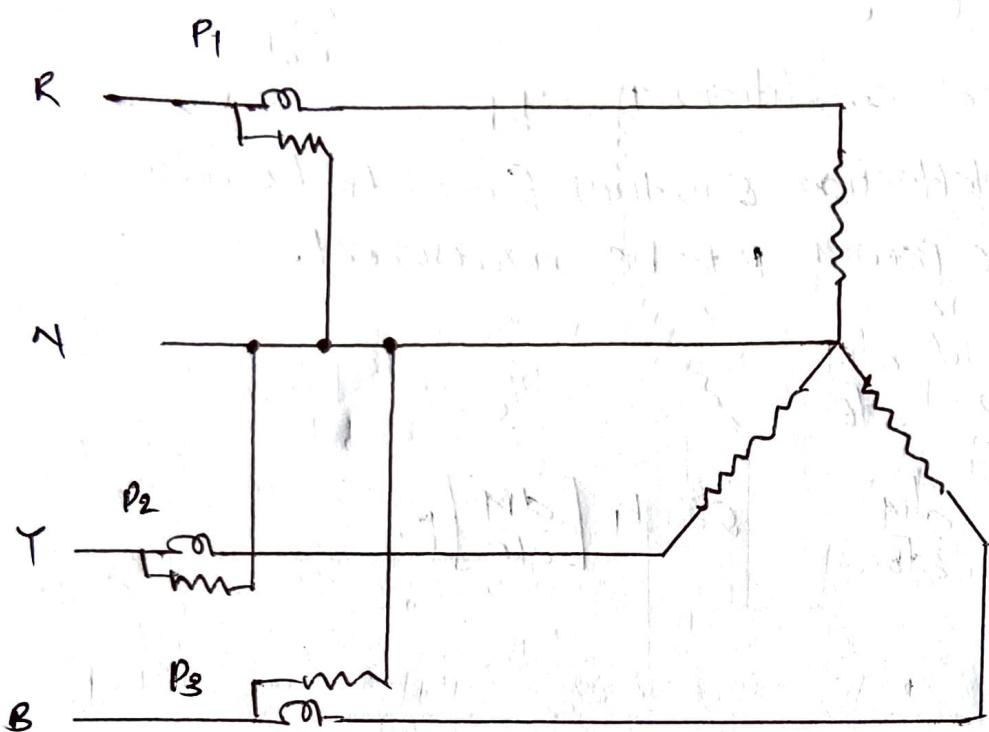
Can be done by 3 methods.

a) Single w/m method.

b) 2 w/m method

c) 3 w/m method.

3w/m Method



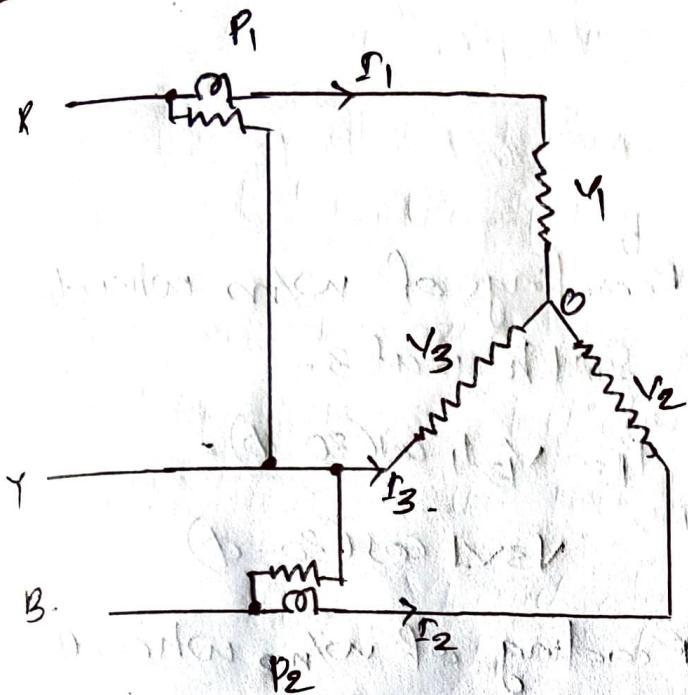
Sum of the instantaneous readings of the wattmeter

$$P = P_1 + P_2 + P_3 = V_1 I_1 + V_2 I_2 + V_3 I_3 \Rightarrow \text{Instantaneous Load.}$$

- * Applied 3ϕ 4 wire system.
- * ~~Res~~ Current coils are connected to respective phases.
- * pressure coils are connected to a common line

2w/m Method

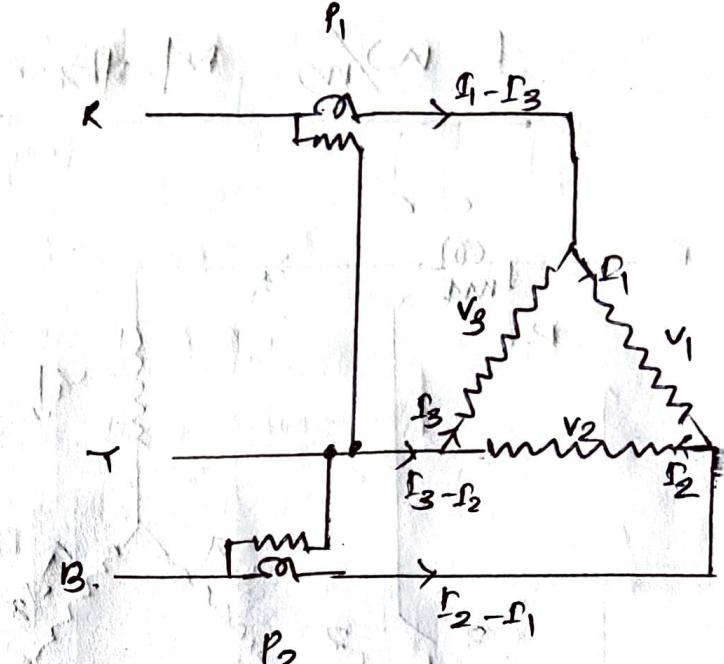
- * Applied to 3ϕ 3 wire system.
- * Can be used in both balanced and unbalanced system.



Star (Y) Connection

Sum of instantaneous readings of 2 w/m.

