

Unit 9

Codes

Outline of Unit 9

- ❑ 9.1 Parity-Check Codes
- ❑ 9.2 Error Detection Capability
- ❑ 9.3 Error Correction Capability
- ❑ 9.4 Generator and Parity Check Matrices

Example: Error Detection

NEW SMART HK ID

CURRENT HONG KONG ID



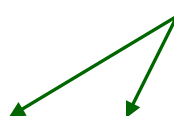
Is this a valid HKID card number?

Weighted Average Mod 11

□ HKID:

	C	6	6	8	6	6	8	(?)
--	---	---	---	---	---	---	---	-----

A=10, B=11, ..., Z=35, Space = 36



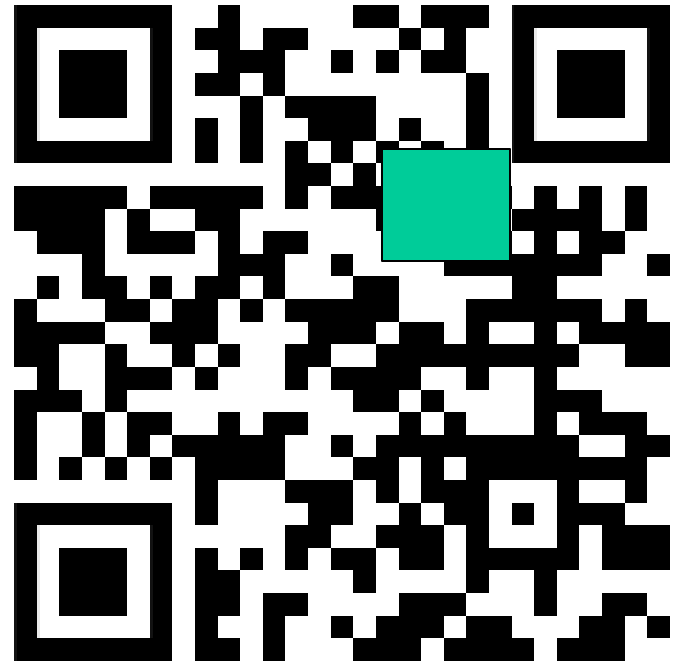
□ Weight: 9 8 7 6 5 4 3 2 1

$$36 \times 9 + 12 \times 8 + 6 \times 7 + 6 \times 6 + 8 \times 5 + 6 \times 4 + 6 \times 3 + 8 \times 2 + x \equiv 0 \pmod{11}$$

