

EGB240: Electronic Design

Assessment 2: Digital Voice Recorder

Executive Summary

The aim of this project is to document the implementation of an input conditioning circuitry for a digital voice recorder on a Teensy 2.0 board. Due to real world limitations implementation of the analogue signal conditioning cannot be done therefore its process is documented.

A series of filters are designed in which the output of the microphone is passed through. The signal is first passed through a high pass filter was used it was then passed through the 6th order Chebyshev filter designed on MATLAB and ltspice.

Before implementation of the filter the components value and other parameters were calculated and simulated in MATLAB. These steps will be covered in more detail with in this design documentation

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Introduction

This report is a technical documentation of the design process, methodology and other research that took place to complete the designated task.

The task given was to simulate input analogue conditioning circuitry for the Digital Voice Recorder (DVR). The task is to be carried out on the Teensy 2.0 microcontroller board which is provided by QUT. The analogue input conditioning circuitry that is designed will be interfaced to the microcontroller via the analogue to digital converter (ADC) peripheral as shown below.

[1] The analogue input condition circuitry is based on the TL974 operational amplifier which has also been provided.

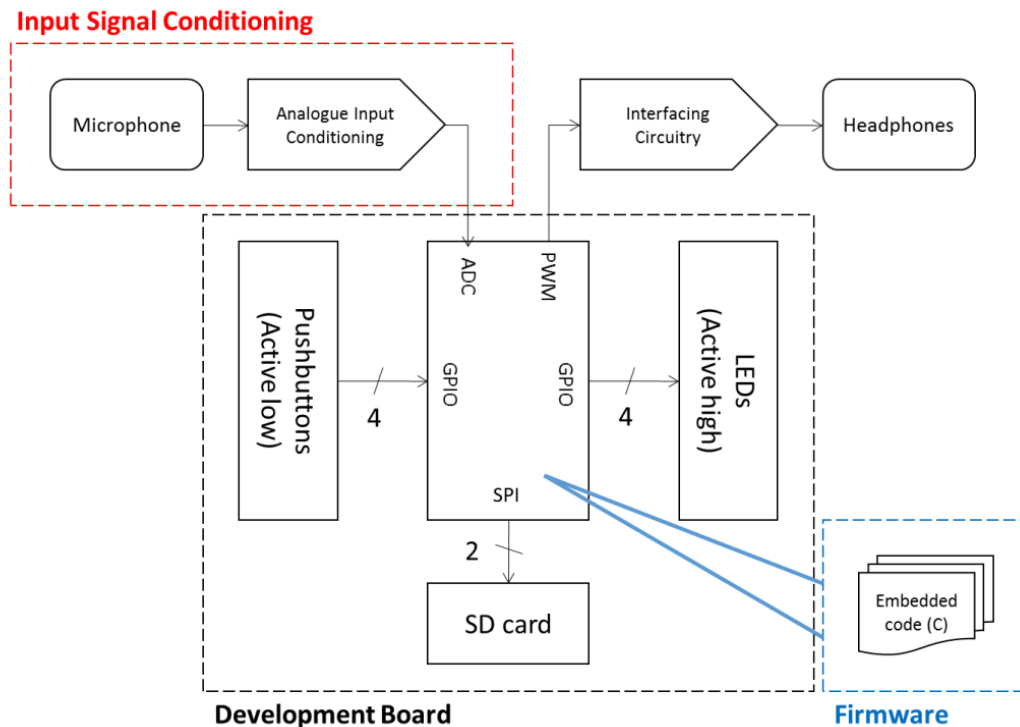


Fig. 1 Development board [1]

Background/Literature review

About Human Speech

The general human hearing range is from 20 – 20kHz the female voice and male voice ranges however are between the ranges 350Hz to 17kHz and 100Hz to 8kHz respectively. Microphones are designed to measure frequencies from as low as 50Hz up to 20kHz. [2] The figures shown below show the characteristics of human hearing and human speech

