

# Chapter 16 二端口网络

## Two-port Networks

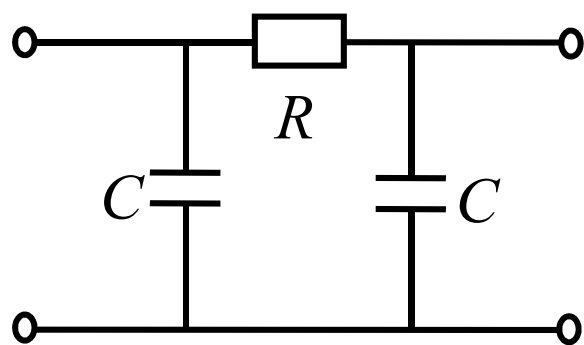
- 16.1 二端口网络的特性 Characteristics of two-ports
- 16.2 二端口网络的参数 Parameters of two-ports
- 16.3 参数之间的关系 Relationships between parameters
- 16.4 二端口网络的等效电路 Equivalent circuits
- 16.5 二端口网络的相互连接 Interconnections of two-ports
- 16.6 带负载的二端口网络 Loaded two-ports

目标：

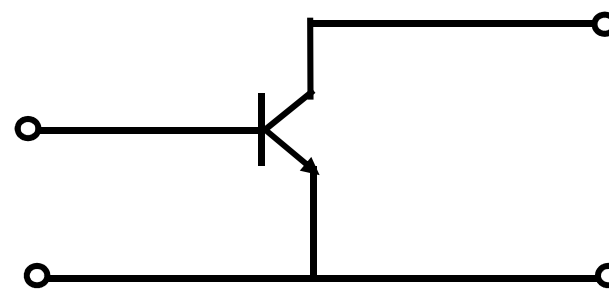
- a. 理解二端口网络特性的描述方法；
- b. 计算、测量二端口网络的任何参数，并可相互转换；
- c. 计算带负载二端口网络的端口电量；
- d. 分析相互连接的二端口网络。

## 16.1 概述

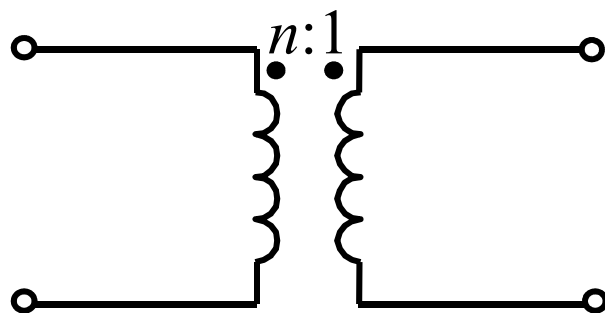
在工程实际中，研究信号及能量的传输和信号变换时，经常碰到如下形式的电路：二端口网络。



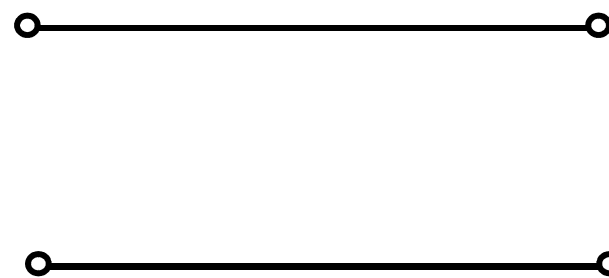
滤波器



三极管



变压器

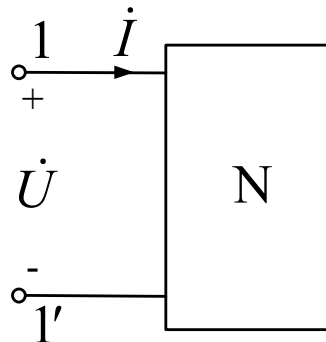


传输线

## 16.2 二端口网络的端口特性方程

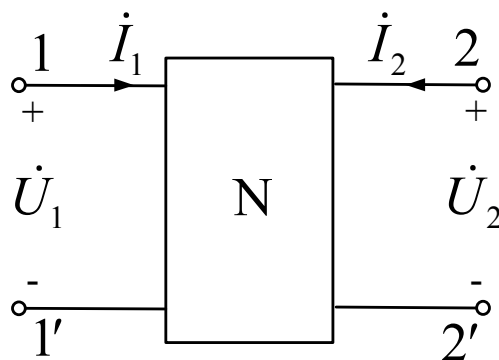
### 1 多端网络端口的定义

➤ 一端口 (port)



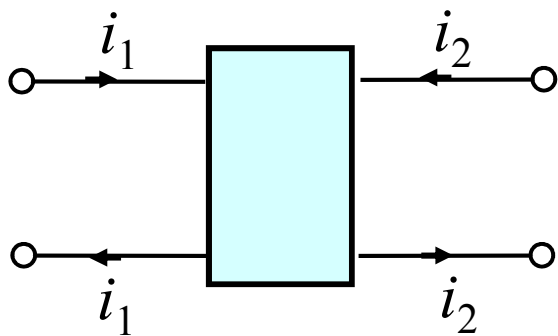
端口由一对端钮构成，且满足如下条件：从一个端钮流入的电流等于从另一个端钮流出的电流。

➤ 二端口 (two-port)

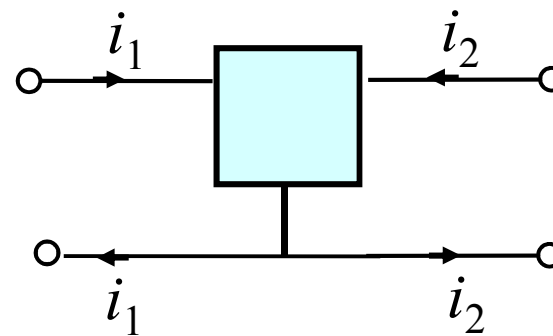


当一个电路与外部电路通过两个端口连接时称此电路为二端口网络。

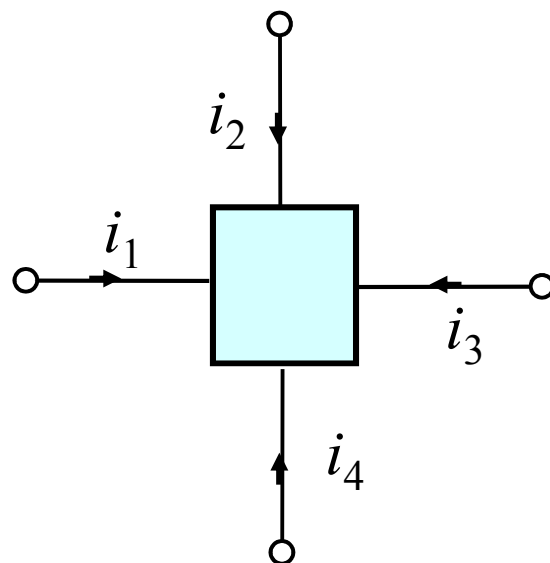
## ► 二端口网络与四端网络



二端口

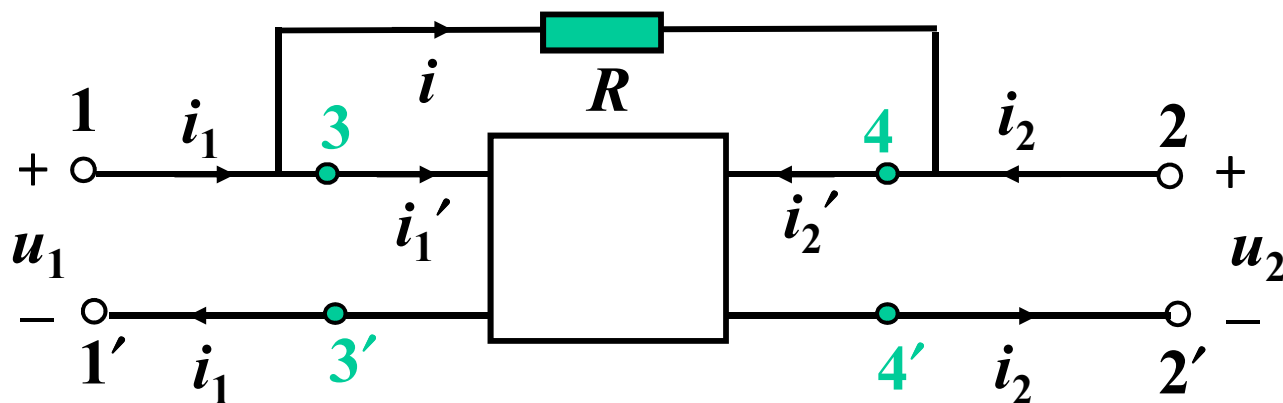


具有公共端的二端口



四端网络，非二端口网络

► 四端二端口的两个端口间若有外部连接，则会破坏原二端口的端口条件。



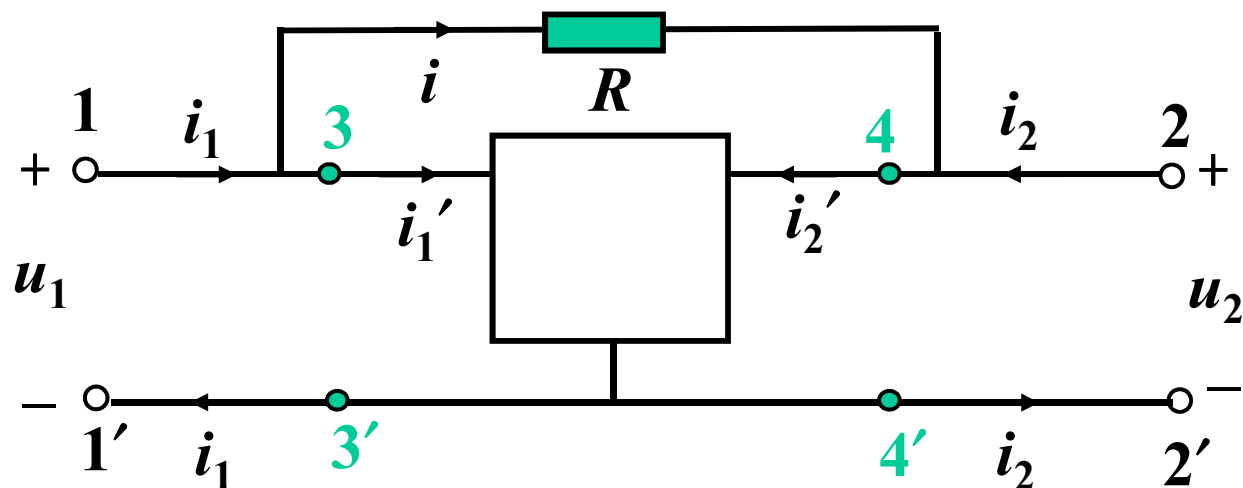
若1-1' 2-2'是二端口网络

3-3' 4-4'不是二端口网络，是四端网络

$$\left. \begin{aligned} i_1' &= i_1 - i \neq i_1 \\ i_2' &= i_2 + i \neq i_2 \end{aligned} \right\} \text{端口条件破坏}$$

思考：若3-3' 4-4'是二端口网络，则1-1' 2-2' 是二端口吗？

➤ 三端二端口的两个端口间若有外部连接，则仍然是二端口，满足端口条件。



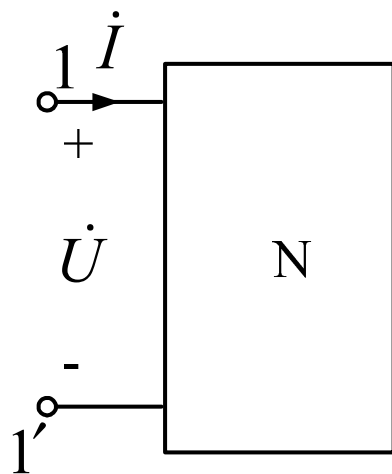
若1-1' 2-2'是二端口网络

3-3' 4-4' 是二端口网络，满足端口条件

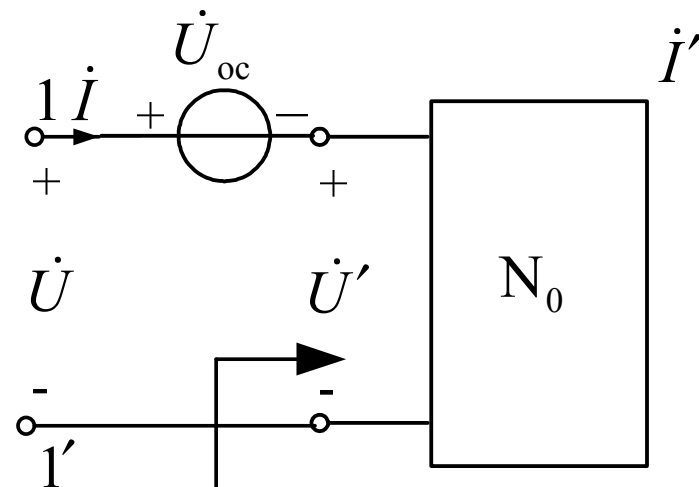
$$\left. \begin{aligned} i_1' &= i_1 - i = i_1 \\ i_2' &= i_2 + i = i_2 \end{aligned} \right\} \text{端口条件成立}$$

