

ECE 863

Analysis of Stochastic Systems

Part IV.1: Power Spectral Density
of Random Processes

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ECE 863

- Reading Assignment
- Section 7.1 – Power Spectral Density
- Section 7.2 – Linear Systems
- Section 7.4 – Optimum Linear Systems

ECE 863: Part IV.1
Page 2
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ECE 863

- Exam 3 is on:
- Wednesday, December 5
- Chapter 7 reading assignment and related
lecture notes are the last material included
in Exam 3

ECE 863: Part IV.1
Page 3
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Power Spectral Density

- The “power spectral density” $S_X(f)$ measures the average energy (or power) of a random process $X(t)$ at the frequency (f)
- Therefore, if the random process $X(t)$ “changes slowly”, then $X(t)$ must have most of its power concentrated at low frequencies, (and vice versa).

ECE 863: Part IV.1
Page 4
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Power Spectral Density

- A slowly (or rapidly) changing random process implies a highly-correlated (or highly-uncorrelated) process across the time axis.
- For a random process we need to measure the "average change"
- The autocorrelation R_X provides a measure of the "average change" for a random process

Power Spectral Density

- Therefore, we expect the power spectral density $S_X(f)$ to be related to the autocorrelation function $R_X(\tau)$, where $X(t)$ is a WSS process.
- We can derive the power spectral density S_X from R_X by using the Fourier Transform

Power Spectral Density

- The power spectral density $S_X(f)$ of the WSS process X_n is the discrete-time Fourier Transform of the autocorrelation $R_X(d)$:

$$S_X(f) = \sum_{d=-\infty}^{\infty} R_X(d) e^{-j2\pi f d}$$

Power Spectral Density

- For any discrete-time WSS random process X_n , the power spectral density $S_X(f)$ is a periodic function of (f) with a period of one:

$$S_X(f+1) = \sum_{d=-\infty}^{\infty} R_X(d) e^{-j2\pi(f+1)d}$$

$$S_X(f+1) = \sum_{d=-\infty}^{\infty} R_X(d) e^{-j2\pi f d} e^{-j2\pi d}$$

$$S_X(f+1) = \sum_{d=-\infty}^{\infty} R_X(d) e^{-j2\pi f d} = S_X(f)$$

Power Spectral Density

- The power spectral density $S_X(f)$ is a "non-negative" function (it is a measure of power):

$$S_X(f) \geq 0 \quad \forall f$$

Power Spectral Density

- The autocorrelation $R_X(d)$ can be computed as the inverse Fourier Transform of the power spectral density function $S_X(f)$
- Since, for discrete-time processes, $S_X(f)$ is a periodic function of (f) . Therefore,

$$R_X(d) = \int_{-1/2}^{1/2} S_X(f) e^{j2\pi f d} df$$

Power Spectral Density

- For a continuous-time WSS process $X(t)$, the power spectral density $S_X(f)$ is the Fourier Transform of the autocorrelation function $R_X(\tau)$:

$$S_X(f) = \int_{-\infty}^{\infty} R_X(\tau) e^{-j2\pi f \tau} d\tau$$

This is known as the Wiener-Khinchin Theorem

Power Spectral Density

- The autocorrelation function $R_X(\tau)$ for a continuous-time WSS process $X(t)$ is the inverse Fourier Transform of the power spectral density $S_X(f)$:

$$R_X(\tau) = \int_{-\infty}^{\infty} S_X(f) e^{j2\pi f \tau} df$$

Power Spectral Density

- Since $S_X(f)$ is the power spectral density, then integrating $S_X(f)$ over all possible frequencies gives the total "average power":

$$\text{Total "Average Power"} = \int_{-\infty}^{\infty} S_X(f) df$$

$$E[(X(t))^2] = \int_{-\infty}^{\infty} S_X(f) df$$

$$R(\tau=0) = \int_{-\infty}^{\infty} S_X(f) df$$

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Page 13

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Power Spectral Density

- If $X(t)$ is a real (i.e. not a complex) process, then $R_X(\tau)$ is a real and symmetric function:

$$R_X(\tau) = R_X(-\tau)$$

- Therefore, the power spectral density $S_X(f)$ is also real and symmetric:

$$S_X(f) = S_X(-f)$$

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Page 14

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Power Spectral Density

- For a real process $X(t)$, the expression for $S_X(f)$ can be simplified by taking advantage of the symmetry of the autocorrelation function $R_X(\tau)$:

$$S_X(f) = \int_{-\infty}^{\infty} R_X(\tau) \cos(2\pi f\tau) d\tau$$

$$-j \int_{-\infty}^{\infty} R_X(\tau) \sin(2\pi f\tau) d\tau$$

$$S_X(f) = \int_{-\infty}^{\infty} R_X(\tau) \cos(2\pi f\tau) d\tau$$

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Page 15

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Example: White Noise

- The "white noise" process $X'(t)$ is generated by taking the derivative of a Wiener process $X(t)$:

$$X'(t) = \frac{dX(t)}{dt}$$

We need to find the power spectral density $S_{X'}(f)$ of $X'(t)$

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Page 16

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Example: White Noise

- Finding the power spectral density of the "white noise" process $X'(t)$ requires computing its autocorrelation function $R_{X'}(t_1, t_2)$
- Since $X'(t)$ is the derivative of $X(t)$, we need to know how to take derivatives of random processes

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Page 17

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Example: White Noise

- Under certain conditions, the autocorrelation function $R_{X'}(\tau)$ of $X'(t)=dX(t)/dt$ can be obtained by taking the partial derivatives of the autocorrelation $R_X(\tau)$ of $X(t)$:

If
$$X'(t) = \frac{dX(t)}{dt}$$

$$\Rightarrow R_{X'}(t_1, t_2) = \frac{\partial^2}{\partial t_1 \partial t_2} R_X(t_1, t_2)$$

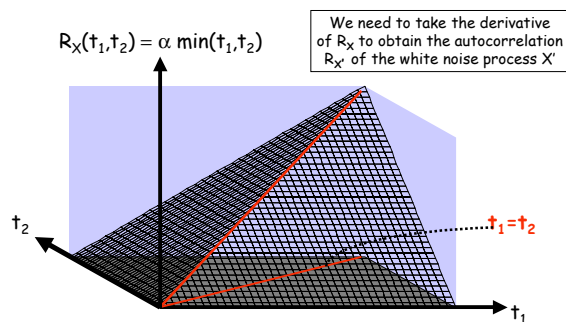
(See Appendix B for more details.)

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Page 18

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Example: White Noise



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Page 19

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Example: White Noise

- Recall that the autocorrelation function for the Wiener process:

$$R_X(t_1, t_2) = \alpha \min(t_1, t_2)$$

For the Wiener process, the partial derivative of $R_X(t_1, t_2)$:

$$R_{X'}(t_1, t_2) = \frac{\partial^2}{\partial t_1 \partial t_2} R_X(t_1, t_2)$$

is zero everywhere except when $t_1 = t_2$

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Page 20

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Example: White Noise

- This leads to the following expression for the autocorrelation $R_{X'}(t_1, t_2)$ of the "white noise" process:

$$R_{X'}(t_1, t_2) = \alpha \delta(t_1 - t_2)$$

$$R_{X'}(\tau) = \alpha \delta(\tau)$$

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Page 21

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Example: White Noise

- Therefore, the white noise process is a WSS process
- Hence, now, we can compute the power spectral density $S_{X'}(f)$:

$$S_{X'}(f) = \int_{-\infty}^{\infty} R_{X'}(\tau) \cos(2\pi f\tau) d\tau$$

$$S_{X'}(f) = \int_{-\infty}^{\infty} \alpha \delta(\tau) \cos(2\pi f\tau) d\tau$$

$$S_{X'}(f) = \alpha \quad \forall f$$

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Page 22

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Example: White Noise

- Therefore, the "white noise" process $X'(t)$ has a "flat" power spectral density that covers all frequencies.
- Consequently, the "white noise" process has infinite power!