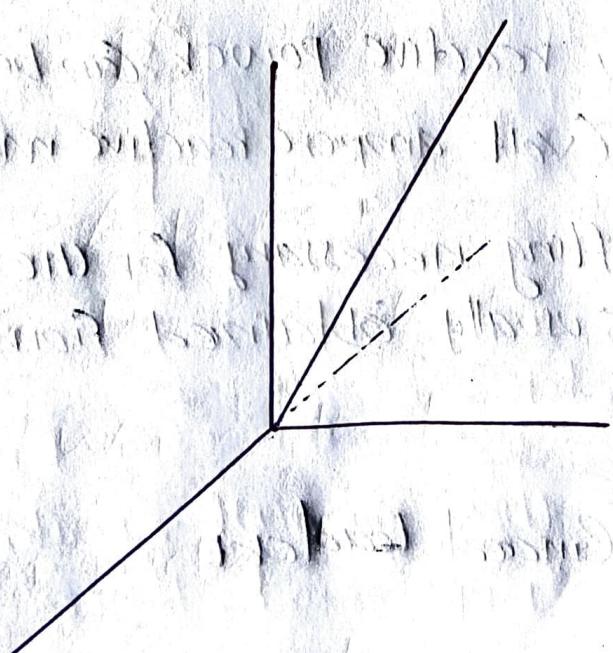
$$\begin{aligned}
 P_1 + P_2 &= V_1 i_1 (V_1 - V_3) + V_2 i_2 (V_2 - V_3) \\
 &= V_1 V_1 + V_2 V_2 - V_3 (V_1 + V_2) \\
 &= V_1 i_1 + V_2 i_2 + V_3 i_3
 \end{aligned}$$



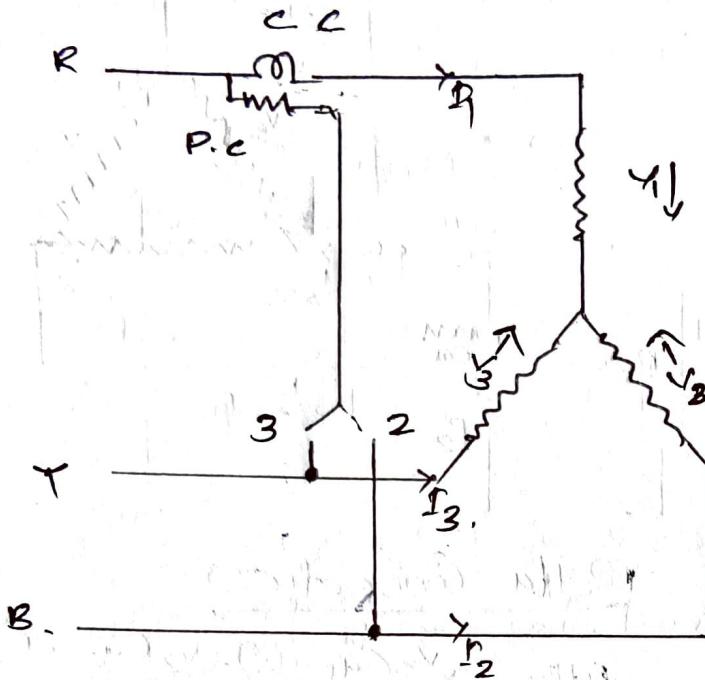
Delta Connection

$$\begin{aligned}
 P_1 + P_2 &= \frac{1}{2} V_3 (i_1 - i_3) + V_2 (i_2 - i_1) \\
 &= V_2 i_2 + V_3 i_3 - V_1 (i_1 + i_3) \\
 &= V_1 i_1 + V_2 i_2 + V_3 i_3
 \end{aligned}$$

The sum of the 2 w/m reading is equal to the power consumed by the load. This is irrespective of whether the load is balanced or unbalanced.



1 w/m Method



Readings of w/m when the switch is at 3.

$$P_1 = V_B I_1 \cos(30 - \phi) =$$

$$\sqrt{3} V I \cos(30 - \phi)$$

Reading of w/m when the switch is at 2

$$P_2 = V_1 I_1 \cos(30 + \phi) = \sqrt{3} V I \cos(30 + \phi)$$

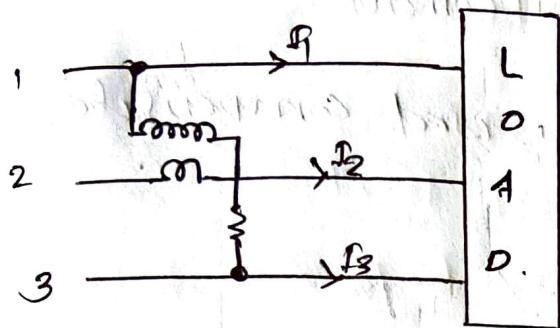
$$\text{Total power} = P_1 + P_2 = \sqrt{3} V I \cos(30 - \phi) + \sqrt{3} V I \cos(30 + \phi).$$

$$= \sqrt{3} V I [\cos(30 - \phi) + \cos(30 + \phi)]$$

$$= \underline{\underline{3 V I \cos \phi}}$$

Measurement of Reactive Power

- * In a Single phase ckt, reactive power can be measured by a Varimeter. (volt-ampere reactive meter).
- * In 3φ ckt's phase shifting necessary for the measurement of reactive power is usually obtained from phase shifting transformers.
- * In case of 3φ balanced load ckt



Reading of $w_{lm} = \sqrt{3} I_2 \cos(\phi_0 + \phi)$.

$$= \sqrt{3} V I \sin \phi,$$

Total reactive Volt-ampere of the cbt

$$= 3 V I \sin \phi$$

$$\phi = \tan^{-1} (\phi_p).$$

Lpf w/mag (electrodynamometer type).

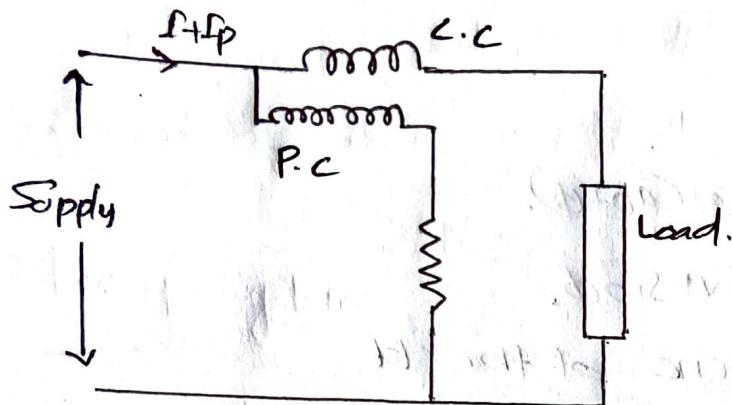
- * Measurement of power with low p.f is difficult with ordinary w_{lm} due to
 - a) T_d is small even at full excitation.
 - b) Errors due to resistance of P.C coil is large at low p.f
- * Special features are added to make a lpf w_{lm} from ordinary w_{lm} .

