

## **In-Class Exercise 2**

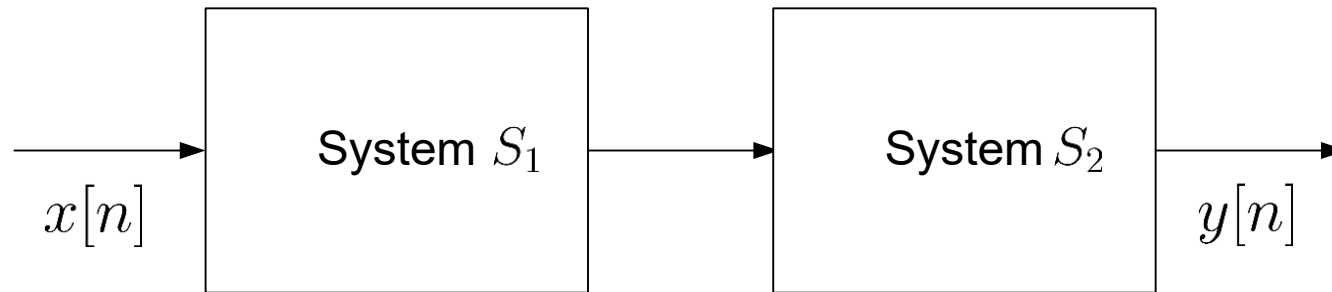
1. Consider a discrete-time system  $S$  with input  $x[n]$  and output  $y[n]$ . This system is obtained through a series interconnection of a system  $S_1$  followed by another system  $S_2$ . The input and output relationships for  $S_1$  and  $S_2$  are:

$$S_1 : y_1[n] = 2x_1[n] + 4x_1[n - 1]$$

$$S_2 : y_2[n] = x_2[n - 2] + \frac{1}{2}x_2[n - 3]$$

where  $x_1[n]$  and  $x_2[n]$  denote inputs and  $y_1[n]$  and  $y_2[n]$  denote outputs.

- (a) Determine the input-output relationship for system  $S$ , i.e., find the equation that relates  $x[n]$  and  $y[n]$ .
- (b) Does the input-output relationship of system  $S$  change if we first pass  $x[n]$  through  $S_2$  and then  $S_1$ ?



2. Determine whether the following discrete-time system, with input signal  $x[n]$  and output signal  $y[n]$  is memoryless, invertible, stable, causal, linear, and/or time-invariant:

$$y[n] = ax[n + 1] + b, \quad 0 < |a| < \infty, \quad 0 < |b| < \infty$$

3. Determine whether the following continuous-time system, with input signal  $x(t)$  and output signal  $y(t)$  is memoryless, invertible, stable, causal, linear, and/or time-invariant:

$$y(t) = \cos [x(t)]$$

4. Consider the following discrete-time system, with input signal  $x[n]$  and output signal  $y[n]$ :

$$y[n] = x[n - 1] - y[n - 1]$$

where  $y[n] = 0$  for  $n < 0$ .

- (a) Determine  $y[n]$  when  $x[n] = \delta[n]$ .
- (b) Determine  $y[n]$  when  $x[n] = u[n]$ .

5. Compute the output  $y[n]$  if the input is  $x[n] = a^n u[n]$  and the linear time-invariant system impulse response is  $h[n] = b^n u[n]$  with  $|b| \geq 1$ . Is the system stable? Why? Is the system causal? Why?

6. Determine  $y[n] = x[n] \otimes h[n]$  where  $x[n]$  and  $h[n]$  are

$$x[n] = \begin{cases} 3, & n = -1 \\ 2, & n = 1 \\ 6, & n = 2 \\ 0, & \text{otherwise} \end{cases}$$

and

$$h[n] = \begin{cases} 2, & n = -1 \\ 4, & n = 0 \\ 7, & n = 1 \\ 0, & \text{otherwise} \end{cases}$$

7. Compute the output  $y(t)$  if the input is  $x(t) = u(t - 3) - u(t - 5)$  and the linear time-invariant system impulse response is  $h(t) = e^{-3t}u(t)$ . Is the system stable? Why? Is the system causal? Why?

8. Compute the impulse response  $h[n]$  for a LTI system which is characterized by the following difference equation:

$$y[n] = x[n - 1] + 2x[n - 2] + 3x[n - 3]$$

9. Define the area under a continuous-time signal  $v(t)$  as:

$$A_v = \int_{-\infty}^{\infty} v(t) dt$$

Show that if  $y(t) = x(t) \otimes h(t)$ , then

$$A_y = A_x \cdot A_h$$

10. Denote  $h[n]$  as the impulse response of a discrete-time linear time-invariant system. If the system is also memoryless, then determine the form of  $h[n]$ .