$$5 + 8 + 9 + 3 + 7 + 2 + 7 + 5 + x \equiv 0 \pmod{11}$$

$$x \equiv -2 \equiv 9 \pmod{11}$$

Example: Error Correction

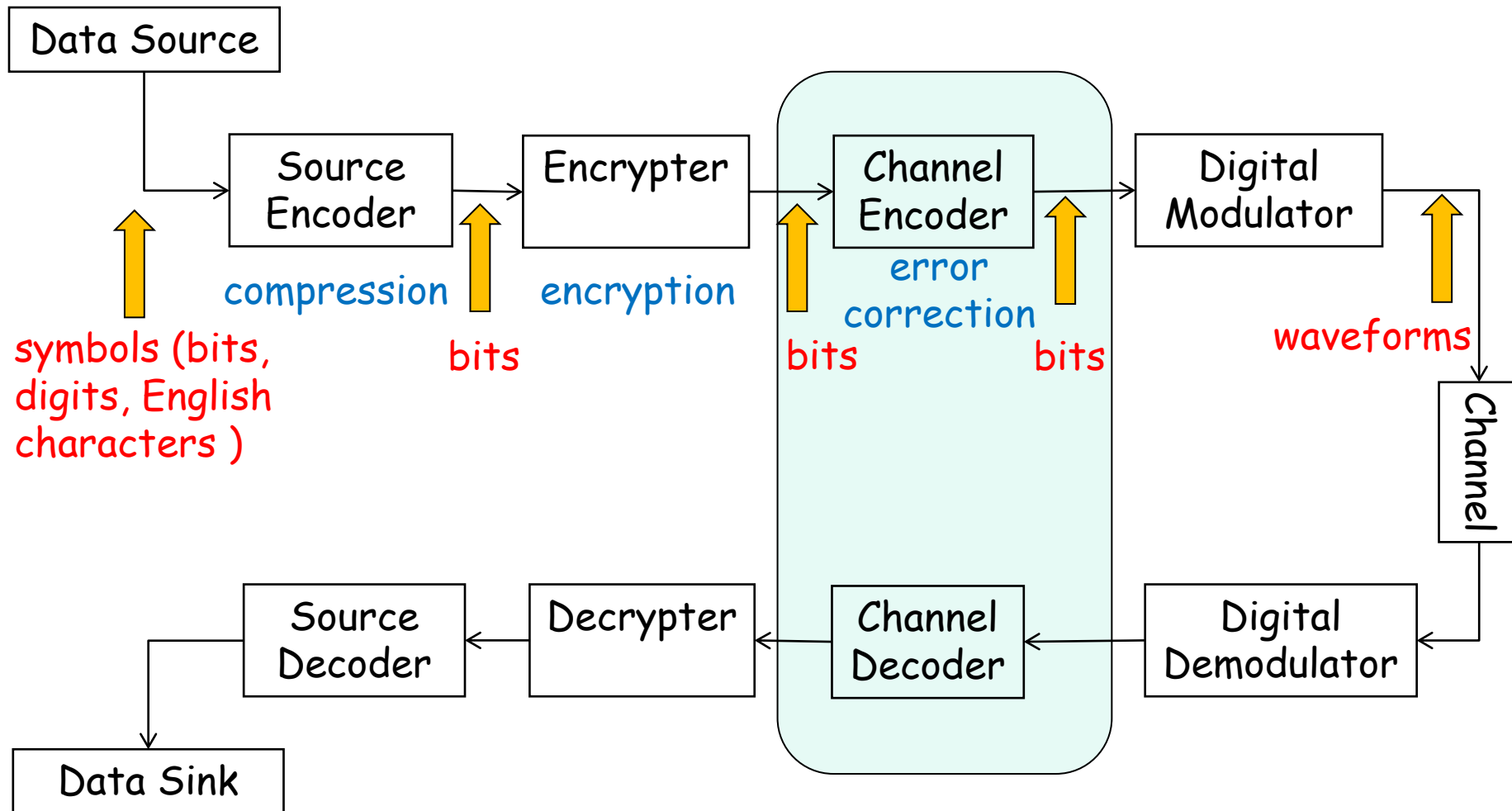


Does it still work?

Unit 9.1

Parity-Check Codes

Digital Communication Systems



Bit Errors due to Noise

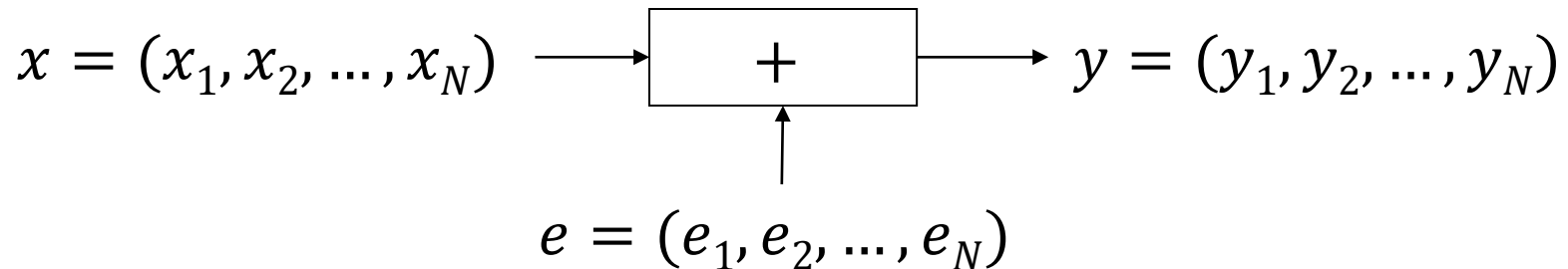
- Suppose N bits are transmitted.



- During the transmissions, bit errors may occur due to noise.
- The probability that a bit error occurs is called the *bit error rate*.

	Twisted Pair	Coaxial Cable	Optical Fiber
Data Rate in Mbps	10	100	1000
Bit Error Rate	10^{-5}	10^{-6}	10^{-9}
Bandwidth	250 kHz	350 MHz	1 GHz

Error Vector



□ If the i -th bit is in error,
then $e_i = 1$; else $e_i = 0$.

□ Hence, for all i ,

$$y_i = x_i + e_i,$$

where **binary addition (+) means logical XOR.**

A	B	A+B
0	0	0
0	1	1
1	0	1
1	1	0

How to Handle Bit Errors?

Two basic strategies:

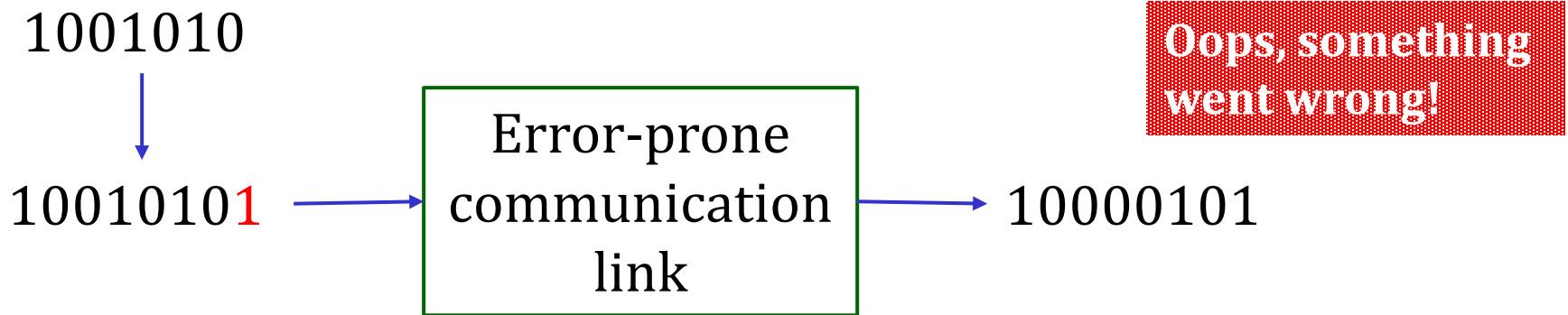
1. Error Correction

- Include enough redundant information along with a data packet to enable the receiver to identify which bits are in error and then *correct* them.

2. Error Detection

- Include only enough redundant information to allow the receiver to detect that an error occurred.
- If error is detected, the receiver may request for *retransmissions*.

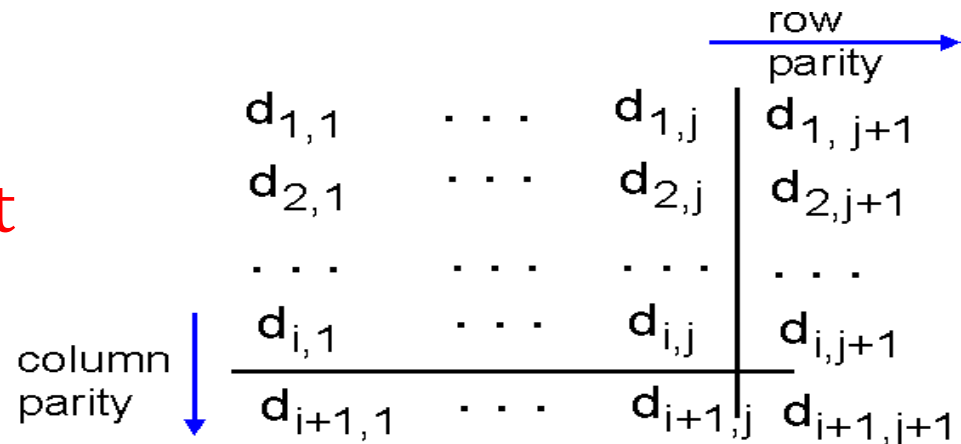
Single Parity Check



- ❑ Suppose there are k information bits.
- ❑ An extra bit (called parity bit) is added for detecting single-bit errors.
 - *Even parity*: add an extra bit so that the total number of '1's is even.
 - Odd parity can be defined similarly.
 - The receiver doesn't know which bit is in error.

Two-Dimensional Parity Check

- This scheme can **detect** and **correct** single-bit errors.



- Even parity is assumed in this example.

1	0	1	0	1	1
1	1	1	1	0	0
0	1	1	1	0	1
0	0	1	0	1	0

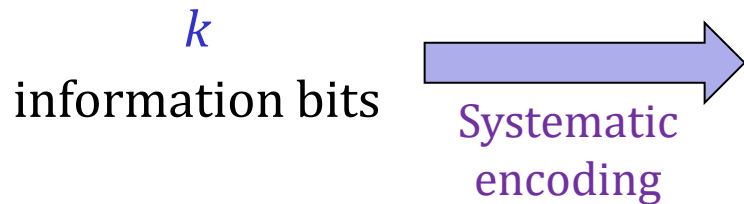
no errors

1	0	1	0	1	1
1	0	1	1	0	0
0	1	1	1	0	1
0	0	1	0	1	0

*correctable
single bit error*

Parity-Check Codes

message u (a row vector)



codeword c (a row vector)

k information bits	$r = n - k$ check bits
-------------------------	---------------------------

□ (n, k) binary code with the following notation:

- k information bits
- r redundant bits
- Codeword length:
$$n = k + r$$
- The code is a set of 2^k codewords.

□ The encoding of a code is **systematic** if the information bits are **embedded** as part of the encoded output.

- The check bits are *not* necessarily after the information bits.

Example: Systematic Codes

(4,3) Even Parity

❑ 8 codewords

0	0	0	0
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	1
1	0	1	0
1	1	0	0
1	1	1	1

Parity bits

$$c_1 = u_1$$

$$c_2 = u_2$$

$$c_3 = u_3$$

$$c_4 = u_1 + u_2 + u_3$$

(5,1) Repetition

❑ 2 codewords

0	0	0	0	0
1	1	1	1	1

Parity bits

Example: A Non-systematic Code

$$u = (u_1, u_2, u_3) \xrightarrow{\text{Encoder}} c = (c_1, c_2, c_3, c_4)$$

□ Encoding:

- $c_1 = u_1$
- $c_2 = u_1 + u_2 + u_3$
- $c_3 = u_2 + u_3$
- $c_4 = u_1 + u_2$

- ## □ It is non-systematic because u_2 and u_3 are not equal to any coded bits.