Pressure Coil Current

Pressure coil cbt is designed to have low value of resistance so current through it is increased and thereby the torque (^c upto 10 times more current).

Compensation for pressure coil Current

" L.P.F w/m. Compensation coil is used compensate error caused by power loss.



Compensation for Inductance of pressure coil

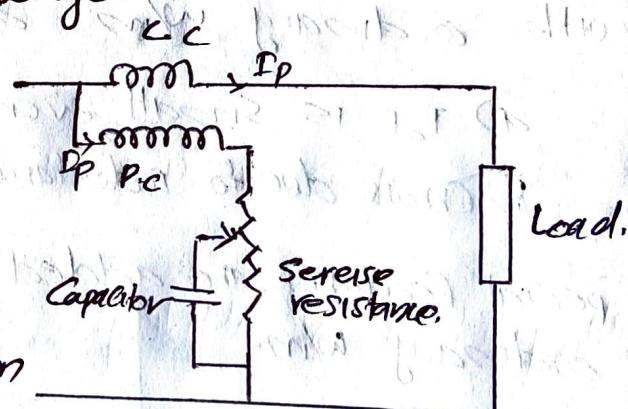
The error caused by pressure coil inductance

→ $\propto \sin \phi \tan \phi$

With low p.f., the value of ϕ is large and therefore the error, is correspondingly large.

Small Control Torque

low p.f. w/m are designed.
with to have a small control torque
so that they give full scale deflection
For p.f as low as 0.1



Electrical Energy meter

A device that measure the amount of electrical energy consumed by electrically powered device.

Unit \rightarrow kWh.

Types of Energy Meter

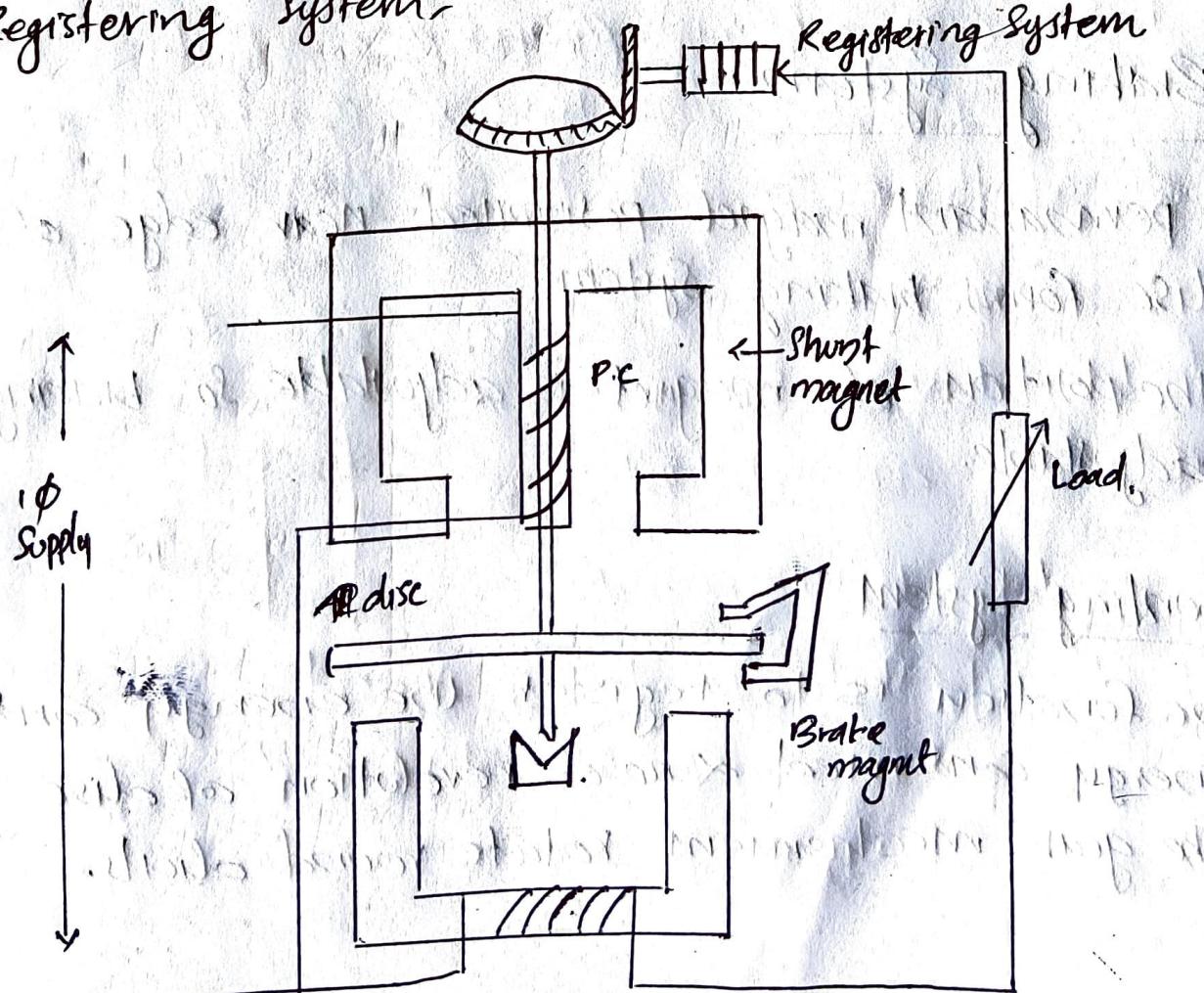
- * Electromechanical induction type meter
- * Electronic meter.

Electro Mechanical Energy Meter

- Measure energy by counting revolution of a rotating disc.
- The disc is non-magnetic but electrically conductive.
- No. of revolutions \propto proportional energy usage.