## **Solution**

1(a)

Let  $w[n]$  be the after output after passing through  $S_1$ :

$$w[n] = 2x[n] + 4x[n - 1]$$

Passing  $w[n]$  through  $S_2$  yields:

$$\begin{aligned} y[n] &= w[n - 2] + \frac{1}{2}w[n - 3] \\ &= (2x[n - 2] + 4x[n - 3]) + \frac{1}{2}(2x[n - 3] + 4x[n - 4]) \\ &= 2x[n - 2] + 5x[n - 3] + 2x[n - 4] \end{aligned}$$



1(b)

Let  $v[n]$  be the after output after passing through  $S_2$ :

$$v[n] = x[n - 2] + \frac{1}{2}x[n - 3]$$

Passing  $w[n]$  through  $S_1$  yields:

$$\begin{aligned} y[n] &= 2v[n] + 4v[n - 1] \\ &= 2 \left( x[n - 2] + \frac{1}{2}x[n - 3] \right) + 4 \left( x[n - 3] + \frac{1}{2}x[n - 4] \right) \\ &= 2x[n - 2] + 5x[n - 3] + 2x[n - 4] \end{aligned}$$

Realizing that  $S_1$  and  $S_2$  are LTI systems and using the commutative property, we can also understand that there will be **no change** in the output.

2.

$$y[n] = ax[n + 1] + b, \quad 0 < |a| < \infty, \quad 0 < |b| < \infty$$

### Memoryless

The system is **not memoryless** because output  $y[n]$  at time  $n$  does not only depend on  $x[n]$  at time  $n$ . In fact,  $y[n]$  depends on input at time  $n + 1$ .

### Invertibility

The system is **invertible**. By reorganizing the equation and we see that  $x[n]$  can be computed from  $y[n]$  using:

$$y[n] = ax[n + 1] + b \Rightarrow x[n] = \frac{y[n - 1] - b}{a}$$

## Stability

If  $x[n]$  is bounded, then  $y[n] = ax[n+1] + b$  is bounded for bounded  $a$  and  $b$ . As a result,  $y[n]$  is bounded and the system is **stable**. In a more rigorous manner, we have:

$$|y[n]| = |ax[n+1] + b| \leq |a| \cdot |x[n+1]| + |b|$$

where  $|x[n+1]| < \infty$  or  $|x[n]| < \infty$  must give  $|y[n]| < \infty$ .

## Causality

It is **not causal** because  $y[n]$  depends on future input, namely,  $x[n+1]$ .

## Linearity

The system outputs for  $x_1[n]$  and  $x_2[n]$  are:

$$y_1[n] = ax_1[n+1] + b \quad \text{and} \quad y_2[n] = ax_2[n+1] + b$$

Consider  $x_3[n] = cx_1[n] + dx_2[n]$ , its system output is then:

$$\begin{aligned} y_3[n] &= ax_3[n+1] + b \\ &= a(cx_1[n+1] + dx_2[n+1]) + b \\ &= acx_1[n+1] + adx_2[n+1] + b \\ &= c(ax_1[n+1] + b) + d(ax_2[n+1] + b) + b - bc - bd \\ &\neq cy_1[n] + dy_2[n] \end{aligned}$$

As a result, this system is **not linear**.

## Time-invariance

First, we have:

$$y[n - n_0] = ax[n - n_0 + 1] + b$$

Consider  $x_1[n] = x[n - n_0]$ , its system output is

$$\begin{aligned} y_1[n] &= ax_1[n + 1] + b \\ &= ax[n - n_0 + 1] + b \\ &= y[n - n_0] \end{aligned}$$

Hence the system is **time-invariant**.

3.

$$y(t) = \cos [x(t)]$$

### Memoryless

The system is **memoryless** because the output  $y(t)$  at time  $t$  only depends on  $x(t)$  at time  $t$ .

### Invertibility

The system is not **invertible**. By reorganizing the equation and we see that  $x(t)$  cannot be computed from  $y(t)$  because there are infinite possibilities of  $x(t)$

$$x(t) = \cos^{-1}(y(t)) + 2n\pi, \quad n = \dots, -1, 0, 1, \dots$$

### Stability

If  $x(t)$  is bounded, then  $y(t) = \cos [x(t)]$  must be bounded because  $1 \geq |\cos [x(t)]|$ .

## Causality

It is **causal** because  $y(t)$  at time  $t$  depends on  $x(t)$  up to time  $t$ .