$$E[(X'(t))^2] = \int_{-\infty}^{\infty} S_{X'}(f) df$$

$$E[(X'(t))^2] = R(\tau=0) = \alpha \delta(0) = \alpha \int_{-\infty}^{\infty} df = \infty$$

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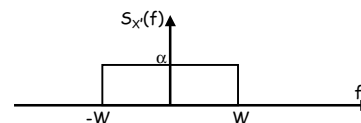
Page 23

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Example: Bandlimited White Noise

- A "bandlimited white noise" process $X'(t)$ has a "flat" power spectral density over a certain range of frequencies: $[-W, W]$

$$S_{X'}(f) = \alpha \quad \forall f \in [-W, W]$$



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Page 24

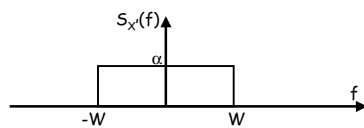
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Example: Bandlimited White Noise

- The "bandlimited white noise" process $X'(t)$ has finite power:

$$E[(X'(t))^2] = \int_{-W}^W S_{X'}(f) df$$

$$E[(X'(t))^2] = \alpha \int_{-W}^W df \quad \boxed{E[(X'(t))^2] = 2W\alpha}$$



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Page 25

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Example: Bandlimited White Noise

- The autocorrelation function $R_{X'}(\tau)$ of the bandlimited white noise process can be computed using the inverse Fourier Transform:

$$R_{X'}(\tau) = \alpha \int_{-W}^W e^{j2\pi f\tau} df = \alpha \int_{-W}^W \cos(2\pi f\tau) df$$

$$\boxed{R_{X'}(\tau) = \frac{(2\alpha)\sin(2\pi W\tau)}{2\pi\tau}}$$

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Page 26

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Example: Bandlimited White Noise

- The autocorrelation function $R_{X'}(\tau)$:

$$R_{X'}(\tau) = \frac{(2\alpha)\sin(2\pi W\tau)}{2\pi\tau}$$

is zero at periodic values of τ :

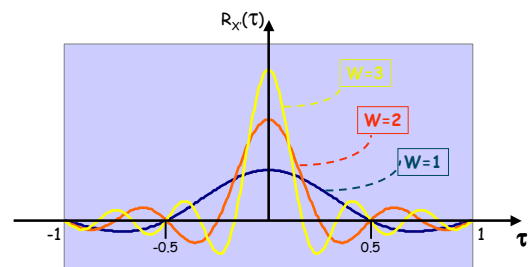
$$R_{X'}(\tau) = 0 \quad \forall \tau = \pm \frac{k}{2W}, \quad k = 1, 2, 3, \dots$$

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Page 27

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Example: Bandlimited White Noise

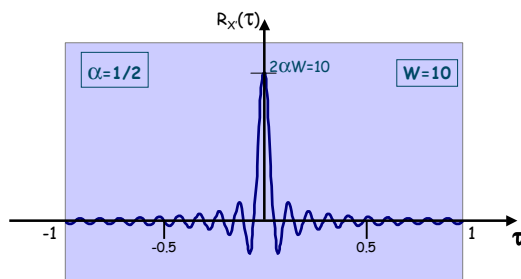


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Page 28

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Example: Bandlimited White Noise



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Page 29

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Example: Discrete-time White Noise

- Let X_n be a zero-mean discrete-time process with the following autocorrelation function:

$$R_X(k) = \sigma_X^2 \delta_k \quad \delta_k = \begin{cases} 1 & k = 0 \\ 0 & k \neq 0 \end{cases}$$

This is a "discrete time" white-noise process

Find the power spectral density $S_X(f)$

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Page 30

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Example: Discrete-time White Noise

- First, we note that X_n is a WSS process. Since:

$$m_X(k) = \text{constant} = 0$$

and $R_X(k) = \sigma_X^2 \delta_k$

The auto-correlation function $R_X(k)$ is a function of the time difference (k) only