Construction:

- * Driving System
- * Moving System
- * Braking System
- * Registering System



Driving System

- * consists of 2 electromagnets.
- * The core laminated with Si steel.
- * one current coil and other voltage coil
- * Electromagnets are called Series and Shunt magnets.

Moving System

- * This consists of an Al disc mounted on a light alloy shaft
- * Disc is mounted b/w Series and Shunt magnets.
- * The pivot is supported by jewel bearings.
- * A gear mechanism is connected to the shaft.

Braking System

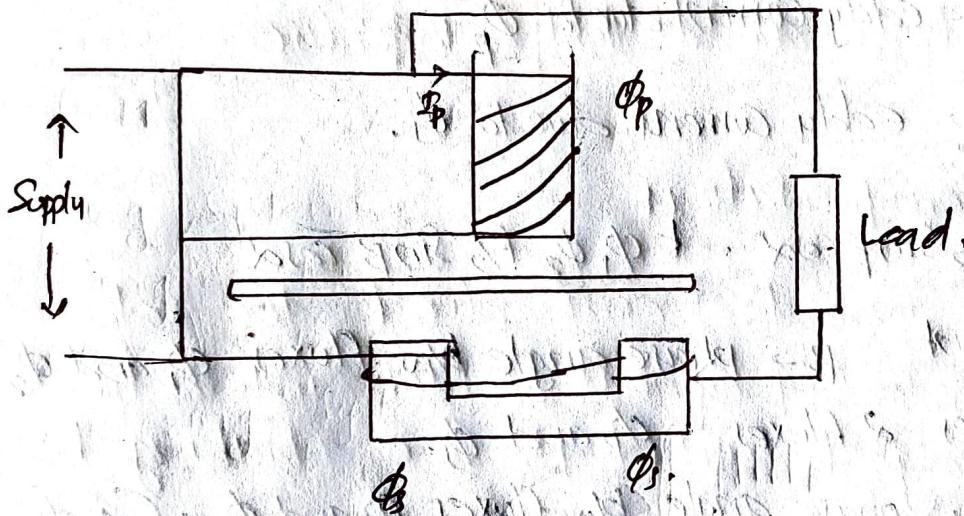
- * A permanent magnet positioned near edge of the Al disc forms braking system.
- * The position of magnet is adjustable so braking also adjustable.

Counting System

- * The function is to register the energy consumed.
- * Energy consumed & no: of revolution of disc
- * The gear mechanism rotate round dial.

Single phase Induction Type Energy Meter

- * Supply voltage is applied across the pressure coil.
- * Pressure coil winding is highly inductive, reluctance to the magnetic act is small.
- * Current through pressure coil $I_p \propto$ Supply Voltage and lags few degree less than 90° .
- * The lag is due to resistance and iron loss.



- * The flux Φ_p goes through the aluminum disc and hence is responsible for production of driving torque.
- * $\Phi_p \propto I$ and is in phase with I .
- * \therefore Flux $\Phi_p \propto$ Voltage and lags by an angle few degrees less than 90° since flux Φ_p is alternating in nature.
- * Load current I flows through current coil and produces counter flux Φ_b .
- * $\Phi_b \propto I_L$ and in phase with I_L .
- * Eddy current I_{es} interacts with both Φ_p and Φ_b for producing different torques.

* These torques are opposite in direction. So the resultant is the difference of these

let V = applied voltage $I_p \rightarrow$ pressure coil current

I = load current. Δ = phase angle b/w V and ϕ_p

ϕ_L = phase angle load Z = impedance of ϕ_L path

E_{ep} = eddy current by ϕ_p E_s = eddy current by ϕ_s

F = frequency α = phase angle of eddy current paths

I_{ep} = eddy current due to ϕ_p

I_{es} = eddy current due to ϕ_s .

$$T_d \propto \phi_1 \phi_2 F/2 \sin \beta \cos \alpha = k \phi_1 \phi_2 F/2 \sin \beta \cos \alpha$$

$k_1 \rightarrow$ constant $\beta \rightarrow$ phase angle b/w fluxes ϕ_1 and ϕ_2

In our cases 2 fluxes ϕ_p and ϕ_s

$\beta =$ phase angle b/w fluxes ϕ_p and $\phi_s = (\alpha - \phi)$

driving torque $T_d = k_1 \phi_p \phi_s F/2 \sin(\alpha - \phi) \cos \alpha$

$\phi_p \propto V$ and $\phi_s \propto I$

$$T_d = k_2 V I F/2 \sin(\alpha - \phi) \cos \alpha$$

If F, I, α are constant.

$$T_d = k_3 V D \sin(\alpha - \phi)$$

At Steady State and Steady Speed.

$$k_v r \sin(\alpha - \phi) = k_v r \cos \phi$$

$\Rightarrow \underline{k \times \text{power}}$

Since, Speed of rotation & power.

Δ should be $= 90^\circ$. Hence ϕ_p must be made lag the Supply Voltage by exactly 90° .

~~No. of revolution = $\int N dt = k \int V R \sin(\alpha - \phi) dt$~~

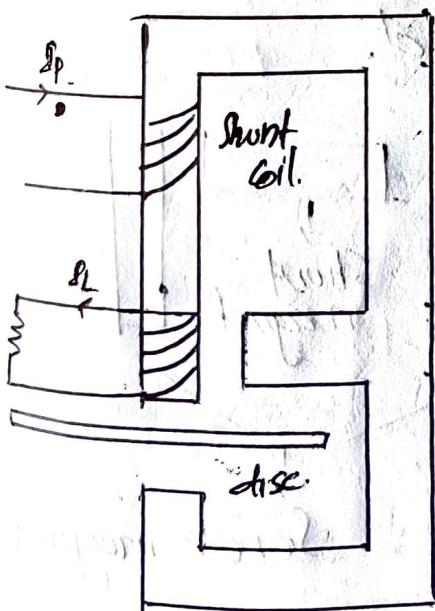
~~= $\int V D \cos \phi dt$~~

~~$\therefore k \int (\text{Power}) dt = \underline{\text{fx energy}}$~~

Lag Adjustment Device

- * The meter will read only if angle α is made equal to 90° .
- * Thus the angle α in Shunt magnet flux ϕ_p and Supply voltage V should be equal to 90° .
- * So the pressure coil should be highly inductive, and has low resistance, iron and core losses are small.