思考：若3-3' 4-4'是二端口网络，则1-1' 2-2' 是二端口吗？

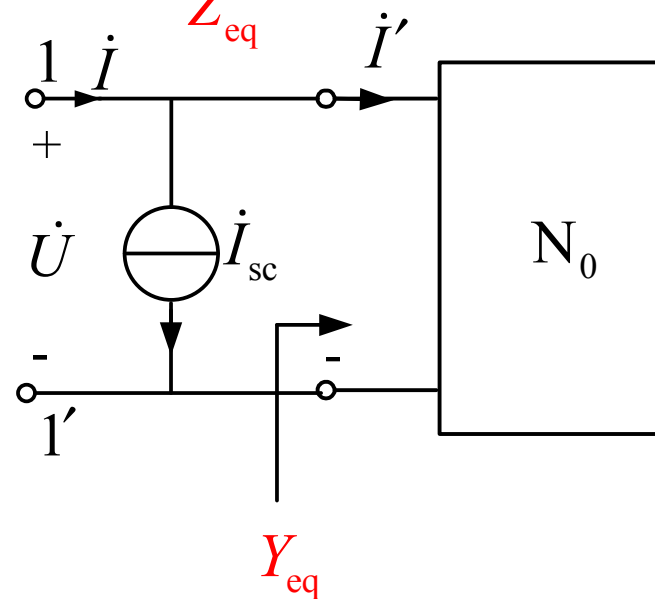
## 2 含源一端口网络



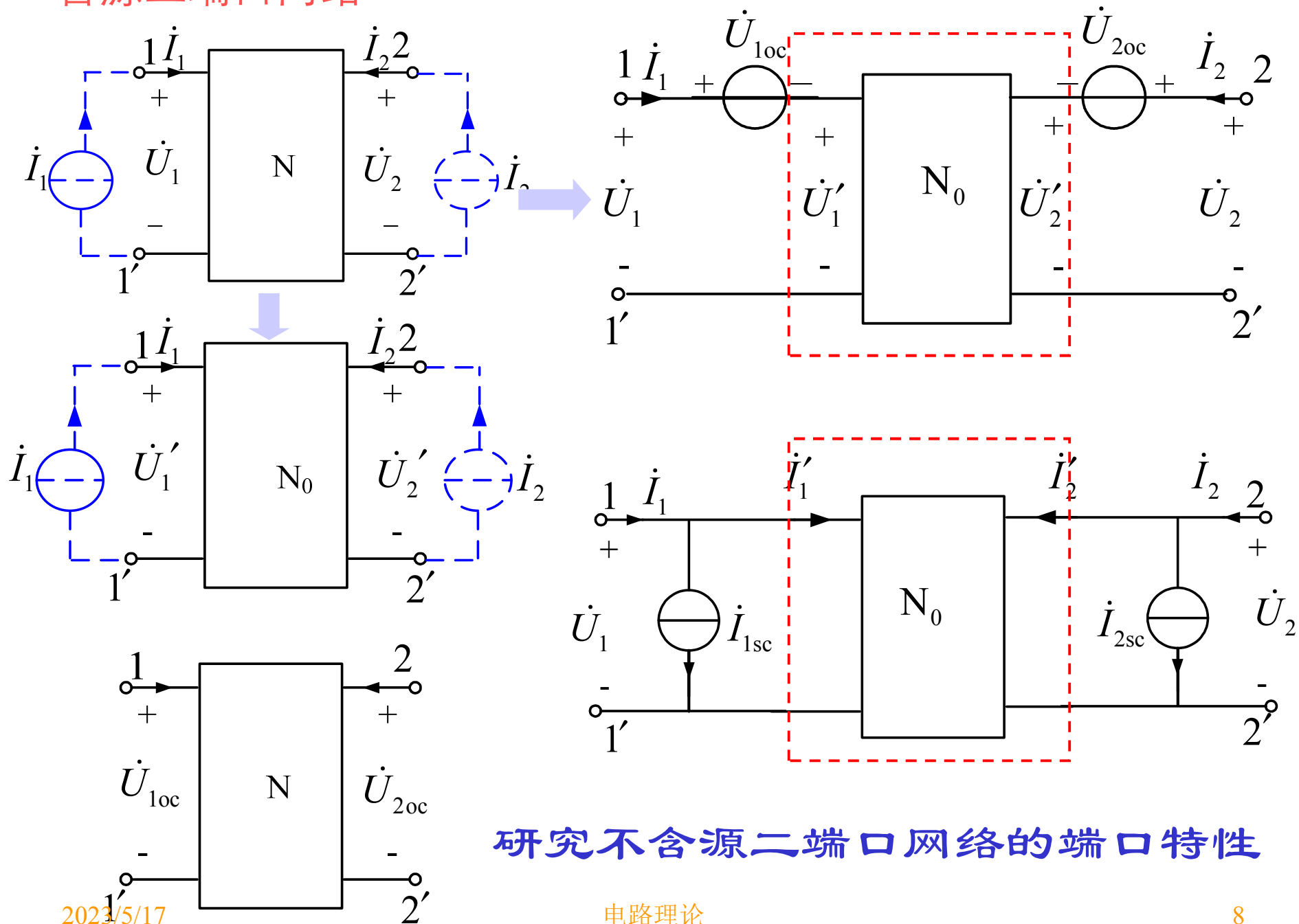
$$\dot{U} = \dot{U}_{oc} + Z_{eq} \dot{I}$$



$$\dot{I} = Y_{eq} \dot{U} + \dot{I}_{sc}$$



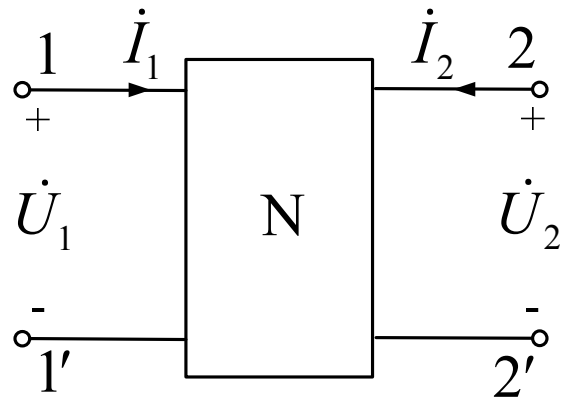
### 3 含源二端口网络



研究不含源二端口网络的端口特性

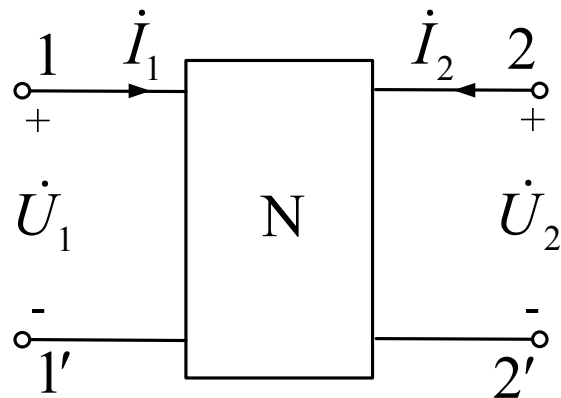


## 4 线性松弛二端口网络的端口特性方程



- 二端口网络的4个端口变量中，只有2个独立变量。他们构成2个端口特性方程
- 任选2个端口变量作为自变量，另外2个作为函数，即可得到二端口网络的一组特性方程；
- 可用六组参数描述二端口网络，六组不同的方程均可表示端口特性。

## 4 线性松弛二端口网络的端口特性方程



$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

阻抗参数方程

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

导纳参数方程

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

$$\begin{cases} \dot{I}_1 = g_{11}\dot{U}_1 + g_{12}\dot{I}_2 \\ \dot{U}_2 = g_{21}\dot{U}_1 + g_{22}\dot{I}_2 \end{cases}$$

混和参数方程

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases} \quad \begin{cases} \dot{U}_2 = A'\dot{U}_1 + B'(-\dot{I}_1) \\ \dot{I}_2 = C'\dot{U}_1 + D'(-\dot{I}_1) \end{cases}$$

传输参数方程

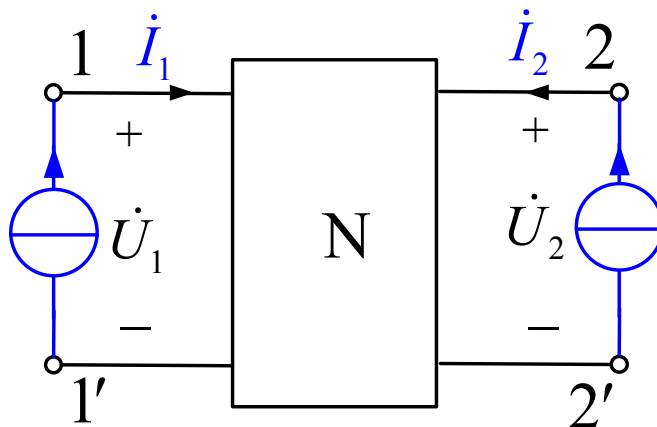
## 4 线性松弛二端口网络的端口特性方程

名称	自变量	方程
阻抗参数方程	$\dot{I}_1, \dot{I}_2$	$\left. \begin{aligned} \dot{U}_1 &= Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 &= Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{aligned} \right\}$
导纳参数方程	$\dot{U}_1, \dot{U}_2$	$\left. \begin{aligned} \dot{I}_1 &= Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 &= Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{aligned} \right\}$
混合参数方程	$\dot{I}_1, \dot{U}_2$	$\left. \begin{aligned} \dot{U}_1 &= h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 &= h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{aligned} \right\}$
逆混合参数方程	$\dot{U}_1, \dot{I}_2$	$\left. \begin{aligned} \dot{I}_1 &= g_{11}\dot{U}_1 + g_{12}\dot{I}_2 \\ \dot{U}_2 &= g_{21}\dot{U}_1 + g_{22}\dot{I}_2 \end{aligned} \right\}$
正向传输参数方程	$\dot{U}_2, \dot{I}_2$	$\left. \begin{aligned} \dot{U}_1 &= A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 &= C\dot{U}_2 + D(-\dot{I}_2) \end{aligned} \right\}$
反向传输参数方程	$\dot{U}_1, \dot{I}_1$	$\left. \begin{aligned} \dot{U}_2 &= A'\dot{U}_1 + B'(-\dot{I}_1) \\ \dot{I}_2 &= C'\dot{U}_1 + D'(-\dot{I}_1) \end{aligned} \right\}$

## 16.3 二端口网络的参数

### 16.3.1 阻抗参数

#### 1 阻抗参数方程 (Z)



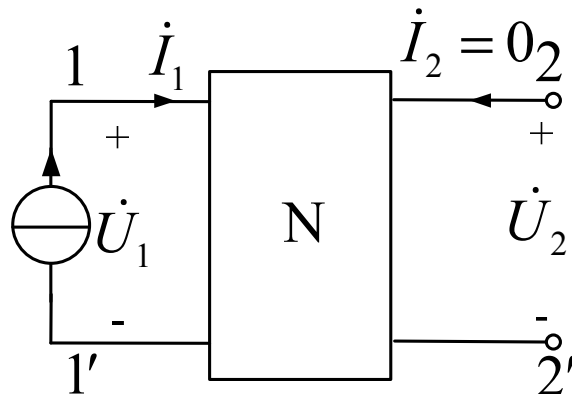
将  $\dot{I}_1$ 、 $\dot{I}_2$  视为激励源， $\dot{U}_1$ 、 $\dot{U}_2$  是在  $\dot{I}_1$ 、 $\dot{I}_2$  共同激励下的响应。  
由叠加定理得

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

上述方程即为  $Z$  参数方程，写成矩阵形式为：

$$\begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{I}_2 \end{bmatrix} \quad [Z] = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \quad \text{称为 } Z \text{ 参数矩阵}$$

## 2. Z 参数计算与测量方法

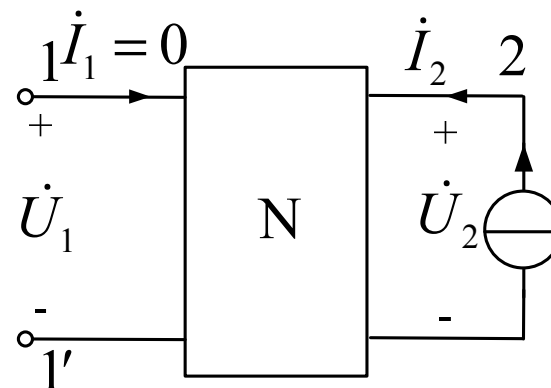


由Z参数方程可得：

$$Z_{11} = \left. \frac{\dot{U}_1}{\dot{I}_1} \right|_{\dot{I}_2=0}$$

$$Z_{21} = \left. \frac{\dot{U}_2}{\dot{I}_1} \right|_{\dot{I}_2=0}$$

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$



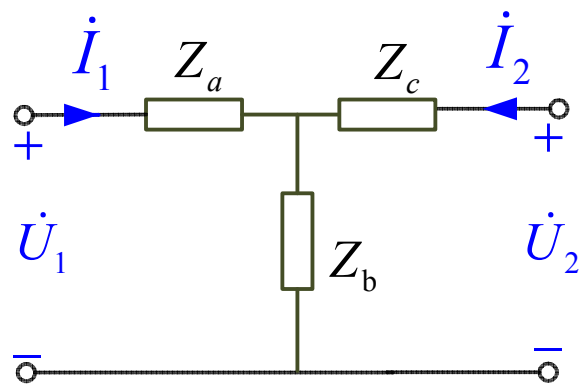
由Z参数方程可得：

$$Z_{12} = \left. \frac{\dot{U}_1}{\dot{I}_2} \right|_{\dot{I}_1=0}$$

$$Z_{22} = \left. \frac{\dot{U}_2}{\dot{I}_2} \right|_{\dot{I}_1=0}$$

3. 互易二端口：N为互易网络时，由互易定理： $Z_{12} = Z_{21}$

### 【例1】



$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

$$Z_{11} = \left. \frac{\dot{U}_1}{\dot{I}_1} \right|_{\dot{I}_2=0} = Z_a + Z_b$$

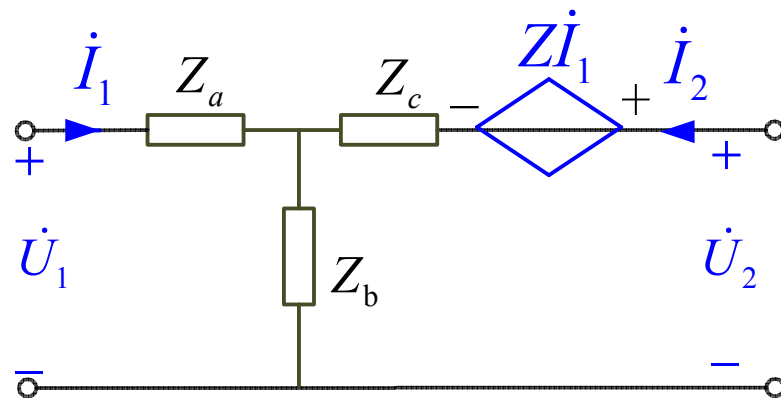
$$Z_{21} = \left. \frac{\dot{U}_2}{\dot{I}_1} \right|_{\dot{I}_2=0} = Z_b$$

$$Z_{12} = \left. \frac{\dot{U}_1}{\dot{I}_2} \right|_{\dot{I}_1=0} = Z_b$$

$$Z_{22} = \left. \frac{\dot{U}_2}{\dot{I}_2} \right|_{\dot{I}_1=0} = Z_b + Z_c$$

