

P.77Inverse Comb Filter $y_t = x_t - R^L x_{t-L}$

$R = 0.999 \quad L = 6$

(a) $H(z) = 1 - R^L z^{-L} \quad Y_t = e^{j\omega t} - R^L e^{j\omega(t-L)}$

(b) $H(\omega) = 1 - R^L e^{j\omega L} = 1 - R^L e^{j6\omega}$

(c) $|H(\omega)| = 1 - R^L (\cos(6\omega) - j\sin(6\omega)) = 1 - 0.99901 (\cos(6\omega) - j\sin(6\omega))$

(d) $H(z) = 1 - R^L z^6 \frac{z^6}{z^6} = \frac{z^6 - R^L}{z^6} \quad z^6 = R^L e^{j6\omega}$

~~$e^{j6\omega} = \cos(6\omega) + j\sin(6\omega)$~~

// zeroes (d) $z^L = R^L e^{j2\pi k / L} \quad k=0, 1, 2, 3, 4, 5$

$Z = R \cdot e^{j2\pi k / 6} = R e^{j\pi k / 3}$

$k=0 = R \quad k=2 = R e^{j2\pi/3} \quad k=4 = R e^{j4\pi/3} \quad k=1 = R e^{j\pi/3} \quad k=3 = R e^{j\pi} \quad k=5 = R e^{j5\pi/3}$

$H(z) = (z-R)(z-R e^{j\pi/3})(z-R e^{j5\pi/3})(z-R e^{j\pi}) (z-R e^{j11\pi/3}) (z-R e^{j7\pi/3}) (z-R e^{j13\pi/3})$

$\hookrightarrow H(z) = \prod_{k=0}^5 (z - R e^{j\pi k / 3})$

~~$H(z)^2 = (z-R)^2 \left[\frac{(z^6 - R^6)}{z^6} \right]^2 = \frac{(z^6 - R^6)^2}{z^{12}}$~~

$\frac{d}{d\omega} H(\omega)^2 = \frac{d}{d\omega} (1 - R^6 \cdot e^{j6\omega})^2 = \frac{d}{d\omega} (1 - R^6 \cdot \cos(6\omega) - j\sin(6\omega))^2$

~~$\frac{d}{d\omega} (1 - R^6 \cos(6\omega) - j\sin(6\omega)) = (1 + R^6 \cos(6\omega)) \rightarrow$~~

$= 1 - R^6 \cos^2(6\omega) + jR^6 \sin^2(6\omega) \quad // \text{conjugate: } (a+bi)(a-bi) = a^2 + b^2$

$H(\omega)^2 = 1 - R^6 \cos(6\omega) - jR^6 \sin(6\omega)$

$(1 - R^6 \cos(6\omega))^2 + (R^6 \sin(6\omega))^2$

$(1 - 2R^6 \cos(6\omega) + R^{12}) + (R^{12} \sin^2(6\omega))$

$1 - 2R^6 \cos(6\omega) + R^{12} (\cos^2(6\omega) + \sin^2(6\omega))$

$\frac{d}{d\omega} (1 - 2R^6 \cos(6\omega) + R^{12}) = \frac{d}{d\omega} (1 - 2R^6 \cos(6\omega) + R^{12}) \quad // \frac{d}{d\omega} (\cos(6\omega)) = -\sin(6\omega) \cdot 6$

$-12R^6 \cdot (-\sin(6\omega) \cdot 6) = 12R^6 \sin(6\omega)$

$\min \rightarrow \sin(6\omega) = 0$

$6\omega = \pi n$

$\omega = \pi/6 n$

① ② factor: $H(z) = (z-R)(z-Re^{j\pi/3})(z-Re^{j2\pi/3})(z-e^{j\pi})(z-Re^{j4\pi/3})(z-Re^{j5\pi/3})$

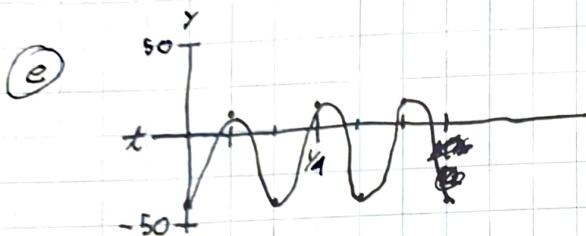
At poles

$$H(z)^{2^n} = 12R^6 \sin(\omega), \text{ Critical } \Theta \approx \frac{\pi}{6} n$$

$$H(z)^{2^n} = 72R^6 \cos(\omega), \text{ min } \Theta 2\pi n, \text{ Max } \Theta \frac{\pi}{6} n$$