* By introducing a magnetic shunt ckt, it is possible to introduce an reactance.



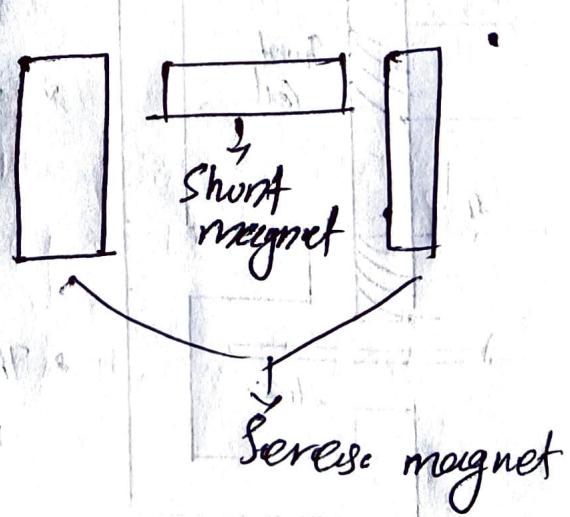
- * The flux ϕ_p in the disc air gap will be created by the combined action of the main mmf.
- * Thus flux ϕ_p will be in phase with resultant main mmf A_{TP} .

Creeping

- * In some meters a slow but continuous rotation is obtained even when there is no current in pressure coil. This is called creeping.
- * The major cause for creeping is over-compensation for friction.
- * In order to prevent creeping, a diametrically opposite holes are drilled in the disc.

Overload Compensation

- * Dynamically induced emf due to the rotation of the disc causes eddy current which interacts with Sereis magnet field to produce a braking torque.
- * Overload Compensating device takes the form of a magnetic shunt for the Sereis magnet core whose permeability decreases at overloaded.



Voltage Compensation

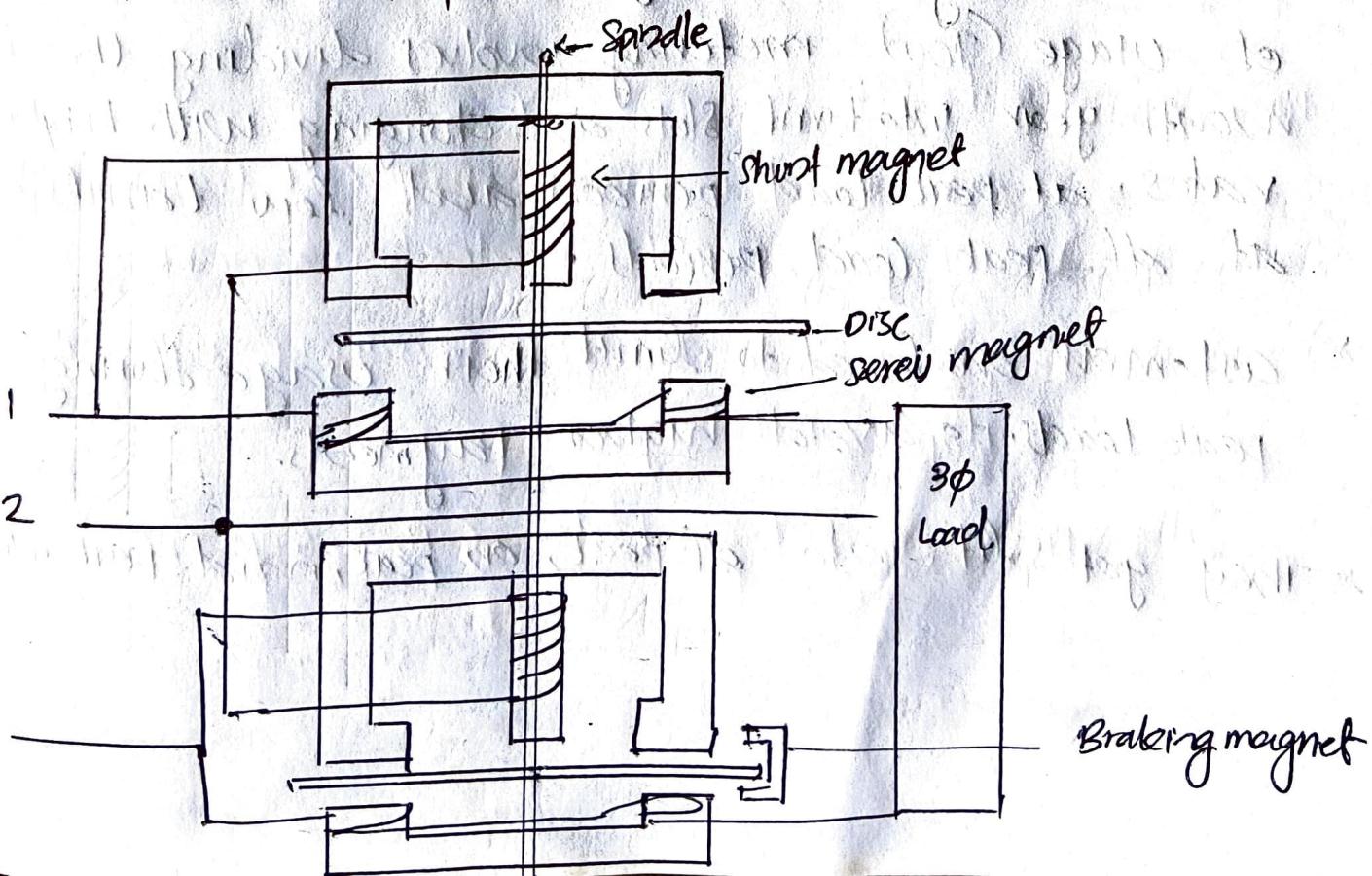
The compensation for voltage variations can be conveniently provided by the increasing the μ_i reluctance of the side limbs of the shunt magnet which is done by providing holes in the side limbs.