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### 【例2】



$$Z_{11} = \left. \frac{\dot{U}_1}{\dot{I}_1} \right|_{\dot{I}_2=0} = Z_a + Z_b$$

$$Z_{21} = \left. \frac{\dot{U}_2}{\dot{I}_1} \right|_{\dot{I}_2=0} = Z_b + Z$$

$$Z_{12} = \left. \frac{\dot{U}_1}{\dot{I}_2} \right|_{\dot{I}_1=0} = Z_b$$

$$Z_{22} = \left. \frac{\dot{U}_2}{\dot{I}_2} \right|_{\dot{I}_1=0} = Z_b + Z_c$$

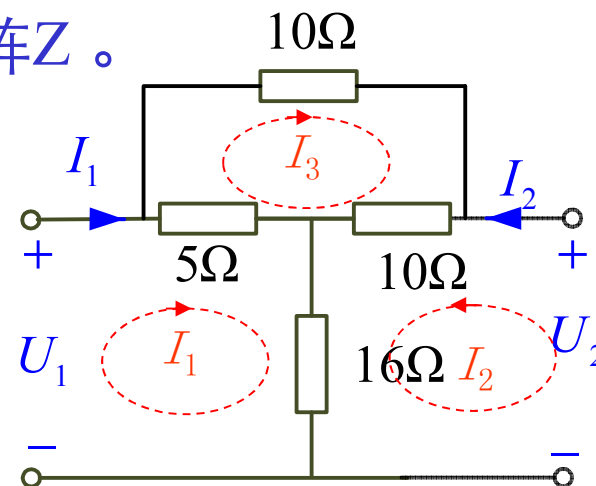
$$Z_{12} \neq Z_{21}$$

电路理论

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【例3】：求图示二端口网络的阻抗参数矩阵 $Z$ 。

解题思路：用网络方程法计算参数。以电流为自变量，列端口电压的方程，并整理系数即可得 $Z$ 参数。



网孔电流方程为：

$$\begin{aligned}U_1 &= (5+16)I_1 + 16I_2 - 5I_3 \\16I_1 + (10+16)I_2 + 10I_3 &= U_2 \\-5I_1 + 10I_2 + (10+10+5)I_3 &= 0\end{aligned}$$

联立求解，消去  $I_3$ ：

$$\begin{aligned}U_1 &= 20I_1 + 18I_2 \\U_2 &= 18I_1 + 22I_2\end{aligned}$$

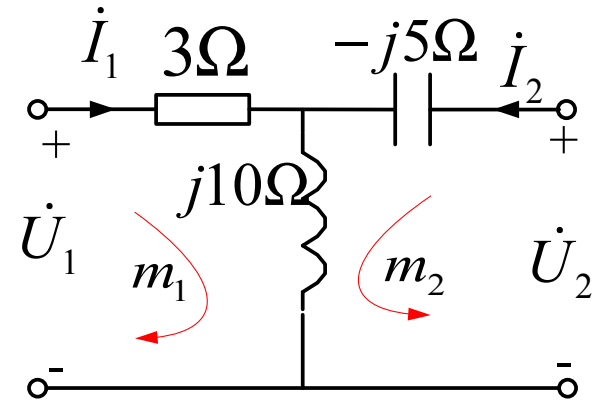
$Z$  参数矩阵方程为：

$$\mathbf{Z} = \begin{bmatrix} 20 & 18 \\ 18 & 22 \end{bmatrix} \Omega$$

思考题：求其他参数矩阵？

【练习】 计算Z参数 (p90, 例16-3-1)

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$



$$Z_{11} = \frac{\dot{U}_1}{\dot{I}_1} \Big|_{\dot{I}_2=0} = (3 + j10) \Omega$$

$$Z_{21} = \frac{\dot{U}_2}{\dot{I}_1} \Big|_{\dot{I}_2=0} = j10 \Omega$$

$$Z_{22} = \frac{\dot{U}_2}{\dot{I}_2} \Big|_{\dot{I}_1=0} = -j5 + j10 = j5 \Omega$$

$$Z_{12} = \frac{\dot{U}_1}{\dot{I}_2} \Big|_{\dot{I}_1=0} = j10 \Omega$$

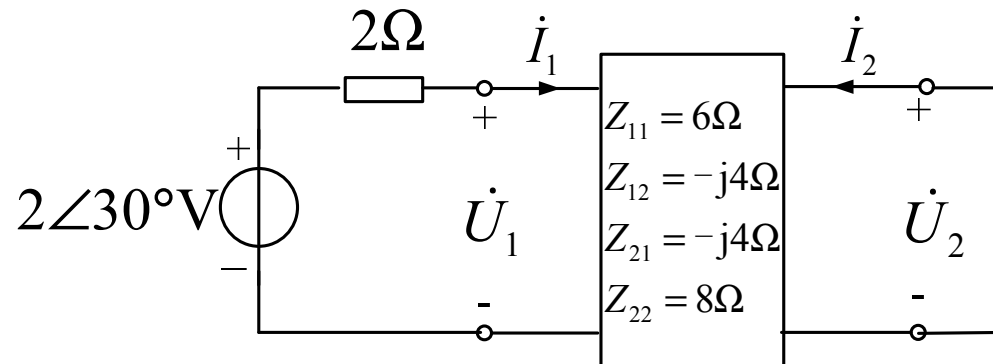
方法2：网络方程

$$\begin{cases} (3 + j10)\dot{I}_1 + j10\dot{I}_2 = \dot{U}_1 \\ j10\dot{I}_1 + (j10 - j5)\dot{I}_2 = \dot{U}_2 \end{cases}$$

$$Z = \begin{bmatrix} 3 + j10 & j10 \\ j10 & j5 \end{bmatrix}$$



例4.计算图中所示电流 $\dot{I}_1$ 、 $\dot{I}_2$



解：端接支路特性方程（VCR）

$$\begin{cases} \dot{U}_1 = 2\angle 30^\circ - 2\dot{I}_1 \\ \dot{U}_2 = 0 \end{cases}$$

二端口网络的参数方程为

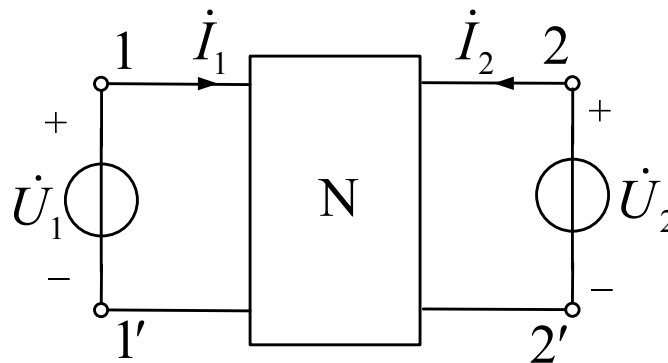
$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 = 6\dot{I}_1 - j4\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 = -j4\dot{I}_1 + 8\dot{I}_2 \end{cases}$$

可以求出

$$\dot{I}_1 = 0.2\angle 30^\circ \text{ A}, \quad \dot{I}_2 = 0.1\angle 120^\circ \text{ A}$$

### 16.3.2 导纳参数

#### 1 导纳参数方程 (Y)



将两个端口各施加一电压源，则端口电流可视为这些电压源的叠加作用产生。

即：

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

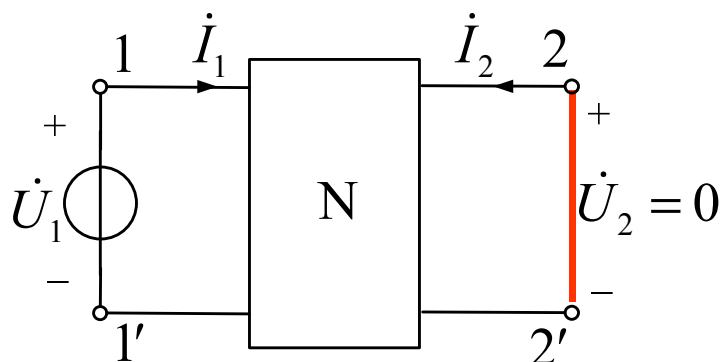
上述方程即为Y参数方程，写成矩阵形式为：

$$\begin{bmatrix} \dot{I}_1 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix} = \mathbf{Y} \begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix}$$

称为Y 参数矩阵.

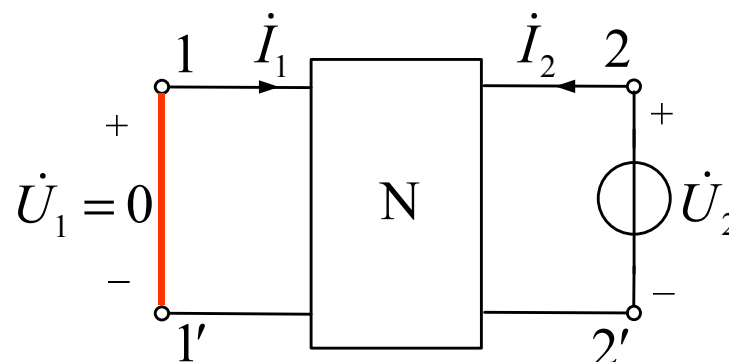
## 2. Y参数的计算和测定

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$



由Y参数方程可得：

$$Y_{11} = \left. \frac{\dot{I}_1}{\dot{U}_1} \right|_{\dot{U}_2=0}$$
$$Y_{21} = \left. \frac{\dot{I}_2}{\dot{U}_1} \right|_{\dot{U}_2=0}$$

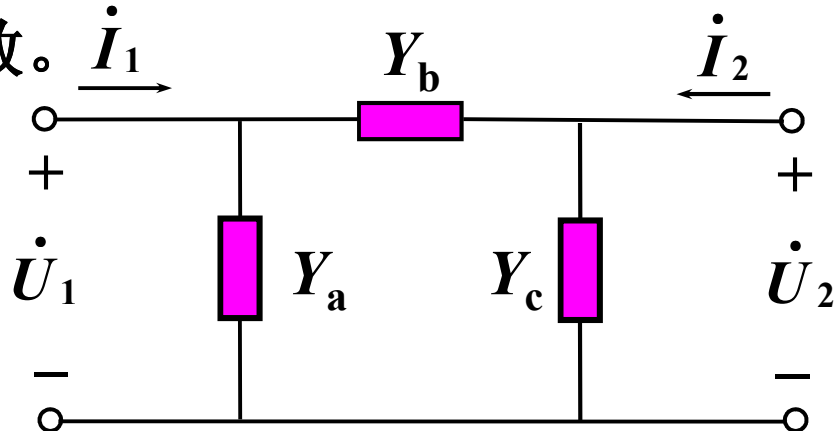


由Y参数方程可得：

$$Y_{12} = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{U}_1=0}$$
$$Y_{22} = \left. \frac{\dot{I}_2}{\dot{U}_2} \right|_{\dot{U}_1=0}$$

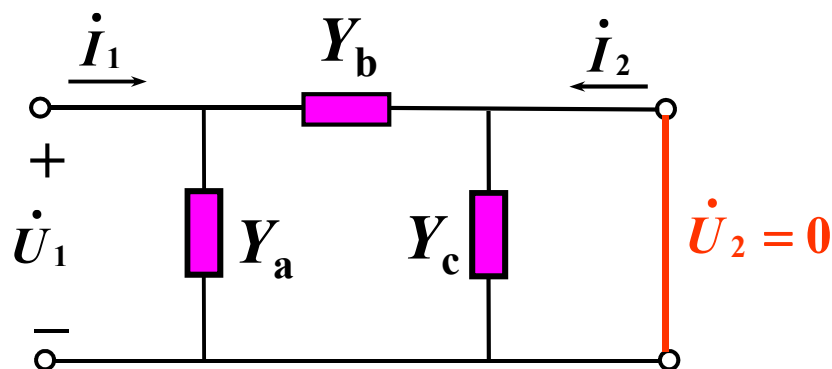
3. 互易二端口：N为互易网络时，由互易定理： $Y_{12} = Y_{21}$

例1. 求  $Y$  参数。  $\dot{I}_1$



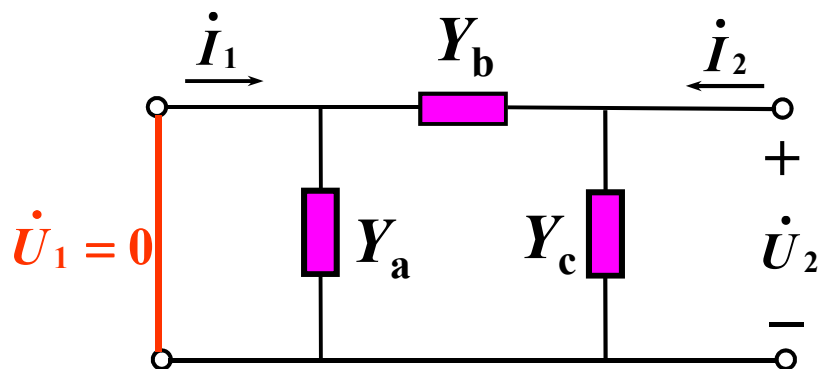
$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

解:



$$Y_{11} = \left. \frac{\dot{I}_1}{\dot{U}_1} \right|_{\dot{U}_2=0} = Y_a + Y_b$$

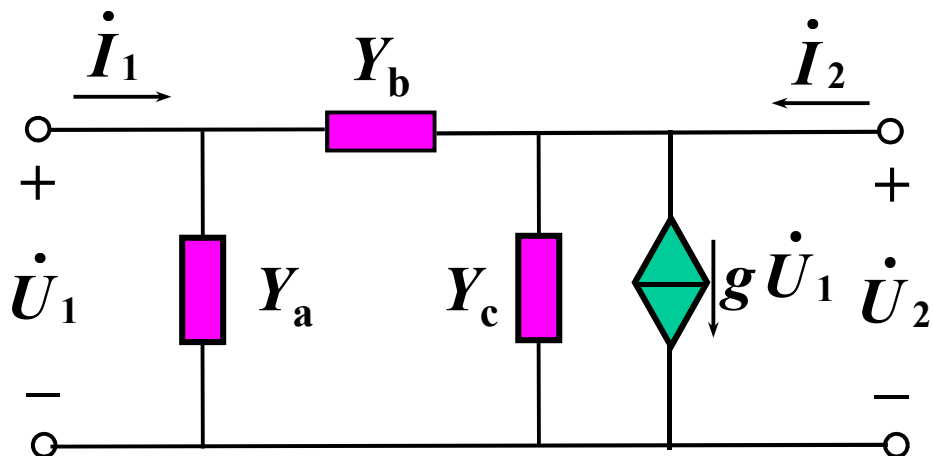
$$Y_{21} = \left. \frac{\dot{I}_2}{\dot{U}_1} \right|_{\dot{U}_2=0} = -Y_b$$



$$Y_{12} = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{U}_1=0} = -Y_b$$

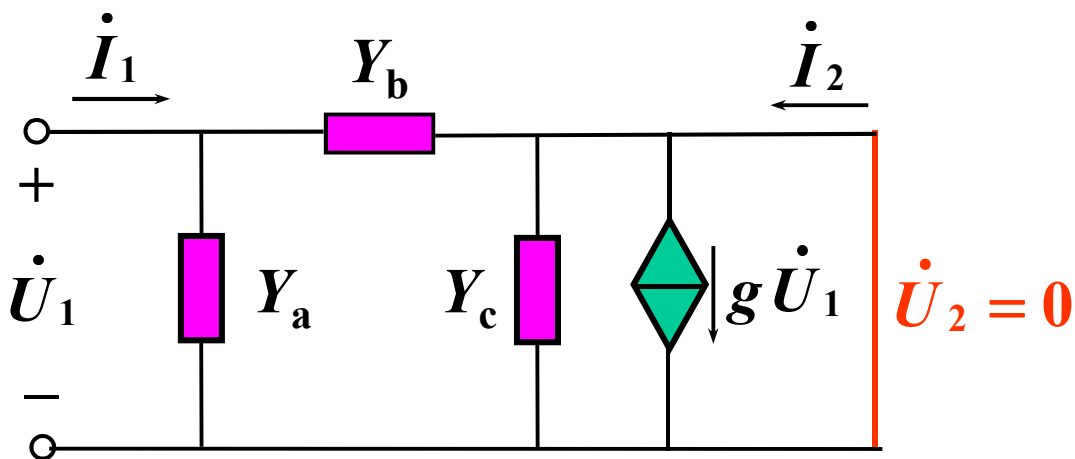
$$Y_{22} = \left. \frac{\dot{I}_2}{\dot{U}_2} \right|_{\dot{U}_1=0} = Y_b + Y_c$$

例2.



