

HW 05 - Mat 320

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11/24/25

1 Inverse Comb Filter

$$y_t = x_t - R^L x_{t-L} \quad R = 0.999 \quad L = 6$$

$$a) H(z) = 1 - R^L z^{-L}$$

$$b) H(\omega) = 1 - R^L e^{j6\omega} = 1 - 0.999 e^{j6\omega}$$

$$c) |H(\omega)| = 1 - R^L (\cos(6\omega) - j \sin(6\omega)) = 1 - 0.999 (\cos(6\omega) - j \sin(6\omega))$$

$$d) H(z) = 1 - R^L z^{-L} = \frac{z^L - R^L}{z^L} \quad z^L = R^L e^{j6\omega}$$

$$e^{j6\omega} = \cos(6\omega) + j \sin(6\omega)$$

// Zeros @ $z^L = R^L$
 $z = R e^{j2\pi/L \cdot k}$

$$z = R \cdot e^{j2\pi/L \cdot k} = R e^{j\pi/3 \cdot k}$$

$$k=0 \rightarrow R \quad k=2 \rightarrow R e^{j2\pi/3} \quad k=4 \rightarrow R e^{j4\pi/3} \quad \frac{2\pi}{\pi/3} = 6 \text{ roots}$$

$$k=1 \rightarrow R e^{j\pi/3} \quad k=3 \rightarrow R e^{j\pi} \quad k=5 \rightarrow R e^{j5\pi/3}$$

$$H(z) = (z - R) (z - R e^{j\pi/3}) (z - R e^{j2\pi/3}) (z - e^{j\pi}) (z - R e^{j4\pi/3}) (z - R e^{j5\pi/3})$$

$$\hookrightarrow H(z) = \prod_{k=0}^5 (z - R e^{j\pi/3 \cdot k})$$

$$H(z)^2 = \left[\frac{(z^6 - R^6)}{z^6} \right]^2 = \frac{(z^6 - R^6)^2}{z^{12}}$$

$$\frac{d}{d\omega} H(\omega)^2 = \frac{d}{d\omega} (1 - R^6 e^{j6\omega})^2 = \frac{d}{d\omega} (1 - R^6 \cos(6\omega) - j \sin(6\omega))^2$$

$$\frac{d}{d\omega} (1 - R^6 \cos(6\omega) - j \sin(6\omega)) = (1 + R^6 \cos(6\omega))$$

$$= 1 - R^6 \cos(6\omega) + j R^6 \sin(6\omega) \quad // \text{conjugate: } (a+bs)(a-bs) = a^2 + b^2$$

$$H(\omega)^* = 1 - R^6 \cos(6\omega) - j R^6 \sin(6\omega)$$

$$(1 - R^6 \cos(6\omega))^2 + (R^6 \sin(6\omega))^2$$

$$(1 - 2R^6 \cos(6\omega) + R^{12} \cos^2(6\omega)) + (R^{12} \sin^2(6\omega))$$

$$1 - 2R^6 \cos(6\omega) + R^{12} (\cos^2(6\omega) + \sin^2(6\omega))$$

$$\frac{d}{d\omega} (1 - 2R^6 \cos(6\omega) + R^{12}) = \frac{d}{d\omega} (1 - 2R^6 \cos(u) + R^{12}) \quad // \frac{d}{d\omega} (\cos(6\omega)) = -\sin(6\omega) \cdot 6$$

$$-2R^6 \cdot (-\sin(6\omega) \cdot 6) = 12R^6 \sin(6\omega)$$

$$\text{Min} \rightarrow \sin(6\omega) = 0$$

$$6\omega = \pi n$$

$$\omega = \pi/6 n$$

① (d) factor: $H(z) = (z-R)(z-Re^{j\pi/3})(z-Re^{j2\pi/3})(z-e^{j\pi})(z-Re^{j4\pi/3})(z-Re^{j5\pi/3})$

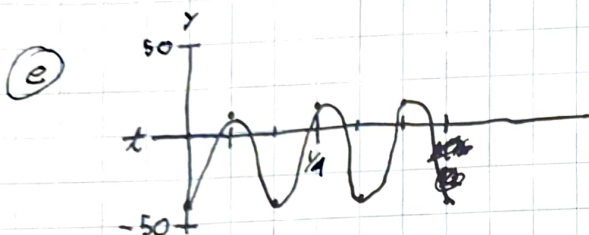
At $\omega = 0$

$H(z)^{2n} = 12R^6 \sin(\omega)$, Critical @ $\pi/6n$

$H(z)^{2n} = 72R^6 \cos(\omega)$, Min @ $2\pi n$, Max @ $\pi/6n$

$\omega \in [0, \pi]$ (fraction of sample rate, dB):

