

Module I

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Network Theorems.

(DC & AC Steady State Analysis)

Linearity and Superposition Principle.

A linear circuit is the circuit whose parameters are constant; ie they don't change with the application of voltage or current.

Eg:- R, L, C.

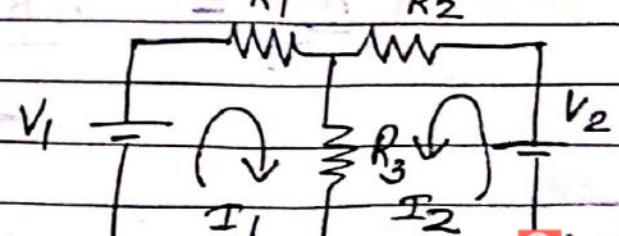
1 Superposition Theorem.

In any linear network, containing two or more sources, the response in any element is equal to the algebraic sum of the responses caused by individual sources acting alone, while the other sources are non-operational.

i.e. while considering the effect of individual sources, the other voltage sources and current sources in the network are replaced by the short circuit and open circuit across their terminals.

Proof

Consider a n/w having two energy Sources  $V_1$  &  $V_2$ .



Suppose we have to find current through  $R_3$ .

By applying mesh analysis.

$$I_{R_3} = I_1 + I_2$$

$$\text{Formula I } V_1 - I_1 R_1 - (I_1 + I_2) R_3 = 0 \Rightarrow V_1 = I_1 (R_1 + R_3) + I_2 R_3$$

$$\text{Formula II } V_2 - I_2 R_2 - (I_2 + I_1) R_3 = 0 \Rightarrow V_2 = I_2 (R_2 + R_3) + I_1 R_3$$

In matrix form,

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} R_1 + R_3 & R_3 \\ R_3 & R_2 + R_3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

$$\Rightarrow I_1 = \frac{V_1 - V_2 R_3}{R_1 + R_3 + R_2}$$

$$I_2 = \frac{V_2 (R_1 + R_3) - V_1 R_3}{R_1 + R_3 + R_2}$$

$$I_{R_3} = I_1 + I_2 = \frac{V_1 (R_2 + R_3) - V_2 R_3}{R_1 + R_3 + R_2}$$

$$I_{R_3} = \frac{V_1 (R_2 + R_3) - V_2 R_3}{R_1 + R_3 + R_2} = \frac{V_1 (R_2 + R_3) - V_1 R_3}{R_1 + R_3 + R_2}$$

$$I_{R_3} = I_1 + I_2 = \frac{V_1 (R_2 + R_3) - V_2 R_3 + V_2 (R_1 + R_3)}{R_1 + R_3 + R_2} - V_1 R_3$$

Expanding second term

$$\frac{I_{R_3}}{R_3} = \frac{V_1 R_2 + V_2 R_1}{R_1 + R_3 + R_2}$$

By Superposition theorem

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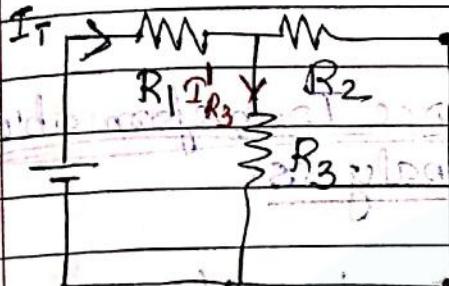
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$$I_{R_3} = I_{R_3} \text{ due to } V_1 + I_{R_3} \text{ due to } V_2$$

$V_1 = 0$

To find  $I_{R_3}$  due to  $V_1$  consider  $V_2 = 0$

Current through  $R_3$



$$I_{R_3} = I_T \times \frac{R_2}{R_2 + R_3}$$

$$I_T = \frac{V_1}{R_{\text{Req}}}$$

$\text{Req} = R_1 + R_2 R_3$

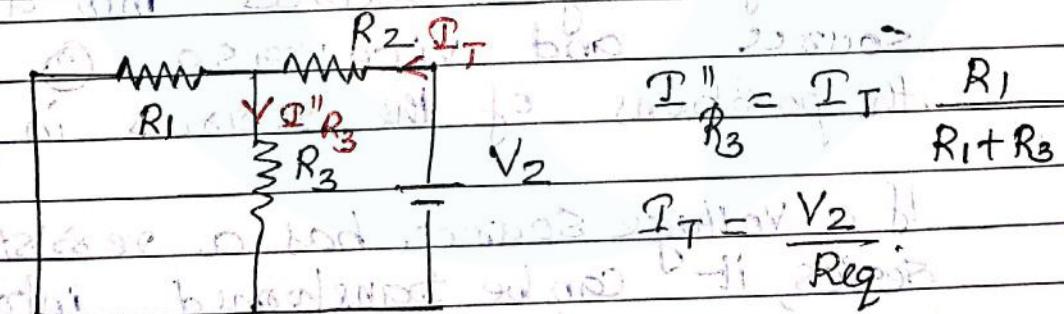
$$\therefore I_{R_3} = \frac{V_1}{R_1 + R_2 R_3} \times \frac{R_2}{R_2 + R_3}$$

$$I_{R_3} = \frac{(R_1 + R_2 R_3)}{R_2 + R_3}$$

$$R_1 R_2 + R_1 R_3 + R_2 R_3$$

in numerator

To find  $I_{R_3}$  due to  $V_2$  consider  $V_1 = 0$



$$I_{R_3}'' = I_T \times \frac{R_1}{R_1 + R_3}$$

$$I_T = \frac{V_2}{R_{\text{Req}}}$$

$$\therefore \text{Req} = R_2 + R_1 R_3$$

$$\begin{aligned} \therefore I_{R_3}'' &= \frac{V_2}{R_2 + R_1 R_3} \times \frac{R_1}{R_1 + R_3} \\ &= \frac{V_2 R_1}{R_1 R_2 + R_2 R_3 + R_1 R_3} \end{aligned}$$

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$$I_{R_3} = \frac{V}{R_3} + \frac{V}{R_3}$$

$$= V_1 R_2 + V_2 R_1 \Rightarrow V = V_1 R_2 + V_2 R_1$$

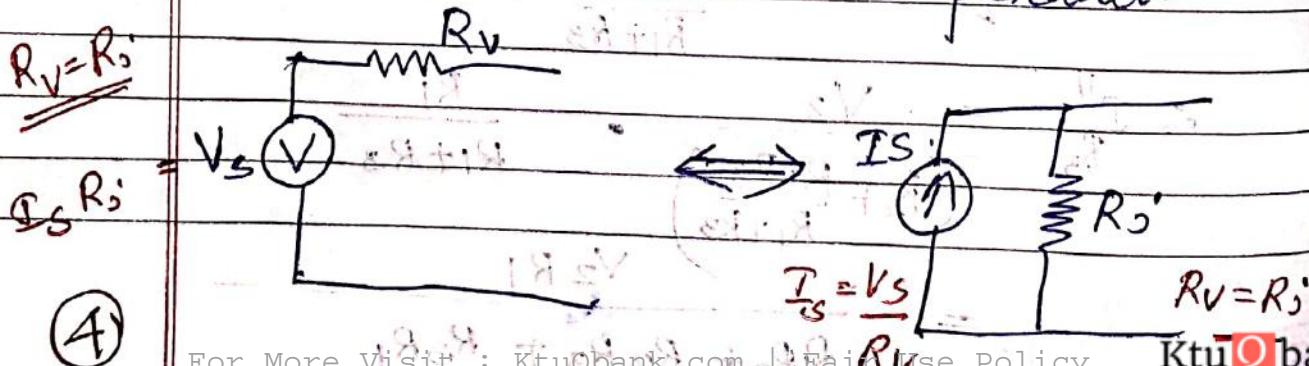
It is same as obtained from mesh analysis.  
This is the proof.

### Application of Source Transformation in electric circuit analysis.

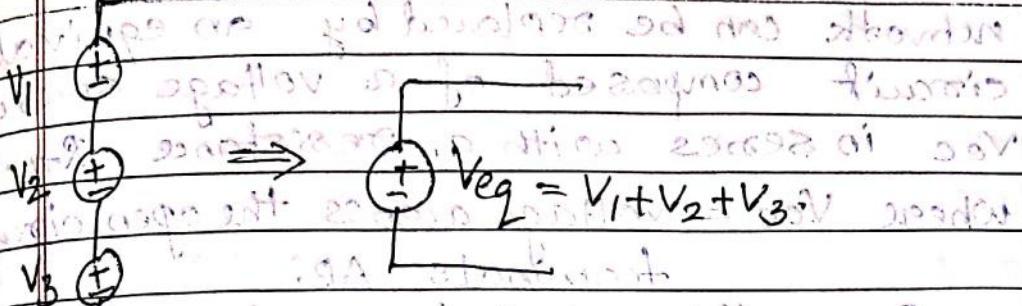
Normally, electric circuits are complicated consisting of many ckt elements. All such circuits can be solved easily by bringing them into the simplest form. This is called network reduction. It includes series ckt concept, parallel ckt concept, Y/A transformation, source transformation.

Source transformation means <sup>①</sup> transformation of voltage source into a current source and vice versa. <sup>②</sup> Shifting the positions of the sources in a n/w.

If a voltage source has a resistance in series, it can be transformed into an equivalent current source with the resistance in parallel.



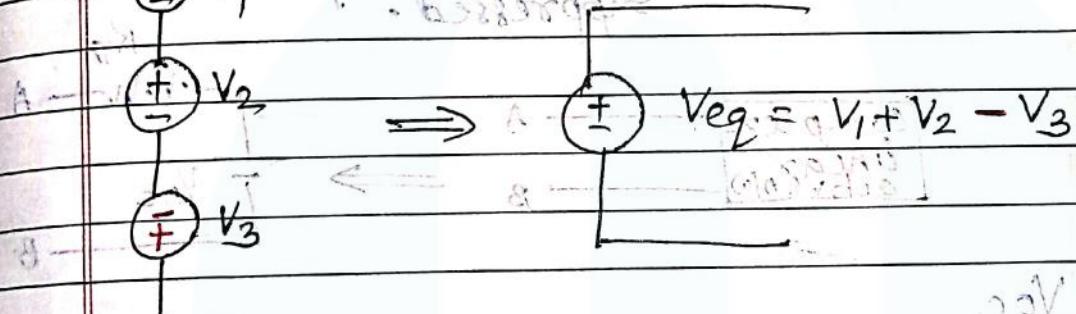
## Applications of Source Transformation.



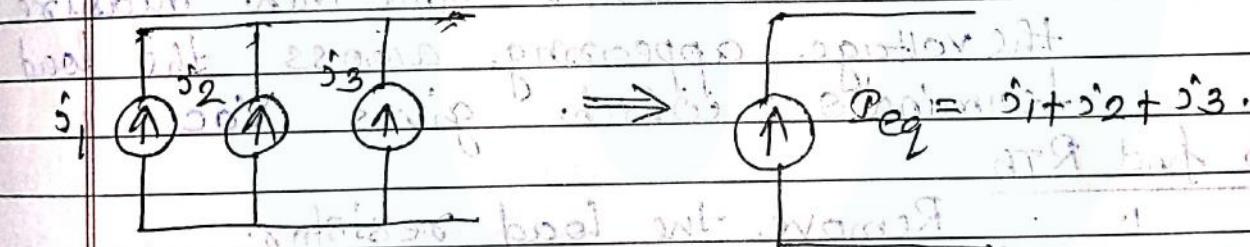
Let's consider the following circuit diagram:

Three voltage sources  $V_1$ ,  $V_2$ ,  $V_3$  are connected in series.

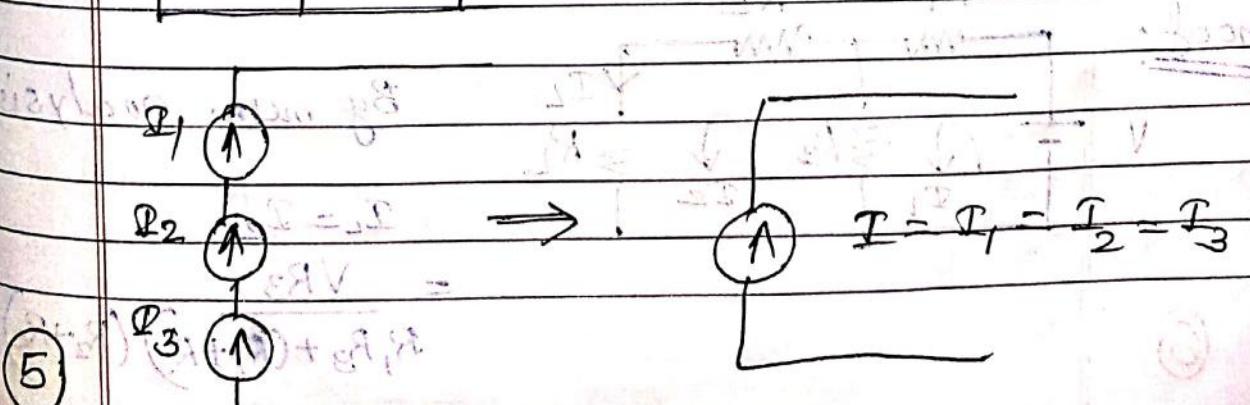
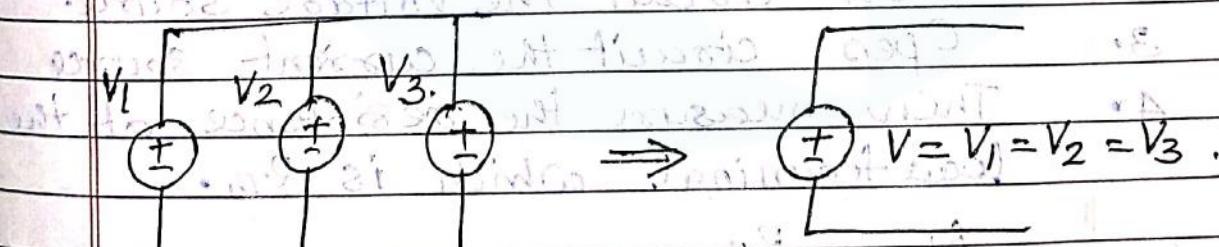
The equivalent voltage  $V_{eq}$  is given by:



Various methods to calculate total current:



Another method to calculate total current:

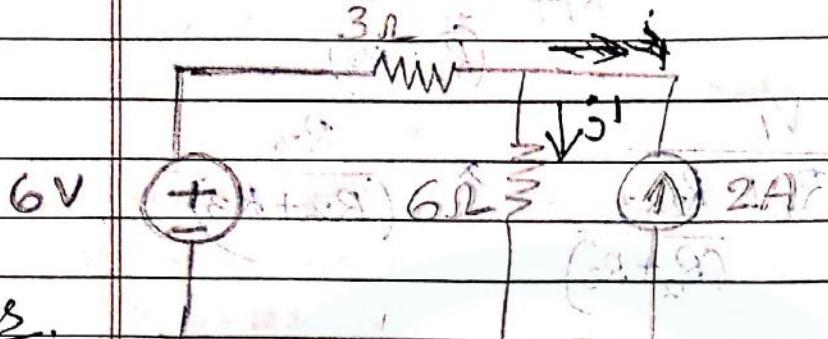


~~DC analysis~~

## Problems

Superposition theorem.

- 1) Find the current in the  $6\Omega$  resistor using the principle of SP.

Ans.

Set the current source to zero ( $2A$ ) i.e open ckt. ie find the current due to  $6V$

$$\text{Circuit diagram for finding current due to } 6V: \text{ A } 6V \text{ source is in series with a } 3\Omega \text{ resistor. This is in parallel with a } 6\Omega \text{ resistor. The current through the } 6\Omega \text{ resistor is } i_1. \text{ The total current } i = \frac{6}{3+6} = \frac{6}{9} \text{ A}$$

To find current due to  $2A$  (short ckt  $6V$ )

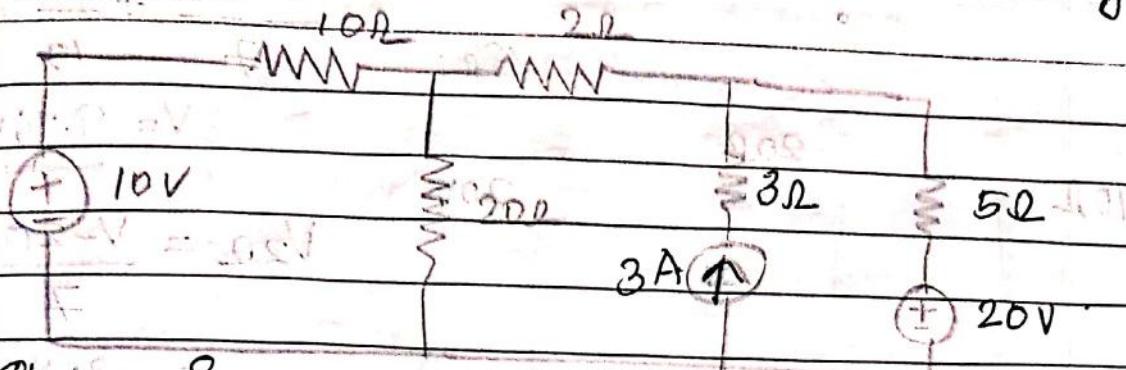
$$\text{Circuit diagram for finding current due to } 2A: \text{ A } 2A \text{ current source is in series with a } 6\Omega \text{ resistor. This is in parallel with a } 3\Omega \text{ resistor. The current through the } 3\Omega \text{ resistor is } i_2. \text{ The total current } i = \frac{2 \times 3}{3+6} = \frac{6}{9} \text{ A}$$

$$\begin{aligned} i &= i_1 + i_2 \\ &= \frac{6}{9} + \frac{6}{9} \end{aligned}$$

$$= \frac{12}{9} = \frac{4}{3} \text{ A.}$$

(1A)

2. Find the voltage across  $2\Omega$  resistor by SP theorem.



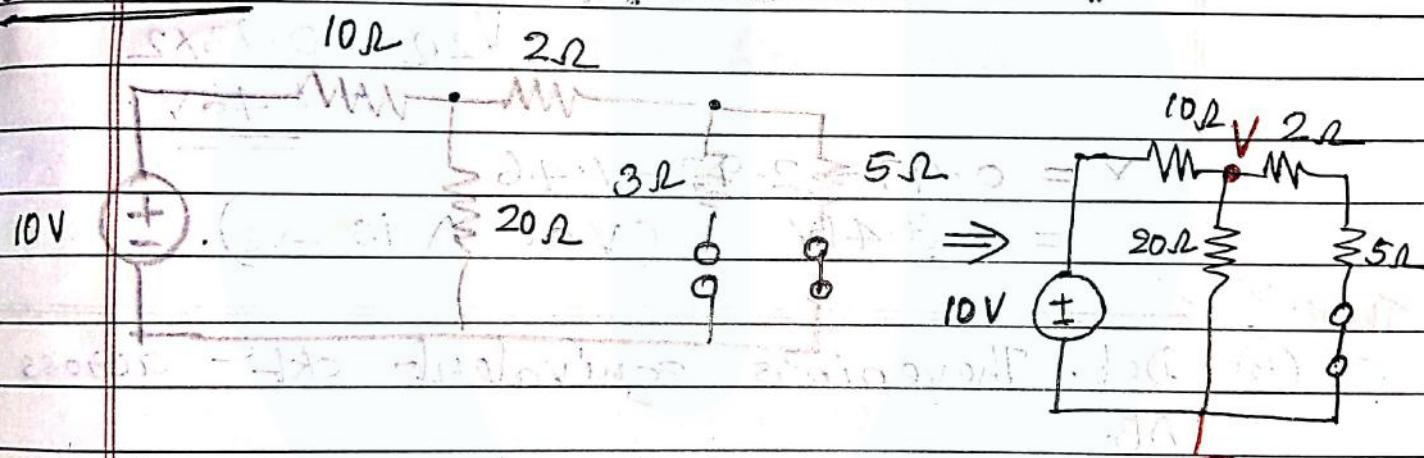
Ans: Three sources.

(10V, 3A, 20V)

Voltage across  $2\Omega$  resistor

= Voltage due to + Due to + Due to 20V

Due to 10V



Applying nodal analysis,

$$\frac{V-10}{10} + \frac{V}{20} + \frac{V}{7} = 0 \Rightarrow V = 3.41V$$

$$V_{2\Omega} = \frac{V \times 2}{2+5} = 0.97V$$

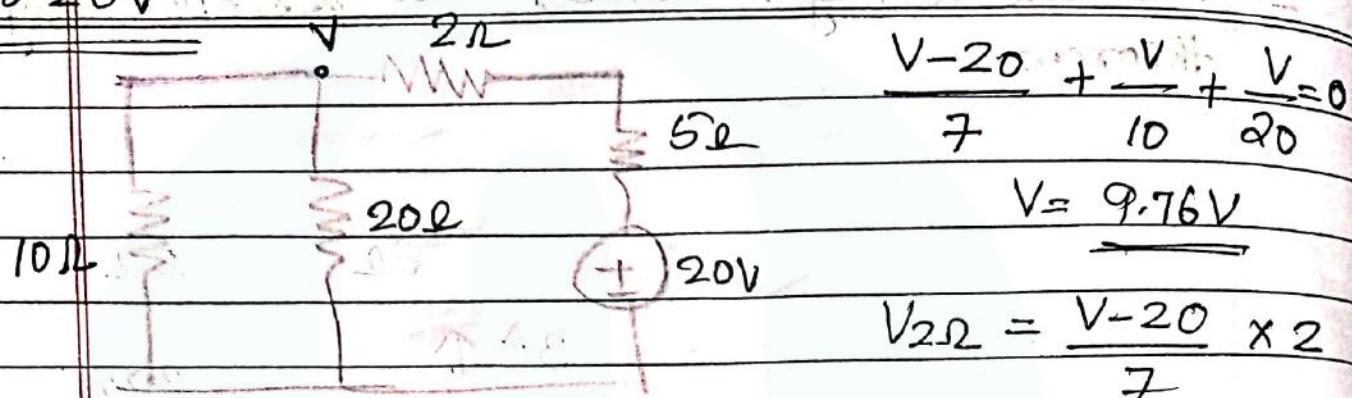
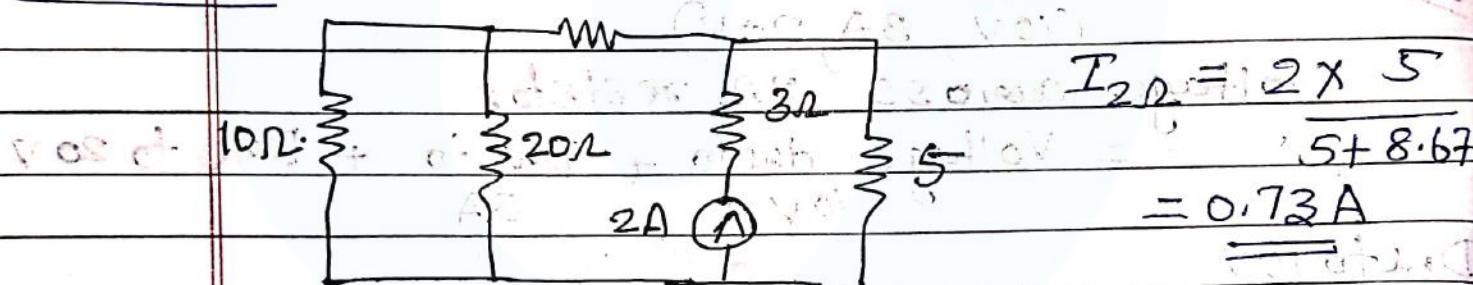
OR

$$I_{2\Omega} = \frac{10 \times 20/3}{\frac{20}{3} + 7} = 4.87A$$

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Due to  $20V$ Due to  $2A$ 

$$V_{2\Omega} = 0.73 \times 2 = 1.46V$$

$$\therefore V = 0.97 - 2.92 - 1.46$$

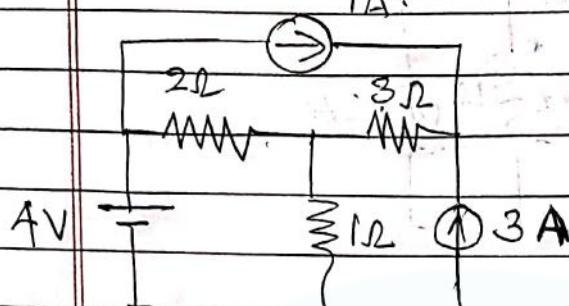
$$= -3.41V \quad (\text{V at A is -ve}).$$

Answers

HW 2 A

Find the current in the  $1\Omega$  resistor.