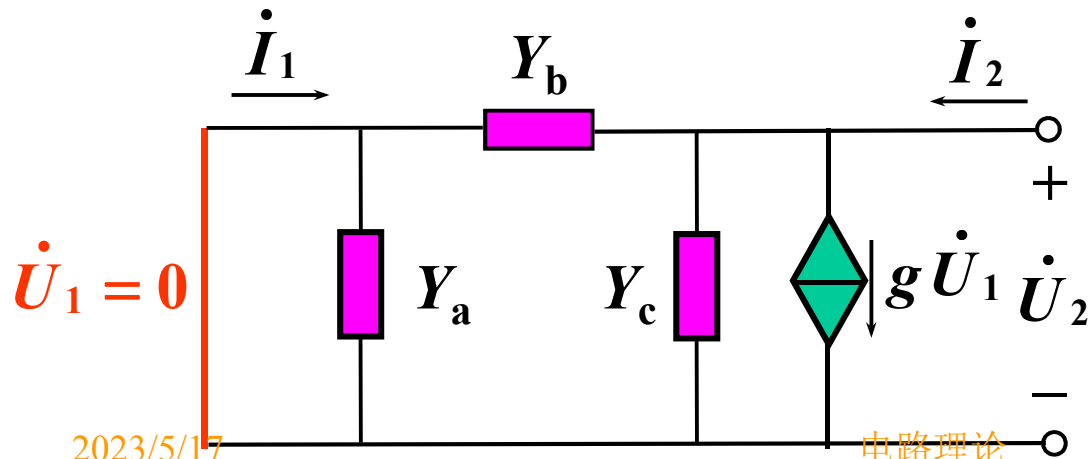
解:

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$



$$Y_{11} = \left. \frac{\dot{I}_1}{\dot{U}_1} \right|_{\dot{U}_2=0} = Y_a + Y_b$$

$$Y_{21} = \left. \frac{\dot{I}_2}{\dot{U}_1} \right|_{\dot{U}_2=0} = -Y_b + g$$



$$Y_{12} = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{U}_1=0} = -Y_b$$

$$Y_{22} = \left. \frac{\dot{I}_2}{\dot{U}_2} \right|_{\dot{U}_1=0} = Y_b + Y_c$$

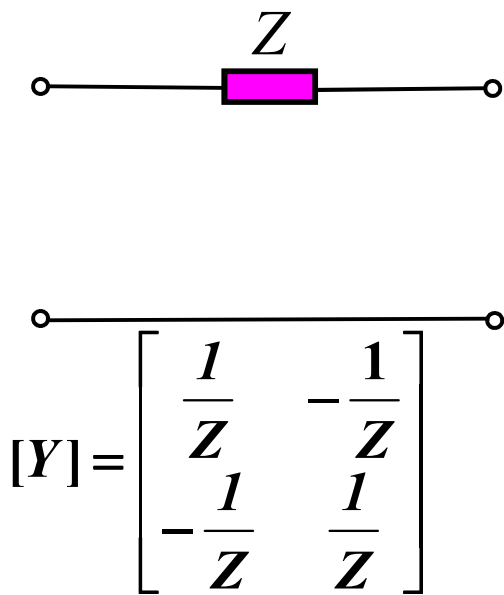
#### 4. 对称二端口

除  $Y_{12} = Y_{21}$  外，若  $Y_{11} = Y_{22}$ ，称为对称二端口。

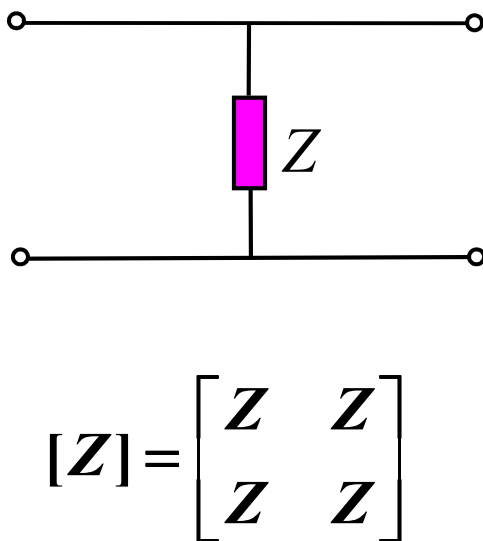
对称二端口是指电路结构左右对称。两个端口电气特性上对称，结构不对称的二端口，其电气特性可能是对称的，这样的二端口也是对称二端口。

5.  $Z$  参数矩阵与  $Y$  参数矩阵互为逆矩阵。即： $Y = Z^{-1}$        $Z = Y^{-1}$

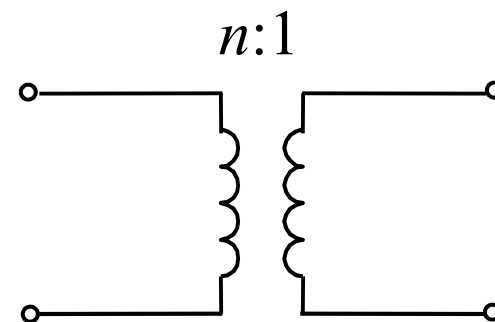
6. 并非所有的二端口均有  $Z, Y$  参数。



$Z$  不存在



$Y$  不存在



$Z, Y$  均不存在

### 16.3.3 混合参数H

$H$  参数和 $G$ 参数称为混合参数， $H$  参数常用于晶体管等效电路。

#### 1. 混合 参数方程

➤  $H$  参数

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

矩阵形式:

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} = H \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} \quad \begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix} = G^{-1} \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} = H \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix}$$

➤  $G$ 参数

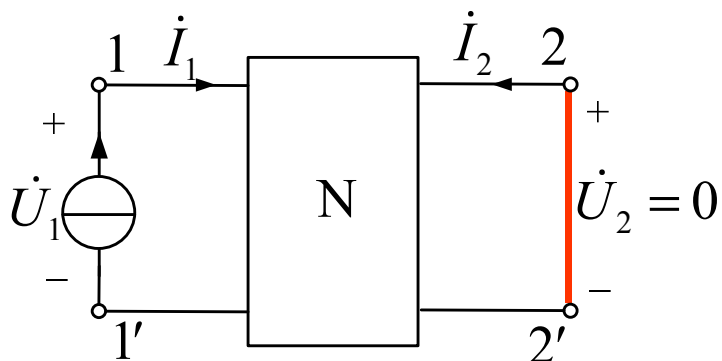
$$\begin{cases} \dot{I}_1 = g_{11}\dot{U}_1 + g_{12}\dot{I}_2 \\ \dot{U}_2 = g_{21}\dot{U}_1 + g_{22}\dot{I}_2 \end{cases}$$

$$\begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix} = G \begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix}$$

$G$ 和 $H$ 存在逆矩阵时，它们互为逆矩阵： $G=H^{-1}$ 或 $H=G^{-1}$

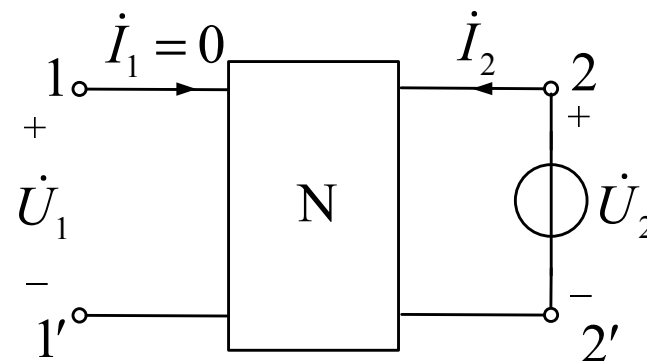
## 2. $H$ 参数的计算与测定

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$



$$h_{11} = \left. \frac{\dot{U}_1}{\dot{\mathbf{I}}_1} \right|_{\dot{U}_2=0}$$

$$h_{21} = \left. \frac{\dot{\mathbf{I}}_2}{\dot{\mathbf{I}}_1} \right|_{\dot{U}_2=0}$$



$$h_{12} = \left. \frac{\dot{U}_1}{\dot{U}_2} \right|_{\dot{\mathbf{i}}_1=0}$$

$$h_{22} = \left. \frac{\dot{\mathbf{I}}_2}{\dot{U}_2} \right|_{\dot{\mathbf{i}}_1=0}$$

3. 互易二端口：N为互易网络时，由互易定理：

$$h_{12} = -h_{21}$$



【例1】.

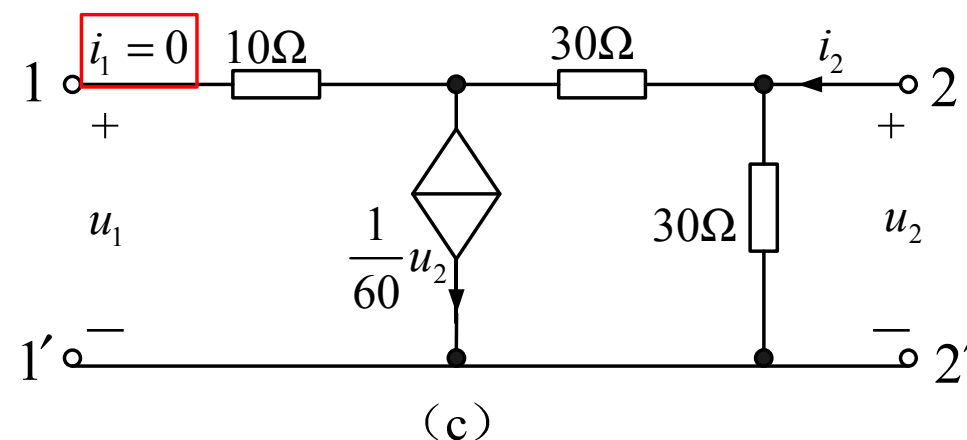
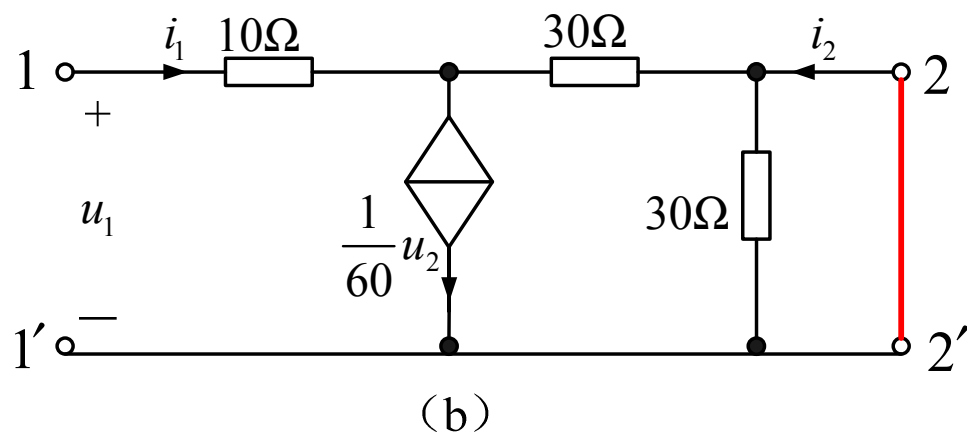
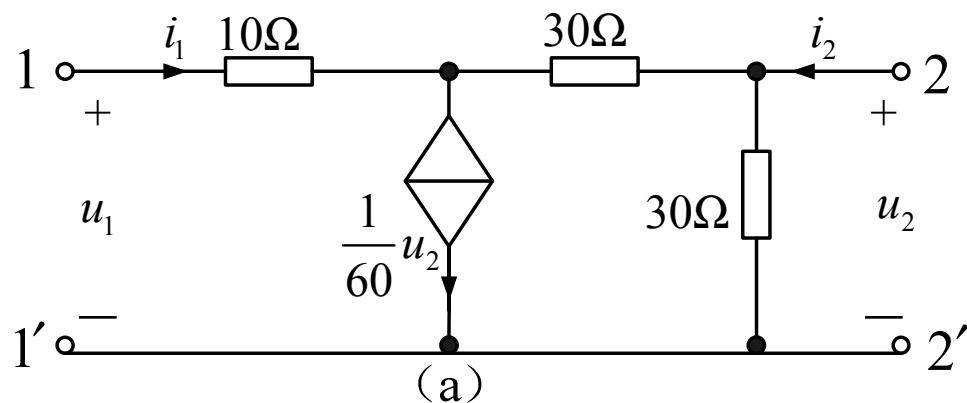
$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