Polyphase Energy Meter

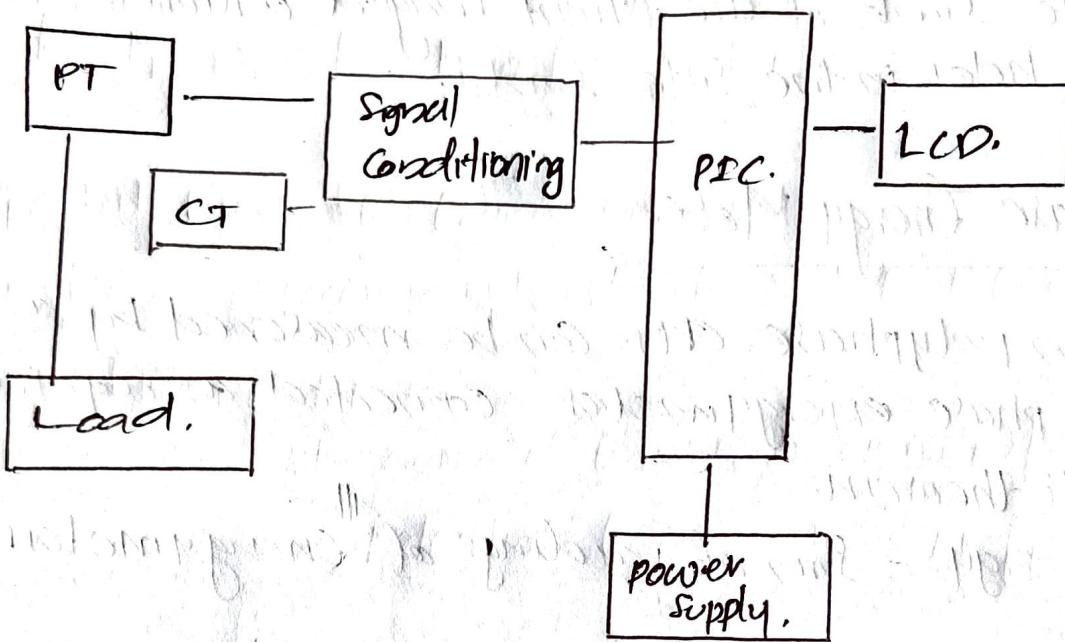
- Energy in polyphase ckt's. can be measured by a group of single phase energy meter connected as required by Blondel's theorem.
- Total energy = sum of readings of ^{all} energy meters

2 Element Energy Meter

2 element E/m used for 3 ϕ 3 wire system.



Digital Energy Meter

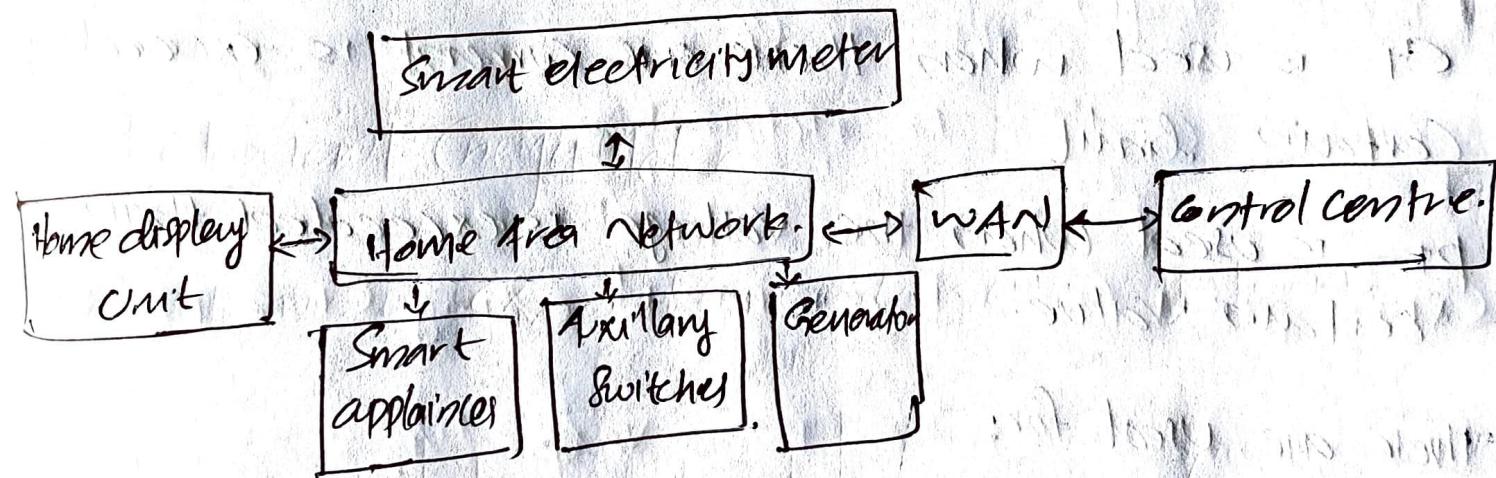


TOD Meter

- * Time of day metering (TOD), also known as time of usage (TOU) metering involves dividing the day, month, year into tariff slots and charging with higher rates at peak load periods and low tariff's rate at off-peak load periods.
- * customers are forced to limit their usage during peak loads, to avoid higher payments.
- * They got split into off-peak, on-peak, mid-peak etc.

Smart Metering

- A smart metering system is built with smart meters, control devices and communication link.
- The key element is combination of metering and intelligence.
- They are introduced for better energy saving, demand management and energy efficiency.
- It will read, max demand, P.F, voltage, current, energy consumption etc.
- It can enable 2 way communication.
- The communication may be wired or wireless.



- Consumer can use the collected data from smart meter to adjust their load and save electricity bill.
- The utilities can use this data to monitor the electricity usage of each customer.
- In home display unit is a touch screen device that communicates with smart meter, that shows a summary of energy usage over the day, week, month.

- * WAN is used for communication b/w premises and control centre

Instrument Transformer

- * Special type of transformers.
- * Used for measurement of V, I, P, P.F etc.
- * 2 types:
 1. Current transformer (CT) [series t/f].
 2. Potential transformer (PT) [parallel t/f].

CT is used when there is current exceeding certain limit

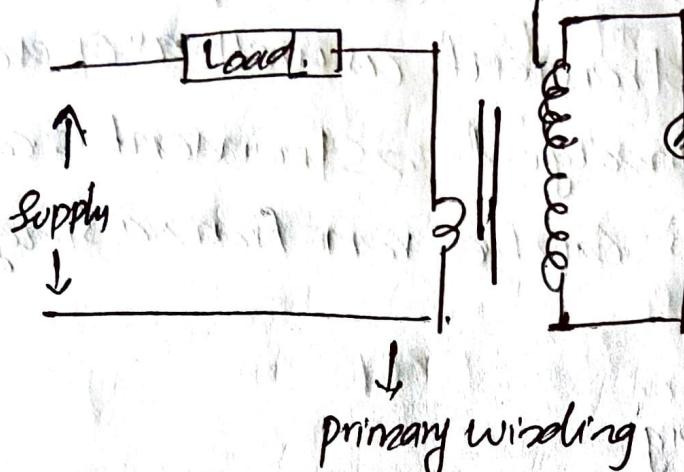
PT is used when the voltage is exceeded than certain value.