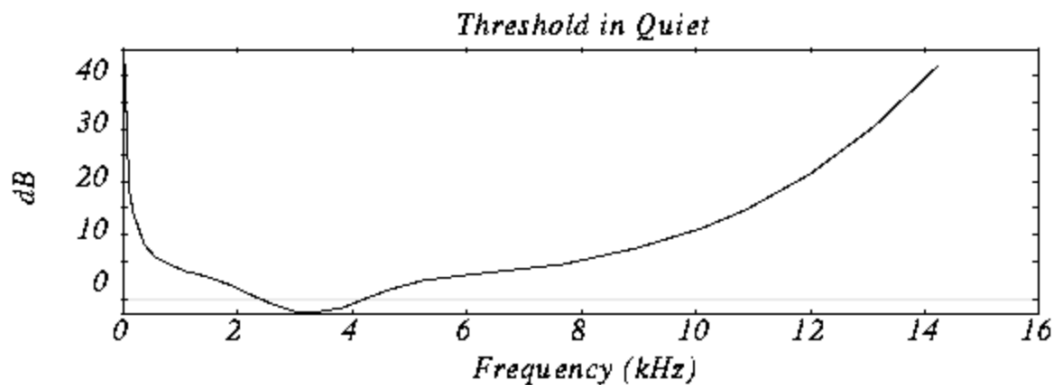


Fig. 2 Human hearing [3]

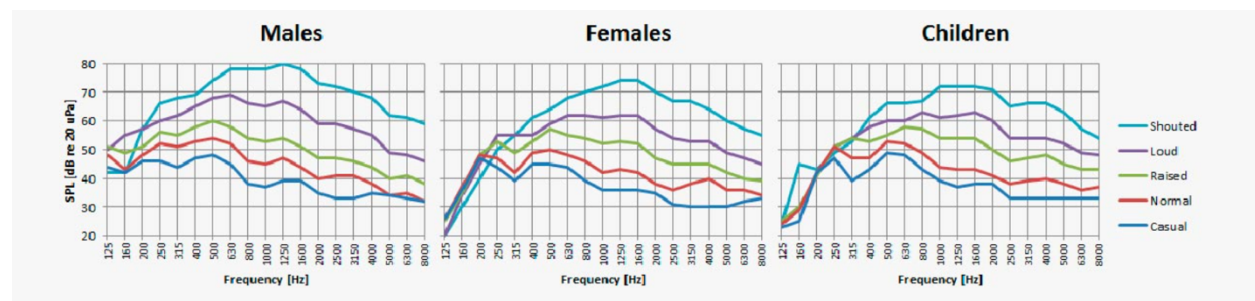


Fig. 3 Human speech [6]

The human voice produces sound which is a form of energy and oscillates and vibrates which can be converted to voltages form using a microphone.

Analog and Digital signals

Analog signals are varying amounts of information against time. These include audio, video and other forms of data. The characteristic of analog waves is that the signals are continuous [4]. Digital signals however are discrete in value and in time and are represented by binary numbers and consistent of different voltage values.

Difference Between Analog And Digital Signal	
Analog Signals	Digital Signals
Continuous signals	Discrete signals
Represented by sine waves	Represented by square waves
Human voice, natural sound, analog electronic devices are few examples	Computers, optical drives, and other electronic devices
Continuous range of values	Discontinuous values
Records sound waves as they are	Converts into a binary waveform.
Only be used in analog devices.	Suited for digital electronics like computers, mobiles and more.

Fig. 4 Differences between signals [5]

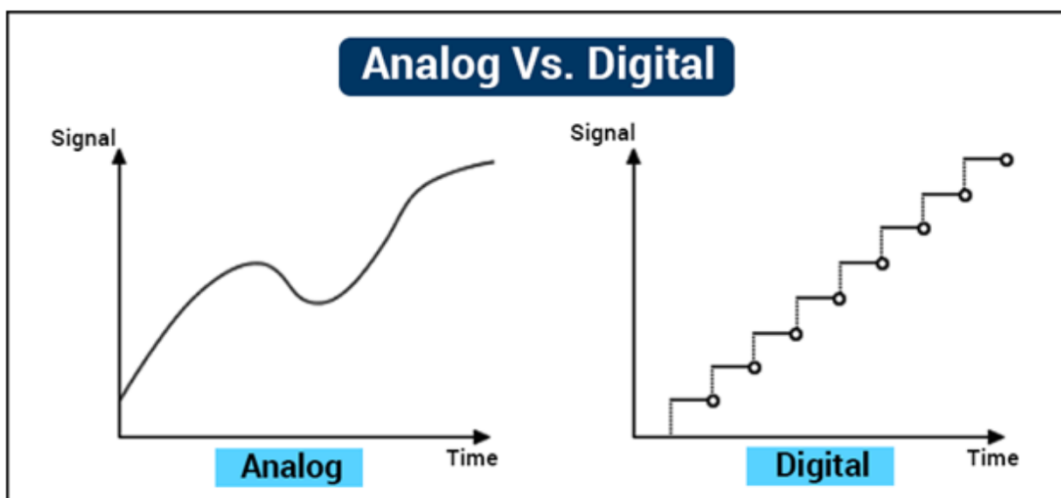


Fig. 5 Comparing analogue and digital signals [5]

