## **EE5410 Signal Processing**

## **Solution for Assignment 2**

1.(a)

Using the relationship between z transform and DTFT, i.e.,  $z=e^{j\omega}$ , we get:

$$H(e^{j\omega}) = \frac{e^{-2j\omega}}{1 + 0.7e^{-j\omega} + 0.1e^{-2j\omega}} \Rightarrow H(z) = \frac{z^{-2}}{1 + 0.7z^{-1} + 0.1z^{-2}} = \frac{z^{-2}}{(1 + 0.5z^{-1})(1 + 0.2z^{-1})}$$

As the DTFT converges, the ROC should include the unit circle. Hence the ROC is |z| > 0.5.

1.(b) 
$$H(z) = \frac{Y(z)}{X(z)} = \frac{z^{-2}}{1 + 0.7z^{-1} + 0.1z^{-2}} \Rightarrow y[n] + 0.7y[n - 1] + 0.1y[n - 2] = x[n - 2]$$

1.(c) 
$$H(z) = -\frac{10}{3} \frac{z^{-1}}{1 + 0.5z^{-1}} + \frac{10}{3} \frac{z^{-1}}{1 + 0.2z^{-1}}$$

With the use of time-shifting property and ROC of |z| > 0.5, we get:

$$h[n] = -\frac{10}{3} \left( -\frac{1}{2} \right)^{n-1} u[n-1] + \frac{10}{3} \left( -\frac{1}{5} \right)^{n-1} u[n-1]$$

2.(a)

$$H(z) = \frac{1 - \frac{1}{5}z^{-1}}{1 - \frac{1}{4}z^{-1} + \frac{5}{24}z^{-2} + \frac{1}{12}z^{-3}}$$

$$z_{[n]}$$

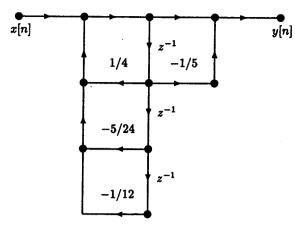
$$z^{-1}$$

$$z^{-1}$$

$$z^{-1}$$

$$z^{-1}$$

2.(b)



2.(c) 
$$H(z) = \left(\frac{1 - \frac{1}{5}z^{-1}}{1 + \frac{1}{4}z^{-1}}\right) \left(\frac{1}{1 - \frac{1}{2}z^{-1} + \frac{1}{3}z^{-2}}\right)$$

$$x[n]$$

$$z^{-1}$$

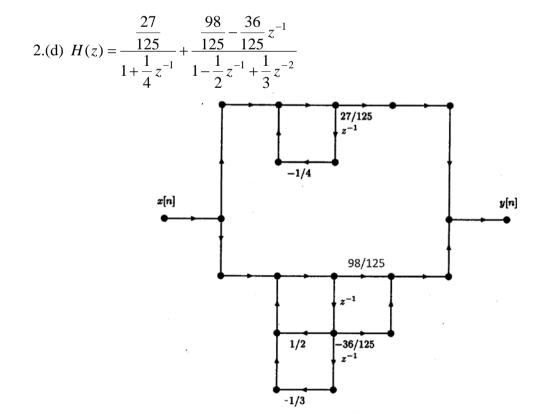
$$-1/4$$

$$-1/5$$

$$1/2$$

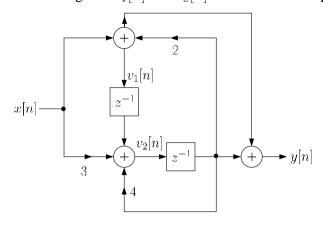
$$z^{-1}$$

$$-1/3$$



3.(a)

We introduce two intermediate signals  $v_1[n]$  and  $v_2[n]$  which are the outputs of unit delays.



We have:

$$v_1[n-1] + 3x[n] + 4v_2[n-1] = v_2[n]$$
  

$$x[n] + 2v_2[n-1] = v_1[n]$$
  

$$y[n] = v_1[n] + v_2[n-1]$$

Taking their z transforms yields

$$z^{-1}V_1(z) + 3X(z) + 4z^{-1}V_2(z) = V_2(z)$$

$$X(z) + 2z^{-1}V_2(z) = V_1(z)$$

$$Y(z) = V_1(z) + z^{-1}V_2(z)$$

From the first two equations, we can solve  $V_2(z)$  in terms of X(z) as:

$$V_2(z) = \frac{3 + z^{-1}}{1 - 4z^{-1} - 2z^{-2}}X(z)$$

Then  $V_1(z)$  is computed as:

$$V_2(z) = \frac{1 + 2z^{-1}}{1 - 4z^{-1} - 2z^{-2}}X(z)$$

Putting the result into the third equation, we get

$$Y(z) = \left(\frac{1 + 2z^{-1}}{1 - 4z^{-1} - 2z^{-2}} + \frac{3z^{-1} + z^{-2}}{1 - 4z^{-1} - 2z^{-2}}\right)X(z) \Rightarrow \frac{Y(z)}{X(z)} = \frac{1 + 5z^{-1} + z^{-2}}{1 - 4z^{-1} - 2z^{-2}}$$

3.(b) 
$$Y(z) \left(1 - 4z^{-1} - 2z^{-2}\right) = X(z) \left(1 + 5z^{-1} + z^{-2}\right)$$
 
$$\Rightarrow y[n] - 4y[n-1] - 2y[n-2] = x[n] + 5x[n-1] + x[n-2]$$

3.(c) 
$$x[n] \xrightarrow{} + \underbrace{} +$$

3.(d)

The system is not stable.

Solving the quadratic equation, the poles of the filter are 4.45 and -0.45. If the system is stable, the ROC should include the unit circle. However, as the system is causal, the ROC is |z| > 4.45, which does not include the unit circle.

4.(a)

According to (10.26), the ideal impulse response is:

$$h_d[n] = \frac{\omega_b}{\pi} \operatorname{sinc}\left(\frac{\omega_b n}{\pi}\right) - \frac{\omega_a}{\pi} \operatorname{sinc}\left(\frac{\omega_a n}{\pi}\right)$$
$$= 0.75 \operatorname{sinc}(0.75n) - 0.25 \operatorname{sinc}(0.25n)$$

For a filter of length 5, we extract  $h_d[-2]$ ,  $h_d[-1]$ ,  $h_d[0]$ ,  $h_d[1]$  and  $h_d[2]$ . The causal filter impulse response is then:

$$h[n] = h_d[n-2] = [-0.3183\ 0\ 0.5\ 0\ -0.3183]$$

With the first element being h[n] at n = 0. As a result,  $H_1(z)$  is:

$$H_1(z) = -0.3183 + 0.5z^{-2} - 0.3183z^{-4}$$

4.(b)

In the implementation, we make use of the symmetry of the impulse response:

$$y[n] = -0.3183(x[n] + x[n-4]) + 0.5x[n-2]$$

As a result, the minimum numbers of additions and multiplications are both equal to 2.