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Page 31

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Example: Discrete-time White Noise

- Therefore, by using:

$$S_X(f) = \sum_{k=-\infty}^{\infty} R_X(k) e^{-j2\pi f k}$$

$$S_X(f) = \sum_{k=-\infty}^{\infty} (\sigma_X^2 \delta_k) e^{-j2\pi f k}$$

$$S_X(f) = \sigma_X^2 \quad \text{For all } -\frac{1}{2} < f < \frac{1}{2}$$

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Page 32

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Cross-Power Spectral Density

- Two processes $X(t)$ and $Y(t)$ have a

"cross-power spectral density" $S_{XY}(f)$

if $X(t)$ and $Y(t)$ are

jointly WSS processes

Cross-Power Spectral Density

- $X(t)$ and $Y(t)$ are jointly WSS processes when the following conditions are satisfied:

$$1) R_X(t_1, t_2) = R_X(t_1 - t_2) = R_X(\tau) \quad \text{i.e. } X(t) \text{ is WSS}$$

$$2) R_Y(t_1, t_2) = R_Y(t_1 - t_2) = R_Y(\tau) \quad \text{i.e. } Y(t) \text{ is WSS}$$

$$3) R_{XY}(t_1, t_2) = R_{XY}(t_1 - t_2) = R_{XY}(\tau)$$

Cross-Power Spectral Density

- If $X(t)$ and $Y(t)$ are jointly WSS processes, then:

$$R_{XY}(\tau) = E[X(t+\tau)Y(t)]$$

$$R_{YX}(\tau) = E[Y(t+\tau)X(t)] = E[X(t)Y(t+\tau)] = R_{XY}(-\tau)$$

$$R_{YX}(\tau) = R_{XY}(-\tau)$$

When computing cross-functions for wide sense stationary processes, we have to use a consistent definition for τ :

$$\tau = t_1 - t_2 \quad \text{OR} \quad \tau = t_2 - t_1$$

Here, and in the book, we use the convention: $\tau = t_1 - t_2$

Cross-Power Spectral Density

- If $X(t)$ and $Y(t)$ are continuous-time jointly WSS processes, then the cross-power spectral densities $S_{XY}(f)$ and $S_{YX}(f)$ are:

$$S_{XY}(f) = \int_{-\infty}^{\infty} R_{XY}(\tau) e^{-j2\pi f\tau} d\tau$$

$$S_{YX}(f) = \int_{-\infty}^{\infty} R_{YX}(\tau) e^{-j2\pi f\tau} d\tau$$

Cross-Power Spectral Density

- Since for jointly WSS processes (discrete- or continuous-time):

$$R_{yx}(\tau) = R_{xy}(-\tau)$$

then

$$S_{yx}(f) = S_{xy}^*(f)$$

where $S_{xy}^*(f)$ is the complex conjugate of $S_{xy}(f)$

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Page 37

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Cross-Power Spectral Density

- Similarly, if X_n and Y_n are discrete-time jointly WSS processes, then the cross-power spectral densities $S_{xy}(f)$ and $S_{yx}(f)$ are:

$$S_{xy}(f) = \sum_{d=-\infty}^{\infty} R_{xy}(d) e^{-j2\pi f d} \quad R_{xy}(d) = E[X_{n+d} Y_n]$$

$$S_{yx}(f) = \sum_{d=-\infty}^{\infty} R_{yx}(d) e^{-j2\pi f d} \quad R_{yx}(d) = E[Y_{n+d} X_n]$$

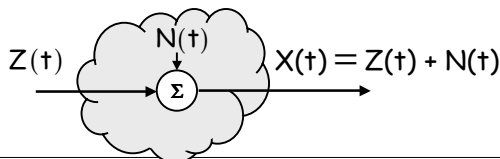
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Page 38