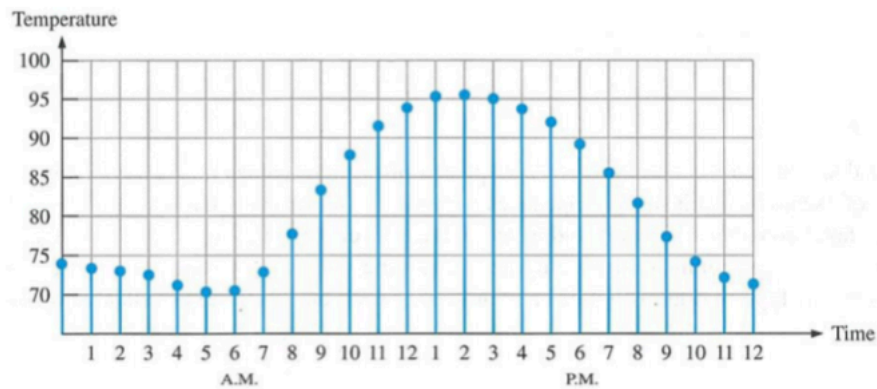


Fig. 6 ADC sampling [1]

Op-amps

Operational amplifiers also known as op-amps are voltage amplifying devices [7]. It's inputs consist of a non-inverting input with voltage (V_+) and an inverting input with voltage (V_-) as shown by the diagrams below.

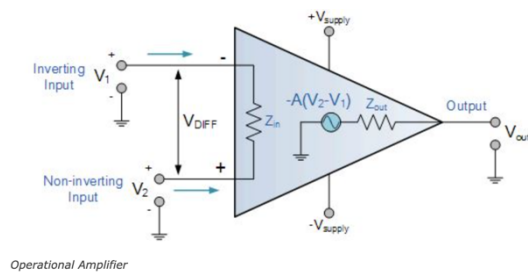


Fig. 7 Operation amplifier [7]

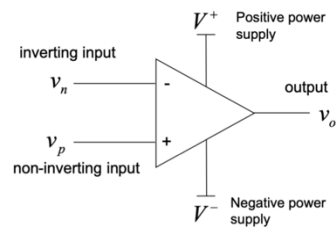
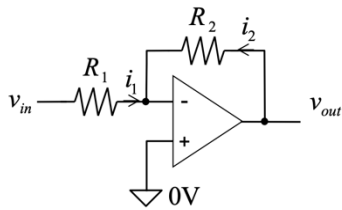


Fig. 8 Operation amplifier [1]

Operation amplifiers can be structured in a few ways such as inverting, non-inverting, summing, differential etc. The ones used in this task are mentioned below.

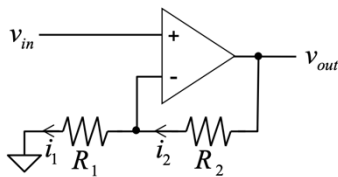
1) Inverting amplifier



- inverting amplifier inverts the input signal and produces a gain equal to $-R_2/R_1$ [1].

Fig. 9 Inverting amplifier [1]

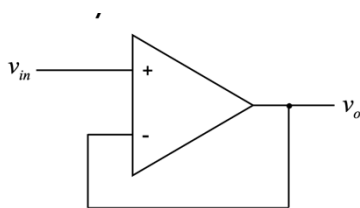
2) Non-inverting amplifier



- It amplifies the input signal without inverting the signal [1].

Fig. 10 Non-inverting amplifier [1]

3) Unity gain follower



- Used as a buffer because of its isolating properties [1].

Fig. 11 Unity gain follower [1]

Filters

Filters can be categorised as

1. Active filters
2. Passive filters

Passive filters are made using passive components such as resistors, capacitors and inductors. These filters cannot amplify a signal. Active filters use active components such as transistors and op-amps and need a power supply so they can amplify the signal [1].