$\omega(0): (0, -44.5)$ $\omega(\pi/6): (1/2, 5.99)$ $\omega(\pi/3): (1/3, -44.5)$
 $\omega(\pi/2): (1/4, 5.99)$ $\omega(2\pi/3): (2/3, -44.5)$ $\omega(5\pi/6): (5/2, 5.99)$
 $\omega(\pi): (1/2, -44.5)$



③ $H(\omega) = 1 - R^6 e^{-j6\omega} = 1 - R^6 (\cos(6\omega) - jR^6 \sin(6\omega)) = 1 - R^6 \cos(6\omega) + jR^6 \sin(6\omega)$

$\theta_\omega = \tan^{-1} \left(\frac{R^6 \sin(6\omega)}{1 - R^6 \cos(6\omega)} \right)$

Phase ④ Min $2\pi n: 0$
Max $\pi/6n: 0$

Phase response at extrema makes sense because $H(z)' = 1$

$$\boxed{2} \quad f_s = 44100 \text{ Hz}$$

$$\boxed{I} \quad B = 2\phi, \quad \cos(\phi) = 2 - \frac{1}{2}(R - \frac{1}{R})$$

$$a) \quad B = 10 \text{ Hz} \quad \gamma = 1000 \text{ Hz}$$

$$\cos(\theta) = \frac{2R}{1+R^2} \cos(\gamma)$$

$$\theta = \frac{20\pi}{441} = 0.1425$$

$$\theta = \frac{2\pi\gamma}{f_s} \quad A_0 = (1-R)^2 \sin(\theta)$$

$$\phi = B/2 = 5 \text{ Hz} \quad \frac{2\pi \cdot 5}{44100} = \frac{\pi}{4410}$$

$$\cos(\phi) = 2 - \frac{1}{2}(R - \frac{1}{R})$$

$$2(2\cos(\phi)) = R - \frac{1}{R} \cdot \frac{R}{R} = 2(2\cos(\phi)) = \frac{R^2+1}{R}$$

$$2R(2\cos(\phi)) - R^2 + 1 = 0 \rightarrow -R^2 + 2(2\cos(\phi)) + 1 = 0$$

$$= R^2 + 2R + 1 = 0$$

$$R = 1.007, [0.9993]$$

$$A_0 = (1 - 0.9993^2) \cdot \sin\left(\frac{20\pi}{441}\right) = 2.022 \text{ E-4}$$

$$y_t = A_0 x_t + 2R \cos(\theta) y_{t-1} - R^2 y_{t-2} \quad \left| \begin{array}{l} A_0 = 2.022 \text{ E-4} \\ R = 0.9993 \\ \theta = 20\pi/441 \end{array} \right.$$

$$2R \cos(\theta) = 1.979$$

$$y_t = 2.022 \text{ E-4} x_t + 1.979 y_{t-1} - 0.9986 y_{t-2}$$

$$s_t: \quad y_0: 2.022 \text{ E-4} \quad y_1: 3.999 \text{ E-4} \quad y_2: 5.8934 \text{ E-4}$$

$$\boxed{II} \quad B = 2(1-R) \rightarrow B=2 \rightarrow 2R \rightarrow B=1 \rightarrow R$$

$$B/2 = 1-R \quad R = 0.9999 \quad \left[\frac{B_{Hz}}{f_s/2} = dS_s \right]$$

$$\theta = 0.1425$$

$$1 - B/2 = R \quad A_0 = 3.219 \text{ E-5}$$

$$y_t = 3.219 \text{ E-5} x_t + 1.979 y_{t-1} - 0.9997 y_{t-2}$$

$$s_t: \quad y_0: 3.219 \text{ E-5} \quad y_1: 6.373 \text{ E-5} \quad y_2: 9.397 \text{ E-5}$$

