Continuous-Time Fourier Transform

- The frequency-domain representation of a periodic signal composed of a weighted sum of sinusoidal signals with harmonically related frequencies is given by its Continuous-Time Fourier Series
- A generalization of this representation is given by the Continuous-Time Fourier Transform (CTFT)

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Continuous-Time Fourier **Transform**

• The CTFT represents the frequency-domain representation of a certain class of aperiodic analog signals as a weighted combination of infinite number of complex exponentials whose frequencies are infinitesimally close to each other with the integral replaced by a

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Continuous-Time Fourier Transform

Definition

• The continuous-time Fourier transform of an aperiodic signal x(t) is given by

$$X(j\Omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\Omega t}dt$$

• The signal x(t) can be recovered from its CTFT $X(j\Omega)$ by the Fourier integral $x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\Omega) e^{j\Omega t} d\Omega$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\Omega) e^{j\Omega t} d\Omega$$

Continuous-Time Fourier **Transform**

- Note: Ω is a real variable denoting the continuous-time angular frequency in radians per sec.
- The inverse CTFT $x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\Omega) e^{j\Omega t} d\Omega$ can be interpreted as a linear weighted sum with weights $X(j\Omega)$ of complex exponentials of the form

$$\frac{1}{2\pi}e^{j\Omega t}d\Omega$$

Continuous-Time Fourier Transform

• We shall denote the CTFT and the inverse CTFT pair in compact form as

$$x(t) \overset{\text{CTFT}}{\Leftrightarrow} X(j\Omega)$$

Example – The CTFT $\mathcal{D}(i\Omega)$ of the unit impulse function $\delta(t)$ is given by

$$\mathcal{D}(j\Omega) = \int_{-\infty}^{+\infty} \delta(t)e^{-j\Omega t}dt = 1$$

Continuous-Time Fourier **Transform**

Example – The CTFT $X(j\Omega)$ of the timeshifted unit impulse function $x(t) = \delta(t - t_o)$

$$X(j\Omega) = \int_{-\infty}^{+\infty} \delta(t - t_o) e^{-j\Omega t} dt = e^{-j\Omega t_o}$$

Continuous-Time Fourier Transform

Example – We determine the CTFT $X(j\Omega)$ of the real-valued analog signal

$$x(t) = e^{-\alpha t} \mu(t)$$
, $0 < \alpha < \infty$

$$X(j\Omega) = \int_{-\infty}^{+\infty} e^{-\alpha t} \mu(t) e^{-j\Omega t} dt = \int_{0}^{+\infty} e^{-\alpha t} e^{-j\Omega t} dt$$
$$= \int_{-\infty}^{+\infty} e^{-(\alpha + j\Omega)t} dt = \frac{1}{\alpha + j\Omega}$$

Commonly Used CTFT Pairs

 $\delta(t) \stackrel{\text{CTFT}}{\Leftrightarrow} 1$

 $1 \stackrel{\text{CTFT}}{\Leftrightarrow} 2\pi \delta(\Omega)$

 $\delta(t-t_o) \stackrel{\text{CTFT}}{\Leftrightarrow} e^{-j\Omega t_o}$

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Commonly Used CTFT Pairs

$$\mu(t) \overset{\text{CTFT}}{\Longleftrightarrow} \pi \delta(\Omega) + \frac{1}{j\Omega}$$

$$e^{-\alpha t}\mu(t) \stackrel{\text{CTFT}}{\Longleftrightarrow} \frac{1}{\alpha + j\Omega}, \mathcal{R}e\{\alpha\} > 0$$

$$e^{j\Omega_o t} \stackrel{\text{CTFT}}{\Leftrightarrow} 2\pi\delta(\Omega - \Omega_o)$$

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Continuous-Time Fourier Transform

- The CTFT of an analog signal is sometimes referred to as the Fourier spectrum, or simply, the spectrum
- In general, the CTFT $X(j\Omega)$ of an analog signal x(t) is a complex function of the real variable Ω in the range $-\infty < \Omega < \infty$ and can be expressed either in rectangular or polar form