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Example: Additive Noise

- A signal $Z(t)$ is corrupted by an additive noise process $N(t)$.
- $Z(t)$ and $N(t)$ are jointly WSS processes with PSD functions $S_Z(f)$, $S_N(f)$, and $S_{ZN}(f)$
- Derive an expression for the PSD $S_X(f)$ of the received process $X(t)$ and the cross-PSD function $S_{ZX}(f)$



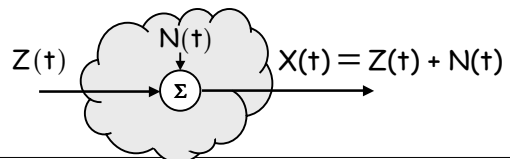
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Page 39

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Example: Additive Noise

- First, we evaluate the PSD function $S_X(f)$
- We start by using the definition of the autocorrelation function $R_X(\tau)$
- Then, we take the Fourier Transform of $R_X(\tau)$ to express $S_X(f)$



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Page 40

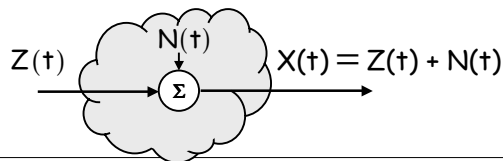
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Example: Additive Noise

- The autocorrelation function $R_X(\tau)$

$$R_X(\tau) = E[X(t+\tau)X(t)]$$

$$R_X(\tau) = E[(Z(t+\tau) + N(t+\tau))(Z(t) + N(t))]$$



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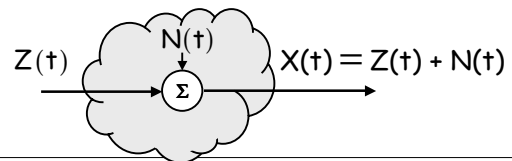
Page 41

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Example: Additive Noise

$$R_X(\tau) = E[(Z(t+\tau) + N(t+\tau))(Z(t) + N(t))]$$

$$R_X(\tau) = E[Z(t+\tau)Z(t) + N(t+\tau)Z(t) + Z(t+\tau)N(t) + N(t+\tau)N(t)]$$



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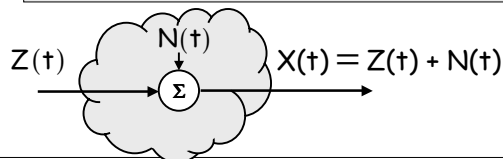
Page 42

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Example: Additive Noise

$$R_X(\tau) = E[Z(t+\tau)Z(t) + N(t+\tau)Z(t) + Z(t+\tau)N(t) + N(t+\tau)N(t)]$$

$$\Rightarrow R_X(\tau) = R_Z(\tau) + R_{NZ}(\tau) + R_{ZN}(\tau) + R_N(\tau)$$



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Page 43

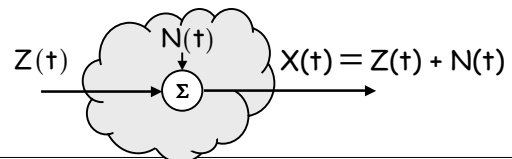
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Example: Additive Noise

- Now, by taking the Fourier transform of $R_X(\tau)$:

$$R_X(\tau) = R_Z(\tau) + R_{NZ}(\tau) + R_{ZN}(\tau) + R_N(\tau)$$

$$\Rightarrow S_X(f) = S_Z(f) + S_{NZ}(f) + S_{ZN}(f) + S_N(f)$$



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Page 44

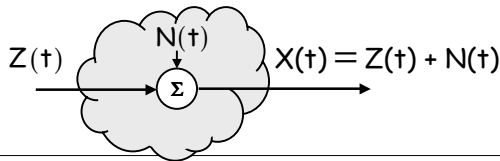
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Example: Independent Additive Noise

- IF $Z(t)$ and $N(t)$ are independent, then:

$$R_{NZ}(\tau) = E[N(t+\tau)Z(t)]$$

$$= E[N(t+\tau)]E[Z(t)]$$



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Page 45

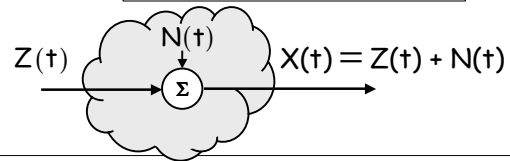
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Example: Independent Additive Noise