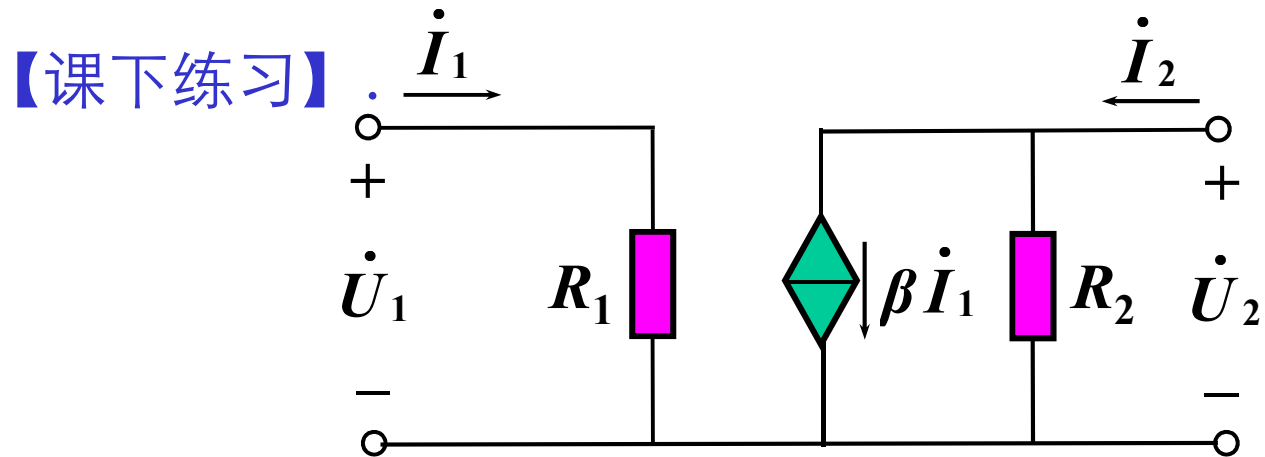
$$h_{11} = \left. \frac{u_1}{i_1} \right|_{u_2=0} = 10 + 30 = 40\Omega$$

$$h_{21} = \left. \frac{i_2}{i_1} \right|_{u_2=0} = \frac{-i_1}{i_1} = -1$$

$$h_{22} = \left. \frac{i_2}{u_2} \right|_{i_1=0} = \frac{\frac{1}{60}u_2 + \frac{1}{30}u_2}{u_2} = \frac{1}{20}\text{S}$$

$$h_{12} = \left. \frac{u_1}{u_2} \right|_{i_1=0} = \frac{u_2 - \frac{1}{60}u_2 \times 30}{u_2} = \frac{1}{2}$$





$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

$$h_{11} = \left. \frac{\dot{U}_1}{\dot{I}_1} \right|_{\dot{U}_2=0} = R_1$$

$$h_{21} = \left. \frac{\dot{I}_2}{\dot{I}_1} \right|_{\dot{U}_2=0} = \beta$$

$$h_{12} = \left. \frac{\dot{U}_1}{\dot{U}_2} \right|_{\dot{I}_1=0} = 0$$

$$h_{22} = \left. \frac{\dot{I}_2}{\dot{U}_2} \right|_{\dot{I}_1=0} = \frac{1}{R_2}$$

### 16.3.4 传输参数T

#### 1. T参数和方程

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} \dot{U}_2 \\ -\dot{I}_2 \end{bmatrix} = \mathbf{T} \begin{bmatrix} \dot{U}_2 \\ -\dot{I}_2 \end{bmatrix}$$

#### 3. T参数的计算或测定

$$A = \left. \frac{\dot{U}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} \quad B = \left. \frac{\dot{U}_1}{-\dot{I}_2} \right|_{\dot{U}_2=0}$$

$$C = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} \quad D = \left. \frac{\dot{I}_1}{-\dot{I}_2} \right|_{\dot{U}_2=0}$$

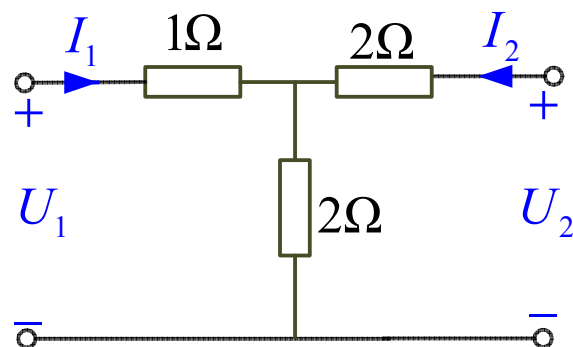
#### 2. T'参数和方程

$$\begin{cases} \dot{U}_2 = A'\dot{U}_1 + B'(-\dot{I}_1) \\ \dot{I}_2 = C'\dot{U}_1 + D'(-\dot{I}_1) \end{cases}$$

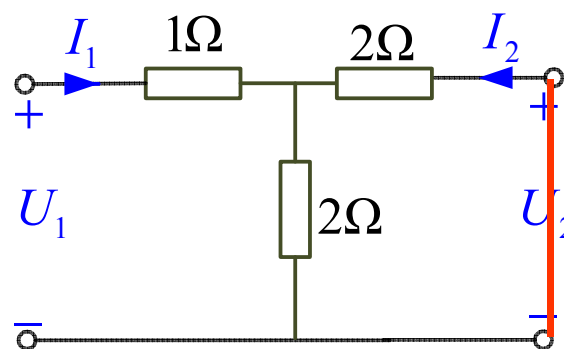
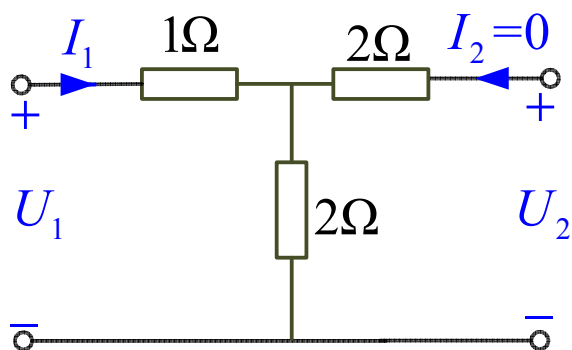
$$\begin{bmatrix} \dot{U}_2 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ -\dot{I}_1 \end{bmatrix} = \mathbf{T}' \begin{bmatrix} \dot{U}_1 \\ -\dot{I}_1 \end{bmatrix}$$

$$\begin{bmatrix} A & -B \\ C & -D \end{bmatrix} \begin{bmatrix} A' & -B' \\ C' & -D' \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

# 【例1】 求T参数



$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$



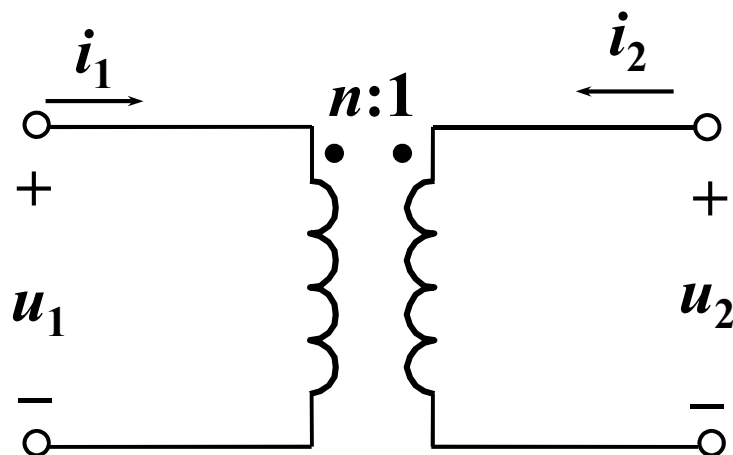
$$A = \left. \frac{U_1}{U_2} \right|_{I_2=0} = \frac{1+2}{2} = 1.5$$

$$C = \left. \frac{I_1}{U_2} \right|_{I_2=0} = 0.5 \text{ S}$$

$$B = \left. \frac{U_1}{-I_2} \right|_{U_2=0} = \frac{I_1[1+(2//2)]}{0.5I_1} = 4 \Omega$$

$$D = \left. \frac{I_1}{-I_2} \right|_{U_2=0} = \frac{I_1}{0.5I_1} = 2$$

【练习】.



$$\begin{cases} u_1 = nu_2 \\ i_1 = -\frac{1}{n}i_2 \end{cases} \quad \text{即} \quad \begin{bmatrix} u_1 \\ i_1 \end{bmatrix} = \begin{bmatrix} n & 0 \\ 0 & \frac{1}{n} \end{bmatrix} \begin{bmatrix} u_2 \\ -i_2 \end{bmatrix}$$

$$\text{则} \quad [T] = \begin{bmatrix} n & 0 \\ 0 & \frac{1}{n} \end{bmatrix}$$

【例2】 求T参数 
$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$

$$\dot{I}_3 = 0$$

$$\dot{U}_3 = \dot{U}_1 = n\dot{U}_2$$

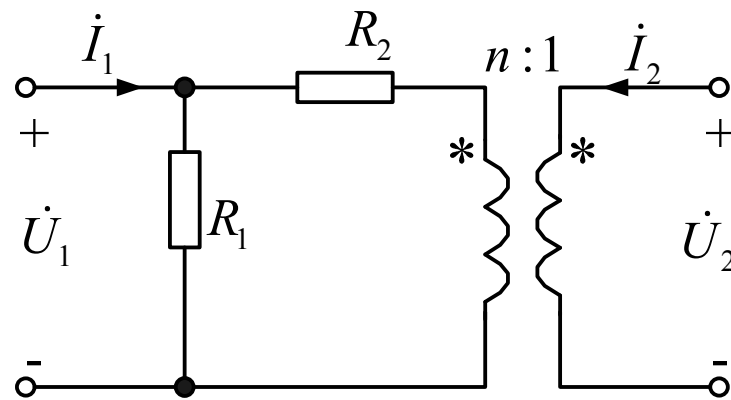
$$A = \left. \frac{\dot{U}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} = n \quad C = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} = \frac{\frac{\dot{U}_1}{R_1}}{\frac{1}{n}\dot{U}_1} = \frac{n}{R_1}$$

$$\dot{U}_3 = 0 \quad \dot{I}_3 = \frac{\dot{U}_1}{R_2} \quad \dot{I}_1 = \frac{\dot{U}_1}{R_1 // R_2}$$

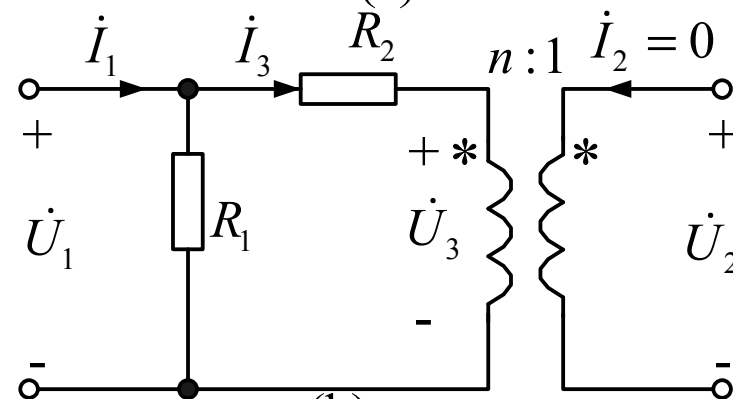
$$\dot{I}_2 = -n\dot{I}_3 = -\frac{n}{R_2}\dot{U}_1$$

$$D = \left. \frac{\dot{I}_1}{-\dot{I}_2} \right|_{\dot{U}_2=0} = \frac{(\frac{1}{R_1} + \frac{1}{R_2})\dot{U}_1}{\frac{n}{R_2}\dot{U}_1} = \frac{R_1 R_2}{n R_1}$$

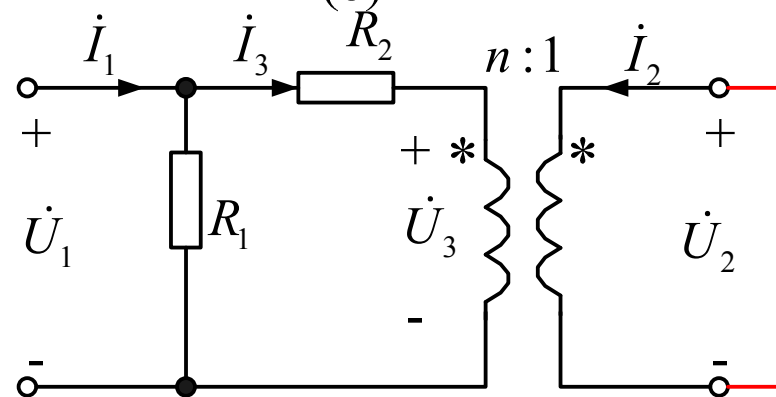
$$B = \left. \frac{\dot{U}_1}{-\dot{I}_2} \right|_{\dot{U}_2=0} = \frac{\dot{U}_1}{\frac{n}{R_2}\dot{U}_1} = \frac{R_2}{n}$$



(a)



(b)



(c)

【例2】 求T参数 
$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$

Method2:

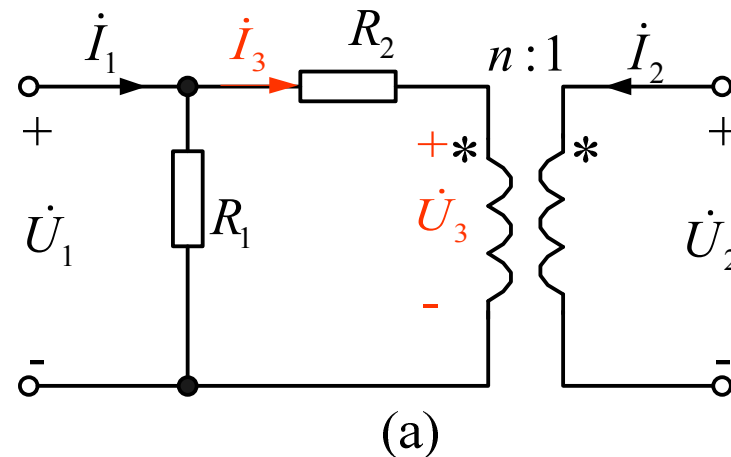
$$\dot{I}_3 = -\frac{1}{n}\dot{I}_2$$

$$\dot{U}_3 = n\dot{U}_2$$

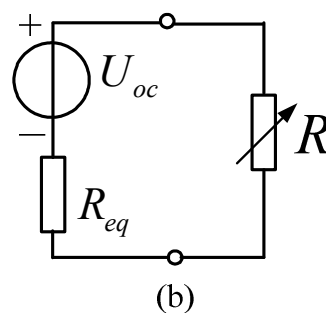
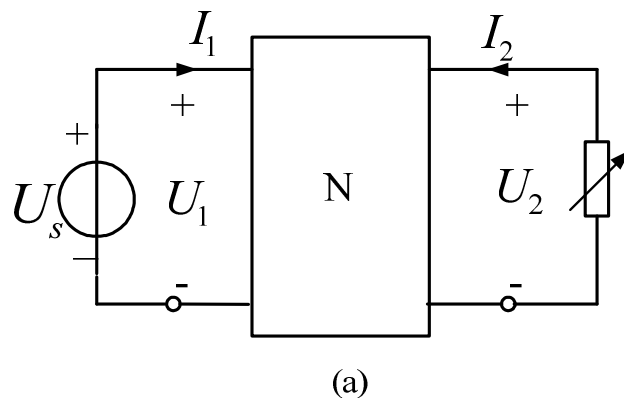
$$\dot{U}_1 = \dot{I}_3 R_2 + \dot{U}_3 = n\dot{U}_2 - \frac{R_2}{n}\dot{I}_2$$

$$\dot{I}_1 = \frac{\dot{U}_1}{R_1} + \dot{I}_3 = \frac{1}{R_1} \left( n\dot{U}_2 - \frac{R_2}{n}\dot{I}_2 \right) - \frac{1}{n}\dot{I}_2$$

$$\begin{cases} \dot{U}_1 = n\dot{U}_2 + \frac{R_2}{n}(-\dot{I}_2) \\ \dot{I}_1 = \frac{n}{R_1}\dot{U}_2 + \left( \frac{R_2}{nR_1} + \frac{1}{n} \right) (-\dot{I}_2) \end{cases}$$



例3：已知：  $U_s=9V$ ，无源双口网络的传输矩阵  $T = \begin{bmatrix} 2.5 & 6 \\ 0.5 & 1.6 \end{bmatrix}$ ，求当R为何值时，R吸收的功率为最大？



$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$

解：

求开路电压：  $U_{oc} = U_2|_{I_2=0} = \frac{U_1}{A} = \frac{9}{2.5} = 3.6V$

求短路电流：  $I_{sc} = -I_2|_{U_2=0} = \frac{U_1}{B} = \frac{9}{6} = 1.5A$

则等效电阻：  $R_{eq} = \frac{3.6}{1.5} = 2.4\Omega$

则  $R=2.4\Omega$  时, 获得最大功率：  $P = \left(\frac{U_{oc}}{R + R_{eq}}\right)^2 R = 1.35W$

思考：直接由T参数求等效电阻？



### 16.3.5 参数间的互换关系

1 各参数间的互换关系：将一种参数方程进行自变量和变量转换，将其变成另外一种参数方程，从而获得两种参数之间的互换关系。

- 双口网络可以用六组参数来表征根据具体情况，可以选用一种合适的参数表示：
- Z参数和Y参数常用于理论的探讨和基本定律得推导中；
- H参数广泛用于低频晶体管电路的分析问题中；
- 有些电路只存在某几种参数

如果知道双口网络的任一参数矩阵，通过对变量的运算，可以求得任何其他的参数矩阵，只要这一矩阵是存在的。

参考书中表 161-3-1 二端口网络六种参数的互换

【例4】：二端口网络的Y参数矩阵为  $Y = \begin{bmatrix} \frac{1}{15} & -\frac{1}{30} \\ -\frac{1}{30} & \frac{1}{15} \end{bmatrix}$ ，求该网络的Z参数矩阵。

解题思路：列写Y参数矩阵方程，整理方程，经过系数比较就可以获得相应的Z参数矩阵。

Y参数矩阵方程为：

$$\begin{cases} \dot{I}_1 = \frac{1}{15}\dot{U}_1 - \frac{1}{30}\dot{U}_2 \\ \dot{I}_2 = -\frac{1}{30}\dot{U}_1 + \frac{1}{15}\dot{U}_2 \end{cases}$$

整理矩阵方程：

$$\begin{cases} \dot{U}_1 = 20\dot{I}_1 + 10\dot{I}_2 \\ \dot{U}_2 = 10\dot{I}_1 + 20\dot{I}_2 \end{cases}$$

Z参数矩阵方程为：

$$\mathbf{Z} = \begin{bmatrix} 20 & 10 \\ 10 & 20 \end{bmatrix} \Omega$$

思考题：求T参数矩阵？

### 16.3.5 参数间的互换关系

Z parameters  $\rightarrow$  T parameters:

$$\begin{cases} \dot{U}_1 = Z_{11} \dot{I}_1 + Z_{12} \dot{I}_2 \\ \dot{U}_2 = Z_{21} \dot{I}_1 + Z_{22} \dot{I}_2 \end{cases}$$

$$\dot{I}_1 = \frac{1}{Z_{21}} \dot{U}_2 + \frac{Z_{22}}{Z_{21}} (-\dot{I}_2)$$

$$\dot{U}_1 = \frac{Z_{11}}{Z_{21}} \dot{U}_2 + \frac{Z_{11}Z_{22} - Z_{12}Z_{21}}{Z_{21}} (-\dot{I}_2)$$

$$AD - BC = \frac{Z_{11}Z_{22}}{Z_{21}^2} - \frac{Z_{11}Z_{22} - Z_{12}Z_{21}}{Z_{21}^2} = \frac{Z_{12}}{Z_{21}} = 1$$

$$\frac{\dot{U}_2|_{\dot{I}_2=0}}{\dot{I}_1} = \frac{\dot{U}_1|_{\dot{I}_1=0}}{\dot{I}_2}$$

$$Z_{21} = Z_{12}$$

互易定理 1

$$\frac{\dot{I}_2|_{\dot{U}_2=0}}{\dot{U}_1} = \frac{\dot{I}_1|_{\dot{U}_1=0}}{\dot{U}_2}$$

$$Y_{21} = Y_{12}$$

互易定理 2

$$\frac{\dot{U}_2|_{\dot{I}_2=0}}{\dot{U}_1} = \frac{-\dot{I}_1|_{\dot{U}_1=0}}{\dot{I}_2}$$

$$g_{12} = -g_{21}$$

$$h_{12} = -h_{21}$$

互易定理 3

### 16.3.5 参数间的互换关系

表16-1 互易及对称二端口网络的参数特点

	互易网络二端口	对称二端口网络	
$Z$	$Z_{12} = Z_{21}$	$Z_{11} = Z_{22}$	$Z_{12} = Z_{21}$
$Y$	$Y_{12} = Y_{21}$	$Y_{11} = Y_{22}$	$Y_{12} = Y_{21}$
$H/G$	$h_{12} = -h_{21}$	$h_{11}h_{22} - h_{12}h_{21} = 1$	$h_{12} = -h_{21}$
$T/T'$	$AD - BC = 1$	$A = D$	$AD - BC = 1$

## 16.4 二端口网络的等效电路模型

两个二端口网络等效：

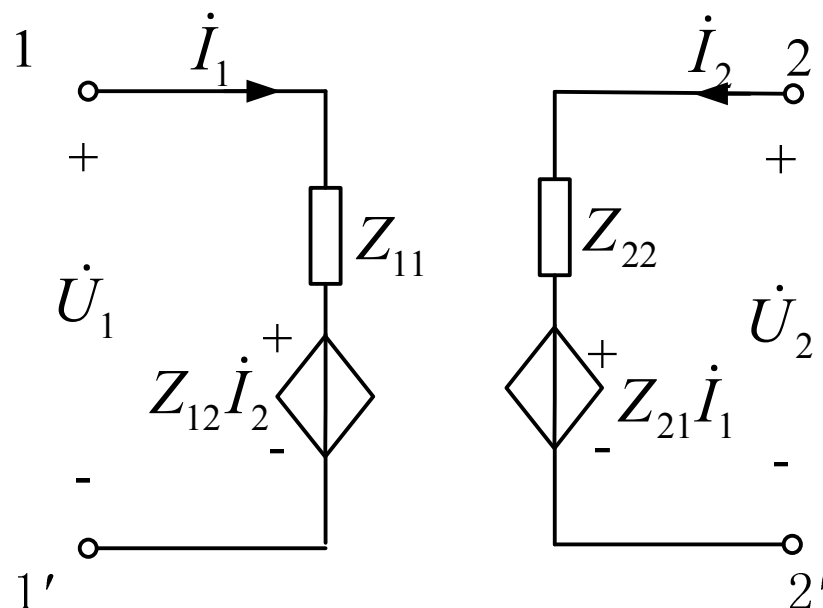
- 是指对外电路而言，端口的电压，电流关系相同。
- 求等效电路即根据给定的参数方程画出电路。

### 1由Z参数方程表示的等效模型

方法1：直接由参数方程得到等效电路。

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

等效电路为：



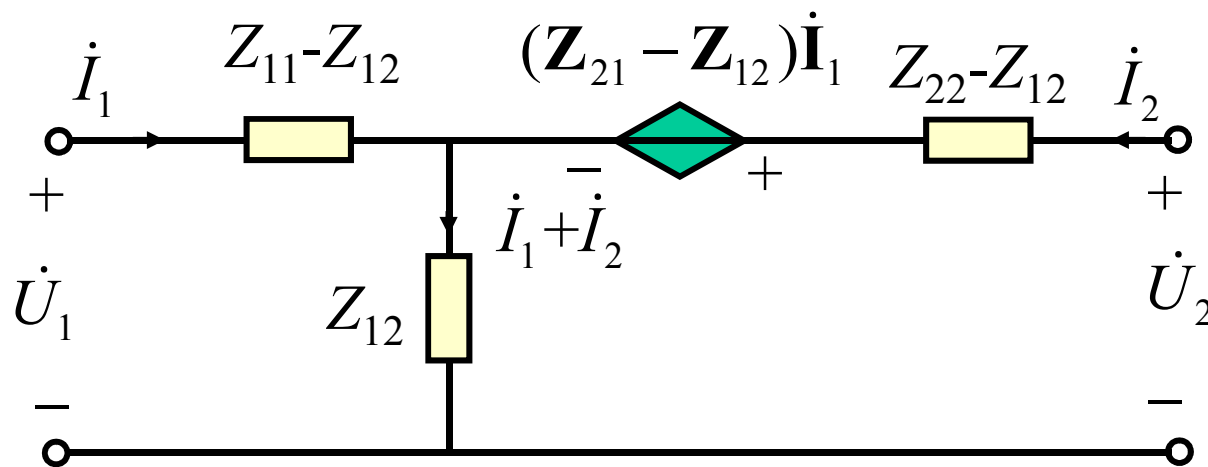
## 1由Z参数方程表示的等效模型

方法2：采用等效变换的方法。

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

$$\dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 + Z_{12}\dot{I}_1 - Z_{12}\dot{I}_1 = (Z_{11} - Z_{12})\dot{I}_1 + Z_{12}(\dot{I}_1 + \dot{I}_2)$$

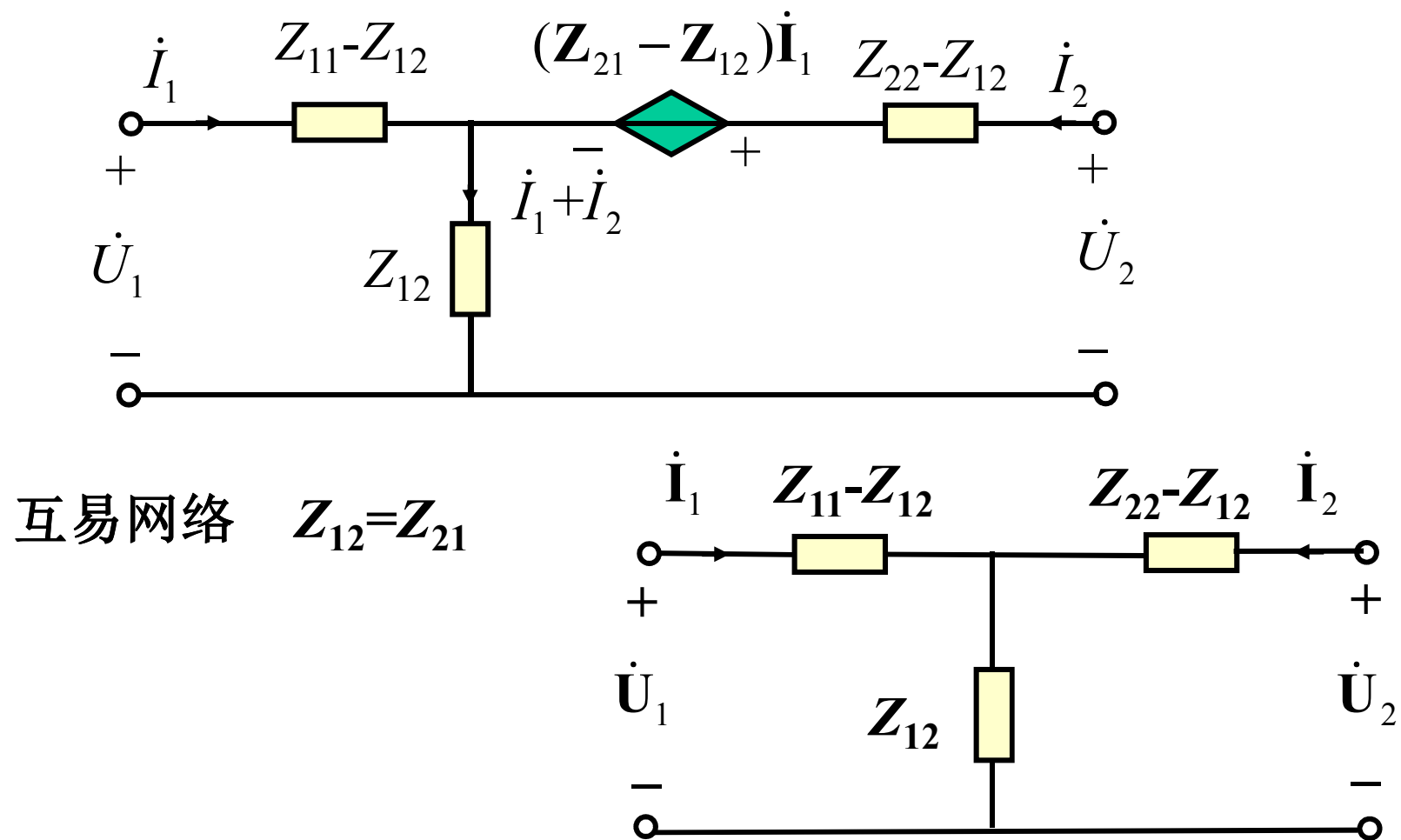
$$\begin{aligned} \dot{U}_2 &= Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 + Z_{12}\dot{I}_1 - Z_{12}\dot{I}_1 + Z_{12}\dot{I}_2 - Z_{12}\dot{I}_2 \\ &= (Z_{22} - Z_{12})\dot{I}_2 + Z_{12}(\dot{I}_1 + \dot{I}_2) + (Z_{21} - Z_{12})\dot{I}_1 \end{aligned}$$