Filters can also be classified as low pass, high pass and band pass filters. For this task we are using an active low pass filter. Specifically, a Chebyshev filter.

Chebyshev and Butterworth filters

Butterworths exhibit a flat passband with no ripple. The roll off is smooth and has a good phase response. Shown below is a diagram of a Butterworth filter [9].

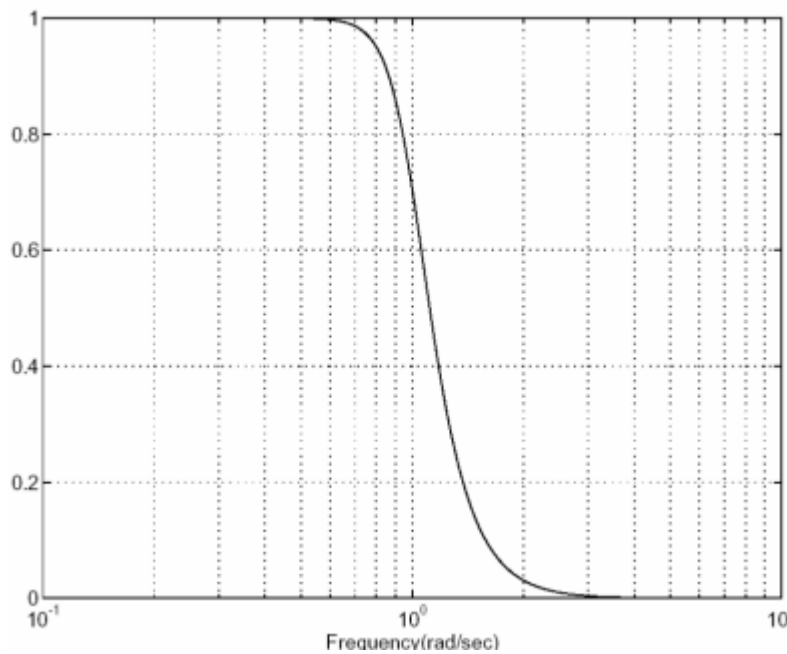


Fig. 12 Butterworth filter response [9]

The frequency response of a Butterworth filter can be calculated by the equation shown below.

$$|H(\omega)| = \frac{1}{\sqrt{1 + \varepsilon^2 \left(\frac{\omega}{\omega_p} \right)^{2n}}}$$

ε = adjustment factor for max. passband attenuation
 ω_p = cut off frequency at edge of passband

Fig. 13 Equation for frequency response of Butterworth filter [1]

Chebyshev filters are analog filters that has a faster roll-off by allowing ripple in the frequency response. As the ripple increases the roll-off becomes sharper. The con of using a Chebyshev is that they have poor phase response.

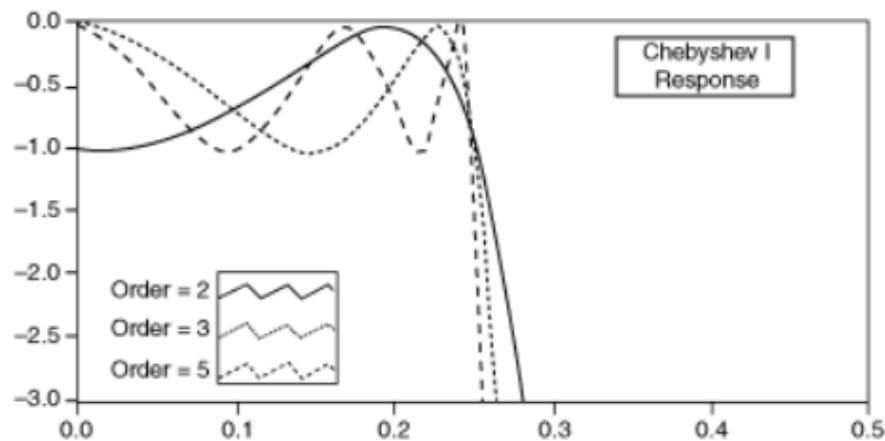


Fig. 14 Chebyshev filter frequency response [8]

The frequency response of a Chebyshev filter can be calculated by the equation below.

$$|H(\omega)| = \frac{1}{\sqrt{1 + \epsilon^2 C_n^2\left(\frac{\omega}{\omega_p}\right)}}$$

Fig. 15 Equation for frequency response of Chebyshev filter [1]

Chebyshev filters produce better results than the Butterworth filter for this task because it has a step roll-off response than a Butterworth filter as shown in the diagram below.