1A.



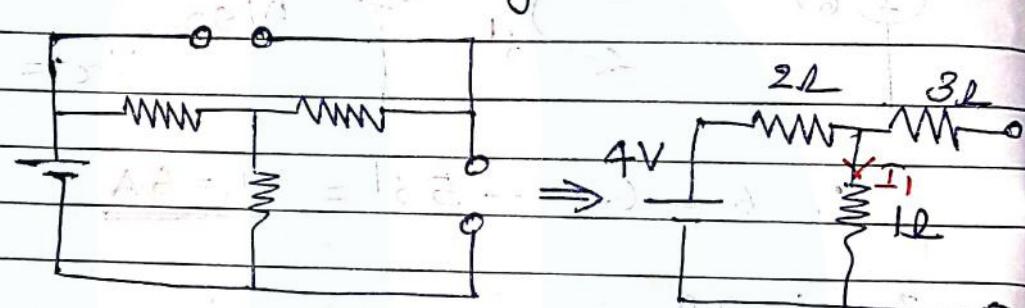
$$\frac{I}{4V} = \frac{1.33}{1\Omega}$$

2A

$$\frac{I}{3\Omega} = \frac{2}{3}$$

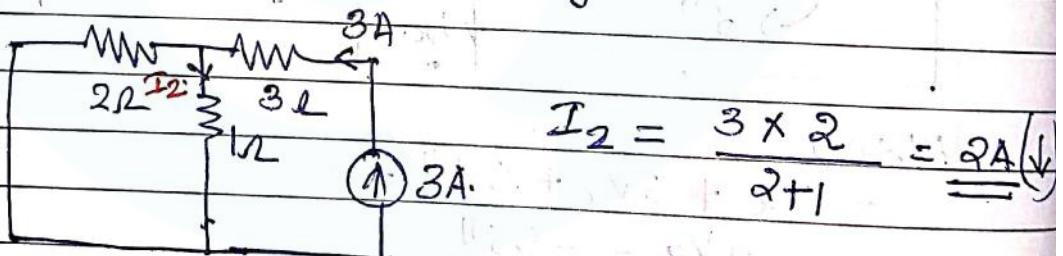
$$I = 4A$$

When the 4V source is acting alone.



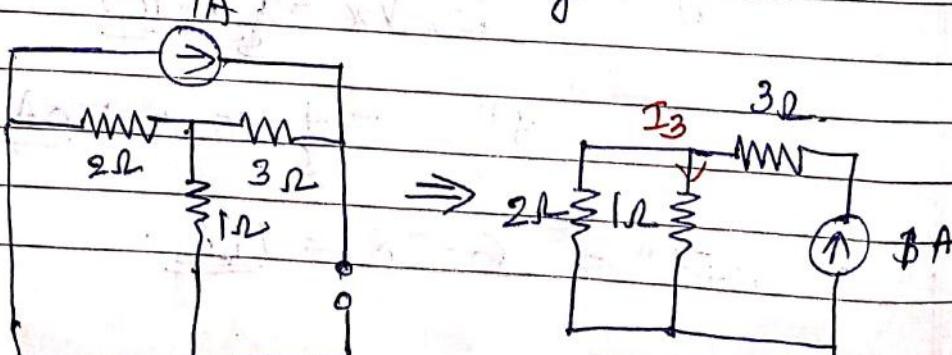
$$I_1 = \frac{4}{2+1} = 1.33A \quad (\downarrow)$$

When the 3A source is acting alone.



$$I_2 = \frac{3 \times 2}{2+1} = 2A \quad (\downarrow)$$

When the 1A source is acting alone.



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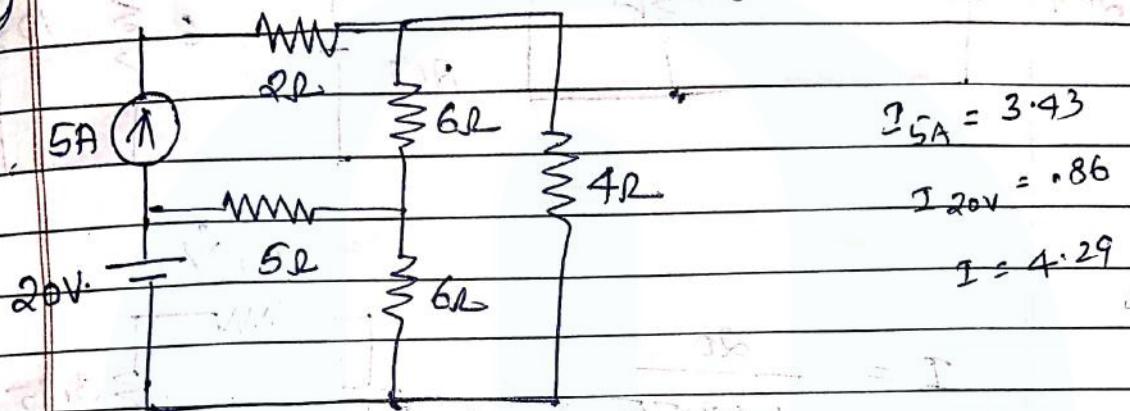
$$I_3 = \frac{1 \times 2}{2+1} = \frac{2}{3} A \quad (\downarrow)$$

Current in the  $1\Omega$  resistor =  $1.33 + 2 + \frac{2}{3}$   
 $= 4A \downarrow$

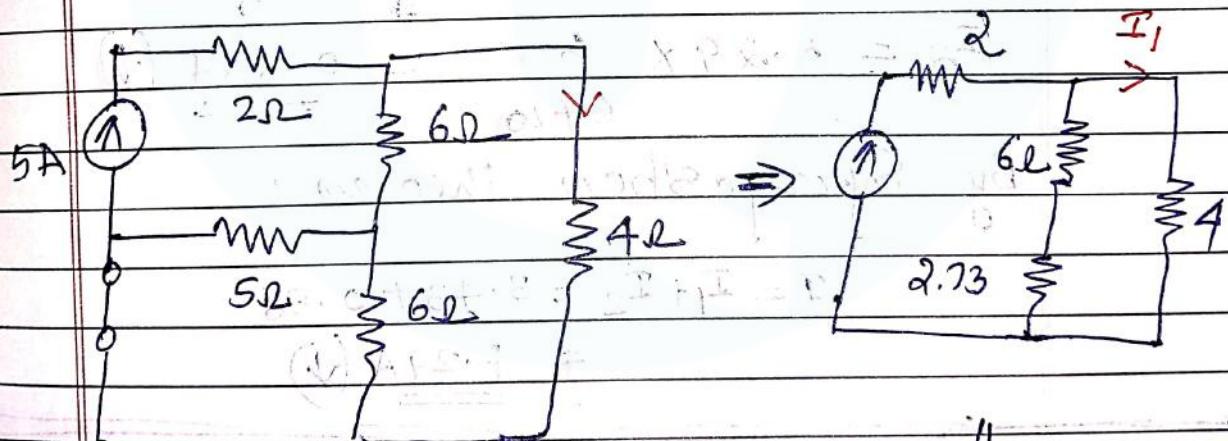
Ans 3

Find the current through the  $4\Omega$  resistor

(5)



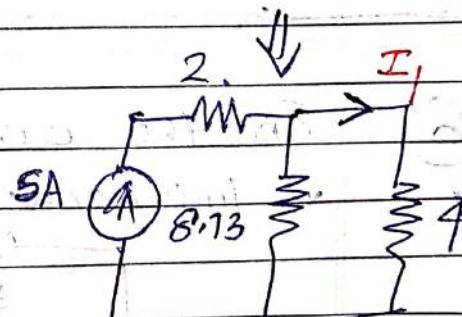
When the 5A source is acting alone.



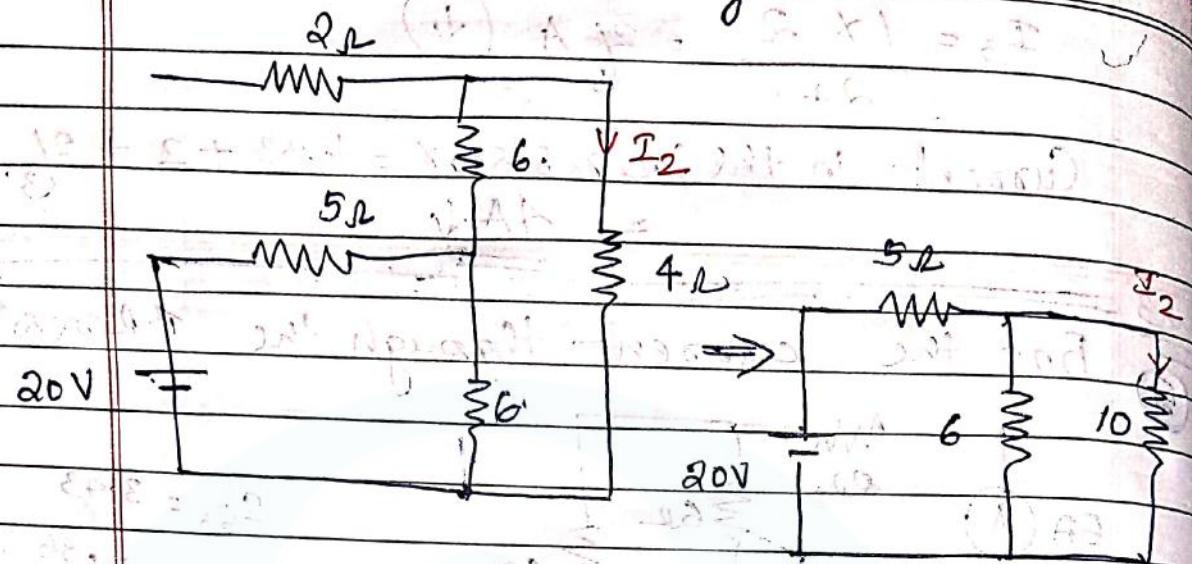
$$I_1 = 5 \times 8.73$$

$$= 8.73 + 4$$

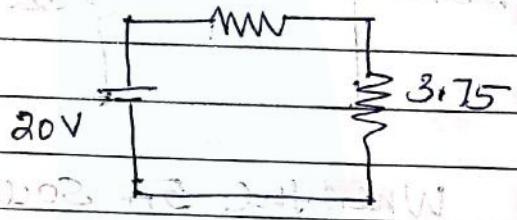
$$= 3.43 A \quad (\downarrow)$$



When the 20V source is acting alone.



$$I = \frac{20}{5+3.75}$$



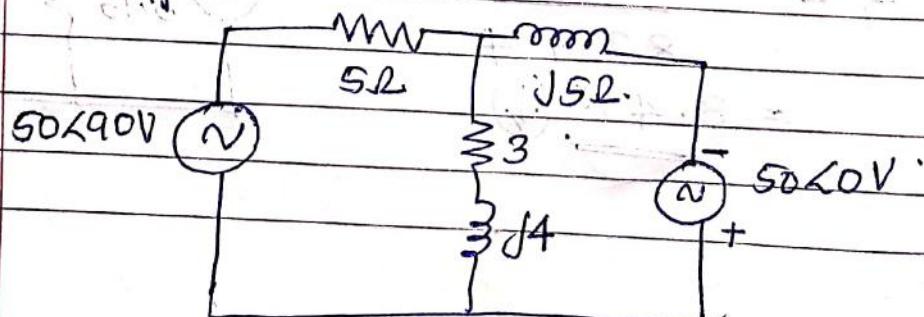
$$I_2 = \frac{2.29 \times 6}{6+10} = 0.86A \quad (\downarrow)$$

By Superposition theorem,

$$\begin{aligned} I &= I_1 + I_2 = 3.43 + 0.86 \\ &= 4.29A \quad (\downarrow) \end{aligned}$$

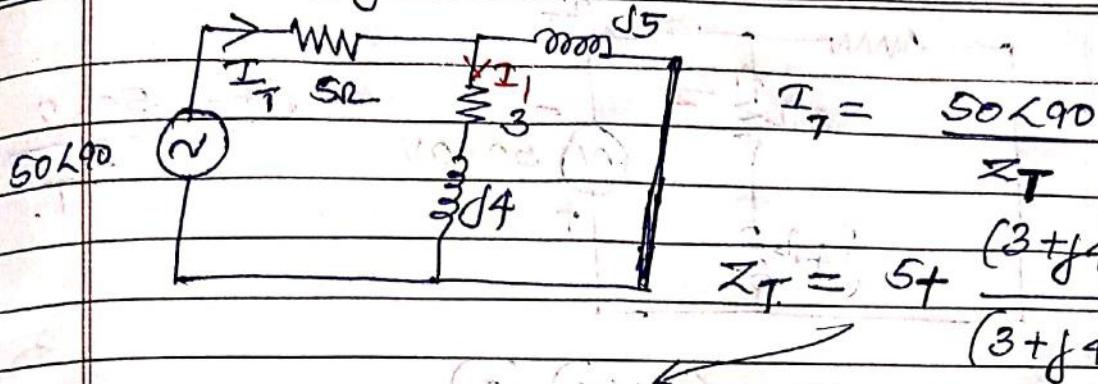
(6)

Find the current through the  $(3+j4)$  Ω impedance.



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When the  $50 \angle 90^\circ$  v source is acting alone,



$$I_T = \frac{50 \angle 90^\circ}{Z_T}$$

$$Z_T = 5 + \frac{(3+j4)(j5)}{(3+j4+j5)}$$

$$Z_T = 5 + \frac{j15 - 20}{3 + j9}$$

$$= 5 +$$

$$= 6.35 \angle 23.2^\circ$$

$$I_T = \frac{50 \angle 90^\circ}{6.35 \angle 23.2^\circ} = 7.87 \angle 66.8^\circ$$

By current division rule,

$$I_1 = \frac{I_T \times j5}{(j5 + 3 + j4)} = I_T \times \frac{5 \angle 90^\circ}{(3 + j9)}$$

$$= (7.87 \angle 66.8^\circ) \times \frac{5 \angle 90^\circ}{(3 + j9)}$$

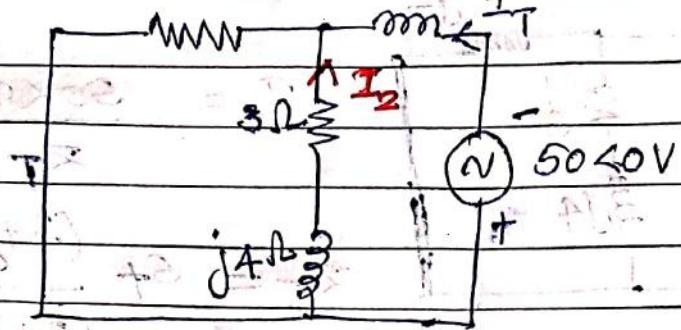
When the  $50 \angle 0^\circ$  v is acting alone.

Ranish

p. 5.16

5Ω

j5



$$I_T = 50\angle 0^\circ$$

$$Z_T = j5 + (3+j4)(5)$$

$$= j5 + \frac{15+j20}{8+j4}$$

$$I_T = \frac{50\angle 0^\circ}{6.74\angle 68.2^\circ} = 7.42\angle -68.2^\circ$$

$$I_2 = I_T \times \frac{5}{(5+3+j4)} = I_T \times \frac{5}{(8+j4)}$$

$$= 4.15 \angle -94.77^\circ$$

By superposition theorem

$$I = I_1 + I_2 = 4.15 \angle 85.3^\circ +$$

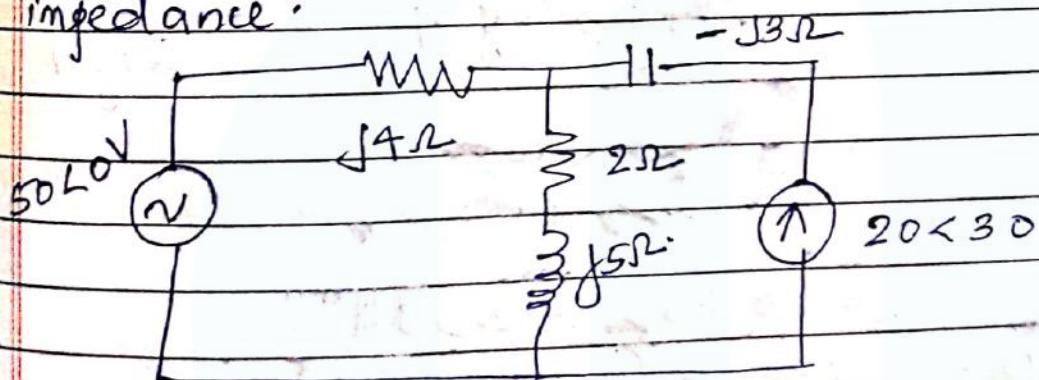
$$4.15 \angle -94.77^\circ$$

$$= 8.31 \angle 85.3^\circ$$

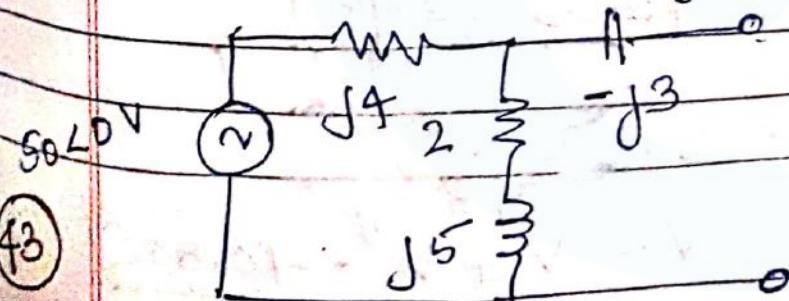
OB.

## Assignment 1 Questions

1. Determine the voltage across the  $(2+j5)\Omega$  impedance:



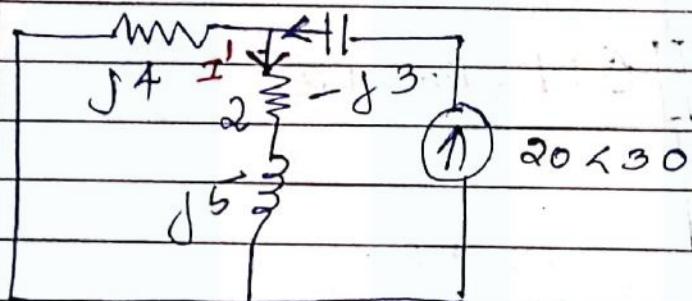
Ans.: Due to  $50\angle 0^\circ$  voltage source.



$$\text{Voltage across } (2+j5) \Omega \\ = \frac{50 \angle 0}{(2+j5) + j4} \times (2+j5)$$

$$V' = \frac{50 \angle 0}{2+j9} (2+j5) = \\ = 29.16 \angle -9.28^\circ$$

Due to  $20 \angle 30^\circ$  source is acting alone.