同一个参数方程，可以画出结构不同的等效电路。

等效电路不唯一。

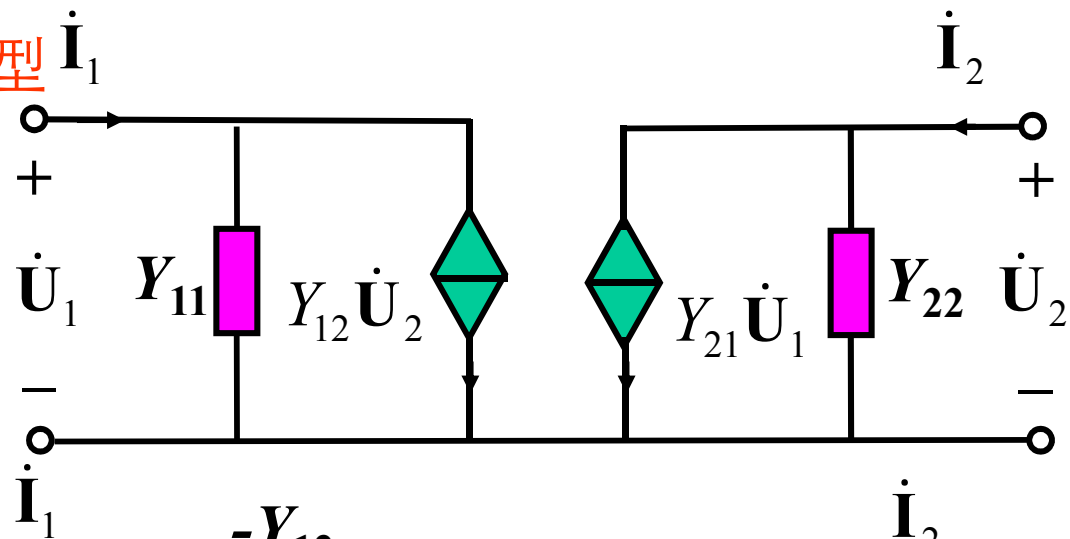
## 1由Z参数方程表示的等效模型



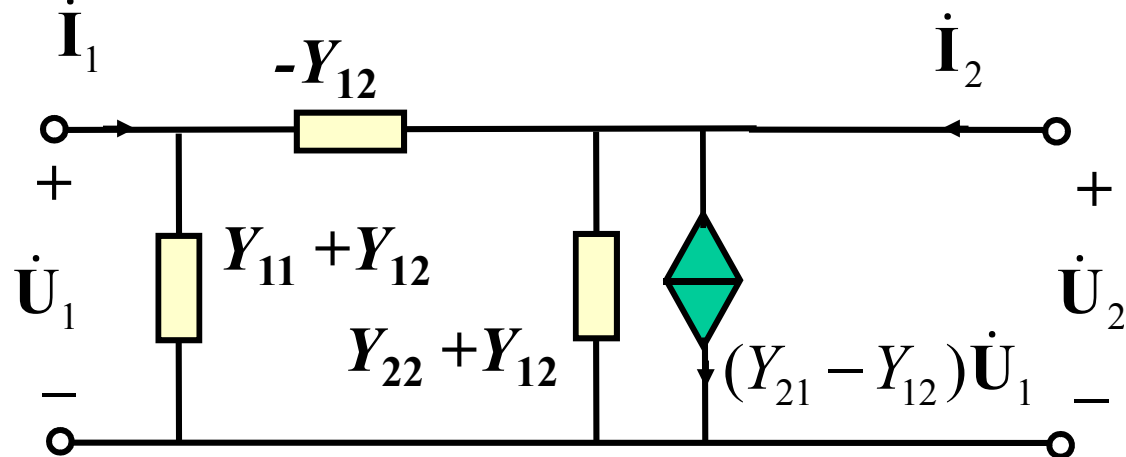
任何二端口网络都有T形等效模型。

## 2 由Y参数方程表示的等效模型

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

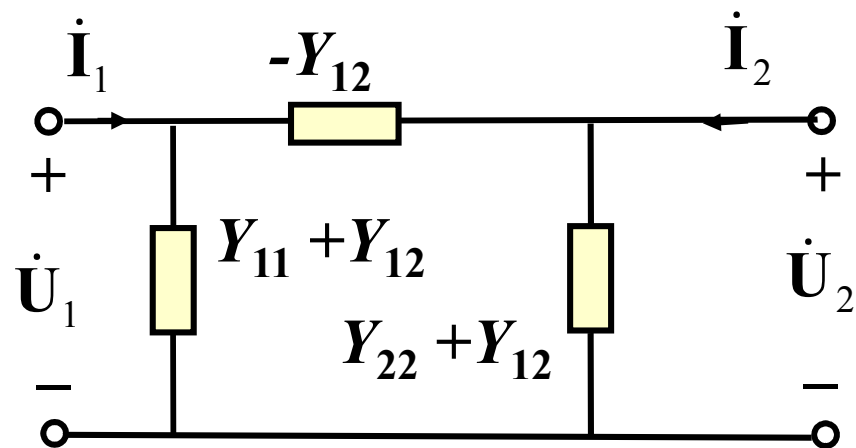


另一种形式



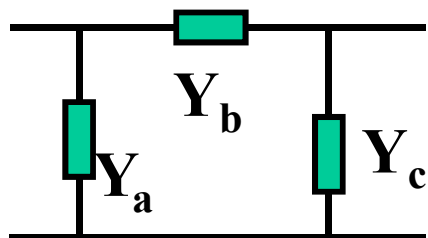
互易网络  $Y_{12}=Y_{21}$

任何二端口网络都有 $\pi$ 形等效模型。

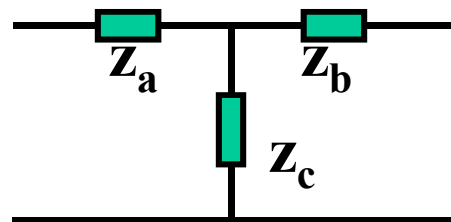




### 3 互易二端口的等效电路



$\pi$ 型等效电路



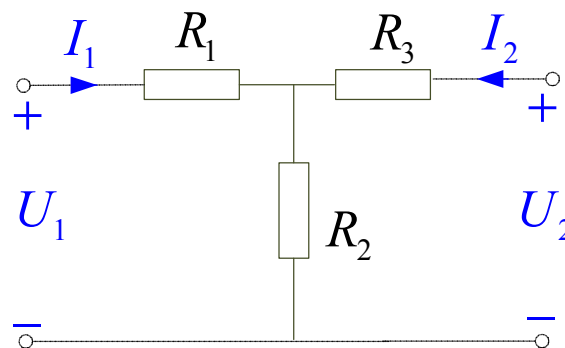
T型等效电路

例：电路如图所示，已知互易二端口网络的正向传输参数为

$$T = \begin{bmatrix} 1.5 & 2.5 \\ 0.5 & 1.5 \end{bmatrix}, \text{ 求该网络的T形等效电路。}$$

解：  $1.5 \times 1.5 - 0.5 \times 2.5 = 1$

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_1 \end{bmatrix} = \begin{bmatrix} 1.5 & 2.5 \\ 0.5 & 1.5 \end{bmatrix} \begin{bmatrix} \dot{U}_2 \\ -\dot{I}_2 \end{bmatrix}$$



$$A = \left. \frac{\dot{U}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} = \frac{R_1 + R_2}{R_2}$$

$$B = \left. \frac{\dot{U}_1}{-\dot{I}_2} \right|_{\dot{U}_2=0} = \frac{\dot{U}_1}{\frac{\dot{U}_1}{R_1 + R_2 // R_3} \times \frac{R_2}{R_2 + R_3}} \bigg|_{\dot{U}_2=0}$$

$$C = \left. \frac{\dot{I}_1}{\dot{U}_2} \right|_{\dot{I}_2=0} = \frac{1}{R_2} = \frac{R_1(R_2 + R_3) + R_2 R_3}{R_2}$$

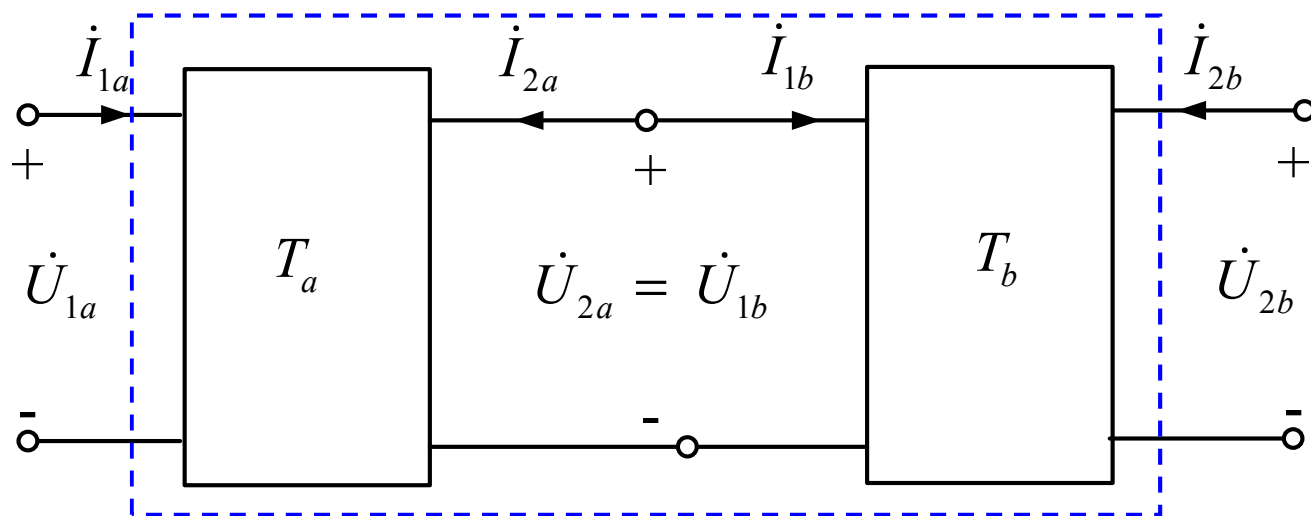
比较系数得  $R_1=1\Omega$ ,  $R_2=2\Omega$ ,  $R_3=1\Omega$

## 16.5 二端口网络的相互连接

- ▶ 二端口网络可以在端口处相互联结构成复杂网络。
- ▶ 可以是串联、并联和级联。

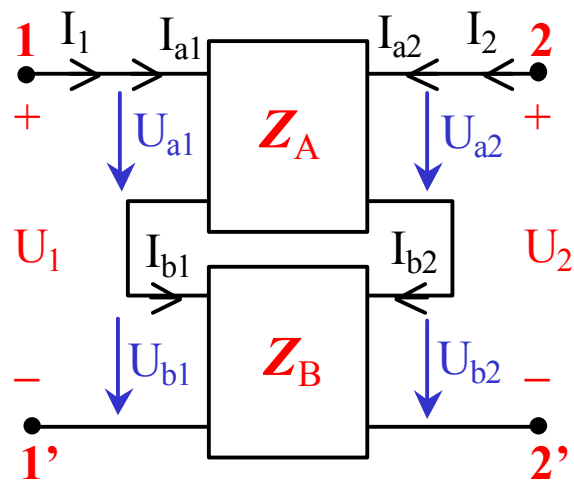
### 16.5.1 级联 Cascade connection

- ▶ 一个网络的输出是另一个网络的输入端。



$$\begin{bmatrix} \dot{U}_{1a} \\ \dot{I}_{1a} \end{bmatrix} = \mathbf{T}_a \begin{bmatrix} \dot{U}_{2a} \\ -\dot{I}_{2a} \end{bmatrix} = \mathbf{T}_a \begin{bmatrix} \dot{U}_{1b} \\ \dot{I}_{1b} \end{bmatrix} = \mathbf{T}_a \times \mathbf{T}_b \begin{bmatrix} \dot{U}_{2b} \\ -\dot{I}_{2b} \end{bmatrix} \quad \mathbf{T} = \mathbf{T}_a \times \mathbf{T}_b$$

## 16.5.2 串联 (Series Connection)

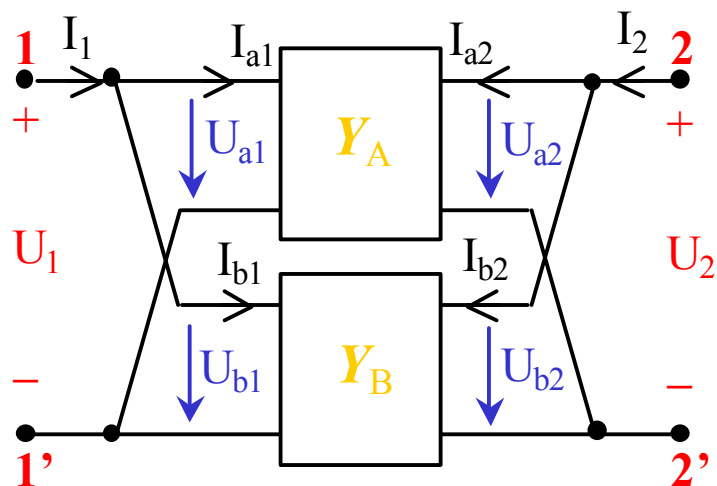


$$I_{a1} = I_{b1} = I_1, \quad I_{a2} = I_{b2} = I_2$$

$$\begin{aligned} \begin{bmatrix} U_1 \\ U_2 \end{bmatrix} &= \begin{bmatrix} U_{a1} \\ U_{a2} \end{bmatrix} + \begin{bmatrix} U_{b1} \\ U_{b2} \end{bmatrix} = \mathbf{Z}_A \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} + \mathbf{Z}_B \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} \\ &= (\mathbf{Z}_A + \mathbf{Z}_B) \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \mathbf{Z} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} \end{aligned}$$

$$\mathbf{Z} = \mathbf{Z}_A + \mathbf{Z}_B$$

### 16.5.3 并联 (Parallel Connection)



$$U_{a1} = U_{b1} = U_1, \quad U_{a2} = U_{b2} = U_2$$

$$\begin{aligned} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} &= \begin{bmatrix} I_{a1} \\ I_{a2} \end{bmatrix} + \begin{bmatrix} I_{b1} \\ I_{b2} \end{bmatrix} = \mathbf{Y}_A \begin{bmatrix} U_1 \\ U_2 \end{bmatrix} + \mathbf{Y}_B \begin{bmatrix} U_1 \\ U_2 \end{bmatrix} \\ &= (\mathbf{Y}_A + \mathbf{Y}_B) \begin{bmatrix} U_1 \\ U_2 \end{bmatrix} = \mathbf{Y} \begin{bmatrix} U_1 \\ U_2 \end{bmatrix} \end{aligned}$$

$$\mathbf{Y} = \mathbf{Y}_A + \mathbf{Y}_B$$

【例1】一电阻二端口N，其传输参数矩阵为  $T = \begin{bmatrix} 2 & 8\Omega \\ 0.5\text{S} & 2.5 \end{bmatrix}$ ，