- Since $Z(t)$ and $N(t)$ are WSS processes, then they have constant means m_Z and m_N :

$$R_{NZ}(\tau) = m_N m_Z \quad \text{Similarly,} \quad R_{ZN}(\tau) = m_Z m_N$$

Therefore, $R_{NZ}(\tau) + R_{ZN}(\tau) = 2 m_Z m_N$



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Page 46

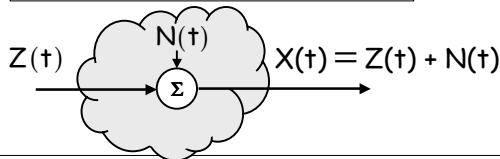
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Example: Independent Additive Noise

- Therefore, when $Z(t)$ and $N(t)$ are independent, then:

$$R_X(\tau) = R_Z(\tau) + R_{NZ}(\tau) + R_{ZN}(\tau) + R_N(\tau)$$

$$\Rightarrow R_X(\tau) = R_Z(\tau) + 2m_Z m_N + R_N(\tau)$$



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Page 47

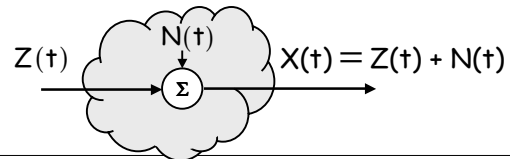
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Example: Independent Additive Noise

- Moreover, IF $N(t)$ is a zero-mean process:

$$R_X(\tau) = R_Z(\tau) + 2m_Z m_N + R_N(\tau)$$

$$\Rightarrow R_X(\tau) = R_Z(\tau) + R_N(\tau)$$



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Page 48

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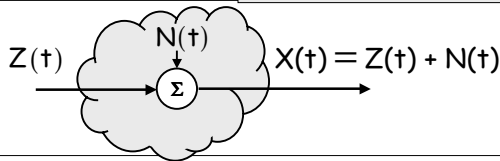
Example: Additive Noise

■ In summary:

$$R_X(\tau) = R_Z(\tau) + R_{NZ}(\tau) + R_{ZN}(\tau) + R_N(\tau)$$

$$R_X(\tau) = R_Z(\tau) + 2m_Z m_N + R_N(\tau) \quad \text{For independent noise}$$

$$R_X(\tau) = R_Z(\tau) + R_N(\tau) \quad \text{For independent \& zero-mean noise}$$



ECE 863: Part IV.1

Page 49

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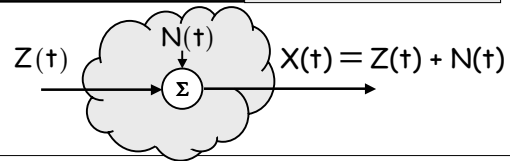
Example: Additive Noise

■ Consequently:

$$S_X(f) = S_Z(f) + S_{NZ}(f) + S_{ZN}(f) + S_N(f)$$

$$S_X(f) = S_Z(f) + 2m_Z m_N \delta(f) + S_N(f) \quad \text{independent noise}$$

$$S_X(f) = S_Z(f) + S_N(f) \quad \text{independent \& zero-mean noise}$$



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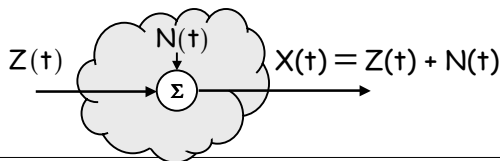
Page 50