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Parts of CTFT

Magnitude and Phase Spectra

• The complex CTFT in polar form is given by

$$X(j\Omega) = |X(j\Omega)|e^{j\theta(\Omega)}$$

where $\theta(\Omega) = \arg\{X(\Omega)\}\$

- $|X(j\Omega)|$ is called the magnitude spectrum
- $\theta(\Omega)$ is called the phase spectrum

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Parts of CTFT

- $|X(j\Omega)|$ and $\theta(\Omega)$ are real functions of Ω
- For a real analog signal

$$|X(-j\Omega)| = |X(j\Omega)|$$

 $\theta(-\Omega) = -\theta(\Omega)$

• Note: Phase spectra can not be uniquely specified for all values of Ω as demonstrated next

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Parts of CTFT

- Define a new CTFT $X_1(j\Omega) = X(j\Omega)e^{j2\pi k}$
- Now, $X_1(j\Omega) = |X(j\Omega)|e^{j\theta(\Omega)}e^{j2\pi k}$ $=|X(j\Omega)|e^{j\theta(\Omega)}=X(j\Omega)$
- For uniqueness, $\theta(\Omega)$ is restricted to the range

$$-\pi < \theta(\Omega) \le \pi$$

called the principal value of phase function

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Parts of CTFT

Real and Imaginary Parts

• The complex CTFT in rectangular form is given by

$$X(j\Omega) = X_{re}(j\Omega) + jX_{im}(j\Omega)$$

where $X_{re}(j\Omega)$ and $X_{im}(j\Omega)$ are real and imaginary parts of $X(j\Omega)$, and are real functions of Ω

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Parts of CTFT

• For a real analog signal,

$$\begin{split} X_{re}(-j\Omega) &= X_{re}(j\Omega) \\ X_{im}(-j\Omega) &= -X_{im}(j\Omega) \end{split}$$

Example – Consider the CTFT $X(j\Omega) = \frac{1}{\alpha + i\Omega}$ of $e^{-\alpha t}\mu(t)$, $\Re(\alpha) > 0$

• Here,
$$|X(j\Omega)| = \left|\frac{1}{\alpha + j\Omega}\right| = \frac{1}{\sqrt{\alpha^2 + \Omega^2}}$$

$$\frac{|\alpha(j\Omega)|}{|\alpha+j\Omega|} = \frac{1}{\sqrt{\alpha^2 + \Omega^2}}$$

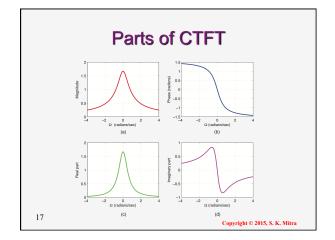
$$\theta(\Omega) = -\tan^{-1}(\Omega/\alpha)$$

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Parts of CTFT

- Note: $|X(-j\Omega)| = \frac{1}{\sqrt{\alpha^2 + (-\Omega)^2}} = \frac{1}{\sqrt{\alpha^2 + \Omega^2}} = |X(j\Omega)|$ $\theta(-\Omega) = -\tan^{-1}(-\Omega/\alpha) = \tan^{-1}(\Omega/\alpha) = -\theta(\Omega)$
- Thus, $|X(j\Omega)|$ is an even function of Ω and $\theta(\Omega)$ is and odd function of Ω , as expected
- Plots of $|X(j\Omega)|$ and $\theta(\Omega)$ shown in the next slide for $\alpha = 0.6$ /sec also verifies the above result

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Parts of CTFT

• Next, we rewrite $X(j\Omega)$ as

XI, we rewrite
$$X(\Sigma)$$
 as
$$X(j\Omega) = \frac{1}{\alpha + j\Omega} = \frac{\alpha - j\Omega}{(\alpha + j\Omega)(\alpha - j\Omega)} = \frac{\alpha - j\Omega}{\alpha^2 + \Omega^2}$$

$$= \left(\frac{\alpha}{\alpha^2 + \Omega^2}\right) - j\left(\frac{\Omega}{\alpha^2 + \Omega^2}\right)$$