These are used for:

- * To isolate measuring ckt from high voltage so as to protect measuring ckt.
- * To make possible of measuring high values of voltage and current.
- * Also used to activate C.R, relays etc.

Measurement By CT

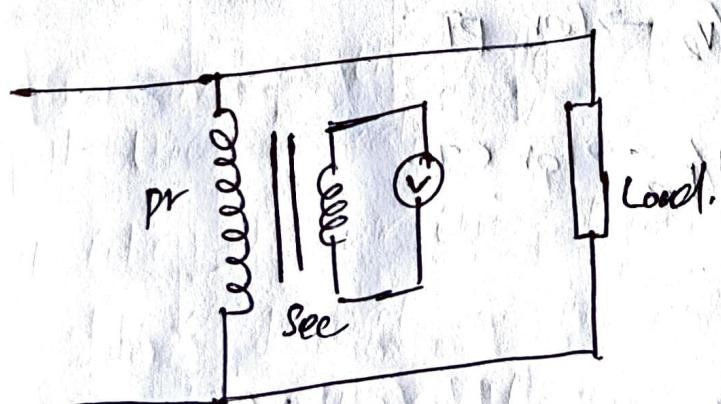
Secondary Winding.



- The primary winding is connected to CT where the current is needed to be measured.
- Secondary winding is connected to ammeter.
- The function of CT is to reduce the current.

Measurement By PT

Primary winding is connected to the voltage side to be measured and Secondary to the Voltmeter.



Advantages of Instrument Transformers

- * the measuring instruments can be placed anywhere for the safety of instrument as well as operator.
- * Replacement of damaged instrument is easy.
- * Several instruments can be run from a single instrument t/c.
- * Moderate size, not big.
- * Measuring ct is isolated from power ct
- * power loss in instrument t/c is very small.

Disadvantages

- * Cannot be used for DC networks.

Expression for Measurement Using CT & PT

$$\text{Transformation ratio (R)} = \frac{V_2}{V_1} = \frac{I_1}{I_2}$$

$$\rightarrow \frac{V_2}{V_1} = \frac{I_1}{I_2} \rightarrow \text{for PT}$$

$$\frac{I_1}{I_2} \rightarrow \text{for CT}$$

Nominal ratio (k_n)

$$\frac{I_1 \text{ rated}}{I_2 \text{ rated}} \rightarrow \text{for CT} \quad \frac{V_1 \text{ rated}}{V_2 \text{ rated}} \rightarrow \text{for PT}$$

Turns ratio (n)

$$= N_2/N_1 \rightarrow \text{for CT}$$

$$N_1/N_2 \rightarrow \text{for PT}$$

Ratio Correction factor :

$$\frac{\text{Transformation ratio}}{\text{normalization}} = R/A_{kn} \rightarrow \text{for CT & PT}$$

Total Sec: burden =

$$\frac{(\text{Sec: induced voltage})^2}{\text{impedance of Sec: coil} \times \text{induced impedance in sec: wdy}}$$

Let

$n \rightarrow$ turns ratio.

$R_s \rightarrow$ resistance to Sec: winding.

$X_s \rightarrow$ reactance to " "

$R_p \rightarrow$ resistance to primary winding.

$X_p \rightarrow$ reactance to "

$E_p \rightarrow$ primary induced voltage

$E_s \rightarrow$ secondary "

$N_p \rightarrow$ no. of primary windings.

$N_s \rightarrow$ no. of secondary "

$V_s \rightarrow$ voltage of "

$I_s \rightarrow$ Secondary winding current

$I_p \rightarrow$ primary " "

$\phi \rightarrow$ working flux of tf

$\delta \rightarrow$ angle b/w E_s and I_s .

$$= \tan^{-1} \left(\frac{r_s + x_e}{r_s + R_e} \right).$$

$\Delta \rightarrow$ phase angle of sec load ckt

$I_o \rightarrow$ exciting current

$I_e \rightarrow$ loss component of exciting current.

$\alpha \rightarrow$ angle b/w I_o and ϕ

$\theta \rightarrow$ phase angle of tf

Transformation Ratio (R) = $\frac{I_p}{I_s}$.

$$= \left[n^2 I_s^2 + 2n I_s I_o \sin(\delta + \alpha) + I_o^2 \right]^{1/2}$$

$$\cdot \frac{I_s}{I_s}$$

$$I_o \ll n I_s \text{ and } I_p \approx n I_s.$$

$$R = \left[n^2 I_s^2 + 2n I_s I_o \sin(\delta + \alpha) + I_o^2 \sin^2(\delta + \alpha) \right]^{1/2} \cdot \frac{I_s}{I_s}$$

$$R \approx n + I_o / I_s \sin(\delta + \alpha)$$

$$\text{expanding, } R = n + \frac{I_m \sin \delta + I_e \cos \delta}{I_s},$$

$$\text{where } I_m = I_o \cos \delta \text{ and } I_e = I_o \sin \delta.$$

As θ is small and I_0 is very small compared to I_s .

phase angle, $\theta = \frac{\pi I_0 \cos(\delta + \alpha)}{n_{IS}} \text{ rad.}$

$\theta = \frac{180}{\pi} \left[\frac{\text{Im cosd - Selind}}{n_{IS}} \right] \text{ degree}$

Errors in Instrument Transformer

a) Ratio Error:

$$\% \text{ ratio error} = \frac{\text{Nominal ratio} - \text{Actual ratio}}{\text{Actual ratio}} \times 100.$$

for normal operation of these instrument tip, the current transformation ratio should be constant within the limits.

b) phase Angle Error:

$$\text{phase angle error, } \theta = \frac{180}{\pi} \left[\frac{\text{Im cosd - Selind}}{n_{IS}} \right] \text{ degree.}$$

phase angle error is due to the no load current or exciting current. thus is the angle by which the secondary current, when reversed, differs in phase from primary current.