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Example: Additive Noise

- Similarly, we can use the definition of the cross-correlation function $R_{ZX}(\tau)$ to derive the cross-PSD function $S_{ZX}(f)$

$$R_{ZX}(\tau) = E[Z(t + \tau)X(t)]$$



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Page 51

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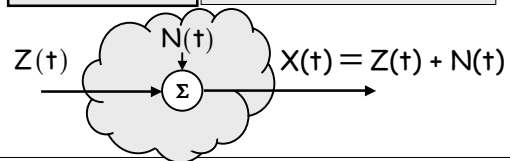
Example: Additive Noise

- As an exercise, show the following :

$$R_{ZX}(\tau) = R_Z(\tau) + R_{ZN}(\tau)$$

$$R_{ZX}(\tau) = R_Z(\tau) + m_Z m_N \quad \text{For independent noise}$$

$$R_{ZX}(\tau) = R_Z(\tau) \quad \text{For independent \& zero-mean noise}$$



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Page 52

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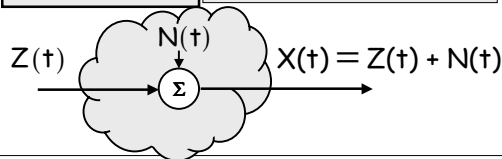
Example: Additive Noise

■ Therefore:

$$S_{ZX}(f) = S_Z(f) + S_{ZN}(f)$$

$$S_{ZX}(f) = S_Z(f) + m_Z m_N \delta(f) \quad \text{For independent noise}$$

$$S_{ZX}(f) = S_Z(f) \quad \text{For independent \& zero-mean noise}$$



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Page 53

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Appendix A

Derivation of the Power Spectral Density

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Page 54

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Power Spectral Density

- We start by looking at the "periodogram" of a discrete-time WSS random process X_n
- If X_0, X_1, \dots, X_{k-1} are k samples from the WSS process X_n , then we can take the discrete-time Fourier Transform of these samples:

$$\tilde{x}_k(f) = \sum_{m=0}^{k-1} X_m e^{-j2\pi f m}$$

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Page 55

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Power Spectral Density

- To estimate the average power that the random process X_n has at frequency (f) , we use the "periodogram estimate" :

$$\tilde{p}_k(f) = \frac{1}{k} |\tilde{x}_k(f)|^2$$

where

$$\tilde{x}_k(f) = \sum_{m=0}^{k-1} X_m e^{-j2\pi f m}$$

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Page 56

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Power Spectral Density

- Taking the expected value of the periodogram:

$$E[\tilde{p}_k(f)] = E\left[\frac{1}{k} |\tilde{x}_k(f)|^2\right]$$

$$E[\tilde{p}_k(f)] = \frac{1}{k} E[\tilde{x}_k(f) \tilde{x}_k^*(f)]$$

ECE 863: Part IV.1

Page 57

Prof. Hayder Radha

Power Spectral Density

$$\begin{aligned} E[\tilde{p}_k(f)] &= \frac{1}{k} E[\tilde{x}_k(f) \tilde{x}_k^*(f)] \\ &= \frac{1}{k} E\left[\sum_{m=0}^{k-1} X_m e^{-j2\pi f m} \sum_{m'=0}^{k-1} X_{m'} e^{j2\pi f m'}\right] \\ &= \frac{1}{k} \sum_{m=0}^{k-1} \sum_{m'=0}^{k-1} E[X_m X_{m'}] e^{-j2\pi f(m-m')} \end{aligned}$$

ECE 863: Part IV.1

Page 58

Prof. Hayder Radha

Power Spectral Density

- Using $E[X_m X_{m'}] = R_X[m, m']$

$$E[\tilde{p}_k(f)] = \frac{1}{k} \sum_{m=0}^{k-1} \sum_{m'=0}^{k-1} E[X_m X_{m'}] e^{-j2\pi f(m-m')}$$

$$= \frac{1}{k} \sum_{m=0}^{k-1} \sum_{m'=0}^{k-1} R_X(m, m') e^{-j2\pi f(m-m')}$$