Message u	Codeword c
000	0000
001	0110
010	0111
011	0001
100	1101
101	1011
110	1010
111	1100

Code Rate

- ❑ The code rate of an (n, k) code is defined as

$$R_c = \frac{k}{n}.$$

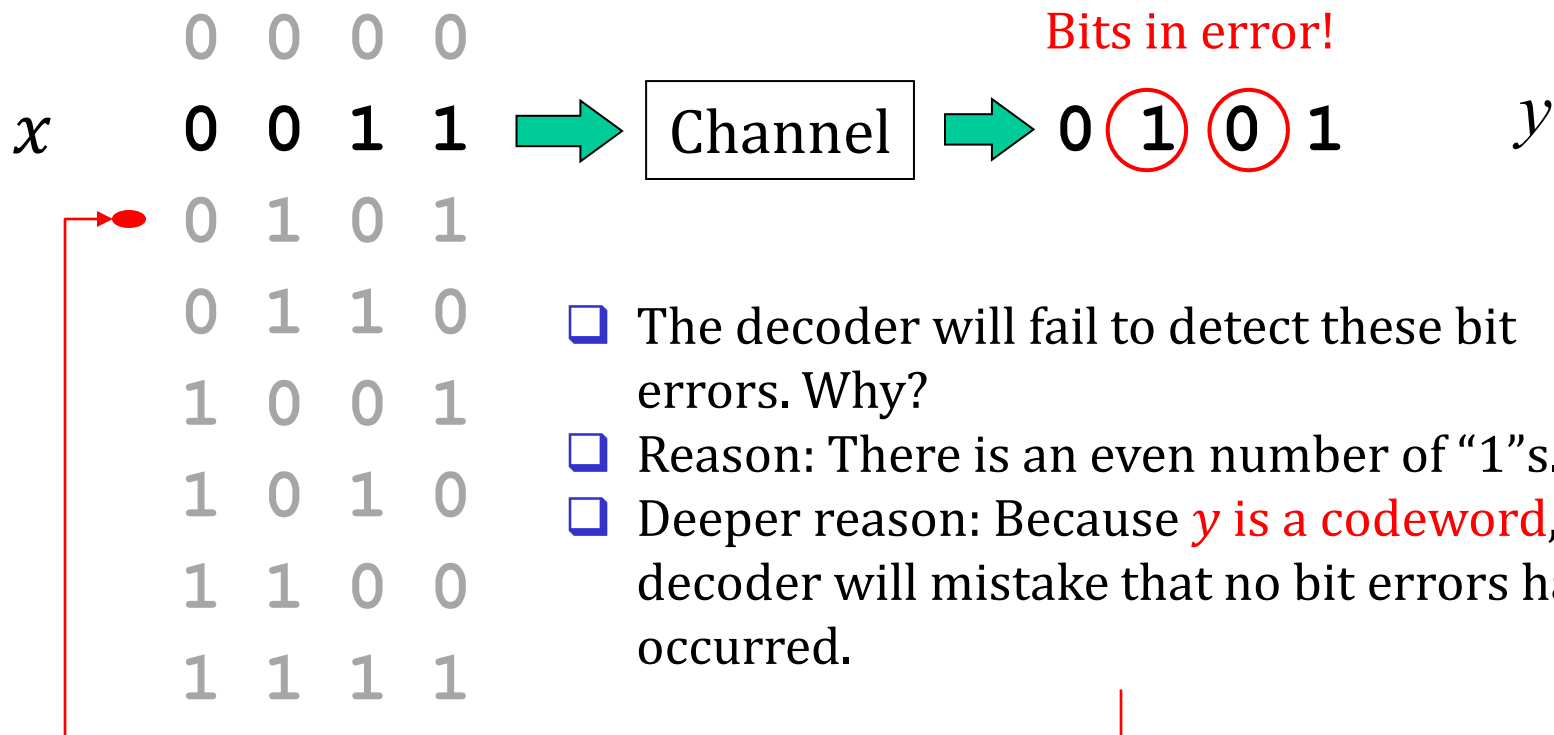
- Proportion of the data stream that is useful.
- ❑ If the raw data rate of a link is W bps, then the effective data rate (or information rate) is $R_c W$ bps.
- ❑ Example:
 - Suppose every 7 bits of data are encoded with single parity check.
 - The encoded output is then transmitted through a link with 1 Mbps.
 - What is the effective data rate?