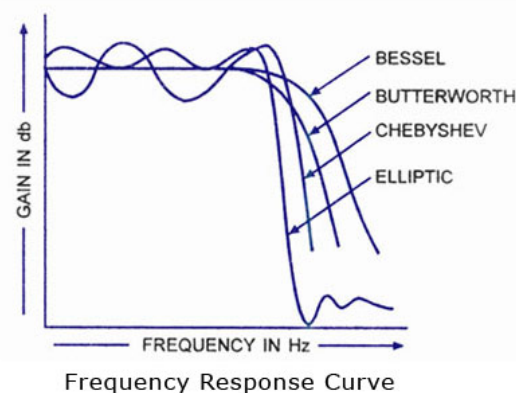


Fig. 16 Graph of different filter types [9]

Task

Input filter

As mentioned above the use of a Chebyshev filter was considered over a Butterworth filter. Specifically, a 6th order filter. This filter was implemented in MATLAB to receive proper values. The equations shown below were used for the MATLAB calculations

Stop band frequency as per Shannon Nyquist theorem,

$$f_s = \frac{\text{sampling frequency}}{2} = \frac{15625}{2} = 7812.5$$

Maximum attenuation in the stop band

$$\text{Attenuation} = 20 \log\left(\frac{1}{256}\right) = 48\text{dB}$$

To find the coefficients to use for the spice simulation the function shown below was used in MATLAB.

```
fsamp = 15.625e3;

fs = fsamp/2;
fp = 3.8e3;

ws = 2*pi*fs;
wp = 2*pi*fp;

Amin = 20*log10(1/2^8);
Amax = 1.5;

[n_butt, wn_butt] = buttord(wp, ws, Amax, Amin, 's');
[n_cheb, wn_cheb] = cheblord(wp, ws, Amax, Amin, 's');

[b,a] = butter(n_butt, wn_butt, 'low', 's');
G = tf(b,a);

figure(1)
h = bodeplot(G);
setoptions(h, 'FreqUnits','Hz');
grid on;
hold;

[b,a] = cheby1(n_cheb, Amax, wn_cheb, 'low', 's');
H = tf(b,a);

h = bodeplot(H);
setoptions(h, 'FreqUnits','Hz');
grid on;

[z, p, k] = tf2zpk(b, a);

a1A = -p(1) - p(2);
a2A = p(1)*p(2);

a1B = -p(3) - p(4);
a2B = p(3)*p(4);

a1C = -p(5) - p(6);
a2C = p(5)*p(6);

a1D = -p(7) - p(8);
a2D = p(7)*p(8);
% A values
wnA = sqrt(a2A);
QA = wnA/a1A;

n = (2*QA)^2;

C1A = 2.7e-9;
C2A = n*C1A;

R1A = 1/(wnA*C1A*sqrt(n));
R2A = R1A;

% B values
wnB = sqrt(a2B);
QB = wnB/a1B;

n = (2*QB)^2;

C1B = 2.7e-9;
C2B = n*C1B;

R1B = 1/(wnA*C1B*sqrt(n));
R2B = R1B;

% C values
wnC = sqrt(a2C);
QC = wnA/a1C;

n = (2*QC)^2;

C1C = 2.7e-9;
C2C = n*C1C;
```

```

R1C = 1/(wnC*C1C*sqrt(n));
R2C = R1C;

% D values

wnD = sqrt(a2D);
QD = wnD/a1D;

n = (2*QD)^2;

C1D = 2.7e-9;
C2D = n*C1D;