ECE 863: Part IV.1

Page 59

Prof. Hayder Radha

Power Spectral Density

- Since X_n is a WSS process, then: $R_X[m, m'] = R_X[m - m']$:

$$E[\tilde{p}_k(f)] = \frac{1}{k} \sum_{m=0}^{k-1} \sum_{m'=0}^{k-1} R_X(m - m') e^{-j2\pi f(m-m')}$$

ECE 863: Part IV.1

Page 60

Prof. Hayder Radha

Power Spectral Density

- If we let $d=m-m'$, then the double summation:

$$E[\tilde{p}_k(f)] = \frac{1}{k} \sum_{m=0}^{k-1} \sum_{m'=0}^{k-1} R_X(m-m') e^{-j2\pi f(m-m')}$$

reduces to the following single summation

$$E[\tilde{p}_k(f)] = \frac{1}{k} \sum_{d=-(k-1)}^{k-1} \{k-|d|\} R_X(d) e^{-j2\pi f d}$$

ECE 863: Part IV.1

Page 61

Prof. Hayder Radha

Power Spectral Density

- Therefore:

$$E[\tilde{p}_k(f)] = \frac{1}{k} \sum_{d=-(k-1)}^{k-1} \{k-|d|\} R_X(d) e^{-j2\pi f d}$$

$$E[\tilde{p}_k(f)] = \sum_{d=-(k-1)}^{k-1} \left\{1 - \frac{|d|}{k}\right\} R_X(d) e^{-j2\pi f d}$$

ECE 863: Part IV.1

Page 62

Prof. Hayder Radha

Power Spectral Density

- Note that we are only considering k samples of the random process X_n
- Considering all samples of the process X_n leads to the power spectral density $S_X(f)$:

$$\begin{aligned} S_X(f) &= \lim_{k \rightarrow \infty} E[\tilde{p}_k(f)] \\ &= \lim_{k \rightarrow \infty} \sum_{d=-(k-1)}^{k-1} \left\{1 - \frac{|d|}{k}\right\} R_X(d) e^{-j2\pi f d} \end{aligned}$$

ECE 863: Part IV.1

Page 63

Prof. Hayder Radha

Power Spectral Density

- Therefore, the power spectral density $S_X(f)$ of the WSS process X_n is the discrete-time Fourier Transform of the autocorrelation $R_X(d)$:

$$S_X(f) = \sum_{d=-\infty}^{\infty} R_X(d) e^{-j2\pi f d}$$

ECE 863: Part IV.1

Page 64

Prof. Hayder Radha

Power Spectral Density

- Moreover, the power spectral density $S_X(f)$ is a "non-negative" function since it results from the "periodogram estimate":

$$S_X(f) = \lim_{k \rightarrow \infty} E[\tilde{p}_k(f)] = \lim_{k \rightarrow \infty} E\left[\frac{1}{k} |\tilde{x}_k(f)|^2\right]$$

$$S_X(f) \geq 0 \quad \forall f$$

Appendix B

Derivatives of Random Processes

Derivative of Random Processes

- The derivative of any random process $X(t)$:

$$X'(t) = \frac{dX(t)}{dt}$$

$$X'(t) = \lim_{\varepsilon \rightarrow 0} \frac{X(t + \varepsilon) - X(t)}{\varepsilon}$$

may exist for some sample functions and may not exist for other sample functions.

Derivative of Random Processes

- Therefore, it is customary to define the "Mean Square Derivative":

$X'(t)$ is the "mean square derivative" of $X(t)$ if the following is satisfied:

$$\lim_{\varepsilon \rightarrow 0} E\left[\left(\frac{X(t + \varepsilon) - X(t)}{\varepsilon} - X'(t)\right)^2\right] = 0$$

Derivative of Random Processes

- It can be shown that if the "mean square derivative" exists, then the autocorrelation function of $X'(t)$ can be obtained by taking the partial derivative of the autocorrelation of $X(t)$:

$$R_{X'}(t_1, t_2) = \frac{\partial^2}{\partial t_1 \partial t_2} R_X(t_1, t_2)$$

(This is a general result that is true
for any process which has a mean-square-derivative;
See section 6.6 for more details.)