By current division rule,

$$I' = \frac{(20 \angle 30^\circ)(j4)}{2+j5+j9} = \frac{(20 \angle 30^\circ)(j4)}{2+j9}$$

$$= 8.68 \angle 42.53^\circ$$

Voltage across  $(2+j5) \Omega$

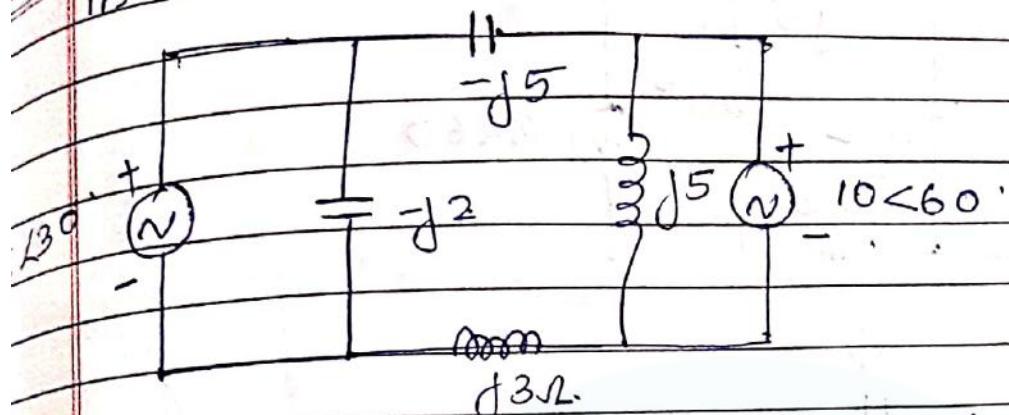
$$V'' = 8.68 \angle 42.53^\circ (2+j5)$$

$$= 46.69 \angle 110.72^\circ$$

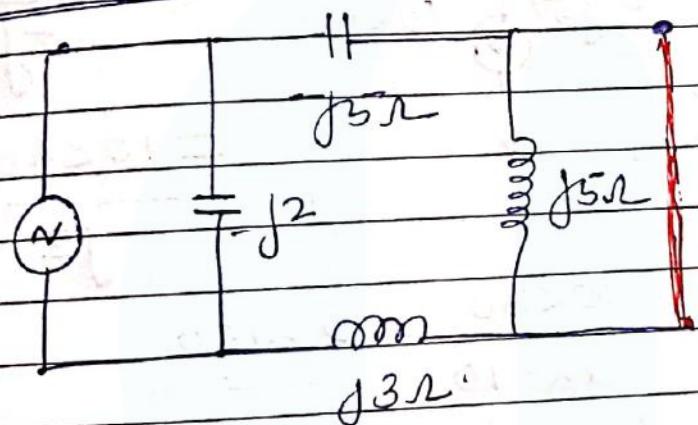
Total voltage  $V = V' + V'' = 29.16 \angle -9.28^\circ + 46.69 \angle 110.72^\circ$



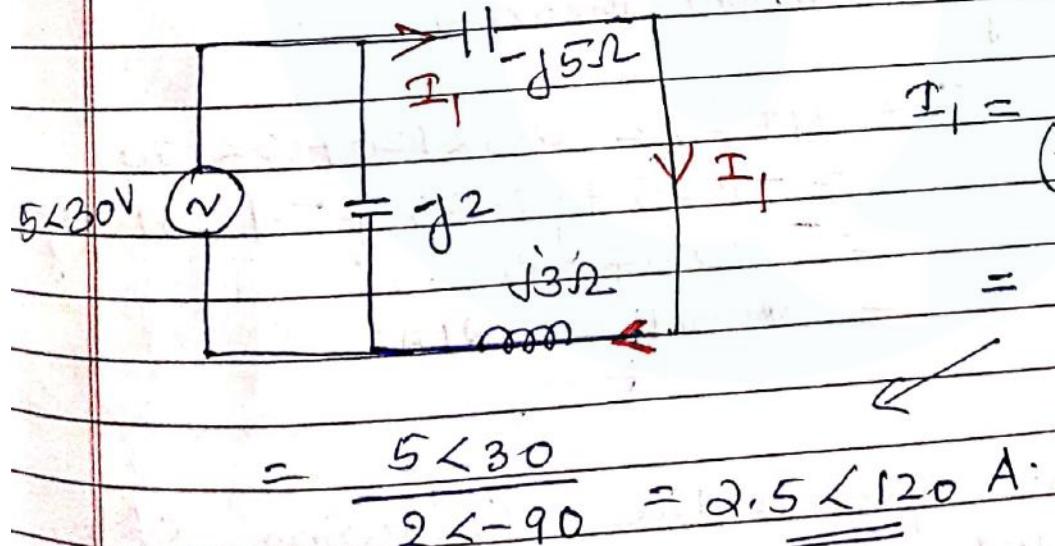
Q2. Find the current through the  $j3\Omega$  reactance in the network.



DS. Due to  $5\angle 30^\circ$  source is acting alone.



When a short ckt is placed across  $j5\Omega$ , it gets shorted.



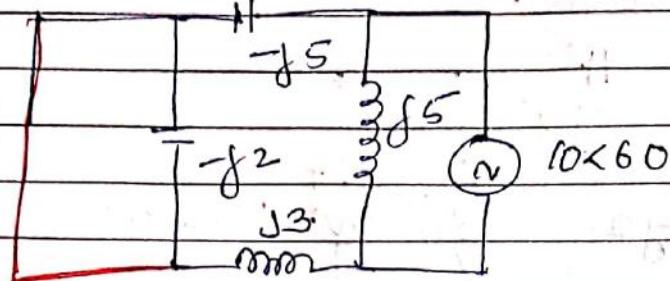
$$I_1 = \frac{5\angle 30}{(-j5 + j3)} = \frac{5\angle 30}{-j2}$$

$$= \frac{5\angle 30}{2\angle -90} = 2.5\angle 120 \text{ A}$$

$| -j5 + j3 |$   
in series  
 $| -j2 |$   
in parallel

Same voltage  
across  
parallel  
branches.

When the  $10 \angle 60^\circ$  source  
is acting alone.



$$\begin{aligned}
 & \text{Downward current } I_2 \text{ through } -j5\Omega \text{ branch} \\
 & \text{Circuit diagram: } \text{Voltage source } 10 \angle 60^\circ \text{ in series with load } 3+j3\Omega \\
 & \text{Current } I_2 = \frac{10 \angle 60^\circ}{-j5 + j3} \\
 & = 10 \angle 60^\circ \angle -j2 \\
 & = \frac{10 \angle 60^\circ}{2 \angle -90^\circ} = 5 \angle 150^\circ
 \end{aligned}$$

By Superposition theorem,

$$\begin{aligned}
 I &= I_1 + I_2 = 2.5 \angle 120^\circ + 5 \angle 150^\circ \\
 &= -1.25 + j2.165 + -4.33 + j0.5 \\
 &= -5.58 + j2.665 \\
 &= 6.31 \angle -6.2^\circ \text{ A}
 \end{aligned}$$

~~KTU~~

3)

With respect to the following ckt

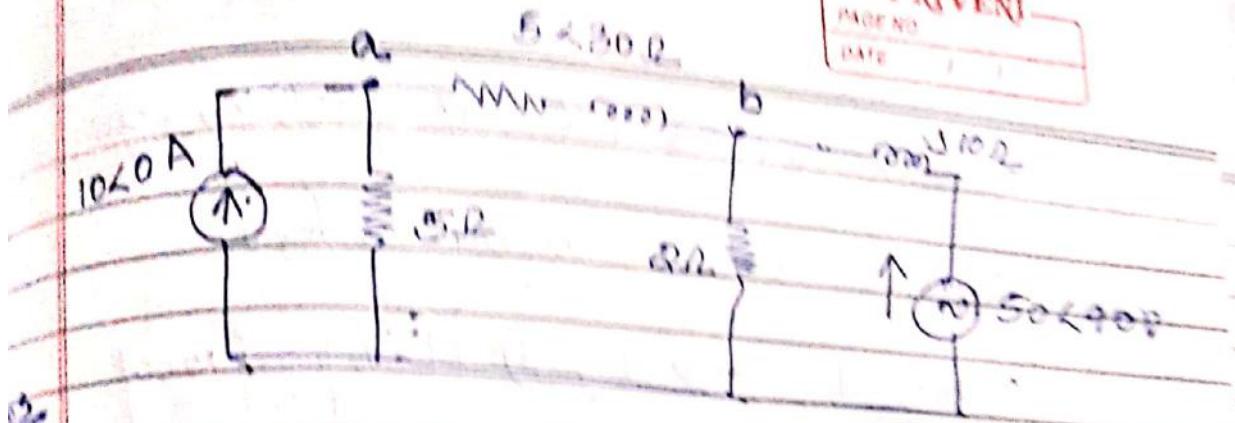
~~sample~~

a) Find the voltages at a and b  
using Superposition theorem.

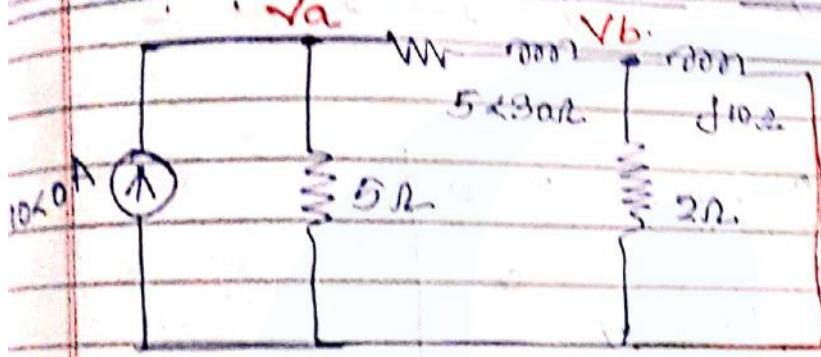
~~Q8~~

b) Obtain the active power dissipated  
in  $5 \angle 30^\circ \Omega$  impedance.

Ab



When the  $10\text{A}$  source is acting alone.



Applying KCL at node a

$$-10\text{A} + \frac{V_a - V_b}{5\text{k}\Omega} + \frac{V_a}{5} = 0$$

$$V_a \left[ \frac{1}{5\text{k}\Omega} + \frac{1}{5} \right] - V_b \left[ \frac{1}{5\text{k}\Omega} \right] = 10\text{A}$$

$$V_a [0.37 - j0.1] = V_b [0.17 - j0.1] = 10\text{A}$$

Applying KCL at node b

$$+ \frac{V_b - V_a}{5\text{k}\Omega} + \frac{V_b}{2} + \frac{V_b}{j10} = 0$$

$$\text{7) } V_a \left[ \frac{-1}{5\text{k}\Omega} \right] + V_b \left[ \frac{1}{5\text{k}\Omega} + \frac{1}{2} + \frac{1}{j10} \right] = 0$$

$$-(0.17-j0.1)V_a + (0.67-j0.2)V_b = 0$$

Express these equations in matrix form,

$$\begin{bmatrix} 0.37-j0.1 & -(0.17-j0.1) \\ -(0.17-j0.1) & 0.67-j0.2 \end{bmatrix} \begin{bmatrix} V_a \\ V_b \end{bmatrix} = \begin{bmatrix} 10 < 0 \\ 0 \end{bmatrix}$$

By Cramer's rule.

$$V_b = \frac{(0.37-j0.1) \cdot 10 < 0}{-(0.17-j0.1)} \cdot 0$$

$$\frac{(0.37-j0.1) \cdot - (0.17-j0.1)}{-(0.17-j0.1) \cdot 0.67-j0.2}$$

$$= (0.37-j0.1) \times 0 + 10 < 0 (0.17-j0.1)$$

$$(0.37-j0.1)(0.67-j0.2) - (0.17-j0.1)(0.17-j0.1)$$

$$= 1.7 - j1$$

$$= \frac{(0.383 \angle -15.12^\circ)(0.699 \angle 16.6^\circ) - (0.1972 \angle -30.47^\circ)(0.1972 \angle -30.47^\circ)}{(0.383 \angle -15.12^\circ)(0.699 \angle 16.6^\circ)}$$

(A8)

$$1.97 \angle -30$$

$$(6.383 \angle -15.12 + j 16.6) - (0.039 \angle -67.94)$$

$$1.97 \angle -30$$

$$0.2677 \angle 1.48 - 0.039 \angle -67.94$$

$$1.97 \angle -30$$

$$0.2676 - 0.0069j - 0.0146 + j 0.036$$

$$1.97 \angle -30 = 1.97 \angle -30$$

$$0.253 + 0.0291j \quad 0.255 \angle 6.56$$

$$= 7.73 \angle -36.56$$

$$V_a = \begin{vmatrix} 10 \angle 0 & -(0.17 - j 0.1) \\ 0 & 0.67 - j 0.2 \end{vmatrix}$$

Δ.

(Δ same  
as above)

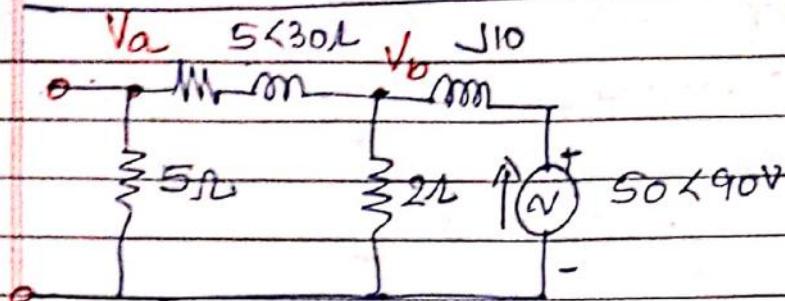
$$= 10 \angle 0 (0.67 - j 0.2) = 6.7 - j 2$$

$$0.255 \angle 6.56 \quad 0.255 \angle 6.56$$

$$= 6.99 \angle -16.6 = 27.41 \angle -23.16$$

$V_a \neq V_b$  due to  $10 \angle 0$  A source ↑

$V_a$  &  $V_b$  due to  $50 \angle 90^\circ V$ .



$$\frac{V_a}{5} + \frac{V_a - V_b}{5\angle 30} = 0 \Rightarrow V_a \left( \frac{1}{5} + \frac{1}{5\angle 30} \right) - \frac{V_b}{5\angle 30} = 0$$

$$\frac{V_b - V_a}{5\angle 30} + \frac{V_b}{2} + \frac{V_b - 50 \angle 90}{j10} = 0$$

$$V_a \left[ \frac{-1}{5\angle 30} \right] + V_b \left[ \frac{1}{5\angle 30} + \frac{1}{2} + \frac{1}{j10} \right] = \frac{50 \angle 90}{j10}$$

In matrix form

$$\begin{vmatrix} \left( \frac{1}{5} + \frac{1}{5\angle 30} \right) & -\frac{1}{5\angle 30} \\ -\frac{1}{5\angle 30} & \frac{1}{5\angle 30} + \frac{1}{2} + \frac{1}{j10} \end{vmatrix} \begin{Bmatrix} V_a \\ V_b \end{Bmatrix} = \begin{Bmatrix} 0 \\ 50 \end{Bmatrix}$$

$$V_a = \frac{\begin{vmatrix} 0 & -\frac{1}{5\angle 30} \\ 5 & \frac{1}{5\angle 30} + \frac{1}{2} + \frac{1}{j10} \end{vmatrix}}{\Delta}$$

(50)

$$\Delta =$$

$$0.2 + 0.2 \angle -30$$

$$-(0.2 \angle 30)$$

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$$-(0.2 \angle -30)$$

$$0.2 \angle -30$$

$$+ 0.5 \angle 0^\circ$$

$$= 0.2 + 0.123 - 0.1j$$

$$= 0.123 \angle -30^\circ$$

$$-(0.2 \angle -30)$$

$$0.123 - 0.1j$$

$$+ 0.5 - j0.1$$

$$0.373 - 0.1j$$

$$-(0.2 \angle -30)$$

$$-0.2 \angle -30$$

$$0.673 - j0.2$$

Cross-Multiplying

$$= (0.385 \angle -15^\circ) (0.702 \angle -16.6^\circ)$$

$$- (0.2 \angle -30) (0.2 \angle -30)$$

$$0.385 \angle -15^\circ$$

$$= 0.246 \angle -31.6^\circ$$

$$0.702 \angle -16.6^\circ$$

$$- 0.04 \angle -60^\circ$$

$$= 0.204 - 0.129j - (7.02 - 0.006j)$$

$$= -6.916 - 0.123j = 6.92 \angle -178^\circ$$

$$\therefore V_a = \frac{\frac{1}{5 \angle 80^\circ} \times 5}{\Delta} = \frac{1 \angle -30^\circ}{6.92 \angle -178^\circ}$$

$$\Delta = 0.145 \angle +175^\circ$$

$$\begin{array}{c} V_a \\ \parallel \\ \begin{array}{|c|c|c|} \hline & 6 & 0 \\ \hline 6 & 1 & 30 \\ \hline 6 & 1 & 30 \\ \hline 0 & & 0 \\ \hline \end{array} \end{array} \quad \begin{array}{c} V_b \\ \parallel \\ \begin{array}{|c|c|c|} \hline & 6 & 0 \\ \hline 6 & 1 & 30 \\ \hline 6 & 1 & 30 \\ \hline 0 & & 0 \\ \hline \end{array} \end{array}$$

$$\begin{array}{r} A \\ \parallel \\ \begin{array}{r} 1 + 540 \\ - 5120 \\ \hline 6 \\ \parallel \end{array} \end{array} \quad \begin{array}{r} B \\ \parallel \\ \begin{array}{r} 1 + 1430 \\ - 6924 \\ \hline 6 \\ \parallel \end{array} \end{array}$$

$$\begin{array}{r} 0.865 = 0.6j + 1 \\ - 6.924 \\ \hline 6.924 - 128 \\ \parallel \end{array} \quad \begin{array}{r} 1.865 - 0.5j \\ - 6.924 \\ \hline 6.924 - 128 \\ \parallel \end{array}$$

$$\begin{array}{r} 128.134 - 8.718 \\ - 6.924 \\ \hline 25.714 \\ \parallel \end{array} \quad \begin{array}{r} 1.865 - 0.5j \\ - 6.924 \\ \hline 6.924 - 128 \\ \parallel \end{array}$$

By Superposition Theorem

$$V_a = 27.414 - 23.16 + 0.145 \angle 175^\circ$$

$$\begin{aligned} &= 26.2510.78j + -0.144 + 0.0126 \\ &= 26.056 - 10.77j \end{aligned}$$

$$V_b = 7.734 - 36.56 + 25.714 \angle 90.23^\circ$$

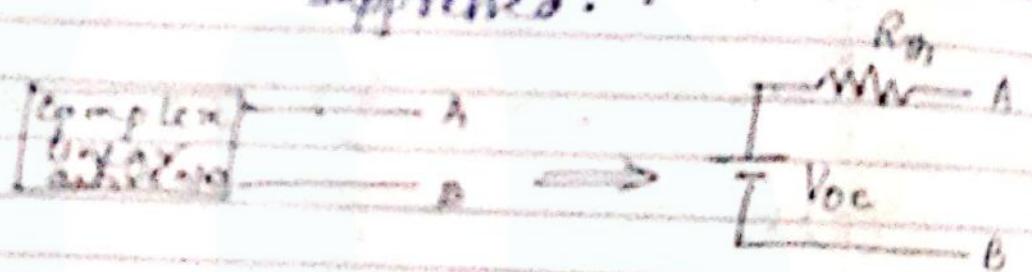
$$\begin{aligned} &= 6.91 - 4.6j - 0.103 + 25.714j \\ &= 6.807 + 21.14j \end{aligned}$$

(62)

## 2. Thvenin's theorem.

At a pair of terminals AB, any linear network can be replaced by an equivalent circuit composed of a voltage source  $V_{oc}$  in series with a resistance  $R_{Th}$ . Where  $V_{oc}$  = voltage across the open circuit terminals AB.

$R_{Th}$  = the equivalent resistance of the network as seen from terminals AB, with all independent sources suppressed.



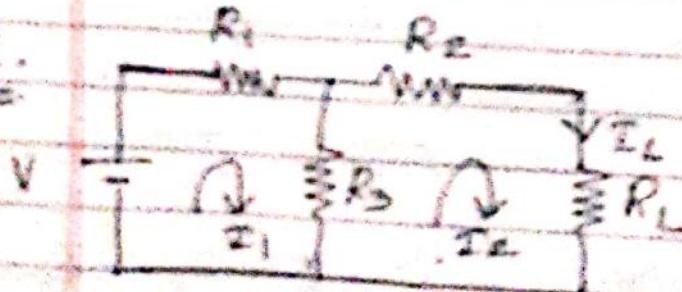
To find  $V_{oc}$

Remove load resistor and then measure the voltage appearing across the load terminals, which gives  $V_{oc}$ .

To find  $R_{Th}$

1. Remove the load resistor.
2. Short circuit the voltage source.
3. Open circuit the current source.
4. Then measure the resistance at the load terminal which is  $R_{Th}$ .