$\omega \in [0, \pi]$ (fraction of sample rate, dB):

$$\begin{aligned} w(0) &: (0, -44.5) & w(\pi/6) &: (\frac{1}{2}, 5.99) & w(2\pi/3) &: (\frac{1}{6}, -11.5) \\ w(\pi/2) &: (\frac{1}{4}, 5.99) & w(5\pi/6) &: (\frac{1}{3}, -14.5) & w(5\pi/6) &: (\frac{5}{12}, 5.99) \\ w(\pi) &: (\frac{1}{2}, -44.5) \end{aligned}$$



④ $H(\omega) = 1 - R^6 e^{-j6\omega} = 1 - R^6 \cos(6\omega) - jR^6 \sin(6\omega) = 1 - R^6 \cos(6\omega) + jR^6 \sin(6\omega)$

$$\Theta_\omega = \tan^{-1} \left(\frac{R^6 \sin(6\omega)}{1 - R^6 \cos(6\omega)} \right)$$

Phase ⑤ $\min 2\pi n : 0$
 $\max \pi/6 n : 0$

Phase response at extrema makes sense because $H(z)^2 = 1$

$$[2] f_s = 44100 \text{ Hz}$$

$$[I] B = 2\phi, \cos(\phi) = 2 - \frac{1}{2}(R - \frac{1}{R})$$

$$\textcircled{a} B = 10 \text{ Hz } \cancel{\pi} = 1000 \text{ Hz}$$

$$\cos(\theta) = \frac{2R}{1+R^2} \cos(\pi)$$

$$\Theta = \frac{20\pi}{441} = 0.1425$$

$$\Theta = \frac{2\pi\pi}{f_s} \quad A_0 = (1-R)^2 \sin(\theta)$$

$$\phi = B/2 = 5 \text{ Hz} \quad \frac{2\pi\pi \cdot S}{44100} = \frac{\pi}{4410}$$

$$\cos(\phi) = 2 - \frac{1}{2}(R - \frac{1}{R})$$

$$2(2-\cos(\phi)) = R - \frac{1}{R} \cdot \frac{R}{R} = 2(2\cos(\phi)) = \frac{R^2 + 1}{R}$$

$$2R(2\cos(\phi)) - R^2 + 1 = 0 \rightarrow -R^2 + 2(2\cos(\phi)) + 1 = 0 \\ = R^2 + 2R + 1 = 0$$

$$R = 1.007, \boxed{0.9993}$$

$$A_0 = (1 - 0.9993^2) \cdot \sin\left(\frac{20\pi}{441}\right) = 2.022 \text{ E-4}$$

$$Y_t = A_0 X_t + 2R \cos(\theta) Y_{t-1} - R^2 Y_{t-2} \quad \left| \begin{array}{l} A_0 = 2.022 \text{ E-4} \\ R = 0.9993 \end{array} \right.$$

$$2R \cos(\theta) = 1.979$$

$$\theta = \frac{20\pi}{441}$$

$$Y_t = 2.022 \text{ E-4} X_t + 1.979 Y_{t-1} - 0.9986 Y_{t-2}$$

$$\delta_t: \quad Y_0 = 2.022 \text{ E-4} \quad Y_1 = 3.999 \text{ E-4} \quad Y_2 = 5.8931 \text{ E-4}$$

$$[II] \quad B = 2(1-R) \rightarrow B = 2 \approx 2R \approx B + 1 = R$$

$$B/2 = 1 - R \quad R = 0.9998 \quad \left| \begin{array}{l} B_{\text{Hz}} / f_{s/2} = \delta s_s \\ \theta = 0.1425 \end{array} \right.$$

$$1 - B/2 = R \quad A_0 = 3.219 \text{ E-5}$$

$$Y_t = 3.219 \text{ E-5} X_t + 1.979 Y_{t-1} - 0.9997 Y_{t-2}$$

$$\delta_t: \quad Y_0 = 3.219 \text{ E-5} \quad Y_1 = 6.373 \text{ E-5} \quad Y_2 = 9.397 \text{ E-5}$$

$$\boxed{2} \quad b) \quad B = 20 \text{ Hz} \quad f = 5000 \text{ Hz}$$

$$= \frac{100\pi}{2205} \approx \frac{100\pi}{2205} \approx 4$$

$$= \frac{\pi}{2205}$$

$$\frac{B}{2} = \phi \quad R^2 - (2 \cdot (2 - \cos(\phi))) + 1 = 0$$

$$\phi = \frac{\pi}{2205}$$

$$R = 0.99993 \quad R = 0.9985$$

$$\text{Methode} \quad \cos(\theta) = \frac{2R}{1+R} \cos(\phi) \quad \cos(\theta) = 0.7563$$