• Hence,
$$X_{re}(j\Omega) = \frac{\alpha}{\alpha^2 + \Omega^2}$$

 $X_{im}(j\Omega) = -\frac{\Omega}{\alpha^2 + \Omega^2}$

Parts of CTFT

Note:

$$\begin{split} X_{re}(-j\Omega) &= \frac{\alpha}{\alpha^2 + (-\Omega)^2} = \frac{\alpha}{\alpha^2 + \Omega^2} = X_{re}(j\Omega) \\ X_{im}(-j\Omega) &= -\frac{(-\Omega)}{\alpha^2 + (-\Omega)^2} = \frac{\Omega}{\alpha^2 + \Omega^2} = -X_{im}(j\Omega) \end{split}$$

• The above properties of the real and imaginary parts can also be seen from the plots given in Slide No. 17 for $\alpha = 0.6/\text{sec}$

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CTFT Computation Using MATLAB

- The function freqs can be used to evaluate a rational CTFT
- The code fragments used to develop the plots shown in Slide No. 17 are given in the next slide

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CTFT Computation Using MATLAB

```
w = -4:0.01:4;
h = freqs(1,[1 0.6],w);
mag = abs(h);
phase = angle(h);
Re = real(h);
Im = imag(h);
```

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t = 0

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Uniform Convergence Conditions

Given by Dirichlet conditions

- (1) x(t) is absolutely integrable
- (2) x(t) has a finite number of maxima and minima, and a finite number of discontinuities

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Uniform Convergence Conditions

Example - $x(t) = e^{-\alpha t}\mu(t)$, $\Re\{\alpha\} > 0$, is absolutely integrable and it has one discontinuity at t = 0 and one maximum at



• Thus, its CTFT $X(j\Omega) = \frac{1}{\alpha + j\Omega}$ converges uniformly

Mean-Square Convergence Conditions

- A finite energy x(t) is square integrable but not absolutely integrable
- Its CTFT exists in a mean-square sense
- Let $X(j\Omega)$ denotes its CTFT with $\bar{x}(t)$ denoting its inverse

• Then $\int_{-\infty}^{+\infty} |\bar{x}(\tau) - x(\tau)|^2 d\tau \to 0$

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CTFT Using Delta Functions

• The CTFT can also be defined using ideal delta functions for a number of analog signals that are neither absolutely summable nor square-summable

Example – Consider x(t) with a CTFT given by

$$X(j\Omega) = 2\pi\delta(\Omega - \Omega_o)$$

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CTFT Using Delta Functions

• The inverse CTFT of the above CTFT is thus given by

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} 2\pi \delta(\Omega - \Omega_o) e^{j\Omega t} dt = e^{j\Omega_o t}$$

obtained using the sampling property of the delta function

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CTFT Properties

- The CTFT has a number of properties that are often useful in determining the CTFTs of various analog signals that one may encounter in some applications
- We list these properties without any proofs

$$x(t) \overset{\text{CTFT}}{\Leftrightarrow} X(j\Omega)$$

$$y(t) \stackrel{\text{CTFT}}{\Leftrightarrow} Y(j\Omega)$$

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CTFT Properties

- Linearity: $\alpha x(t) + \beta y(t) \stackrel{\text{CTFT}}{\iff} \alpha X(j\Omega) + \beta Y(j\Omega)$
- Conjugation: $x^*(t) \stackrel{\text{CTFT}}{\leftrightarrow} X^*(j\Omega)$
- Time-reversal: $x(-t) \stackrel{\text{CTFT}}{\Leftrightarrow} X(-j\Omega)$
- Time-shifting: $x(t-t_o) \stackrel{\text{CTFT}}{\leftrightarrow} e^{-j\Omega t_o} X(j\Omega)$

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CTFT Properties

• Frequency-shifting:

$$e^{j\Omega_o t} x(t) \stackrel{\text{CTFT}}{\Leftrightarrow} X(j(\Omega - \Omega_o))$$

• Differentiation-in-frequency:

$$(-jt)x(t) \overset{\text{CTFT}}{\longleftrightarrow} \frac{X(j\Omega)}{d\Omega}$$

• Differentiation-in-time: $\frac{dx(t)}{dt} \stackrel{\text{CTFT}}{\Leftrightarrow} j\Omega X(j\Omega)$