Proof:



By mesh analysis,

$$I_1 = I_2$$

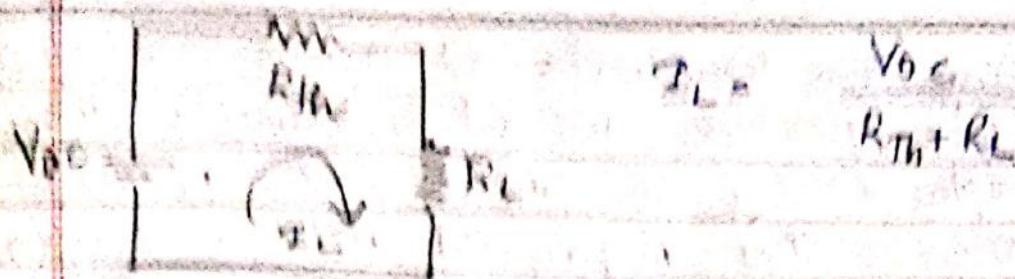
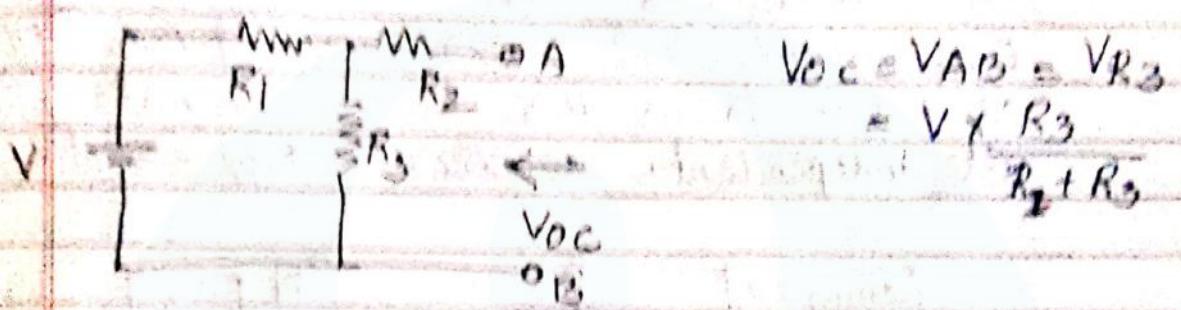
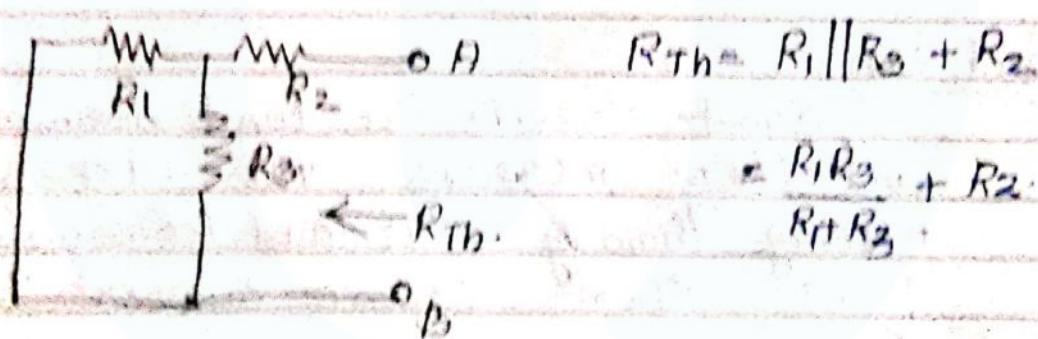
$$= \frac{VR_3}{R_1R_3 + (R_1 + R_2)(R_2 + R_3)}$$

$$R_1R_3 + (R_1 + R_2)(R_2 + R_3)$$

(6)

By Theorem

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$$I_L = \frac{V_{oc}}{R_{Th} + R_L} = \frac{V R_3 / (R_1 + R_3)}{\frac{R_1 R_3}{R_1 + R_3} + R_2 + R_L}$$

$$= \frac{V R_3}{(R_1 + R_3) + R_1 R_3 + R_2 R_1 + R_2 R_3}$$

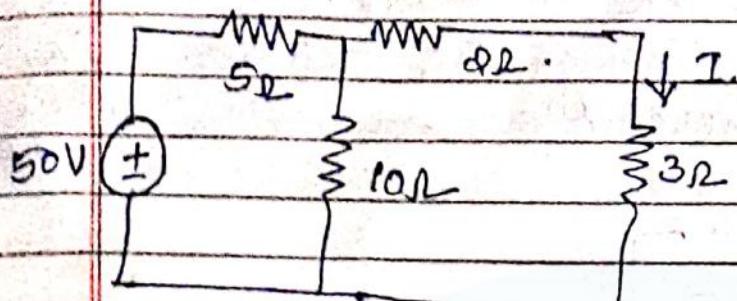
$$(R_1 + R_3)$$

$$= \frac{R_1 R_2 + R_2 R_3 + R_3 R_1}{R_1 R_2 + R_2 R_3 + R_3 R_1 + R_L (R_1 + R_3)}$$

④

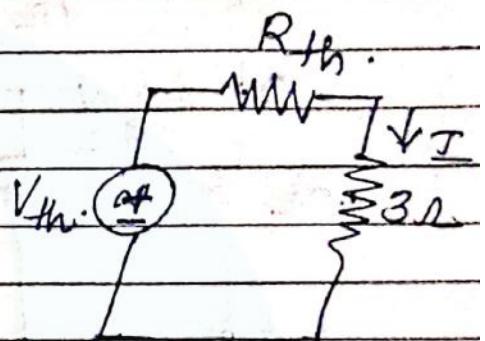
This is the proof

Using Thevenin's theorem, find the current through  $3\Omega$  resistor.

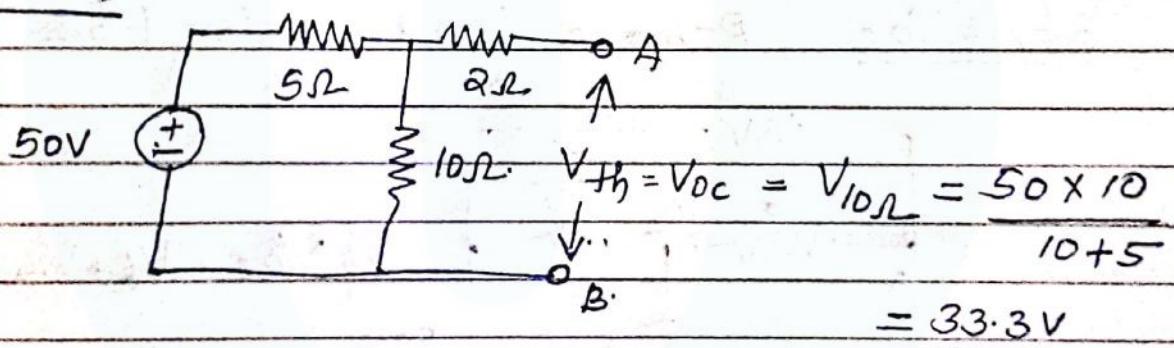
Ans

Thevenin's equivalent Ckt

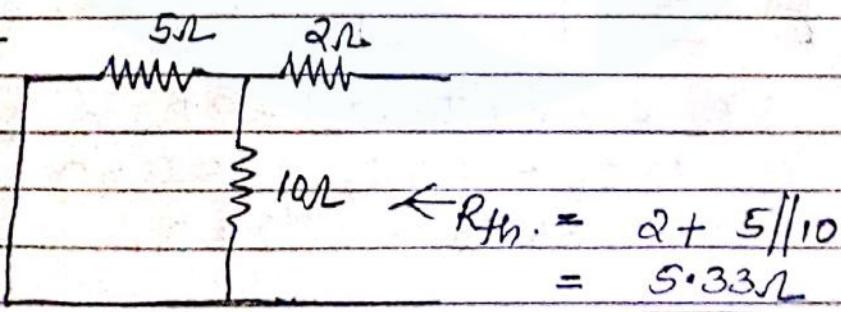
$$I = \frac{V_{th}}{R_{th} + R_L}$$



To find  $V_{th}$



To find  $R_{th}$



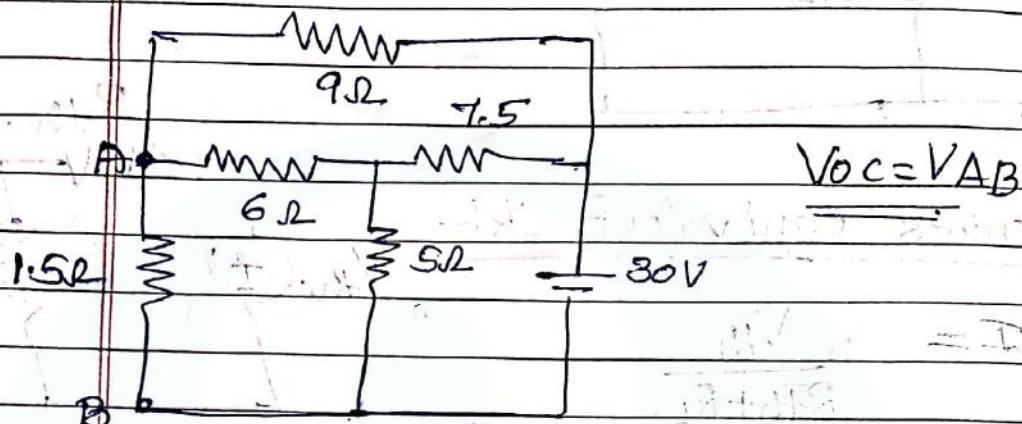
$$I_L = \frac{33.3}{5.33 + 2} = \underline{\underline{4A}}$$

For the ckt, find ① the  $V_{oc}$  at AB.

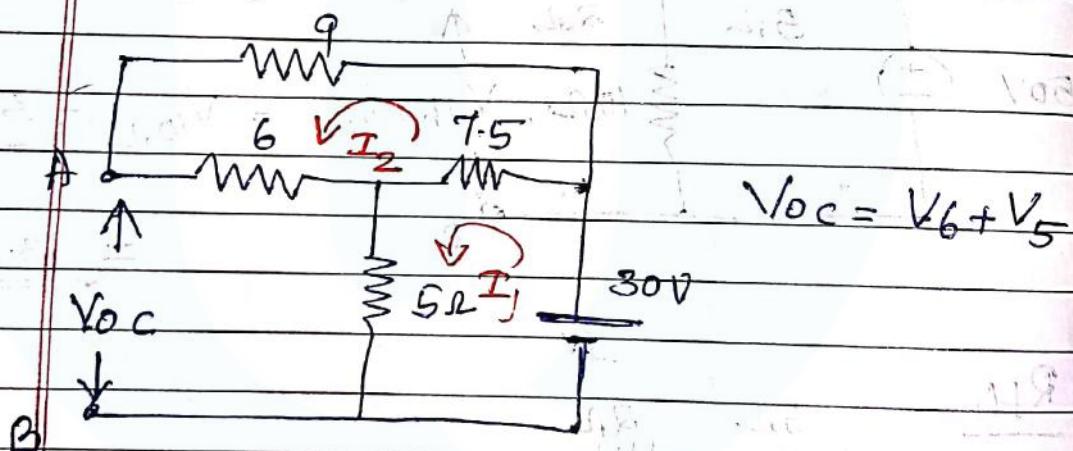
② the looking back resistance at AB.

③ the current through the resistance

Draw the Thevenin's equivalent ckt



To find  $V_{oc}$ , Remove load resistor  $1.5\Omega$



$$6I_2 + 7.5(I_2 - I_1) + 9I_2 = 0$$

$$15I_2 + 7.5I_2 - 7.5I_1 = 0$$

$$22.5I_2 - 7.5I_1 = 0$$

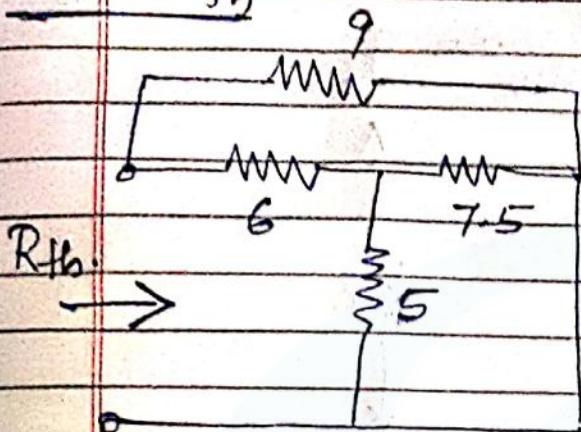
$$30 - 7.5(I_1 - I_2) - 5I_1 = 0$$

$$30 = 12.5I_1 - 7.5I_2$$

$$\Rightarrow \underline{\underline{I_1 = 3A}}, \underline{\underline{I_2 = 1A}}$$

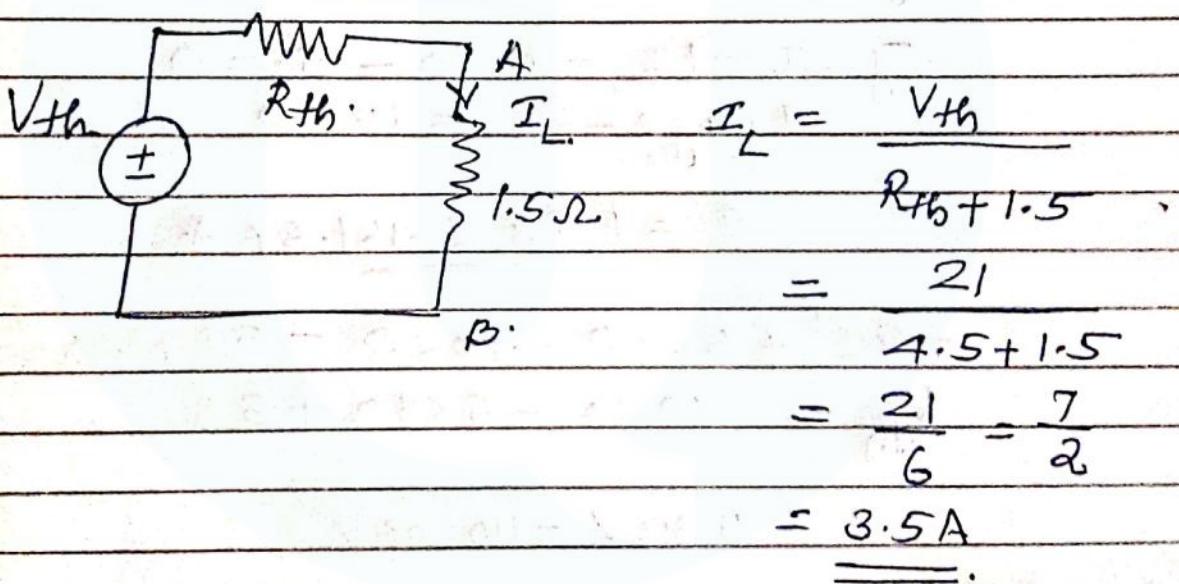
$$V_{oc} = I_1 \times 5 + I_2 \times 6 = 3 \times 5 + 1 \times 6 = 15 + 6 \\ = \underline{\underline{21V}}$$

To find  $R_{th}$

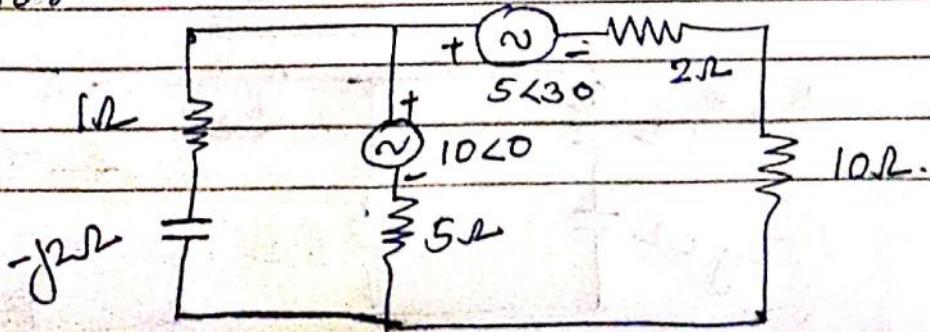


$$(7.5 \parallel 5 + 6) \parallel 9 \\ = \underline{\underline{4.5\Omega}}$$

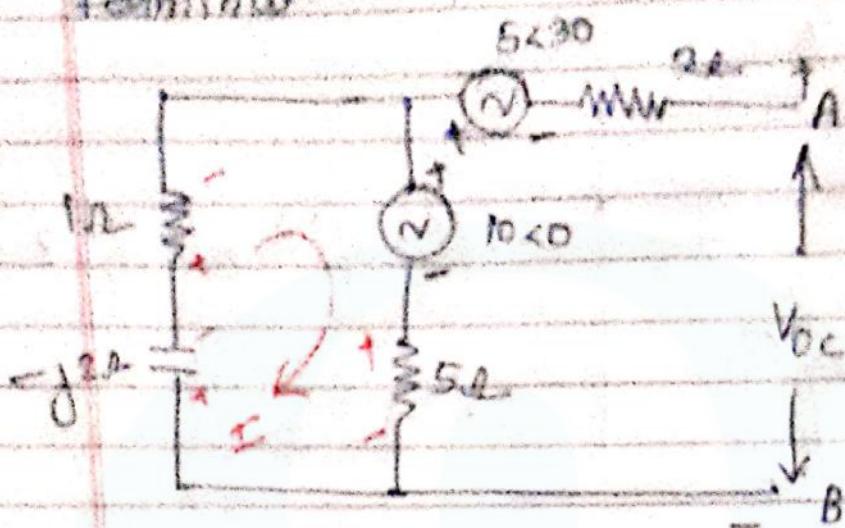
Thevenin's eqt ckt



Draw the Thevenin's eqt ckt.  
Find the current through the 10 ohm resistor.



To find  $V_{TH}$  - Remove  $10\Omega$  resistor, then measure voltage across open circuited terminals.



Applying KVL to the mesh:

$$-\jmath 2I - \mathbf{I} - 10\angle 0^\circ - 5I = 0$$

$$\mathbf{I} (+\jmath 2 - 6) = 10\angle 0^\circ$$

$$\mathbf{I} = 1.58 \angle -161.57^\circ$$

$$-2I^1 + 5\angle 30^\circ - 10\angle 0^\circ - 5I + V_{TH} = 0$$

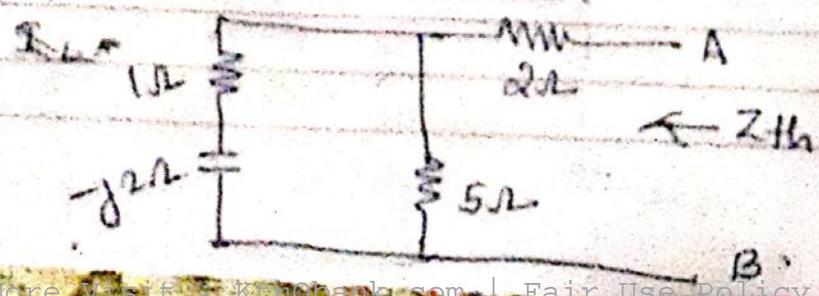
$$V_{TH} = 10\angle 0^\circ - 5\angle 30^\circ + 5I$$

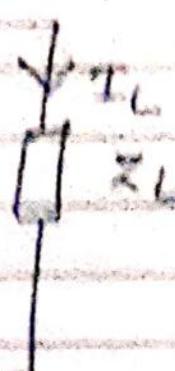
$$= 5.32 \angle -110.06^\circ$$

To find  $Z_{TH}$

$$Z_{TH} = \frac{5(1-\jmath 2)}{6+\jmath 1-\jmath 2}$$

$$= 3.48 \angle -21.84^\circ$$



Z<sub>1h</sub>V<sub>1h</sub>Y<sub>1h</sub>T<sub>1h</sub>V<sub>1h</sub>Z<sub>1h</sub> + Z<sub>L</sub>

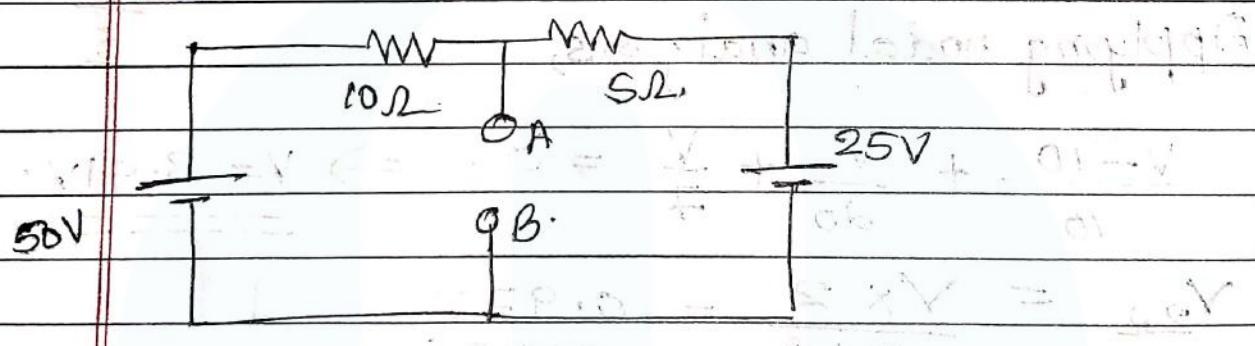
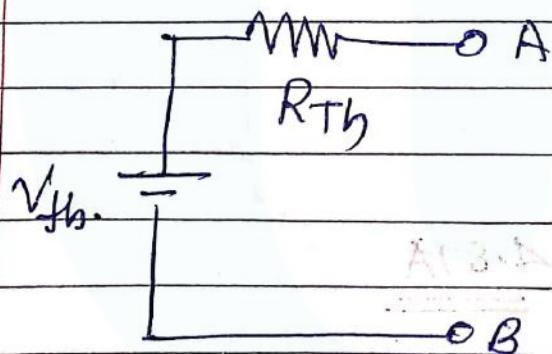
S 38 L - 110.06

3.48 L - 21.04 + 10

= 0.4 L - 10.4 - 67

~~Thevenin's~~

- ③ Det. Thevenin's equivalent circuit across AB.

Ans.

16

Given the load resistance, voltage across open-circuited load terminals is  $V_{th}$ .



$$\begin{aligned} 35 - 10I &= 5I \Rightarrow 25 = 15I \\ I &= \frac{25}{15} = \frac{5}{3} A \\ V_{th} &= 10 \times \frac{5}{3} = \underline{\underline{16.7 V}} \end{aligned}$$

$$V_{th} = 10 - 10I = 10 + 25 \times \frac{5}{3} = \underline{\underline{33.3 V \times V_m}}$$



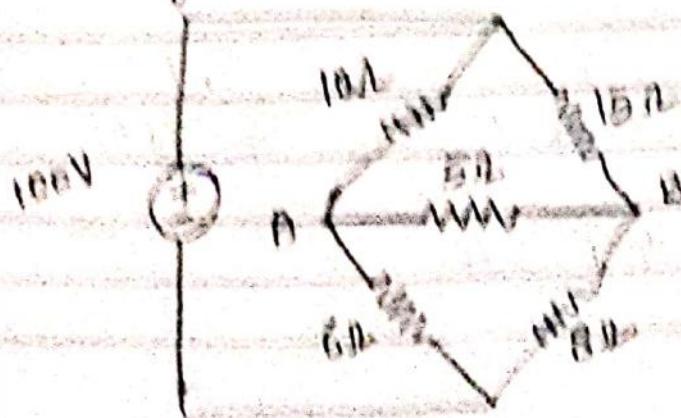
$$R_{th} = 10 \parallel 5 = \underline{\underline{3.33 \Omega}}$$

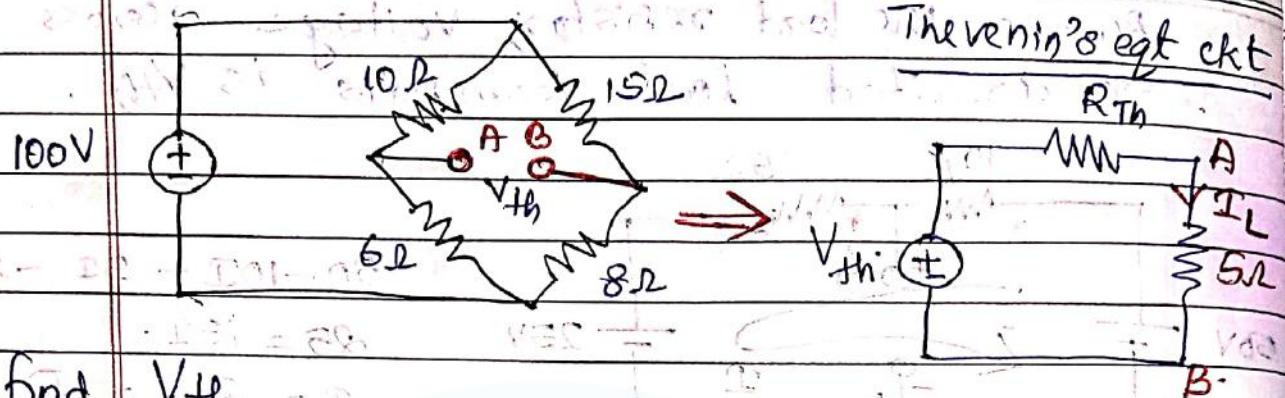
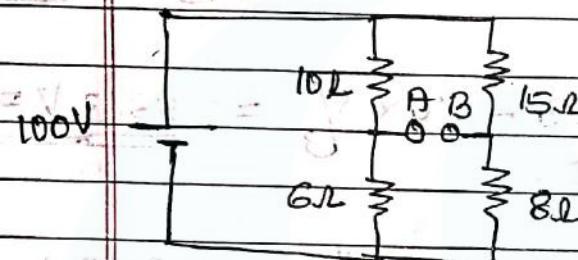
$R_{th}$ ,  
 $V_m$  &

$V_{th}$

Q.B

- ④ Using Thevenin's Theorem, find the current through the 5Ω resistor.