Methods to minimise Error (ct).

Ratio error mainly depends upon the working component of current and phase angle error. depends upon magnetising component of the current.

1. In order to minimise these errors (I_w and ϕ_m) needed to be minimised. This could be only done by materials of having high permeability, short magnetic path and large cross-section areas. we can use.

Hot rolled Si

CR GO Steel Si Steel.

Ni iron alloys.

The construction of core has min. no. of joints, for that we just made cores as the shape of either

a) Ring

b) Spiral

- 2) By providing suitable turn ratio of the secondary can be reduced by one or two turns.

3. Leakage reactance should also increases the ratio error. Therefore a windings should be placed as close as can.

4. If the current on the secondary is too large, it should be reduced by putting a shunt either of side. It also reduces phase angle error.

Application of CT

- * CT's are used in panel boards of subst power stations and sub stations to measure very high current.
- * CT are widely used in power measuring ckt.
- * Also used as an actuator in C.B, relays etc.

Application of PT

- * used to measure high voltage by means of low range $\sqrt{3} \text{ m.v.}$.
- * used in over voltage protection to activate the ckt.

PT

a) Construction:

- * Basically PT is a 2 winding t/f. primary is connected to L.v and Secondary is connected to H.v.
- * Core is shell type.
- * Insulation is provided b/w L.v and H.v winding.
- * Dry type P.T's are used upto 6.6kv and for much we will use immersed type.

b) Features:

- * Output of PT is very small and size is large, no problem of temperature
- * Size of core is larger compared to power t/f
- * material of core should be highly permable to reduce iron loss and ratio error, phase angle error.
- * primary and Secondary are co-axial to reduce leakage reactance.

Difference B/w CT and PT

- * CT is also called Series t/f, Secondary of CT is virtually short circuted when primary is energised.
- * PT is also called parallel t/f, and Secondary is open

- Under normal condition, N_1 of PT is constant. Flux density and the exciting current of PT varies in small range where I_p and excitation of primary varies by a wide range
- Current in CT is independent of load condition while on PT is dependent on its secondary load burden
- Primary of PT is connected to full line voltage, while of CT there is a small voltage

Methods to Minimise Error in PT

- In order to minimise the error, no load component I_m and I_w must kept very low.
 - By reducing winding resistance and leakage reactance (X_L).
 - By providing suitable turn ratio
- Q A CT has a single turn on primary and 200 turns on secondary winding. The secondary winding of 5A is passing through a secondary burden of 1Ω resistance. The required flux is set up in the core by emf of 80V. The frequency is 50Hz and net cross-section area. Calculate the ratio and phase angle And also the flux density.

Ane.

$$N_1 = 1 \quad N_2 = 200$$

$$Z_2 = 12$$

$$\text{turn ratio} = N_2/N_1 = 200/1 = 200$$

Voltage induced in Secondary, $E_2 = I_2 Z_2 = 5 \times 22 = 5V$

$$\text{Also, } I_1 = k \cdot I_2$$

The working component of no load current is neglected, $I_{w1} = 0$.

Now, magnetizing component of no load current is neglected. $I_{w2} = 0$.

Now magnetizing component of no load current

$$= \text{magnet primary turns. } I_{m1} = 80/1 = 80A$$

Secondary Winding Current, $I_2 = 5A$

Secondary reverse current, $I_1 = k \cdot I_2 = 200 \times 5 = 1000A$

Now primary Current

$$I_1 = [(I_m^2) + (I_1^2)]^{1/2}$$

$$= [80^2 + 1000^2]^{1/2}$$

$$= \underline{\underline{1003.2A}}$$

Actual transformation ratio = $1003.2/5 = 200.64$

phase angle $\phi = \tan^{-1}(I_m/I_1) = \tan^{-1}(80/1000) = 40.34^\circ$

From emf eqn:

$$e = 4.44 F \phi_{max} N_2$$

$$25 = 4 \cdot 4 \times 50 \times \phi_m \times 200,$$

$$\frac{25}{4 \cdot 4 \times 50 \times 200} = \phi_m \rightarrow 0.1126 \times 10^{-3} \text{ wb.}$$

Now area of core, $A = 1000 \text{ mm}^2$

$$1000 \times 10^{-6} \text{ m}^2 = 10^{-3} \text{ m}^2 //$$

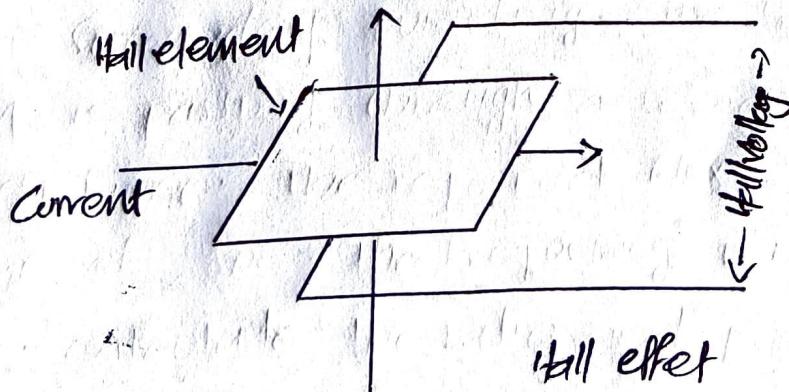
$$B_{max} = \frac{\phi_{max}}{Area}$$

$$= \frac{0.1126 \times 10^{-3}}{1000 \times 10^{-6}}$$

$$= 0.1126 \text{ wb/mm}^2$$

Hall Effect Multiplier

- The current is passed through the current coil which produces a magnetic field proportional to the current i . This field is \perp to the hall effect element.
- A current i_p is proportional to the voltage passed through hall element in a direction \perp to field which is limited by the multiplier resistance R_s .



The O/p Voltage of the Hall effect multiplier.

$$V_H = R_H \times \frac{IB}{t}$$

$R_H \rightarrow$ Hall G-efficient; $\propto -m/A - wbmt^2$

$B \rightarrow$ Flux density; $\propto wb/mc$

$t \rightarrow$ thickness of hall element

now $B \propto I$ and $I_{sp} = V/R_s \propto V$. Hence $V_H \propto V$

* O/p Voltage of Hall effect multiplier \propto P_{int}