## Linearity

The system outputs for  $x_1(t)$  and  $x_2(t)$  are:

$$y_1(t) = \cos [x_1(t)] \quad \text{and} \quad y_2(t) = \cos [x_2(t)]$$

Consider  $x_3(t) = ax_1(t) + bx_2(t)$ , its system output is then:

$$\begin{aligned} y_3(t) &= \cos [x_3(t)] \\ &= \cos [ax_1(t) + bx_2(t)] \\ &\neq a \cos [x_1(t)] + b \cos [x_2(t)] \\ &= ay_1(t) + by_2(t) \end{aligned}$$

As a result, this system is **not linear**.

## Time-invariance

First, we have:

$$y(t - t_0) = \cos [x(t - t_0)]$$

Consider  $x_1(t) = x(t - t_0)$ , its system output is

$$\begin{aligned} y_1(t) &= \cos [x_1(t)] \\ &= \cos [x(t - t_0)] \\ &= y(t - t_0) \end{aligned}$$

Hence the system is **time-invariant**.



4(a)

When  $x[n] = \delta[n]$  and using  $y[-1] = 0$ , we start with  $n = 0$ :

$$y[0] = \delta[-1] - y[-1] = 0$$

$$y[1] = \delta[0] - y[0] = 1$$

$$y[2] = \delta[1] - y[1] = -1$$

$$y[3] = \delta[2] - y[2] = 1$$

We may deduce the general form of  $y[n]$  as:

$$y[n] = (-1)^{n-1}u[n-1]$$

4(b)

When  $x[n] = u[n]$  and using  $y[-1] = 0$ , we start with  $n = 0$ :

$$y[0] = u[-1] - y[-1] = 0$$

$$y[1] = u[0] - y[0] = 1$$

$$y[2] = u[1] - y[1] = 0$$

$$y[3] = u[2] - y[2] = 1$$

We may deduce the general form of  $y[n]$  as:

$$y[n] = \begin{cases} 1, & n = 1, 3, 5, \dots \\ 0, & \text{otherwise} \end{cases}$$

5.

Using (3.11), we have:

$$\begin{aligned}y[n] &= \sum_{m=-\infty}^{\infty} x[m]h[n-m] \\&= \sum_{m=-\infty}^{\infty} a^m u[m] b^{m-n} u[n-m] \\&= \sum_{m=0}^{\infty} a^m b^{m-n} u[n-m] \\&= \sum_{k=-\infty}^{\infty} a^{n-k} b^k u[k], \quad k = n - m \\&= a^n \sum_{k=-\infty}^n (a^{-1}b)^k u[k]\end{aligned}$$

Since  $u[k] = 0$  for  $k < 0$ ,  $y[n] = 0$  for  $n < 0$ .

For  $n \geq 0$ , we then have:

$$\begin{aligned} y[n] &= a^n \sum_{k=0}^n (a^{-1}b)^k \\ &= a^n \frac{1 - (a^{-1}b)^{n+1}}{1 - a^{-1}b} \\ &= \frac{a^{n+1} - b^{n+1}}{a - b} \end{aligned}$$

Combining the results, we have:

$$y[n] = \frac{a^{n+1} - b^{n+1}}{a - b} u[n]$$

Since  $\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=0}^{\infty} |b| = \infty$ , the system is **not stable**.  
Moreover, the system is **causal** because  $h[n] = 0$  for  $n < 0$ .

6.

Using (3.11) again, we have:

$$\begin{aligned} y[n] &= \sum_{m=-\infty}^{\infty} x[m]h[n-m] \\ &= x[-1]h[n+1] + x[1]h[n-1] + x[2]h[n-2] \end{aligned}$$

**Try**  $n = -2$ :

$$y[-2] = x[-1]h[-1] + x[1]h[-3] + x[2]h[-4] = 3 \cdot 2 = 6$$

**Try**  $n = -1$ :

$$y[-1] = x[-1]h[0] + x[1]h[-2] + x[2]h[-3] = 3 \cdot 4 = 12$$

Try  $n = 0$ :

$$y[-1] = x[-1]h[1] + x[1]h[-1] + x[2]h[-2] = 3 \cdot 7 + 2 \cdot 2 = 25$$

Compute  $y[n]$  for other values of  $n$  and combine the results, we get:

$$y[n] = \begin{cases} 6, & n = -2 \\ 12, & n = -1 \\ 25, & n = 0 \\ 20, & n = 1 \\ 38, & n = 2 \\ 42, & n = 3 \\ 0, & \text{otherwise} \end{cases}$$

7.