AnsRemove  $5\Omega$  resistorTo find  $V_{th}$ 

$$V_{th} = V_{AB}$$

$$= V_6 - V_8$$

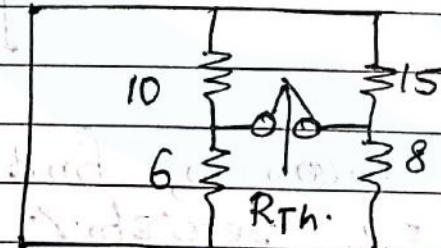
$$I_6 = \frac{100}{10+15} = \frac{100}{25} = 4A$$

$$I_8 = \frac{100}{23} = 4.35A$$

$$V_8 = I_8 \times 8 = 4.35 \times 8 = 34.8V$$

$$V_{AB} = 37.5 - 34.8$$

$$= 2.7V$$

To find  $R_{th}$ 

$$\frac{6 \times 10}{6+10} + \frac{15 \times 8}{15+8} = R_{th}$$

$$= 8.97 \Omega$$

$$I_L = \frac{V_{oc}}{R_{th} + R_L} = \frac{2.7}{8.97 + 5} = 0.193A$$

(18)

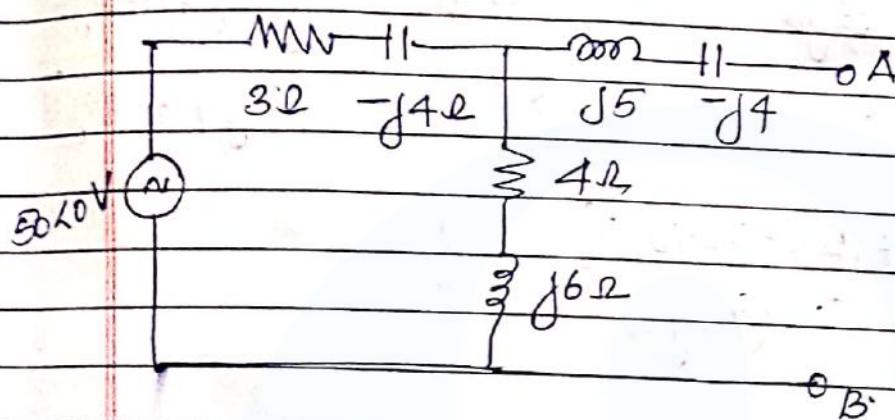
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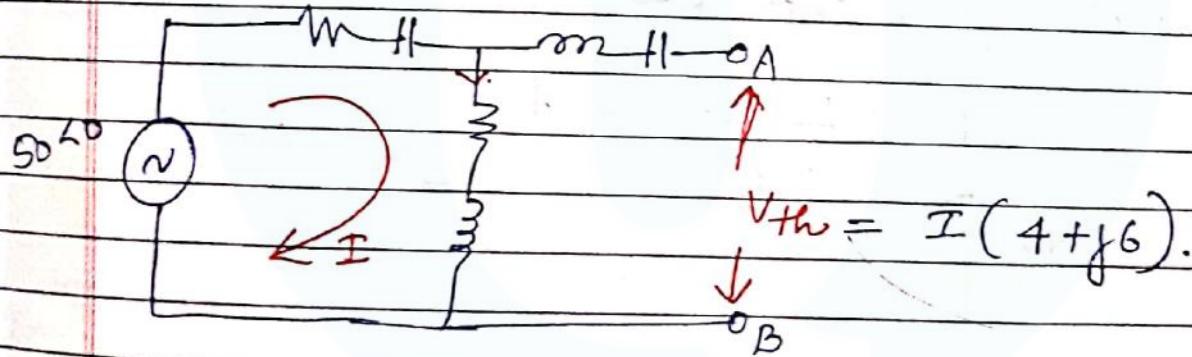
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Thevenin's theorem.Assignment Qn 3

- 1) Obtain Thevenin's equivalent network for the terminals A & B.



Ans

To find  $V_{th}$ 

$$I = \frac{50 \angle 0}{3 - j4 + 4 + j6} = \frac{50 \angle 0}{7 + j2}$$

$$= 6.87 \angle -15.95^\circ$$

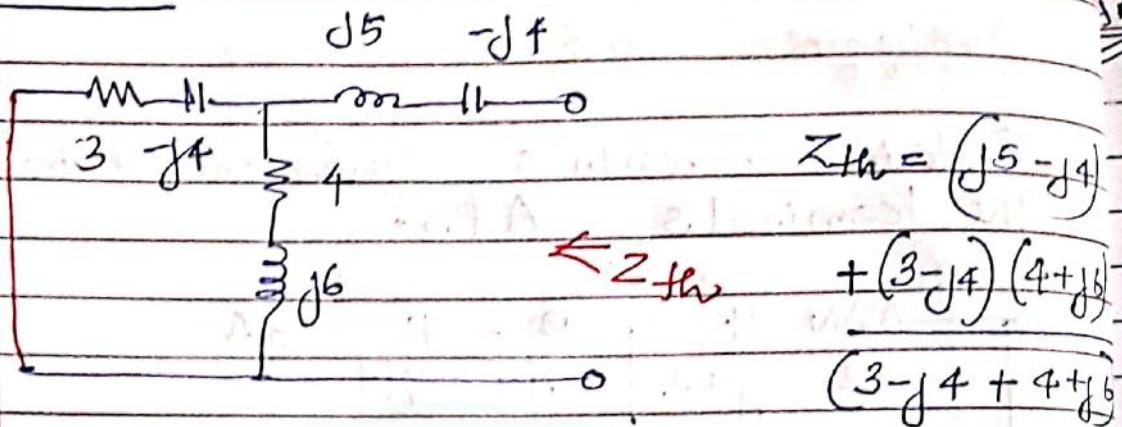
$$V_{th} = 6.87 \angle -15.95^\circ \times (4 + j6)$$

$$= 6.87 \angle -15.95^\circ \times 7.21 \angle 56.31^\circ$$

$$= 49.53 \angle 40.36^\circ$$

(53)

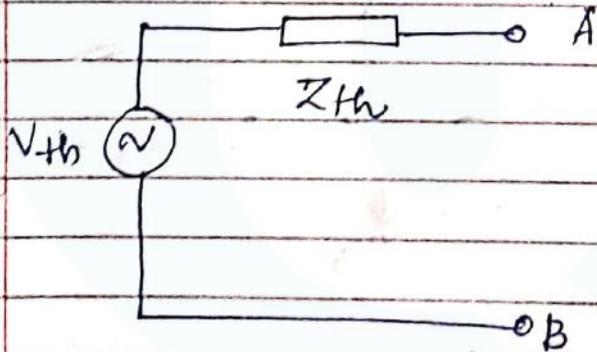
To find  $Z_{th}$



$$= j1 + \frac{(5 \angle -53.13^\circ)(7.2 \angle 56.3^\circ)}{(7 + j2)} \rightarrow 7.28 \angle 15.44^\circ$$

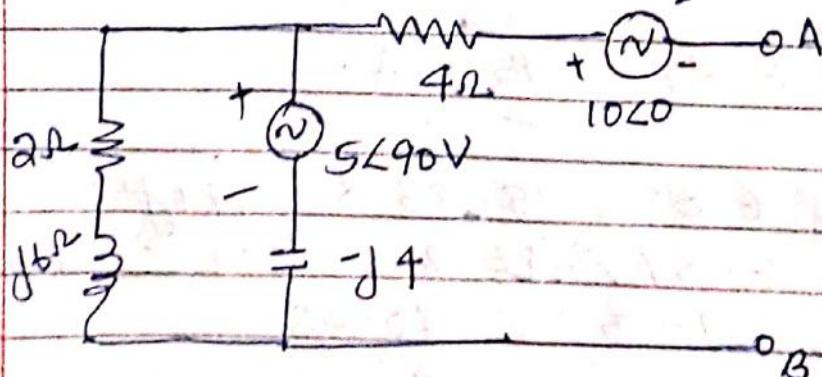
$$\begin{aligned} &= j1 + 4.95 \angle -12.27^\circ \\ &= j1 + 4.837 - 1.05j \end{aligned} \quad = 4.83 \angle -1.13^\circ$$

Thevenin's equivalent ckt



②

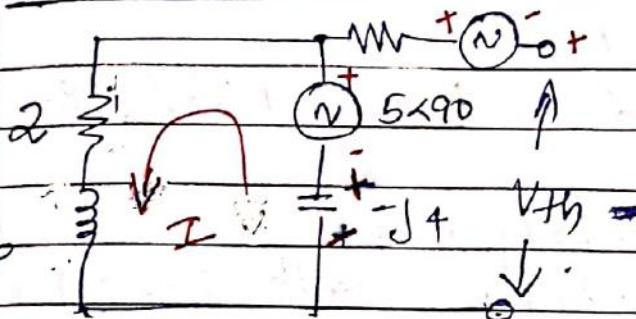
Obtain Thevenin's equivalent ckt



6A.

~~Ans~~Find  $V_{Th}$ 

10∠0°



$$-\bar{j}4I + 5\angle 90^\circ - 2I$$

$$-\bar{j}6I = 0$$

$$5\angle 90^\circ - 2I - \bar{j}2I = 0$$

$$I = \frac{5\angle 90^\circ}{2 + j2}$$

$$= 1.77 \angle 45^\circ A$$

$$V_{Th} = 5\angle 90^\circ - \bar{j}4I + 10\angle 0^\circ = 0$$

$$V_{Th} = 10\angle 0^\circ + 5\angle 90^\circ - 4jI$$

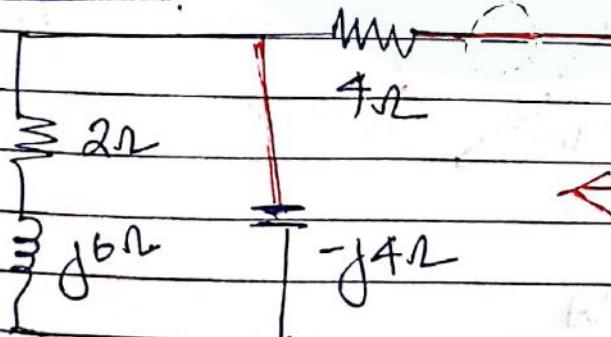
$$= -10 + 5j + 5j - 4j [1.77 \angle 45^\circ]$$

$$= -10 + 5j + 4\angle 90^\circ (1.77 \angle 45^\circ)$$

$$= -10 + 5j + 7.08 \angle 135^\circ$$

$$= -10 + 5j - 5.01 + 5.01j$$

$$= -15.01 + 10.01j = 18.04 \angle 146.31^\circ$$

To find  $Z_{Th}$ 

$$\leftarrow Z_{Th} = \left[ \frac{(2+j6)}{-j4} \right] + 4$$

$$= \frac{(2+j6) + j4}{2+j6 - j4} + 4 = \frac{-8j + 24}{2+j2} + 4$$

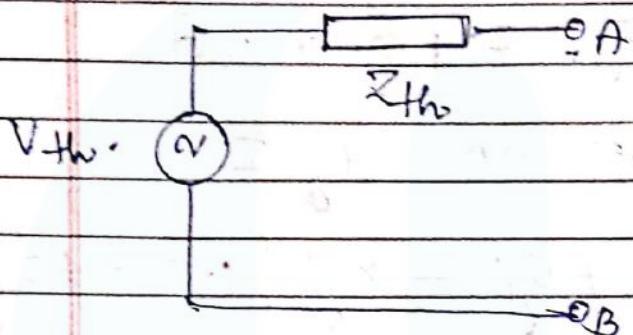
$$= 25.29 < 93.9$$

$$\frac{2.83 < 45}{+ 4} = 8.936 < 98.9 + 4$$

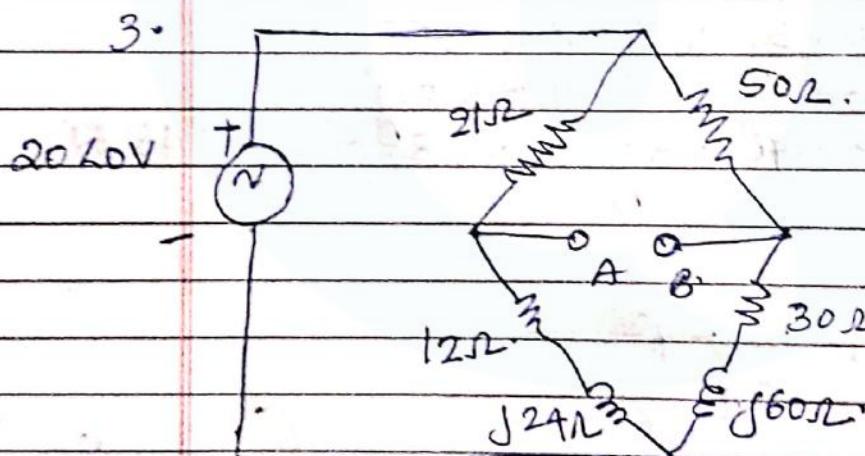
$$= 5.874 + 6.734j + 4$$

$$= 9.874 + 6.734j = 11.95 < -44.93$$

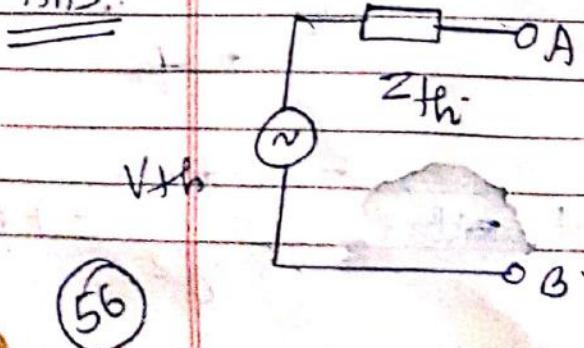
Thevenin's equivalent ckt:



Assignment Qn 4

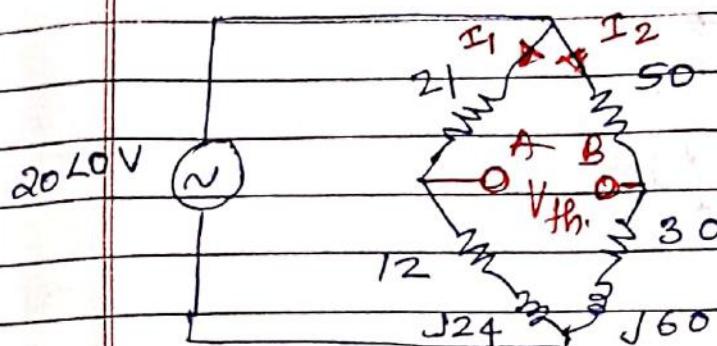


Ans.



So find  $V_{Th}$  &  $Z_{Th}$

To find  $V_{th}$ .



$$21, (12+j24) - \text{in series.}$$

$$50, (30+j60) - \text{"}$$

These 2 branches are parallel.

So  $I_1, I_2 \rightarrow 2$  currents.

$$I_1 = \frac{20\angle 0}{21+12+j24} = \frac{20\angle 0}{33+j24} = \frac{20\angle 0}{40.8\angle 36.02}$$

$$= 0.49 \angle -36.02 \text{ A}$$

$$I_2 = \frac{20\angle 0}{50+30+j60} = \frac{20\angle 0}{80+j60} = \frac{20\angle 0}{100\angle 36.86}$$

$$= 0.2 \angle -36.86$$

$$V_{th} = 21I_1 - 50I_2$$

$$= 21 \times 0.49 \angle -36.02 - 50 \times 0.2 \angle -36.86$$

$$= 10.29 \angle -36.02 - 10 \angle -36.86$$

$$= 8.322 - 6.05j - 8 + 6j$$

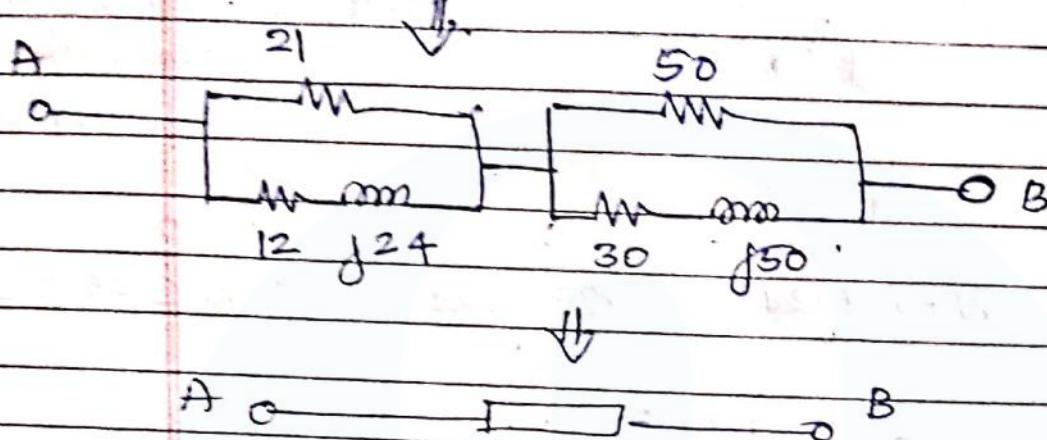
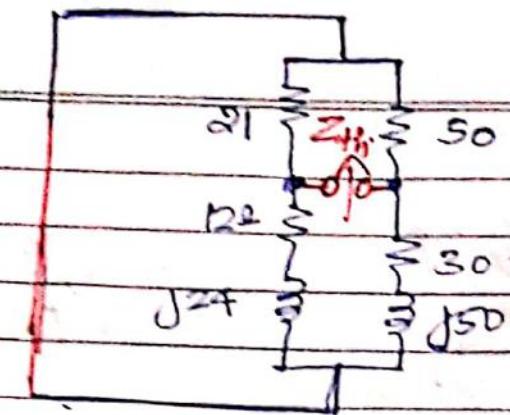
$$= 0.322 - 0.65j = 0.33 \angle 171.12 \text{ V}$$

OR

$$V_{th} = (12+j24) I_1 - (30+j60) I_2$$

$$= 0.33 \angle 171.12 \text{ V}$$

To find  $Z_{th}$ . Shoot circuit  $20\angle 0 \text{ V}$



$$Z_{th} = \frac{21}{(12+j24)} + \frac{50}{(30+j50)}$$

$$= \frac{21(12+j24)}{21+12+j24} + \frac{50 \times (30+j50)}{50+30+j50}$$

$$= \frac{252+j504}{33+j24} + \frac{1500+j2500}{50+j50}$$

$$= 563.49 \angle 61.1^\circ$$

$$= 40.80 \angle 36.03^\circ + 2915.48 \angle 59.04^\circ$$

$$= 13.811 \angle 25.07^\circ + 30.9 \angle 27.03^\circ$$

$$= 12.51 + 5.852j + 27.524 + 14.043j$$

$$= 40.034 + 19.895j$$

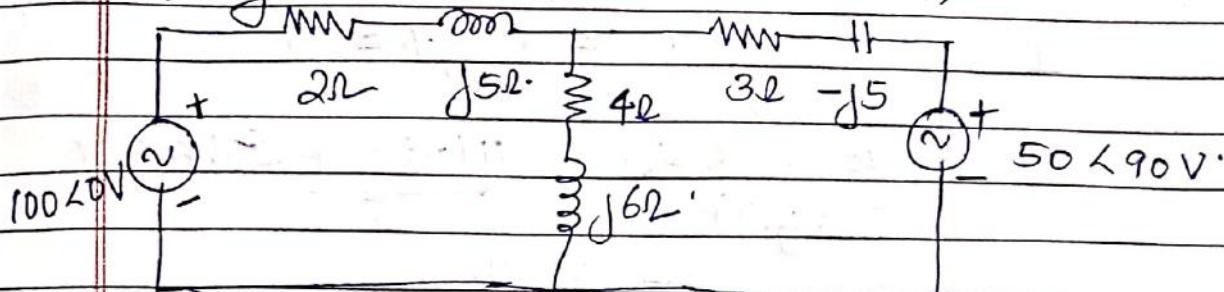
$$= 44.710 \angle 26.8^\circ$$

58

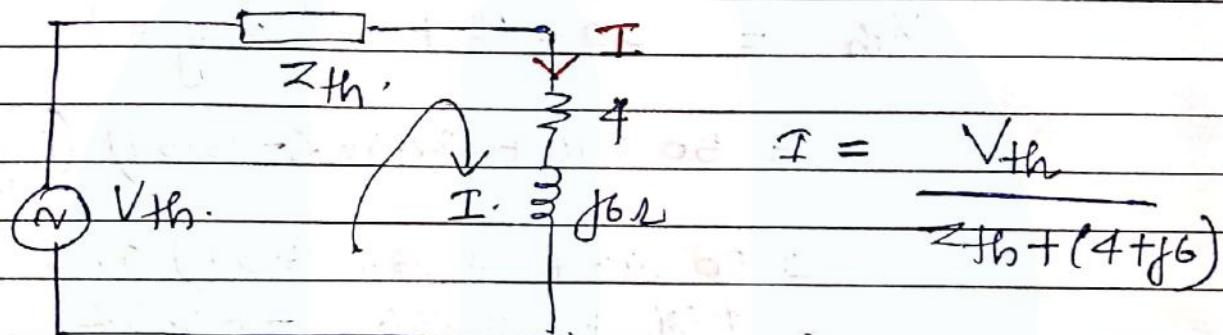
Assignment Qn. 5

Find the current through.