Unit 9.2

Error Detection Capability

Error Detection Failure

□ (4, 3) Even Parity Check Code:



Hamming Distance

- ❑ Hamming distance is a useful concept for analyzing the error correction/detection capability of a code.
- ❑ The **Hamming distance** $d(x, y)$ of two vectors, x and y , is defined as the number of bits that they are different.
- ❑ The **Hamming weight** $w(x)$ of a vector x is defined as the number of 1's in x .
- ❑ Example:
 - $x = (0, 0, 1, 1, 1), \quad w(x) = 3.$
 - $y = (0, 1, 1, 0, 1), \quad w(y) = 3.$
 - $d(x, y) = 2.$

Error Detection Failure

- ❑ If the Hamming distance between two codewords is d , then it will require d errors to convert one codeword into the other.

- ❑ Example:

○ Codeword 1:	1	0	0	0	1	0	0	1
			↓	↓	↓			
○ Codeword 2:	1	0	1	1	0	0	0	1

} Hamming distance = 3

- An *error detection failure* occurs if the above three bits are in error.

Error Detection Capability

Definition: The minimum distance, d_{\min} , of a code is the *smallest* Hamming distance between *all pairs* of distinct codewords in the code.

- ❑ Error detection capability of a code depends on its d_{\min} .
- ❑ It is guaranteed that error can be detected if number of bits in error is less than or equal to

$$s = d_{\min} - 1.$$

Examples (revisited)

(4,3) Even Parity

❑ Eight codewords:

0	0	0	0
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	1
1	0	1	0
1	1	0	0
1	1	1	1

(5,1) Repetition

❑ Two codewords:

0	0	0	0	0
1	1	1	1	1

What is d_{\min} of each of these codes?

How many bit errors does each code guarantee to detect?

Unit 9.3

Error Correction Capability

Decoding Rule for Error Correction



❑ *Nearest-Neighbor Decoding*: The decoder picks a codeword that is closest to y in terms of Hamming distance.

○ In other words, find $x \in C$ which minimizes $d(x, y)$, where C is the set of all codewords.

- Tie is broken arbitrarily.
- a.k.a minimum-distance decoding

❑ Example: (5, 1) Repetition Code

○ How many bit errors can the code correct?

Error Correction Capability

- ❑ Error correction capability of a code depends on its d_{\min} .
- ❑ Error can be corrected if no. of bits in error is less than or equal to

$$t = \left\lfloor \frac{d_{\min} - 1}{2} \right\rfloor,$$

- where $\lfloor x \rfloor$ is the floor operator which denotes the largest integer less than or equal to x .
- ❑ Example: (6, 1) Repetition Code
 $d_{\min} = 6, \quad t = \lfloor 2.5 \rfloor = 2.$

Code Rate of (5, 1) Repetition

- ❑ $C = \{00000, 11111\}$.
- ❑ $d_{\min} = 5, t = 2$. (correct all double-bit errors)
- ❑ Code rate $R = \frac{k}{n} = \frac{1}{5}$.
 - For each information bit, we need to transmit 5 bits.
 - For example, if transmission rate equals 1 Mbps, then we can transmit 200 kbps of useful information.
- ❑ What if we want to convey information faster?



Classwork (Repetition with an Extra Parity)

$$u = (u_1, u_2) \quad \Rightarrow \quad \boxed{\text{Encoder}} \quad \Rightarrow \quad c = (c_1, c_2, c_3, c_4, c_5)$$

□ $c_1 = c_3 = u_1$

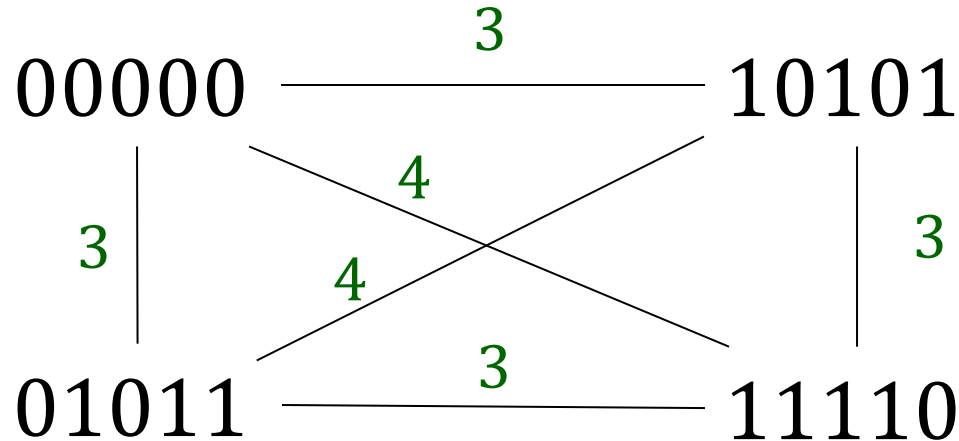
□ $c_2 = c_4 = u_2$

□ $c_5 = u_1 + u_2$

Message u	Codeword c
00	
01	
10	
11	

- Complete the table.
- Determine d_{\min} .
- How many errors can it correct?