Using the convolution for continuous-time case, we have:

$$\begin{aligned} y(t) &= \int_{-\infty}^{\infty} h(\tau)x(t-\tau)d\tau = \int_{-\infty}^{\infty} e^{-3\tau}u(\tau) [u(t-3-\tau) - u(t-5-\tau)] d\tau \\ &= \int_0^{\infty} e^{-3\tau}u(t-3-\tau)d\tau - \int_0^{\infty} e^{-3\tau}u(t-5-\tau)d\tau \end{aligned}$$

Let the first and second components be  $y_1(t)$  and  $y_2(t)$  such that  $y(t) = y_1(t) - y_2(t)$ .

$$\begin{aligned} y_1(t) &= \int_0^{\infty} e^{-3\tau}u(t-3-\tau)d\tau, \quad \lambda = t-3-\tau \\ &= \int_{t-3}^{-\infty} e^{-3(t-3-\lambda)}u(\lambda)d(-\lambda) \\ &= \int_{-\infty}^{t-3} e^{-3(t-3-\lambda)}u(\lambda)d\lambda \end{aligned}$$

When  $t - 3 < 0$  or  $t < 3$ , the integral will only involve the zero part of  $u(\lambda)$  because  $u(\lambda) = 0$  for  $\lambda < 0$ . Hence  $y_1(t) = 0$  for  $t < 3$ . For  $t > 3$ , we have:

$$\begin{aligned} y_1(t) &= \int_0^{t-3} e^{-3(t-3-\lambda)} d\lambda = e^{-3(t-3)} \int_0^{t-3} e^{3\lambda} d\lambda \\ &= e^{-3(t-3)} \cdot \frac{1}{3} e^{3\lambda} \Big|_0^{t-3} = e^{-3(t-3)} \cdot \frac{1}{3} \left( e^{3(t-3)} - 1 \right) = \frac{1}{3} \left( 1 - e^{-3(t-3)} \right) \end{aligned}$$

That is,

$$y_1(t) = \frac{1}{3} \left( 1 - e^{-3(t-3)} \right) u(t - 3)$$

Similarly,  $y_2(t)$  is:

$$y_2(t) = \frac{1}{3} \left( 1 - e^{-3(t-5)} \right) u(t - 5)$$



Combining the results yields:

$$y(t) = \frac{1}{3} \left(1 - e^{-3(t-3)}\right) u(t-3) - \frac{1}{3} \left(1 - e^{-3(t-5)}\right) u(t-5)$$

or

$$y(t) = \begin{cases} 0, & t < 3 \\ \frac{1}{3} \left(1 - e^{-3(t-3)}\right) u(t-3), & 3 < t < 5 \\ \frac{1}{3} \left(e^{-3(t-5)} - e^{-3(t-3)}\right), & t > 5 \end{cases}$$

Since  $\int_{-\infty}^{\infty} |h[t]| dt = \int_0^{\infty} e^{-3t} dt < \infty$ , the system is **stable**.  
Moreover, the system is **causal** because  $h(t) = 0$  for  $t < 0$ .

8.

Using (3.12), we have:

$$\begin{aligned} y[n] &= \sum_{m=-\infty}^{\infty} h[m]x[n-m] \\ &= \cdots h[0]x[n] + h[1]x[n-1] + h[2]x[n-2] + h[3]x[n-3] + \cdots \end{aligned}$$

It is seen that only  $h[1]$ ,  $h[2]$  and  $h[3]$  are nonzero. That is, the impulse response is:

$$h[n] = \delta[n-1] + 2\delta[n-2] + 3\delta[n-3]$$

9.

$$\begin{aligned} A_y &= \int_{-\infty}^{\infty} y(t) dt = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x(\tau) h(t - \tau) d\tau dt \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x(\tau) h(\lambda) d\tau d(\lambda + \tau), \quad \lambda = t - \tau \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x(\tau) h(\lambda) d\tau d\lambda \\ &= \left( \int_{-\infty}^{\infty} x(\tau) d\tau \right) \left( \int_{-\infty}^{\infty} h(\lambda) d\lambda \right) \\ &= A_x \cdot A_h \end{aligned}$$

10.

Let  $x[n]$  and  $y[n]$  be the system input and output, respectively. Expanding the convolution formula yields:

$$\begin{aligned} y[n] = x[n] \otimes h[n] &= \sum_{m=-\infty}^{\infty} h[m]x[n-m] \\ &= \cdots h[-2]x[n+2] + h[-1]x[n+1] + \\ &\quad h[0]x[n] + h[1]x[n-1] + h[2]x[n-2] + \cdots \end{aligned}$$

If the system is memoryless,  $y[n]$  at time  $n$  only depends on  $x[n]$  at the same time, implying that

$$y[n] = x[n] \otimes h[n] = h[0]x[n]$$

That is,  $h[n] = K\delta[n]$  where  $K = h[0]$  is a constant.