$(4+j6)\Omega$  impedance is the network.  
Using Thevenin's theorem.

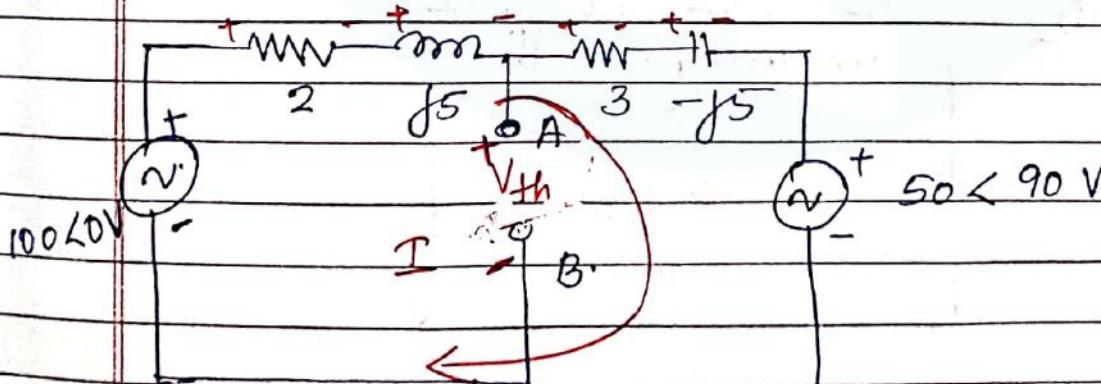


Ans Thevenin's equivalent circuit.



So find out  $V_{th}$  &  $Z_{th}$ .

To find  $V_{th}$ .



Applying KVL to the mesh,

$$100∠0 - 2I - j5I - 3I - j5I - 50∠90 = 0$$

$$100∠0 - 50∠90 = 5I$$

$$\frac{100 \angle 0 - 50 \angle 90}{5} = I$$

$$I = \frac{100 - j50}{5} = 20 + j10 \angle 20^\circ$$

$$= 40 \angle 25^\circ \text{ A} \quad 111.8034 \angle 26.57^\circ$$

$$= 22.86 \angle -26.57^\circ$$

$$V_{th} - 3I + j5I - 50 \angle 90 = 0$$

Ans

$$V_{th} = 50 \angle 90 + I(3 - j5)$$

$$= 50 \angle 90 + (22.86 \angle -26.57) (3 - j5)$$

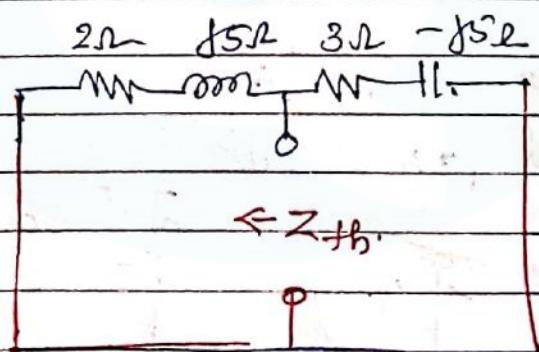
$$= 50 \angle 90 + (22.86 \angle -26.57) 5.83 \angle -59^\circ$$

$$= +50j + 130.36 \angle -85.57^\circ$$

$$= 50j + 10.069 - 129.97j$$

$$= 10.069 - 79.97j$$

$$= 79.33 \angle -82.88^\circ V$$

To find  $Z_{th}$ 

$$Z_{th} = \frac{(2+j5)(3-j5)}{2+j5+3-j5} = 6.28 \angle 9.16^\circ$$

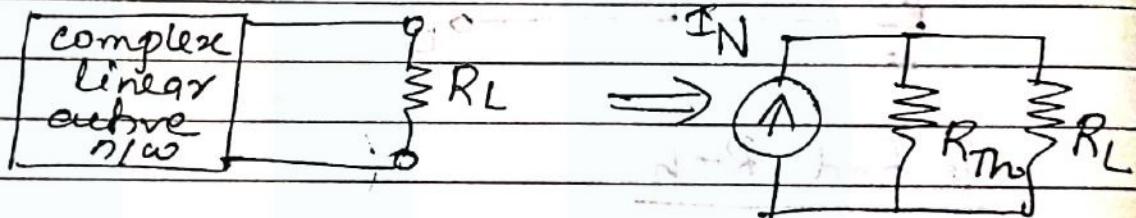
(6)

$$I_{through (4+j6)} = \frac{V_{th}}{Z_{th} + (4+j6)} = 6.52 \angle -117.61^\circ$$

### 3. Norton's theorem.

Any two terminals of a n/w containing linear, passive and active elements may be replaced by an equivalent current source  $I_N$  in parallel with a resistance  $R_{Th}$ , where  $I_N$  is the current flowing through a short circuit placed across the terminals AB.

$R_{Th}$  = equivalent resistance of the n/w as seen from the two terminals with all independent sources suppressed.



To find  $I_N$

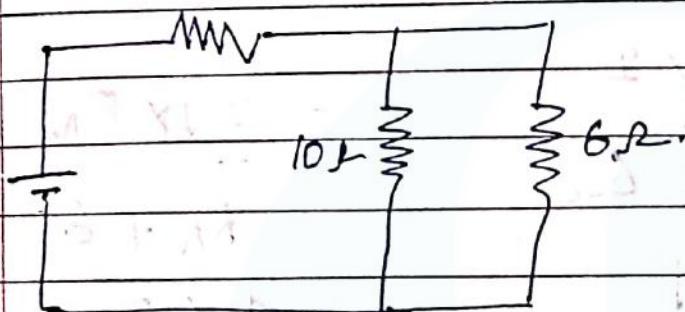
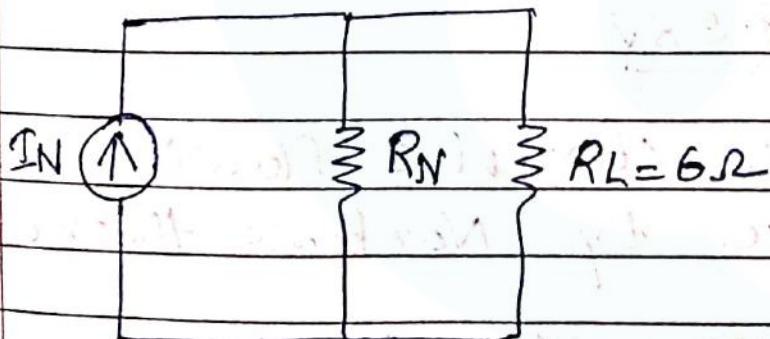
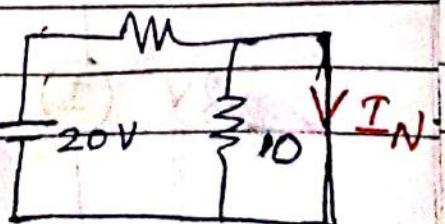
Short circuit the load terminal and then measure the current passing through the load terminal that gives short circuit current  $I_N$  at AB.

To find  $R_{Th}$

- ① Remove  $R_L$
- ② Short circuit the voltage source
- ③ Open circuit the current source
- ④ Then measure the resistance at the load terminal that gives  $R_{Th}$

Nothon's theorem

i) Find the voltage across the  $6\Omega$  resistor using Nothon's theorem.

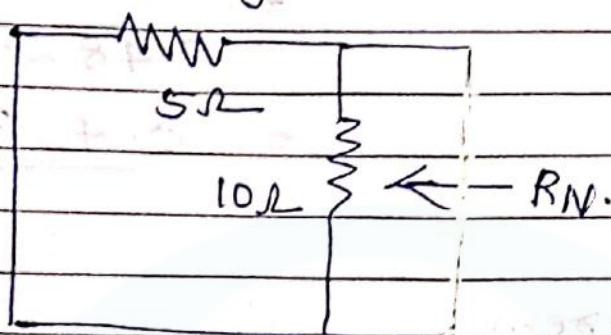
 $5\Omega$ AnsNothon's equivalent cktTo find  $I_N$ .Short Circuit  $R_L$ 

$$I_N = \frac{20}{5} = 4A$$

To find  $R_N$

Remove  $R_L$ , shorted out.

Voltage source shorted.



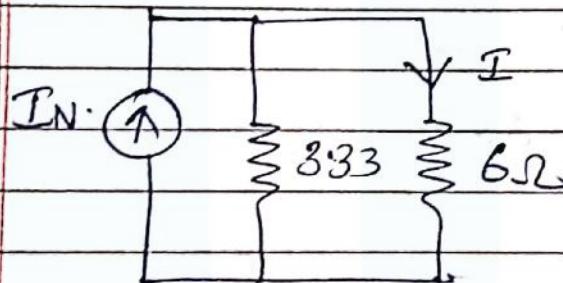
$$R_N = 5 // 10$$

$$= 5 \times 10$$

$$5 + 10$$

$$= 3.33\Omega$$

Nothor's equivalent circuit



$$I = I_N \times R_N$$

$$R_N + 6$$

$$= 4 \times 3.33$$

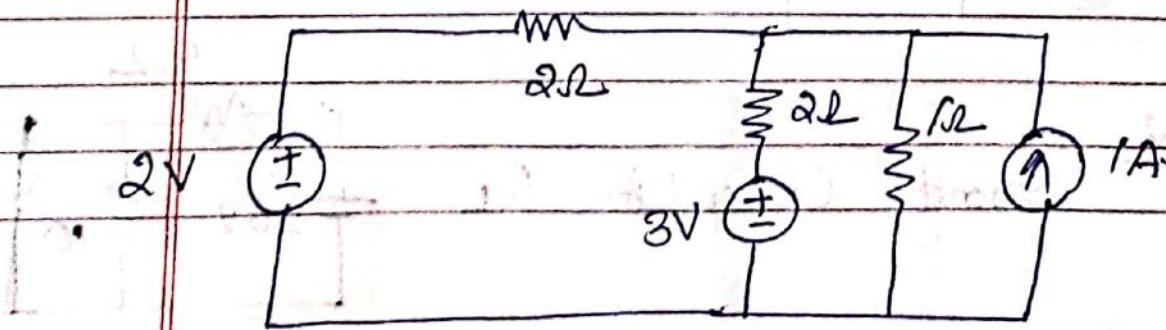
$$3.33 + 6$$

$$= 1.43A$$

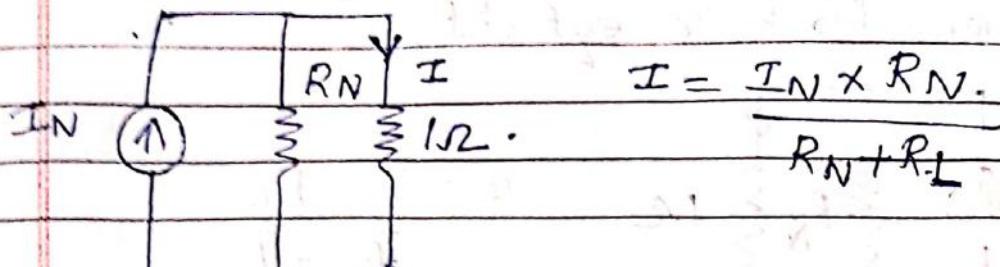
$$V_{6\Omega} = 1.43 \times 6$$

$$= 8.58V$$

- (2) Evaluate the current I flowing through  $1\Omega$  resistance by Norton's theorem

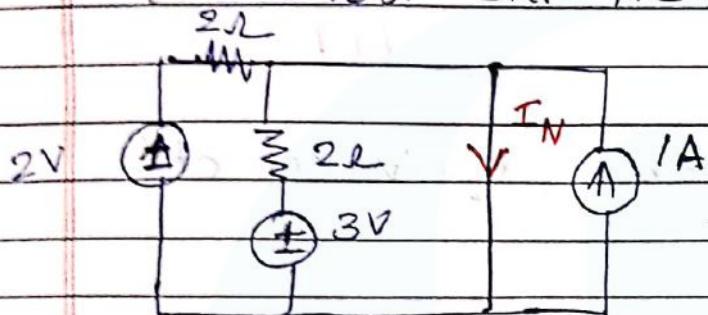


Noether's eq.t ckt

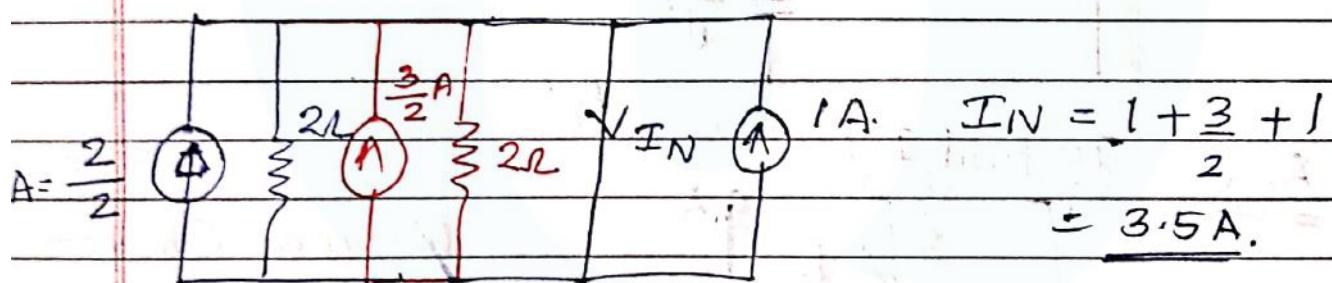


To find  $I_N$

Short ckt  $R_L$ .

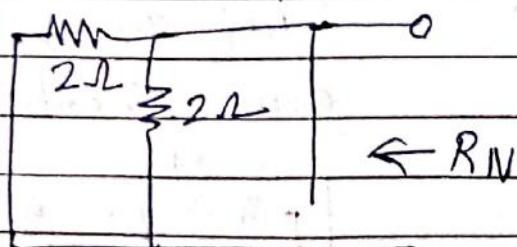


We can convert 2 voltage sources to current source.

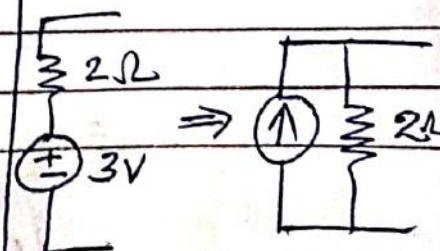
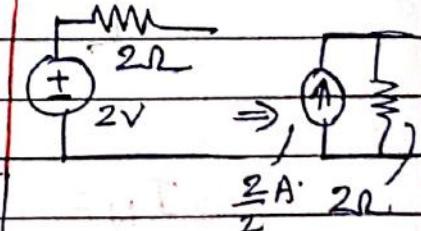


To find  $R_N$

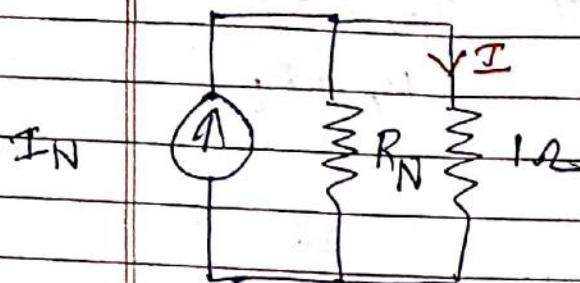
Using source transformation



$$R_N = 2//2 = \frac{2 \times 2}{2+2} = 1\Omega$$



From Norton's eqt ckt,

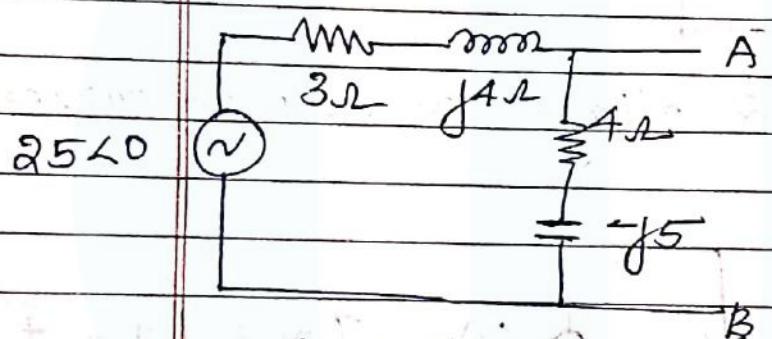


$$I = \frac{IN}{RN + I}$$

$$= \frac{3.5 \times 1}{1+1} = 1.75A$$

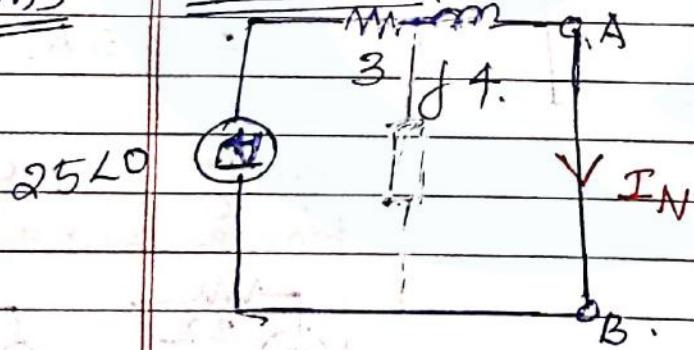
~~AC Analysis~~

- ③ Obtain Norton's equivalent ckt across AB.



Ans

To find  $I_N$



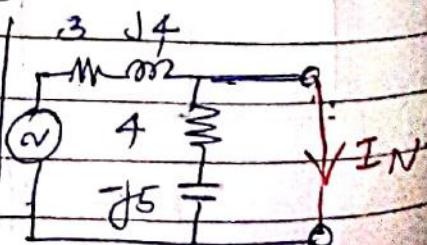
When a short circuit is placed across  $(4-j5)\Omega$

impedance, it gets shorted

$$I_N = \frac{25 \angle 0}{3+j4}$$

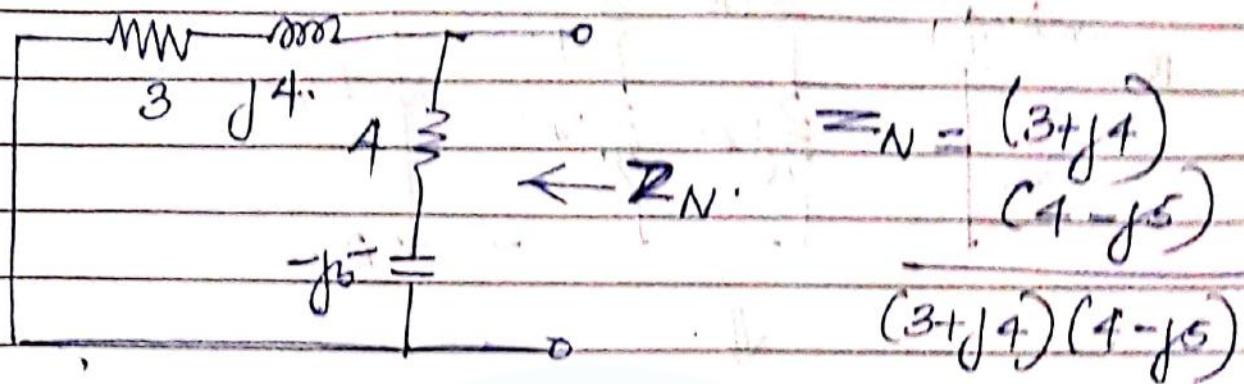
$$= \frac{25 \angle 0}{5 \angle 53.13}$$

$$= 5 \angle -53.13 A.$$



To find  $Z_N$ .

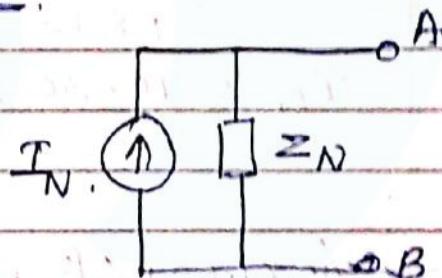
Remove load.



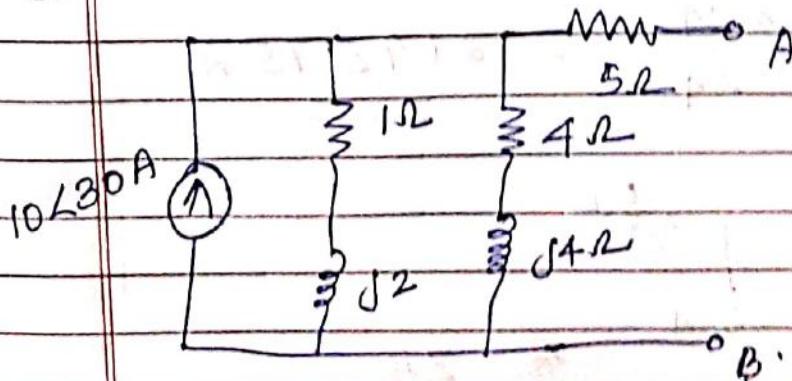
$$Z_N = \frac{(5 \angle 53.13^\circ)(6.4 \angle -51.1^\circ)}{(7 - j1)}$$

$$= \frac{5 \times 6.4}{7.07} \angle (53.13^\circ + 81.3^\circ)$$

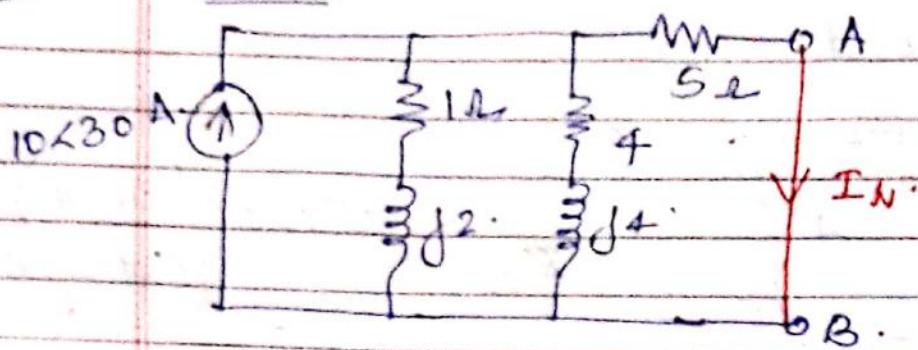
$$= 4.53 \angle 9.92^\circ$$

Nofton's eqt ckt

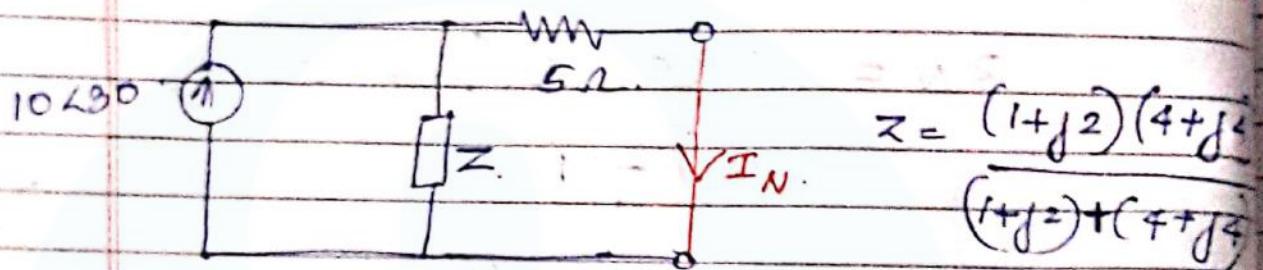
(4) Obtain Norton's eqt ckt. across A B.