Solution



- $R = \frac{2}{5}, d_{\min} = 3, t = 1.$
 - For example, if transmission rate equals 1 Mbps, then we can transmit 400 kbps of useful information.
- Comparison with (5, 1) repetition code:
 - More efficient in communications (less redundancy)
 - Weaker error correction capability

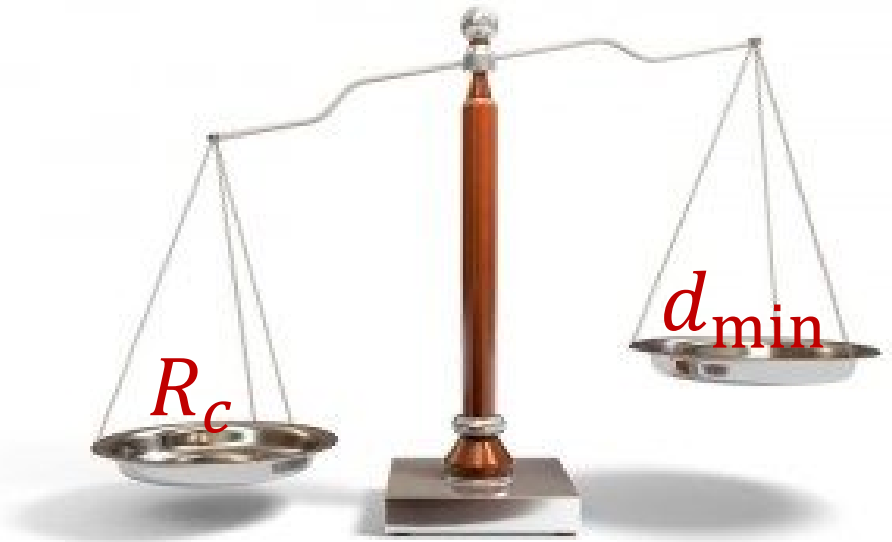
Performance Measures

□ Code rate

- The higher the value of R_c , the more efficient the coding scheme is, which means the higher the effective data rate can be achieved.

□ Minimum distance

- The larger the value of d_{\min} , the higher the error detection/correction capability.



Unit 9.4

Generator and Parity-Check Matrices

Binary Linear Codes

□ A *linear* code is defined by the *generator matrix*, G .

- $k \times n$ matrix
- Each entry is 0 or 1.

□ Encoding is done by :

$$c = uG.$$



(u and c are *row* vectors.)

□ For *systematic encoding*,

$$G = [I_k \mid P].$$

- I_k is the $k \times k$ identity matrix.
- G is said to be in *standard form*.
- In general, G need *not* contain the identity matrix, and the corresponding code is non-systematic.

Example: (5, 4) Even Parity

- ❑ Message $u = [u_1 \ u_2 \ u_3 \ u_4]$.
- ❑ Add a parity c_5 so that there is an *even number* of 1's in every codeword.
- ❑ In matrix form, $c = uG$, where

$$G = \begin{bmatrix} 1 & 0 & 0 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 \\ 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 1 \end{bmatrix}_{k \times n}$$

(G is the generator matrix.)

$$c_1 = u_1$$

$$c_2 = u_2$$

$$c_3 = u_3$$

$$c_4 = u_4$$

$$c_5 = u_1 + u_2 + u_3 + u_4$$

Encoding: An Injective, Linear Mapping

- Encoding of a linear code is a **linear function**:

$$f: \mathbb{B}^k \rightarrow \mathbb{B}^n,$$

where

- $f(u) = uG$;
 - \mathbb{B}^m is the set of all binary m -vectors.
- The mapping should be **injective**.
 - That means, no two inputs map to the same output.
 - Remark: G needs to be of full row rank, (i.e., the rows are linearly independent).

- Example:

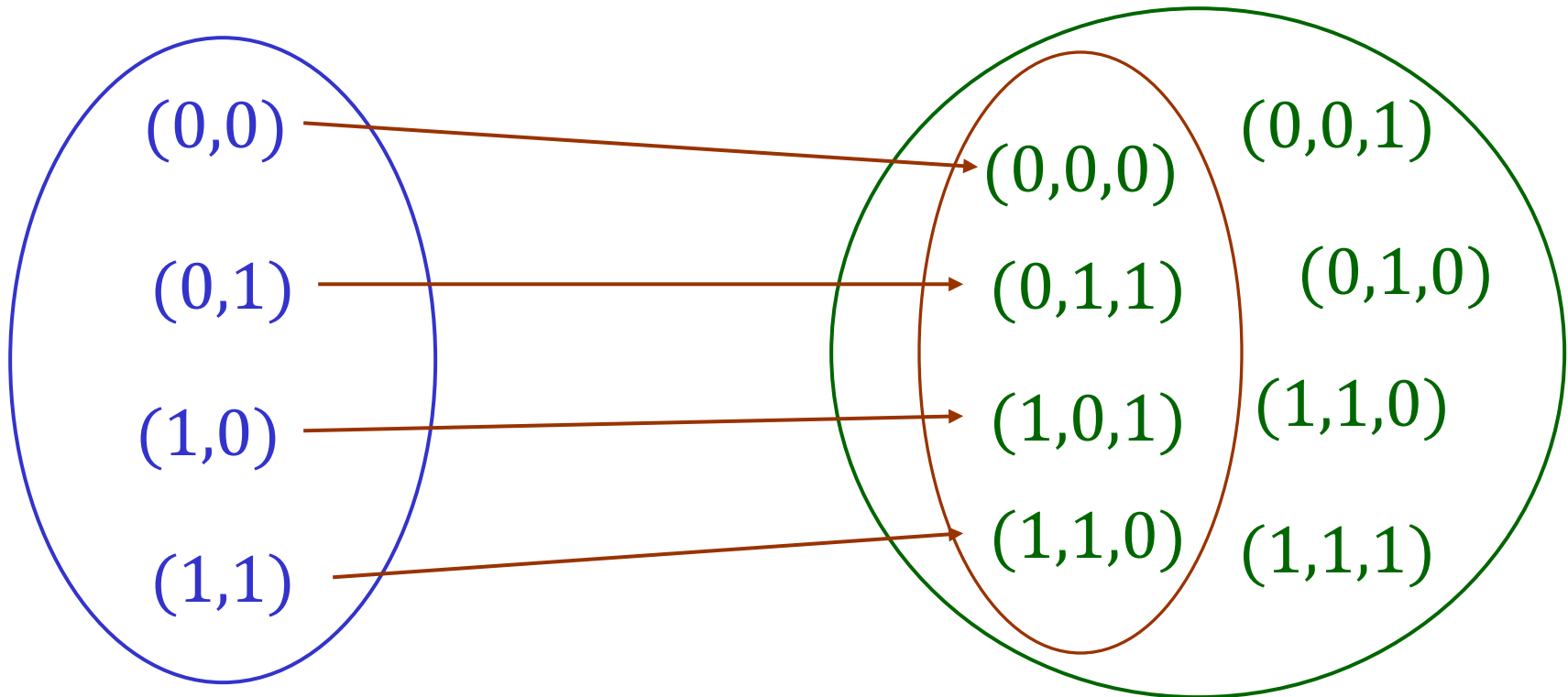
- $G = \begin{bmatrix} 1 & 1 & 0 & 0 \\ 0 & 1 & 1 & 1 \\ 1 & 0 & 1 & 1 \end{bmatrix}.$

- Consider two messages
 - $u_1 = [1 \ 1 \ 0]$
 - $u_2 = [0 \ 0 \ 1]$
- Both of them map to the same codeword:
 - $u_1 G = [1 \ 0 \ 1 \ 1]$
 - $u_2 G = [1 \ 0 \ 1 \ 1]$
- Ambiguity arises in decoding.

(3,2) Even Parity

Domain: \mathbb{B}^2

Co-domain: \mathbb{B}^3



The range (i.e., the code) is a vector subspace of \mathbb{B}^3 .

Linear Code is a Subspace of \mathbb{B}^n

□ (Closed under vector addition)

- For any binary linear code, the *sum* (i.e., XOR) of any two codewords is a *codeword*.
 - Consider two codewords, c_u and c_v .
 - $c_u + c_v = uG + vG = (u + v)G$, which is a codeword.

□ (Closed under scalar multiplication)

- Any codeword c multiplied by a *scalar* (i.e., 0 or 1) is also a codeword.
 - $c \times 1 = c$ (a codeword)
 - $c \times 0 = \mathbf{0}$ (a codeword due to zero-in zero-out)
- Note that *binary multiplication is the same as logical AND*.

Fast Method to Determine d_{\min}

Theorem: Given any linear code,

$$d_{\min} = \text{min. weight of all non-zero codewords}$$

Proof: Pick two distinct codewords x and y .

$$d(x, y) = w(x + y) \quad (x_i + y_i = 1 \text{ if the two bits are different})$$

$$= w(z) \text{ for some } z \in \mathcal{C}. \quad (\text{property of linear code})$$

Note that z is non-zero since $x \neq y$. *Q.E.D.*

- Verify it using the (7,4) Hamming code in the previous slide!