$$\theta = 0.7132$$

$$\begin{aligned} A_0 &= (1-R^2) \cdot \sin(\theta) \\ &= 1.861 \cdot 10^{-3} \end{aligned} \quad y_t = 1.861 \cdot 10^{-3} x_t + 1.5109 y_{t-1} - 0.9972 y_{t-2}$$

$$S_t: \quad y_0 = 1.861 \cdot 10^{-3} \quad y_1 = 2.812 \cdot 10^{-3} \quad y_2 = 2.391 \cdot 10^{-3}$$

$$\boxed{2} \quad B = 2(1-R) \quad R = 0.99929 \quad \cos(\theta) = 0.7563$$

$$\theta = 0.71279$$

$$A_0 = 9.314 \cdot 10^{-4} \quad 2 \cdot R \cdot \cos(\theta) = 1.512$$

$$y_t = 9.314 \cdot 10^{-4} x_t + 1.512 y_{t-1} - 0.9986 y_{t-2}$$

$$S_t: \quad y_0 = 9.314 \cdot 10^{-4} \quad y_1 = 1.408 \cdot 10^{-3} \quad y_2 = 1.199 \cdot 10^{-3}$$

$$[3] \quad y_t = \sqrt{2} (0.99^t) \sin\left(\frac{\pi}{4}(t+1)\right)$$

$$R = 0.99$$

$$\theta = \frac{\pi}{4}$$

$$\cos\left(\frac{\pi}{4}\right) = \frac{\sqrt{2}}{2}$$

$$= 2 \cdot \cos(\theta) \cdot R = 2 \cdot \frac{\sqrt{2}}{2} \cdot 0.99$$

$$\therefore y_t = s_t + \sqrt{2} \cdot 0.99 \sqrt{2} y_{t-1} - 0.9801 y_{t-2}$$

[4] $R \approx e^{-B/2}$, show very close to eqn 3.6:

$$R \approx 1 - B/2$$

$$e^{-B/2} = \left[e^x = 1 + x + \frac{x^n}{n!} \dots \right]$$

$$e^{-B/2} \approx 1 + \frac{(-B/2)^1}{1} + \frac{(-B/2)^2}{2} + \frac{(-B/2)^3}{6} \left[e^x = \sum_{k=0}^{\infty} \frac{x^k}{k!} \right]$$

$$\approx 1 - \frac{B}{2} + \frac{B^2}{8} - \frac{B^3}{48} \dots$$

\hookrightarrow terms after $k=1$ are negligibly impactful
on the pole radius approximation for very
small bandwidths.

first-order
approximation

\therefore can reasonably be approximated as

$$e^{-B/2} \approx 1 - B/2$$

Impulse response R^t can be written as $e^{t \ln(R)}$

For $R \approx 1$ (pole close to unit circle), $\ln(R) \approx R-1$

Using above $R \approx 1 - B/2$, $\ln(R) \approx (1 - B/2) - 1$

$$\ln(R) \approx -B/2 \quad \therefore e^{-B/2 t} \propto e^{t \ln(R)}$$

* Using Taylor series expansion of \ln around 1:

$$\ln(1+u) = u - \frac{u^2}{2} + \frac{u^3}{3} \dots$$

$$u = x-1$$

$$\ln(R) = u \quad \begin{matrix} \rightarrow \\ \text{first-order} \\ \text{approximation} \end{matrix}$$

$$\ln(R) \approx R-1$$

5 a) For what range of values of R & θ is the discrepancy greatest between the true peak frequency ω_0 & pole angle θ ?

- As R is very close to 1, peak frequency ($\bar{\omega}$)

$$\omega_0(\bar{\omega}) \text{ is very close to } \theta \quad \cos(\bar{\omega}) = \frac{1+R^2}{2R} \cos(\theta)$$

\therefore the discrepancy occurs as R moves away from 1
can also vary by pole angle θ but is primarily affected by value of R .

b) When θ is small, ω_0 is shifted to higher or lower frequencies?

- As described on p. 91, $\cos(\bar{\omega})$ & $\cos(\theta)$ are used to make it easier to see the directly proportional relationship of ω_0 & θ .

When θ is small $\cos(\bar{\omega})$, ω_0 is shifted to lower frequencies.