Ans Find  $I_N$  - Short cut AB.



↓.



$$= (2.236 \angle 63.43^\circ) (5.657 \angle 45^\circ)$$

$$Z_N = \frac{12.649 \angle 108.43^\circ}{7.81 \angle 50.19^\circ} = 1.6196 \angle 58.24^\circ$$

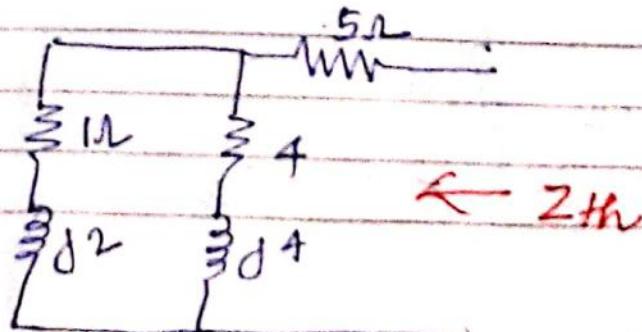
$$I_N = 10 \angle 30^\circ \times 1.62 \angle 58.24^\circ$$

$$1.62 \angle 58.24^\circ + 5$$

$$= \frac{(10 \times 1.62) \angle (30 + 58.24)}{0.8527 + 1.377j + 5} = \frac{16.2 \angle 88.24}{5.8527 + 1.377j}$$

$$= \frac{16.2 \angle 88.24}{6.011 \angle 13.24^\circ} = 2.69 \angle 75^\circ$$

To find  $Z_N$



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Read No.

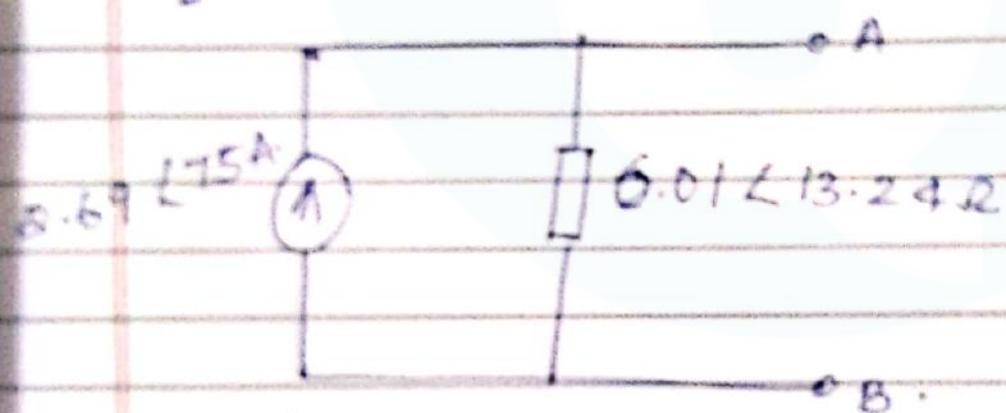
Date / /

$$Z_N = 5 + \frac{(1+j^2)(4+j^4)}{(4j^2) + (4+j^4)}, \quad 5 + j \cdot 6.196158 \cdot 2$$

$$= 5 + 0.852 + 1.372j = 5.852 + j1.372$$

$$= \underline{6.01 \angle 13.24^\circ}$$

Eg &amp; Ckt.



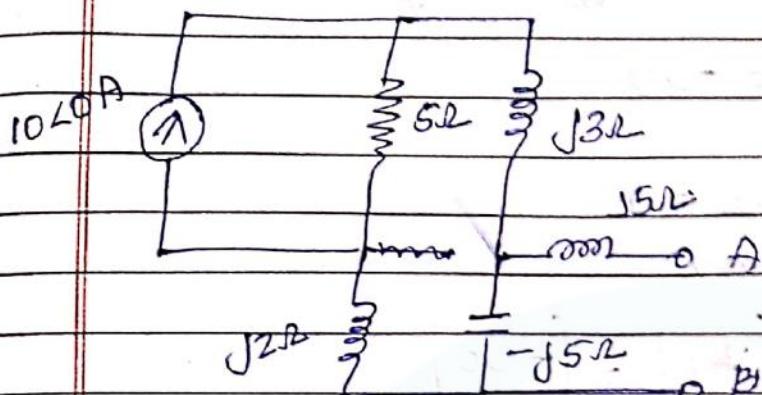
Assignment Qn: 7

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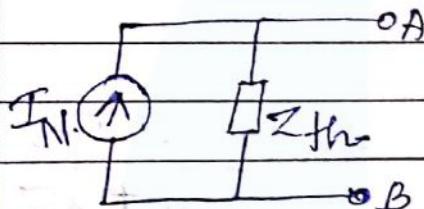
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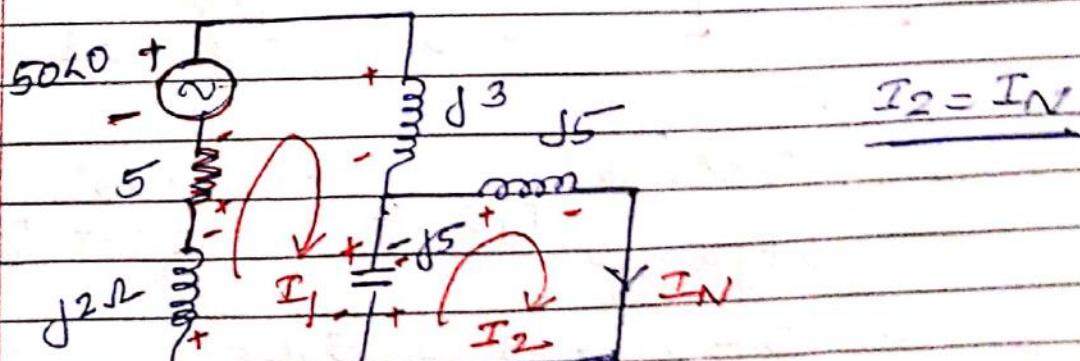
Obtain the Norton's equivalent ckt.

Ans.

Norton's equivalent ckt.

So find out  $I_N$  &  $Z_{th}$ To find  $I_N$ 

By Source transformation, the n/w can be redrawn as,



~~Explain meta analysis.~~

$$-422_1 - 52_1 - 50 - j32_1 - j5(7_1 - 7_2) = 0 =$$

$$-52_1 - j52_2 = -90$$

$$-j52_2 - 2 = -52_2 = 0$$

In matrix form

$$\begin{bmatrix} -5 & 75 & I_1 \\ -75 & 0 & I_2 \end{bmatrix} = \begin{bmatrix} -50 \\ -90 \end{bmatrix}$$

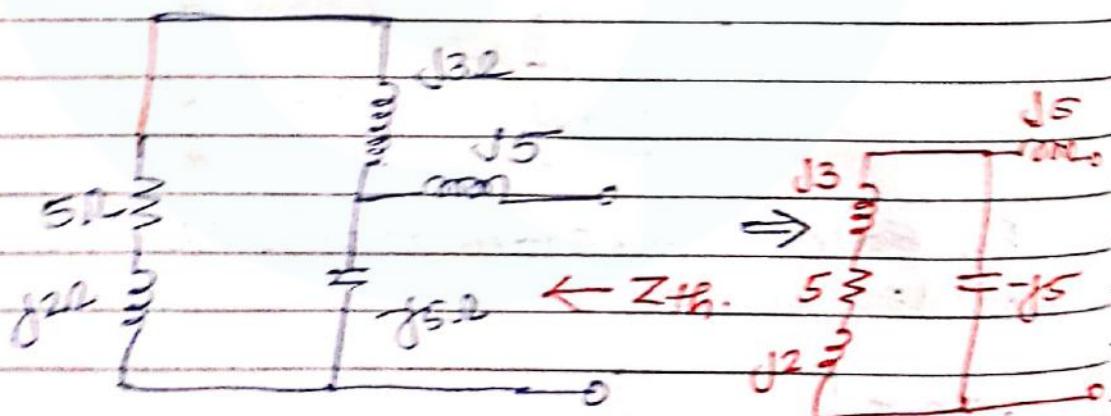
$$\begin{bmatrix} -5 & 75 & I_1 \\ -75 & 0 & I_2 \end{bmatrix} = \begin{bmatrix} -50 \\ -90 \end{bmatrix}$$

$$I_2 = \begin{bmatrix} -5 & -50 \\ -75 & 0 \end{bmatrix} = \frac{-50}{+75} = -\underline{\underline{50/-75}}$$

$$\begin{bmatrix} -5 & -50 \\ -75 & 0 \end{bmatrix} = \frac{-50}{+75} = +\underline{\underline{75}}$$

$$\begin{bmatrix} -5 & -50 \\ -75 & 0 \end{bmatrix} = \frac{-50}{+75} = -\underline{\underline{50}} = 10L - \underline{\underline{90}}$$

To find  $Z_{th}$



$$Z_{th} = \frac{(5+j5)(-j5)}{5+j5-j5} + j5 = \frac{-j25+25}{5} + j5$$

$$= 5 - 5j + j5 = \underline{\underline{5j}}$$

(62)

# A. Maximum Power Transfer

TERMINAL  
BLOCKS

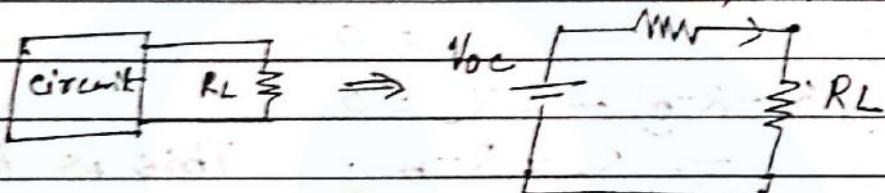
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Q5

In dc circuits, maximum power is transferred from a source to the load when the load  $Z_L$  is made equal to the resistance of the  $h_{11}$  as viewed from the load terminals, with load removed and all the sources replaced by their internal resistances.

By Thévenin's theorem,

$R_{Th}$ ,  $I_1$



Here,  $V_{oc}$  = open circuit voltage.

$R_{Th}$  =  $R$  measured b/w terminals AB

with  $R_L$  removed and sources replaced by their internal resistance

Proof

The current supplied to  $R_L$

$$I_L = \frac{V_{oc}}{R_{Th} + R_L}$$

$$\begin{aligned} \text{Power delivered to } R_L &= P_L = I_L^2 R_L \\ &= \left( \frac{V_{oc}}{(R_{Th} + R_L)} \right)^2 R_L \quad (1) \end{aligned}$$

For a given ckt,  $V_{oc}$  &  $R_{Th}$  are constant.

$\therefore$  Power delivered to load depends on  $R_L$ .

In order to find the value of  $R_L$  for which the value of  $P_L$  is maximum, differentiate eqn (1) w.r.t to  $R_L$ .

(9)

$$\frac{dP_L}{dR_L} = 0$$

$$\frac{d}{dR_L} \left( \frac{V_{oc}}{(R_{Th} + R_L)}^2 \right) = 0$$

$$\frac{V_{oc}^2}{dR_L} \left( \frac{R_L}{(R_{Th} + R_L)^2} \right) = V_{oc}^2 \left[ \frac{(R_{Th} + R_L)^2}{-R_L^2(R_{Th} + R_L)} \right] \\ (R_{Th} + R_L)^4$$

$$(R_L + R_{Th}) - 2R_L = 0 \\ \Rightarrow R_L = R_{Th} . \text{ This is the proof.}$$

$$\therefore P_{max} = \frac{V_{oc}^2}{(R_{Th} + R_L)} R_L = \frac{V_{oc}^2}{(R_{Th} + R_{Th})^2} R_{Th}$$

$$\frac{V_{oc}^2}{4R_{Th}^2} R_{Th} = \frac{V_{oc}^2}{4R_{Th}}$$

This theorem is used to find the value of load resistance for which there would be maximum amount of power transfer from source to load.

Under the condition of maximum power transfer, the  $\eta$  is only 50%, as only one half of the total power generated is dissipated in the internal resistance  $R_{Th}$  of the cell.

i.e:

$$\eta = \frac{O/P}{I/P} = \frac{I^2 R_L}{I^2 (R_L + R_{Th})}$$

(10)

At max. power transfer condition,

$$R_L = R_{Th}$$

$$\eta = \frac{I^2 R_L}{2 I^2 R_L} = \frac{1}{2} = 50\%$$

Load voltage

is one half of the open circuited voltage at the load terminals.

$$\text{Load voltage} = \frac{V_{oc}}{(R_L + R_{Th})} \cdot R_L \quad (\text{using voltage division rule})$$

$$= \frac{V_{oc}}{2R_L} \cdot R_L$$

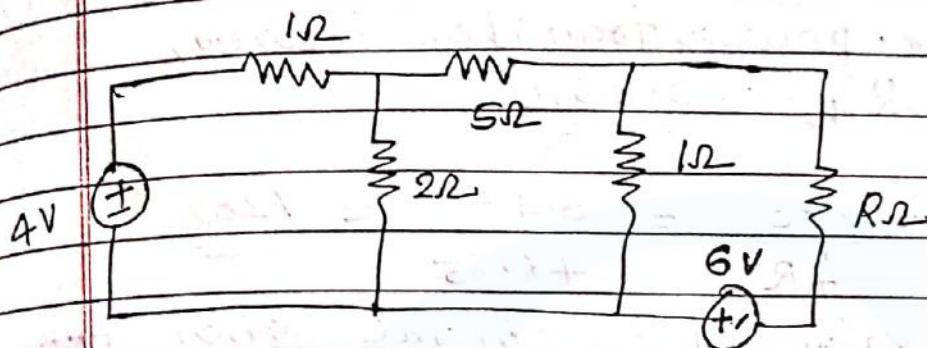
$$= \frac{V_{oc}}{2}$$

Steps

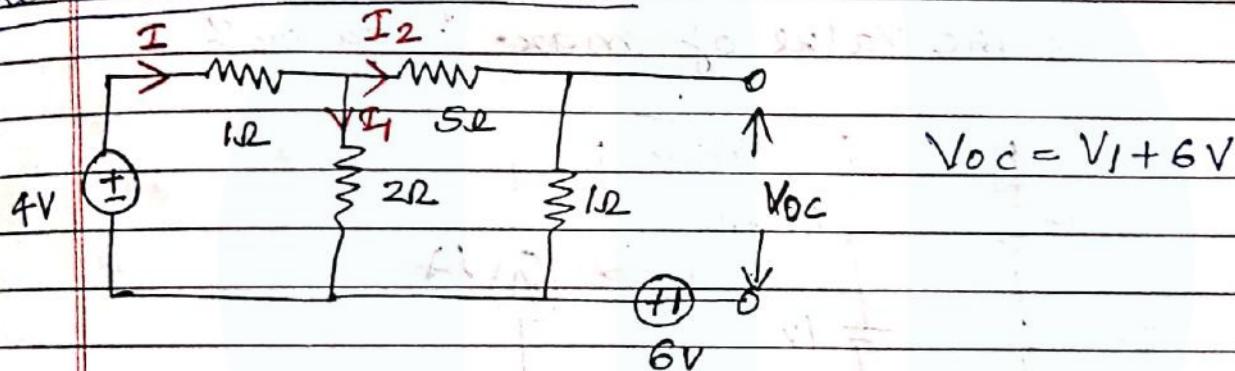
- ① Remove the load resistor  $R_L$  and find  $R_{Th}$  of the source network looking through the open circuited load terminals.
- ② As per max. power transfer theorem, this  $R_{Th}$  is the load resistance of the no ie  $R_L = R_{Th}$  that allows max. power transfer.
- ③ Find the thvenin's voltage  $V_o$  across the open circuited load terminals.
- ④ Max. power transfer,  $V_o^2 / 4R_{Th}$ .

Maximum power transfer

- (3) Find the value of  $R$ , such that maximum power transfer takes place. What is the amount of this power?



Remove the load resistor,  $R$



$$R = \frac{6 \times 2}{8} + 1 = \frac{12}{8} + 1 = \frac{20}{8} \Omega.$$

$$I = \frac{4}{R} = \frac{4}{\frac{20}{8}} = \frac{4 \times 8}{20} = \underline{\underline{\frac{8}{5}}} \text{ A}$$

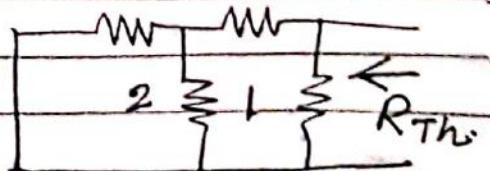
$$I_2 = \frac{\frac{8}{5} \times 2}{2+5+1} = \frac{16}{5} \div 8 = \frac{16}{5} \times \frac{1}{8} = \underline{\underline{\frac{2}{5}}} \text{ A}$$

$$V_{1\Omega} = \frac{2}{5} \times 1 = \frac{2}{5} \text{ V.}$$

$$V_{oc} = \frac{2}{5} + 6 = \frac{32}{5} = \underline{\underline{6.4}} \text{ V}$$

To find  $R_{th}$

$$R_{th} = (2 \parallel 1 + 5) \parallel 1$$



$$= \left( \frac{2+1}{3} + 5 \right) // 1 = \left( \frac{2}{3} + 5 \right) // 1 = \frac{17}{3} \times 1 \\ = 0.85 \Omega$$

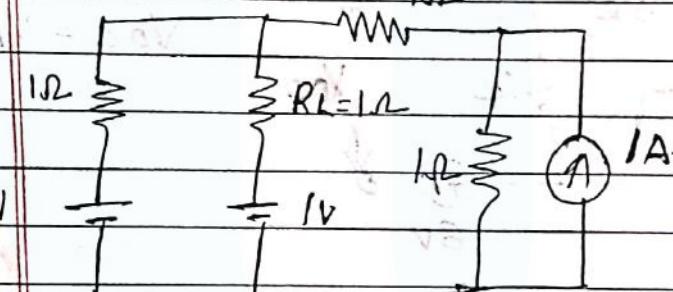
By max. power transfer theorem,

$$R = R_{Th} = 0.85 \Omega$$

$$P_{max} = \frac{V_{oc}^2}{4R} = \frac{6.4^2}{4 \times 0.85} = 1200$$

Q2) What value of  $R$  in the given ckt will receive the max. power? what is the value of max. power?

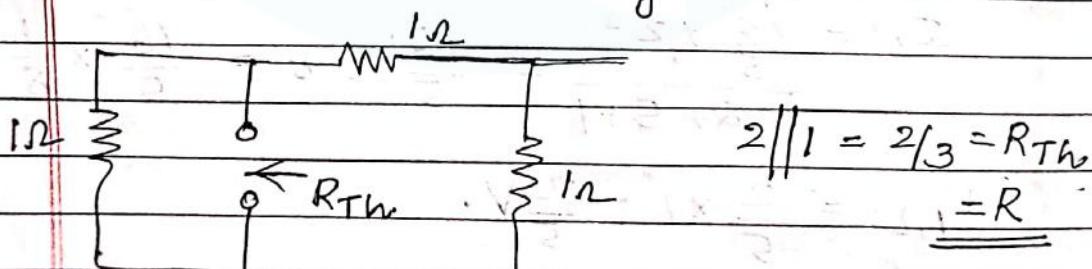
10...



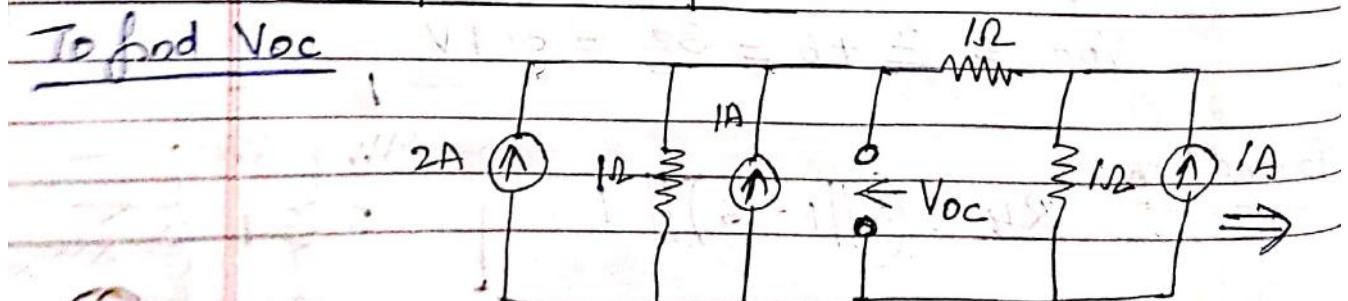
Ans.

To find  $R_{Th}$

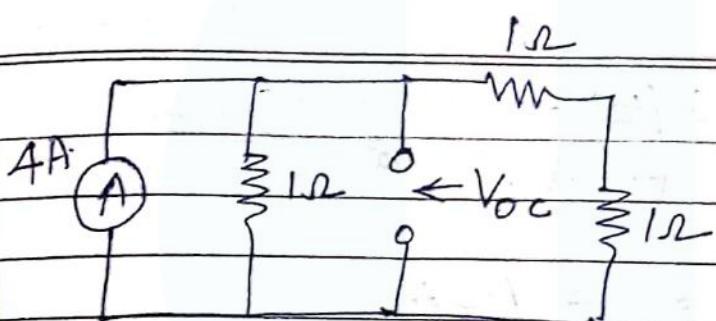
Remove  $R_L$ , open ckt current source & short ckt voltage source.