[2] ⑥

$$B = 20 \text{ Hz}$$

$$= 9.15 \text{E-24}$$

$$= \pi/2205$$

$$\gamma = 5000 \text{ Hz}$$

$$100\pi/441 = \gamma$$

[I]

$$B/2 = \phi$$

$$\phi = \frac{\pi/2205}{\pi/2205}$$

$$R^2 - (2 \cdot (2 - \cos(\phi)) + 1) = 0$$

$$R = 0.9993 \quad R = 0.9985$$

W₂₂

$$\cos(\theta) = \frac{2R}{1+R} \cos(\gamma)$$

$$\cos(\theta) = 0.7563$$

$$\theta = 0.7132$$

$$A_0 = (1-R^2) \cdot \sin(\theta)$$

$$y_t = 1.861 \text{E-3}$$

$$y_t = 1.861 \text{E-3} x_t + 1.5104 y_{t-1} - 0.9972 y_{t-2}$$

$$s_t: y_0 = 1.861 \text{E-3} \quad y_1 = 2.812 \text{E-3} \quad y_2 = 2.391 \text{E-3}$$

[II]

$$B = 2(1-R)$$

$$R = 0.99929$$

$$\cos(\theta) = 0.7563$$

$$\theta = 0.71279$$

$$A_0 = 9.314 \text{E-4}$$

$$2 \cdot R \cdot \cos(\theta) = 1.512$$

$$y_t = 9.314 \text{E-4} x_t + 1.512 y_{t-1} - 0.9986 y_{t-2}$$

$$s_0: y_0 = 9.314 \text{E-4} \quad y_1 = 1.408 \text{E-3} \quad y_2 = 1.199 \text{E-3}$$

$$[3] \quad y_t = \sqrt{2} (0.99^t) \sin\left(\frac{\pi}{4} (t+1)\right)$$

$$R = 0.99 \quad \theta = \pi/4$$

$$\cos(\pi/4) = \sqrt{2}/2$$

$$= \frac{2 \cdot \cos(\theta) \cdot R}{\sqrt{2} (0.99)} = \frac{2 \cdot \frac{\sqrt{2}}{2} \cdot 0.99}{\sqrt{2}} = 0.99$$

$$\therefore y_t = S_t + \sqrt{2} \cdot 0.99 \sqrt{2} y_{t-1} - 0.9801 y_{t-2}$$

$$[4] \quad R \approx e^{-B/2}, \text{ show very close to eqn 3.6:}$$

$$R \approx 1 - B/2$$

$$e^{-B/2} \approx 1 + \frac{(-B/2)}{1} + \frac{(-B/2)^2}{2} + \frac{(-B/2)^3}{6} + \dots$$

$$e^x = \sum_{k=0}^{\infty} \frac{x^k}{k!}$$

$$\approx 1 - \frac{B}{2} + \frac{B^2}{8} - \frac{B^3}{48} \dots$$

→ terms after $k=1$ are negligibly impactful on the pole radius approximation for very small bandwidths.

→ first-order approximation

\therefore can reasonably be approximated as

$$e^{-B/2} \approx 1 - B/2$$

Impulse response R^t can be written as $e^{t \ln(R)}$

For $R \approx 1$ (pole close to unit circle), $\ln(R) \approx R-1$

Using above $R \approx 1 - B/2$, $\ln(R) \approx (1 - B/2) - 1$

$$\ln(R) \approx -B/2 \quad \therefore e^{-B/2 t} \approx e^{t \ln(R)}$$

* Using Taylor Series expansion of \ln around 1:

$$\ln(1+u) = u - \frac{u^2}{2} + \frac{u^3}{3} \dots$$

$$u = x-1$$

$$\ln(R) = u$$

$$\ln(R) \approx R-1$$

→ first-order approximation

5 a) For what range of values of R & θ is the discrepancy greatest between the true peak frequency θ_0 & pole angle θ ?

- As R is very close to 1,

$\theta_0(\omega)$ is very close to θ

Peak frequency (ω)

$$\cos(\omega) = \frac{1+R^2}{2R} \cos(\theta)$$

\therefore the discrepancy occurs as R moves away from 1
can also vary by pole angle θ but is primarily affected by value of R .

b) When θ is small, θ_0 is shifted to higher or lower frequencies?

- As described on p. 91, $\cos(\omega)$ & $\cos(\theta)$ are used to make it easier to see the directly proportional relationship of θ_0 & θ .

When θ is small $\cos(\omega)$, θ_0 is shifted to lower frequencies.