(1)求其T型等效电路

(2)若端口1接 $U_s=6\text{V}$ 、 $R_1=2\Omega$ 的串联支路，端口2接电阻 $R$ （1），求 $R=?$ 时可使其上获得最大功率，并求此最大功率值。

(3)若端口1接电压源 $u_s=6+10\sin t$  V与电阻 $R_1=2\Omega$ 的串联支路，端口2接 $L=1\text{H}$ 与 $C=1\text{F}$ 的串联支路（图2），求电容 $C$ 上电压的有效值。

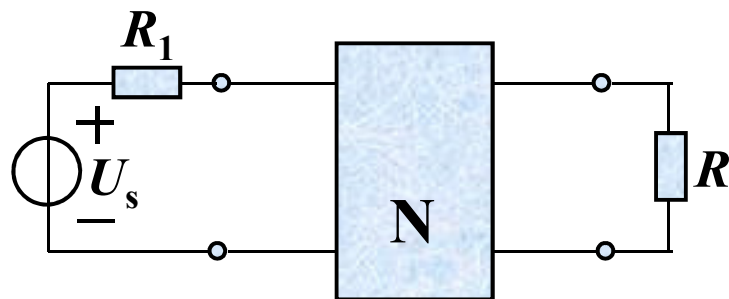


图1

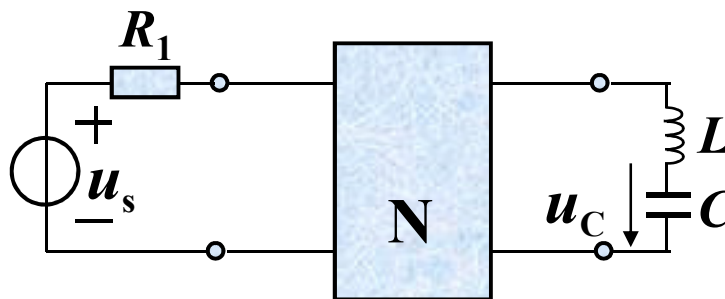
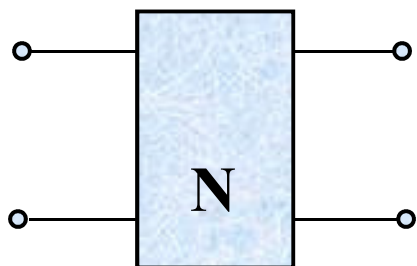
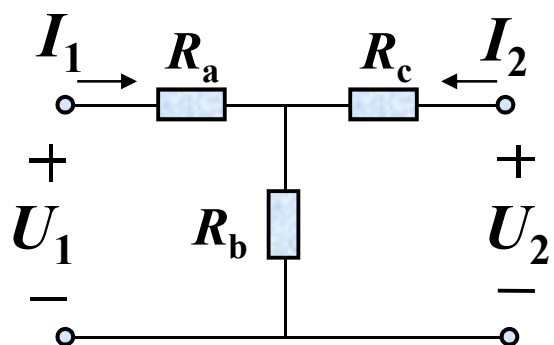


图2

# (1) 求T型等效电路



$$T = \begin{bmatrix} 2 & 8\Omega \\ 0.5\text{ S} & 2.5 \end{bmatrix} \quad \begin{cases} U_1 = AU_2 - BI_2 \\ I_1 = CU_2 - DI_2 \end{cases}$$

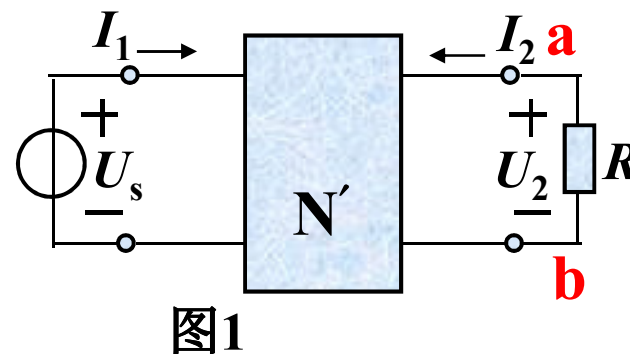
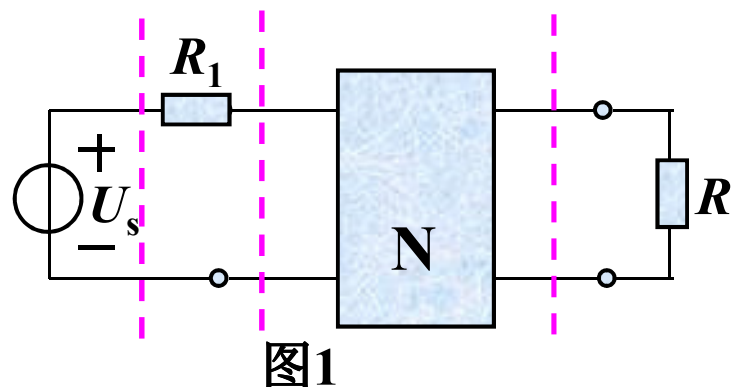


$$\begin{cases} A = \frac{U_1}{U_2} \Big|_{I_2=0} = \frac{R_a + R_b}{R_b} = 2 \\ C = \frac{I_1}{U_2} \Big|_{I_2=0} = \frac{1}{R_b} = 0.5 \\ D = \frac{I_1}{-I_2} \Big|_{U_2=0} = \frac{R_c + R_b}{R_b} = 2.5 \end{cases}$$

解之得：

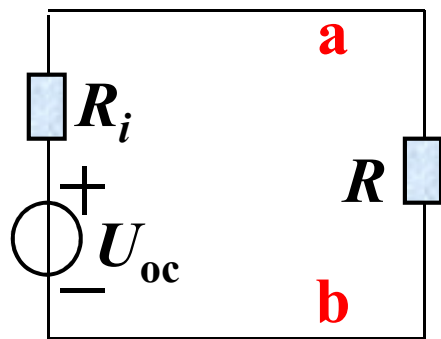
$$\begin{cases} R_a = 2\Omega \\ R_b = 2\Omega \\ R_c = 3\Omega \end{cases}$$

(2)若端口1接 $U_s=6V$ 、 $R_1=2\Omega$ 的串联支路，端口2接电阻 $R$ （图1），求 $R=?$ 时可使其上获得最大功率，并求此最大功率值。



$$T = \begin{bmatrix} 2 & 8\Omega \\ 0.5S & 2.5 \end{bmatrix} \quad T' = \begin{bmatrix} 1 & 2 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 2 & 8 \\ 0.5 & 2.5 \end{bmatrix} = \begin{bmatrix} 3 & 13\Omega \\ 0.5S & 2.5 \end{bmatrix} \quad \begin{cases} U_s = 3U_2 - 13I_2 \\ I_1 = 0.5U_2 - 2.5I_2 \end{cases}$$

计算a、b以左电路的戴维南等效电路：



$$U_{oc} = U_2 \Big|_{I_2=0} = \frac{U_s}{3} = \frac{6}{3} = 2V \quad R_i = \frac{U_2}{I_2} \Big|_{U_s=0} = \frac{13}{3} = 4.33\Omega$$

$R$  等于 $R_i$ 时其上功率最大，此时最大功率为：

$$P_{\max} = \frac{U_{oc}^2}{4R_i} = \frac{4}{4 \times 4.33} = 0.231W$$



(3)若端口1接电压源 $u_s=6+10\sin t$  V与电阻 $R_1=2\Omega$ 的串联支路，端口2接 $L=1\text{H}$ 与 $C=1\text{F}$ 的串联支路（图2），求电容 $C$ 上电压的有效值。

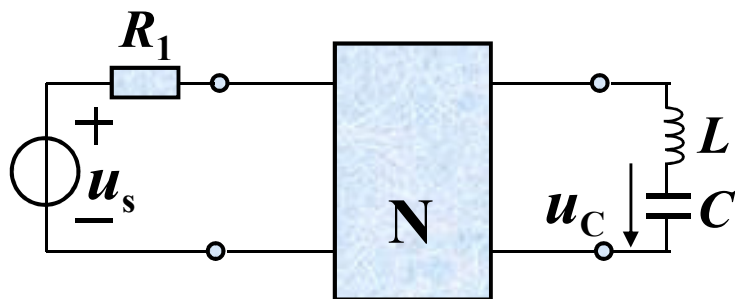
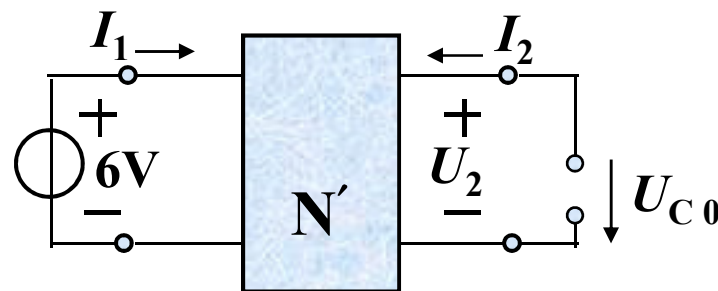


图2

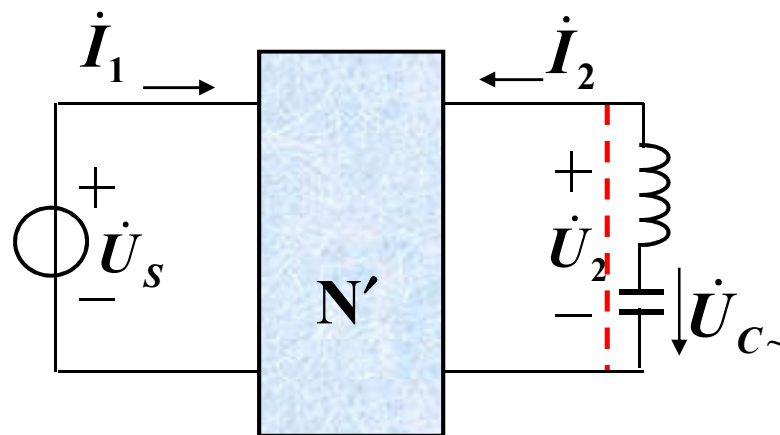


(1) 6V电压源单独作用， $L$ 短路、 $C$ 开路

$$\begin{cases} 6 = 3U_2 - 13I_2 \\ I_1 = 0.5U_2 - 2.5I_2 \end{cases} \quad U_{c0} = U_2 \big|_{I_2=0} = 2V$$

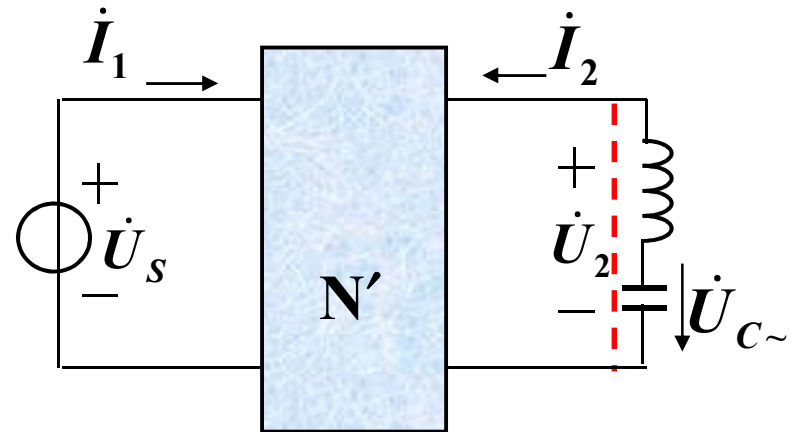
(2) 正弦电源单独作用， $L$ 、 $C$ 发生串联谐振相当于短（ $U_2=0$ ）

$$\begin{cases} \dot{U}_s = 3\dot{U}_2 - 13\dot{I}_2 & \dot{U}_s = \frac{10}{\sqrt{2}} \angle 0^\circ \text{ V} \\ \dot{I}_1 = 0.5\dot{U}_2 - 2.5\dot{I}_2 & \dot{U}_2 = 0 \end{cases}$$



$$\begin{cases} \dot{U}_s = 3\dot{U}_2 - 13\dot{I}_2 & \dot{U}_s = 10/\sqrt{2} \angle 0^\circ \text{ V} \\ \dot{I}_1 = 0.5\dot{U}_2 - 2.5\dot{I}_2 & \dot{U}_2 = 0 \end{cases}$$

$$\dot{I}_2 = \frac{\dot{U}_s}{-13} = -\frac{7.07 \angle 0^\circ}{13} = 0.544 \angle 180^\circ \text{ A}$$



$$\dot{U}_c = -j \frac{1}{\omega C} (-\dot{I}_2) = 0.544 \angle 90^\circ \text{ V} \quad u_c(t) = 0.544\sqrt{2} \sin(t + 90^\circ) \text{ V}$$

$$u_c = U_{c0} + u_c(t) = 2 + 0.544\sqrt{2} \sin(t + 90^\circ) \text{ V}$$

**有效值**  $U_c = \sqrt{2^2 + 0.544^2} = \sqrt{4.296} = 2.073 \text{ A}$

**计划学时：5学时；课后学习15学时**

**作业：**

**16-2/含源二端口端口方程**

**16-11,16-16,16-26 /参数**

**16-30/级联**

**16-36， 16-38/综合应用**