To find  $V_{oc}$



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$$= \frac{4 \times 2}{3} \times 1$$

$$V_{oc} = \frac{8}{3} \times 1 \text{ V}$$

$$P_{max} = \frac{V_{oc}^2}{4R_{Th}} = \frac{(8/3)^2}{4 \times 2/3}$$

Reciprocity

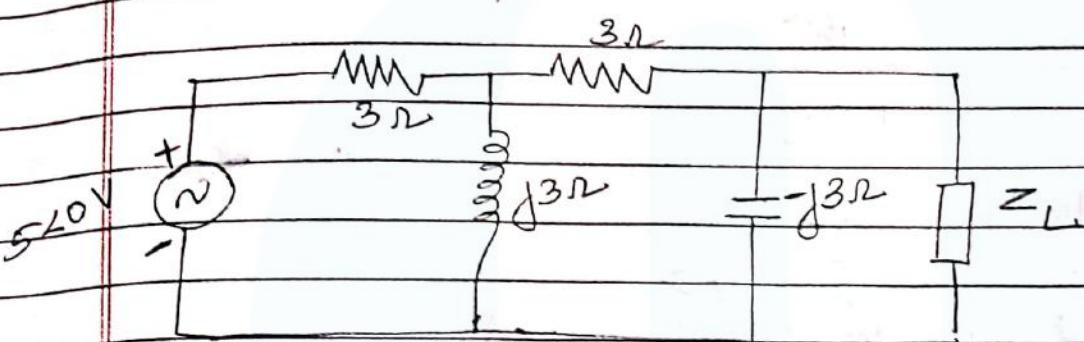
Maximum power transfer

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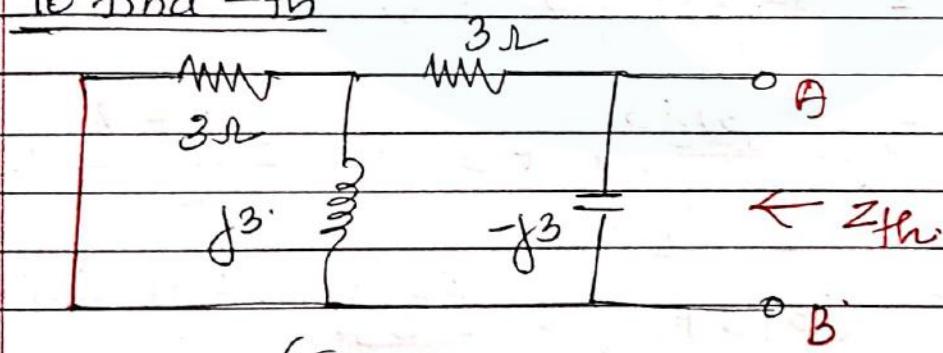
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- ③ Find the impedance  $Z_L$  so that maximum power can be transferred to  $i_L$ . Find maximum power.



Ans By maximum power transfer theorem,  $Z_L$  should be the complex conjugate of the source impedance. ( $Z_{th}$ )

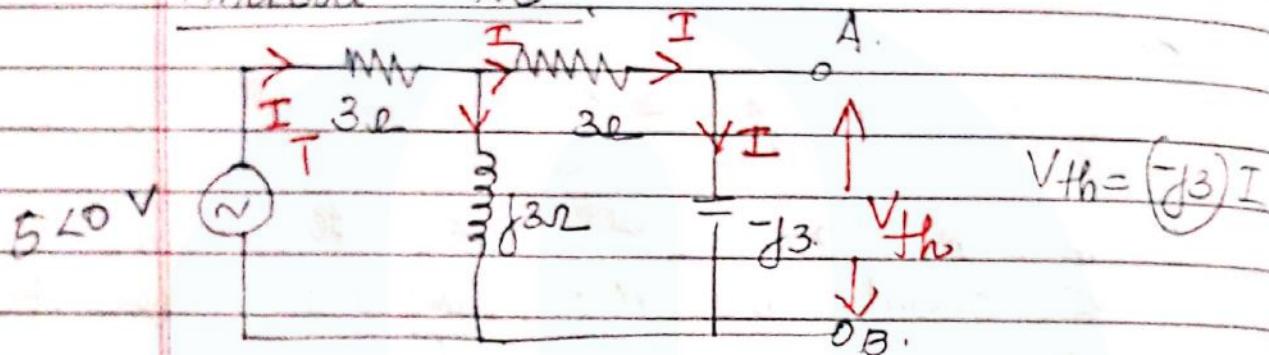
To find  $Z_{th}$



$$\begin{aligned}
 &= -\frac{(1+j1)}{3+j4 - j3(1+j1)} = \frac{12-j9}{(6+j1)} = \frac{15x-36-8j}{6+08+9+j1} \\
 &= \frac{15x-36-8j}{6+08+9+j1}
 \end{aligned}$$

To find  $P_{max} = \frac{V_{TH}^2}{4R_L}$

Find out  $V_{TH}$ :



$$Z_{eq} = (3-j3) \parallel j3 + 3$$

$$= \frac{(3-j3)j3}{(3-j3)+j3} + 3 = \frac{j9+9}{3} + 3$$

$$= j3+3+3 = \underline{\underline{6+j3}} = 6.71 \angle 26.57^\circ$$

$$I_T = \frac{5∠0}{6+j3} = 0.75 \angle -26.57^\circ$$

$$I = I_T \times \frac{j3}{j3+3-j3} = \frac{(0.75 \angle -26.57^\circ) \angle 90^\circ}{3 \angle 0^\circ}$$

$$= 0.75 \angle -26.57^\circ 63.43 A$$

6A

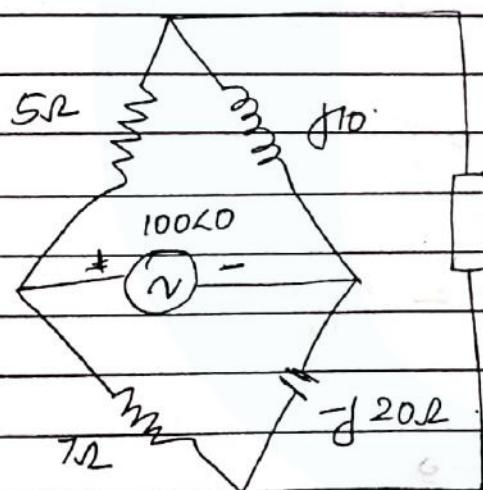
$$V_{th} = (-j3) (0.75 \angle 63.43^\circ) = 2.24 \angle -26.57^\circ V$$

$$P_{max} = \frac{|V_{th}|^2}{4R_L} = \frac{2.24^2}{4 \times 1.8} = 0.710$$

By max power transfer, load impedance should be equal to complex conjugate of source impedance.

$$Z_L = Z_{th}^* = 1.8 + j2.4 \Omega$$

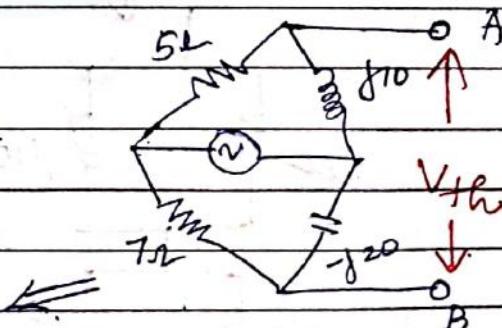
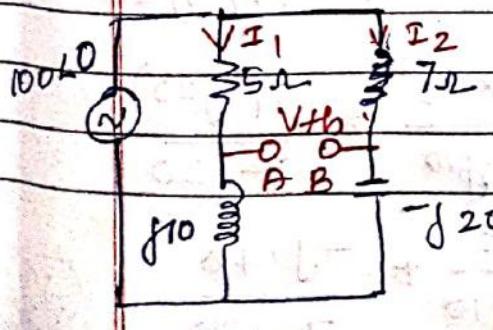
- (4) Find the value of  $Z_L$  for maximum power transfer and find max. power.



$Z_L$  for max. power transfer will be, complex conjugate of  $Z_{th}$ .

Ans. To find  $V_{th}$

Redrawing the circuit



$$I_1 = \frac{100 \angle 0}{5+j10} = 8.94 \angle -63.43^\circ$$

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$$I_2 = \frac{100 \angle 0}{7-j20} = 4.72 \angle 70.7^\circ$$

$$V_{th} = (V_A - V_B) = (8.94 \angle -63.43^\circ) j10 - (4.72 \angle 70.7^\circ) (-j20)$$

$$= (8.94 \angle -63.43^\circ) (10 \angle 90^\circ) + (4.72 \angle 70.7^\circ) (20 \angle +90^\circ)$$

$$= 89.4 \angle 26.57^\circ + 94.4 \angle 160.7^\circ$$

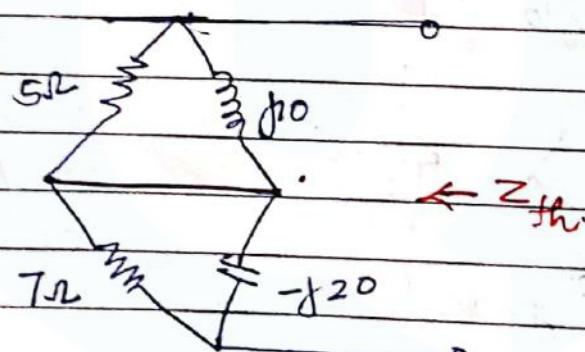
$$= 93.26 + j16.55 - 96.35 - j11.55$$

$$= -3.09 + j4.95$$

$$= 5.87 \angle 21.74^\circ$$

$$= 71.76 \angle 97.3^\circ$$

To find  $Z_{th}$



$$Z_{th} = 5 \parallel j10$$

$$= 5 \parallel j10 + 7 \parallel -j20$$

$$= \frac{5 \times j10}{5+j10} + \frac{7(-j20)}{7-j20}$$

$$= \frac{j50}{5+j10} + \frac{-j140}{7-j20}$$

(65)

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$$= \frac{50\angle 90}{11.18 \angle 63.43} + \frac{140\angle -90}{21.19 \angle -70.7} = (10.23 - j0.18)$$

For maximum power transfer, the  $Z_L$  should be complex conjugate of the source impedance.

$$\underline{Z_L} = 10.23 + j0.18 \cdot 1$$

$$P_{max} = \frac{V_{th}^2}{4R_L} = \frac{71.76^2}{4 \times 10.23} = \underline{125.84 \text{ V}}$$

## 5. Reciprocity Theorem

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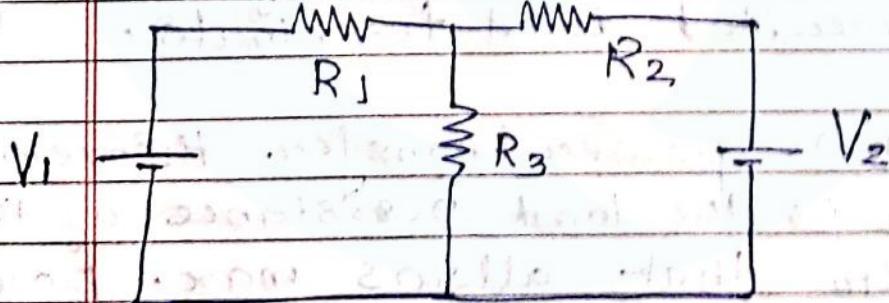
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In any linear network containing linear bilateral resistances & energy sources, the ratio of voltage 'V' introduced in one mesh to the current 'I' in any second mesh is the same as when obtained if the positions of V & I are interchanged other voltage sources are removed.

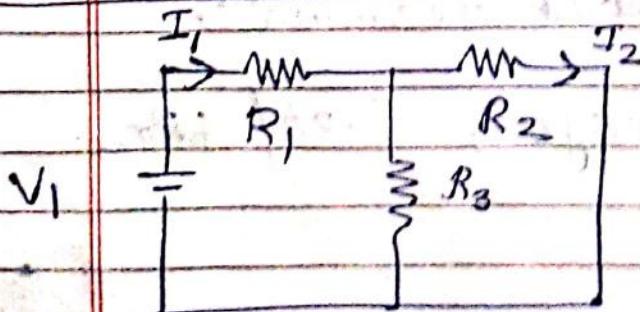
This theorem leads to a shunt feed reactance  $R_T$  where  $V$  is the voltage in one mesh &  $I$  is the current in any second mesh.

$$R_{T12} = R_T \cdot \frac{V_1}{I_2}$$

Proof Consider a network consists of resistors  $R_1, R_2$  &  $R_3$ .



To calculate  $R_{T12}$ ,  $V_2 = 0$



$$R_{T12} = V_1 / I_2$$

$$R_{eq} = R_1 + R_2 // R_3$$

(12)

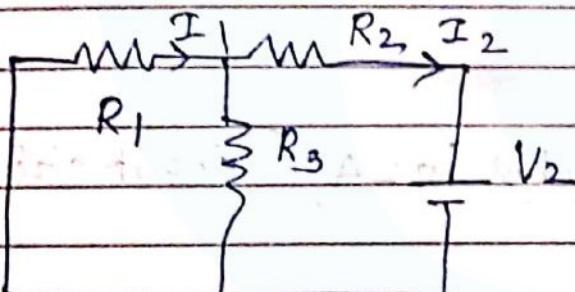
$$I_2 = I_1 \times \frac{R_3}{R_2 + R_3}$$

$$I_1 = \frac{V_1}{R_{eq}} = \frac{V_1}{R_1 + \frac{R_2 R_3}{(R_2 + R_3)}}$$

$$I_2 = \frac{V_1}{R_1 + \frac{R_2 R_3}{(R_2 + R_3)}} \times \frac{R_3}{(R_2 + R_3)} \\ = \frac{V_1 R_3}{R_1 R_2 + R_2 R_3 + R_1 R_3}$$

$$R_{T_{1,2}} = \frac{V_1}{I_2} = \frac{R_1 R_2 + R_2 R_3 + R_3 R_1}{R_3}$$

To calculate  $R_{T_{21}}$ ,  $V_1 = 0$



$$R_{eq} = R_1 \parallel (R_3 + R_2)$$

$$R_{T_{21}} = \frac{V_2}{I_1}$$

$$I_1 = -I_2 \times \frac{R_3}{R_2 + R_3} = -\left(\frac{V_2}{R_{eq}}\right) \frac{R_3}{R_1 + R_3}$$

$$= \frac{-V_2 R_3}{\frac{R_1 R_3}{R_1 + R_3} + R_2} = \frac{-V_2 R_3}{R_1 R_2 + R_2 R_3 + R_3 R_1}$$

$$R_{T_{21}} = \frac{R_1 R_2 + R_2 R_3 + R_3 R_1}{R_3} \quad | \text{ i.e. } \quad R_{T_{12}} = R_{T_{21}}$$

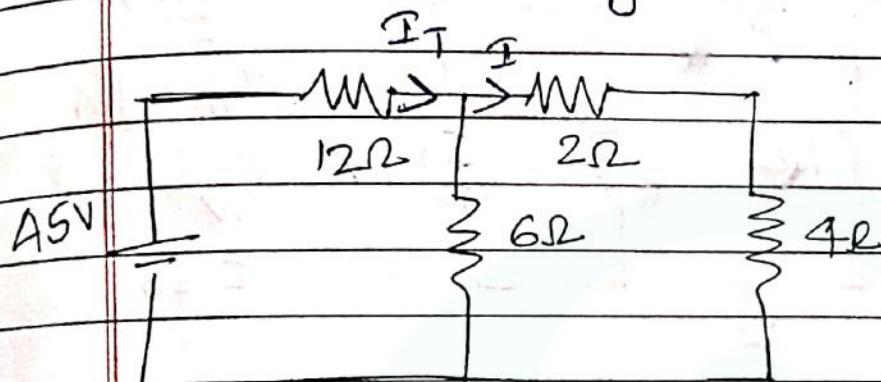
This is the proof

T &amp; Th

4x2/3

~~Reciprocity~~

q. For the net, find the value of current I and also verify the reciprocity theorem.

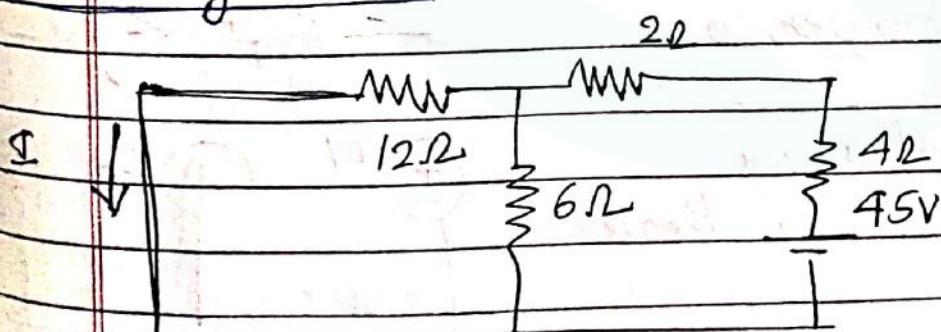


$$I = I_T \times \frac{6}{6+2+4}, \text{ Req} = [(2+4) \parallel 6] + 12$$

$$I_T = \frac{V}{\text{Req}} = \frac{45}{15} = 3A$$

$$I = 3 \times \frac{6}{12} = 1.5A$$

Interchange V & I



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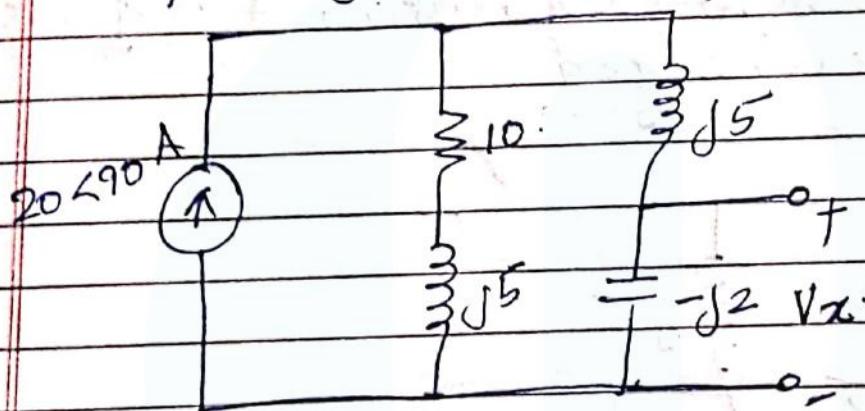
$$R_T = 6 + \left( 6 // 12 \right) = 6 + \frac{72}{18} = \underline{\underline{10\Omega}}$$

$$I_T = \frac{45}{10} = \underline{\underline{4.5A}}.$$

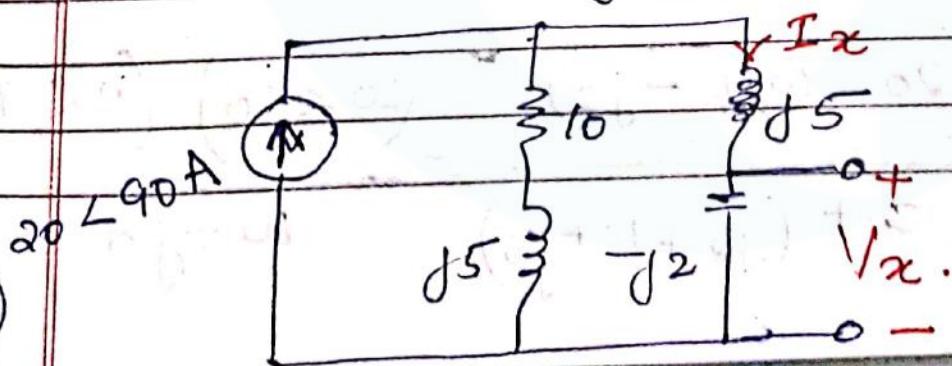
$$I = \frac{4.5 \times 6}{6 + 12} = \underline{\underline{1.5A}}$$

## Reciprocity theorem:

Find the voltage  $V_x$  & then verify reciprocity theorem.



~~Ans~~ Calculate  $V_x$  when excitation & response are interchanged.



Find  $I_x$ , by current division rule!

$$I_x = \frac{20 \angle 90^\circ \times (10 + j5)}{(10 + j5) + (j5 - j2)}$$

$$= (20 \angle 90^\circ) (11.18 \angle 26.57^\circ)$$

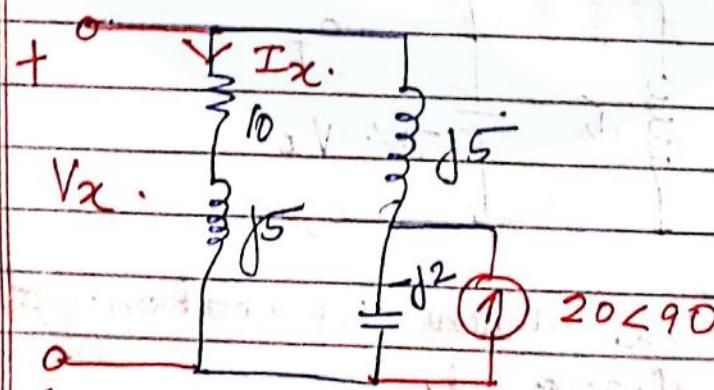
$$= \frac{20 \times 11.18}{12.81} \angle (90 + 26.57 - 38.66)^\circ$$

$$= 17.46 \angle 77.91^\circ A.$$

$$V_{xc} = (-j2) I_{xc} = (2 \angle -90^\circ) (17.46 \angle 77.91^\circ)$$

$$= 34.92 \angle -12.09^\circ V$$

Find  $V_{xc}$  when excitation & response are interchanged



$$I_x = \frac{(20 \angle 90^\circ)(-j2)}{-j2 + (j5 + 10 + j5)} = \frac{(20 \angle 90^\circ)(-j2)}{10 + j8}$$

(6x)

$$-j2 + (j5 + 10 + j5)$$

$$10 + j8$$

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$$= \frac{40\angle 0}{12.81\angle 38.66} = 3.12 \angle -38.66 \text{ A.}$$

$$V_x = I_x \times (10 + j5)$$

$$= (3.12 \angle -38.66) (11.18 \angle 26.57)$$

$$= 34.8816 \angle -12.09 \text{ V}$$

Since,  $V_x$  is same in both cases,  
reciprocity theorem is verified.