Examples

□ (5, 2) repetition with extra parity

- Codewords: 00000, 01011, 10101, 11110
- The minimum weight of non-zero codewords is 3.
- $d_{\min} = 3$

□ (5, 4) Even Parity

- Codewords: 00000, 00011, 00101, 00110, 01001, 01010, 01100, 01111, 10001, 10010, 10100, 10111, 11000, 11011, 11101, 11110.
- The minimum weight of non-zero codewords is 2
- $d_{\min} = 2$

Example: (5,4) Even Parity (cont'd)

- Clearly, any codeword satisfies

$$c_1 + c_2 + c_3 + c_4 + c_5 = 0.$$

- This is called the **parity-check equation**.

- Represented in matrix form,


$$c H^T = 0,$$

where

$$H = [1 \ 1 \ 1 \ 1 \ 1]$$

is called the **parity-check matrix**.

This equation can be used to check whether a given vector is a codeword or not.



Parity-Check Matrices

□ Let G be a generator matrix of an (n, k) code.

- It is a $k \times n$ matrix.

□ A parity-check matrix H is an $r \times n$ matrix satisfying $G H^T = 0$.

- It is not unique.
- Recall $r = n - k$ is the number of redundant bits.

□ For a systematic code, the generator matrix is of the form

$$G = [I_k \mid A].$$

□ A parity-check matrix is given by

$$H = [A^T \mid I_r].$$

- Caution: This applies only to binary codes.

Re-visit (Repetition with an Extra Parity)

- $c_1 = c_3 = u_1, c_2 = c_4 = u_2, c_5 = u_1 + u_2$
- This is a systematic matrix, which can be expressed as

$$c = uG = [u_1 \quad u_2] \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 1 & 1 \end{bmatrix}. \quad \boxed{G \text{ is } k \times n}$$

- According to the previous slide, the parity-check matrix can be expressed as

$$H = \begin{bmatrix} 1 & 0 & 1 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \\ 1 & 1 & 0 & 0 & 1 \end{bmatrix}. \quad \boxed{H \text{ is } r \times n}$$

- All codewords satisfy the parity-check equation:
$$cH^T = 0.$$

Re-visit (Repetition with an Extra Parity)

$$\square \quad cH^T = [c_1 \quad c_2 \quad c_3 \quad c_4 \quad c_5] \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} = 0.$$

\square Parity check equations:

1. $c_1 + c_3 = 0$ (repetition)
2. $c_2 + c_4 = 0$ (repetition)
3. $c_1 + c_2 + c_5 = 0$. (c_5 is parity)

Error Detection by Checking r Parities

$$x = (x_1, x_2, \dots, x_N) \xrightarrow{\text{Channel}} y = (y_1, y_2, \dots, y_N)$$

- The receiver computes $s = yH^T$ to check the parities.
 - s is an r -vector, which corresponds to r parity-check equations.
 - If $s_i \neq 0$, then the i -th parity-check equation does not hold.
- s is called the **syndrome**.

$$s \begin{cases} = 0 & \text{no error is detected.} \\ \neq 0 & \text{error is detected.} \end{cases}$$

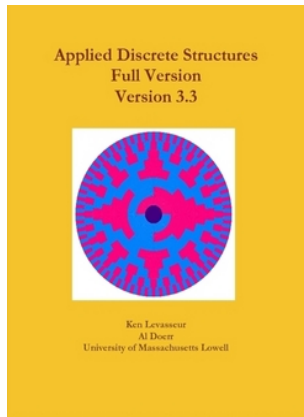
Re-visit (Repetition with an Extra Parity)

- Suppose $y = (0, 1, 0, 0, 1)$ is received.
- Compute the syndrome:

$$s = yH^T = [0 \quad 1 \quad 0 \quad 0 \quad 1] \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} = [0 \quad \textcolor{red}{1} \quad 0].$$

- Since the syndrome is non-zero, error is detected.
- Only the **second** parity-check equation does not hold, i.e.,
 1. $c_1 + c_3 = 0$
 2. $c_2 + c_4 \neq 0$ (i.e., either c_2 or c_4 is in error)
 3. $c_1 + c_2 + c_5 = 0$.
- If one bit is in error, then it must be c_4 .
 - Otherwise, if c_2 is in error, the third equation cannot hold.

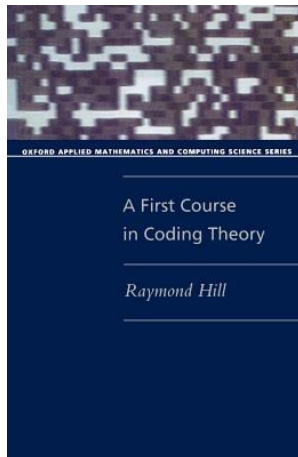
Recommended Reading



- ❑ Section 15.5, K. Levasseur and A. Doerr, *Applied Discrete Structures*, lulu.com, 2017.

- Available online:

- <http://faculty.uml.edu/klevasseur/ads/>



- ❑ Chapters 5-7, R. Hill, *A First Course in Coding Theory*, Oxford 1986.