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CTFT Properties

• Integration:

$$\textstyle \int_{-\infty}^t x(\tau) d\tau \overset{\text{CTFT}}{\Longleftrightarrow} \frac{dX(j\Omega)}{j\Omega} + \pi X(0) \delta(\Omega)$$

- Example Recall $\mu(t) = \int_{-\infty}^{t} \delta(\tau) d\tau$
- Now the CTFT of $\delta(t)$ is given by

$$\mathcal{D}(j\Omega) = 1$$

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CTFT Properties

• Using the integration property of the CTFT we arrive at the CTFT of $\mu(t)$:

$$\mathcal{M}(j\Omega) = \frac{\mathcal{D}(j\Omega)}{j\Omega} + \pi \mathcal{D}(0)\delta(\Omega)$$
$$= \frac{1}{j\Omega} + \pi \delta(\Omega)$$

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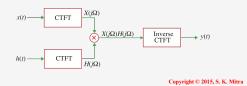
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CTFT Properties

• Convolution:

$$x(t) \circledast y(t) \Leftrightarrow X(j\Omega)Y(j\Omega)$$

• Implementation of the convolution integral using the CTFT-based approach



CTFT Properties

• Modulation:

$$x(t)y(t) \overset{\text{CTFT}}{\Longleftrightarrow} \frac{1}{2\pi} \int\limits_{-\infty}^{\infty} X(j\Psi) Y\big(j(\Omega - \Psi)\big) d\Psi$$

· Parseval's Relation:

$$\int_{-\infty}^{+\infty} |g(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |G(j\Omega)|^2 d\Omega$$

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CTFT Properties

• Example – We determine the inverse CTFT x(t) of

 $X(j\Omega) = \begin{cases} 1, & |\Omega| \le \Omega_c \\ 0, & |\Omega| > \Omega_c \end{cases}$ • The total energy of X(t) obtained using the Parseval's relation is

$$E_x = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\Omega)|^2 d\Omega = \frac{1}{2\pi} \int_{-\Omega_c}^{\Omega_c} d\Omega = \frac{\Omega_c}{\pi}$$

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CTFT Properties

- Hence, x(t) is a square-integrable analog
- The inverse CTFT of $X(j\Omega)$ is $x[n] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\Omega) e^{j\Omega t} d\Omega$ $= \frac{1}{2\pi} \int_{-\Omega}^{\Omega_c} e^{j\Omega t} d\Omega = \frac{\sin(\Omega_c t)}{\pi t}$

$$= \frac{1}{2\pi} \int_{\Omega}^{\Omega_c} e^{j\Omega t} d\Omega = \frac{\sin(\Omega_c t)}{\pi t}$$

• Note: x(t) is not absolutely integrable

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Band-Limited Analog Signals

 An ideal band-limited analog signal has a CTFT $X(j\Omega)$ that is zero outside a finite frequency range $\Omega_1 \leq |\Omega| \leq \Omega_2$, that is

$$X(j\Omega) = \begin{cases} 0, & 0 \leq \left|\Omega\right| < \Omega_1 \\ 0, & \Omega_2 < \left|\Omega\right| < \infty \end{cases}$$

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Band-Limited Analog Signals

- · An analog signal is said to have a bandlimited spectrum if most of its energy is concentrated in a given finite frequency range
- Lowpass Analog Signal has most of its energy in the low frequency range

$$0 \le |\Omega| < \Omega_c < \infty$$

The frequency range from dc to Ω_c is called its bandwidth

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Band-Limited Analog Signals

• Highpass Analog Signal – has most of its energy in the high frequency range

$$0 < \Omega_c \le |\Omega| < \infty$$

The frequency range from Ω_c to ∞ is called its bandwidth

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Band-Limited Analog Signals

• Bandpass Analog Signal – has most of its energy in the mid frequency range

$$0<\Omega_{c1}\leq \left|\Omega\right|\leq \Omega_{c2}<\infty$$

The frequency range $\Omega_{c2} - \Omega_{c1}$ is called its bandwidth

• A precise definition of the bandwidth is given by the percent of the total energy in the specified frequency range which depends on applications