R1D = 1/(wnA*C1D*sqrt(n));
R2D = R1D;

fprintf(1, 'R1A = %g, R2A = %g, C1A = %g, C2A = %g \n', R1A, R2A, C1A, C2A);
fprintf(1, 'R1B = %g, R2B = %g, C1B = %g, C2B = %g \n', R1B, R2B, C1B, C2B);
fprintf(1, 'R1C = %g, R2C = %g, C1C = %g, C2C = %g \n', R1C, R2C, C1C, C2C);
fprintf(1, 'R1D = %g, R2D = %g, C1D = %g, C2D = %g \n', R1D, R2D, C1D, C2D);

%New values
R1A = 2700;
R2A = R1A;
R1B = 7500;
R2B = R1B;
R1C = 12000;
R2C = R1C;
R1D = 13000;
R2D = R1D;

C2A = 68*10^-9;
C2B = 10*10^-9;
C2C = 3.9*10^-9;
C2D = 2.7*10^-9;

% Check

K1test = 1;
numA = [0 0 K1test/(R1A*R2A*C1A*C2A)];
denA = [1 (1/(R1A*C2A) + 1/(R2A*C2A) + (1-K1test)/(R2A*C1A)) 1/(R1A*R2A*C1A*C2A)];
HtestA = tf(numA, denA);

K2test = 1;
numB = [0 0 K2test/(R1B*R2B*C1B*C2B)];
denB = [1 (1/(R1B*C2B) + 1/(R2B*C2B) + (1-K2test)/(R2B*C1B)) 1/(R1B*R2B*C1B*C2B)];
HtestB = tf(numB, denB);

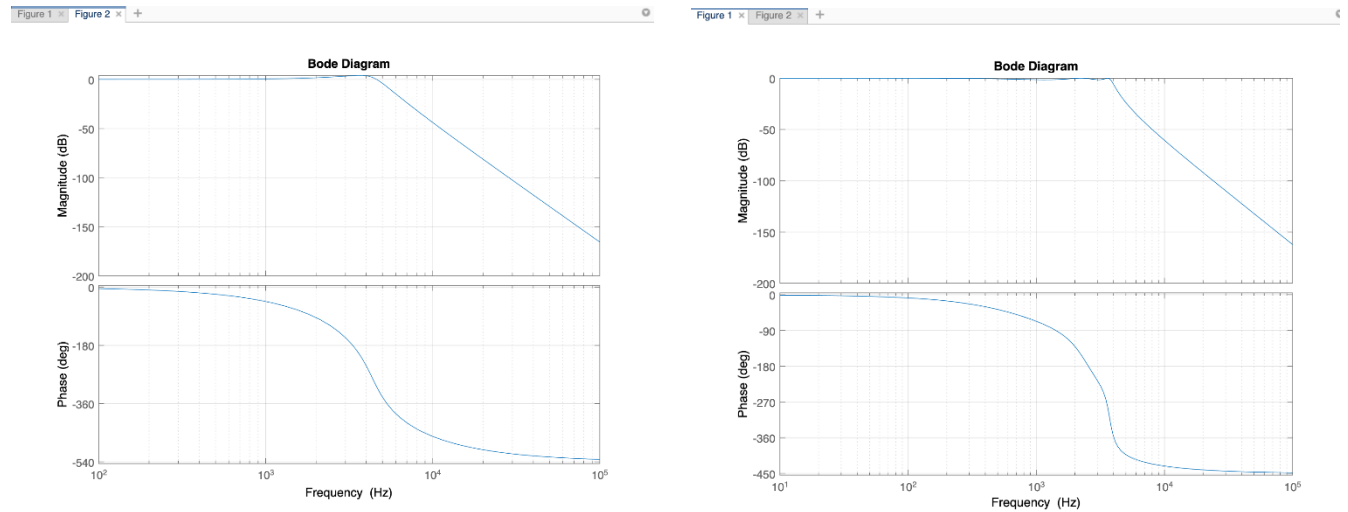
K3test = 1;
numC = [0 0 K3test/(R1C*R2C*C1C*C2C)];
denC = [1 (1/(R1C*C2C) + 1/(R2C*C2C) + (1-K3test)/(R2C*C1C)) 1/(R1C*R2C*C1C*C2C)];
HtestC = tf(numC, denC);

%K4test = 1;
%numD = [0 0 K4test/(R1D*R2D*C1D*C2D)];
%denD = [1 (1/(R1D*C2D) + 1/(R2D*C2D) + (1-K4test)/(R2D*C1D)) 1/(R1D*R2D*C1D*C2D)];
%HtestD = tf(numD, denD);

Htest = HtestA*HtestB*HtestC; %HtestD;

figure(2);
h = bodeplot(Htest);
setoptions(h, 'FreqUnits', 'Hz');
grid on;

```



Figs. 17 MATLAB simulations

As shown by the images above the cheby function was used to calculate the coefficients to be used with the filter along with the components values which were determined using the excel sheet found in appendix A.

LTspice simulation

The calculated component values were set up in the LTspice simulation and shown in the figure below. The first figure shows an overview of the 6th order chebyshev filter and the following figures show its components in more detail.

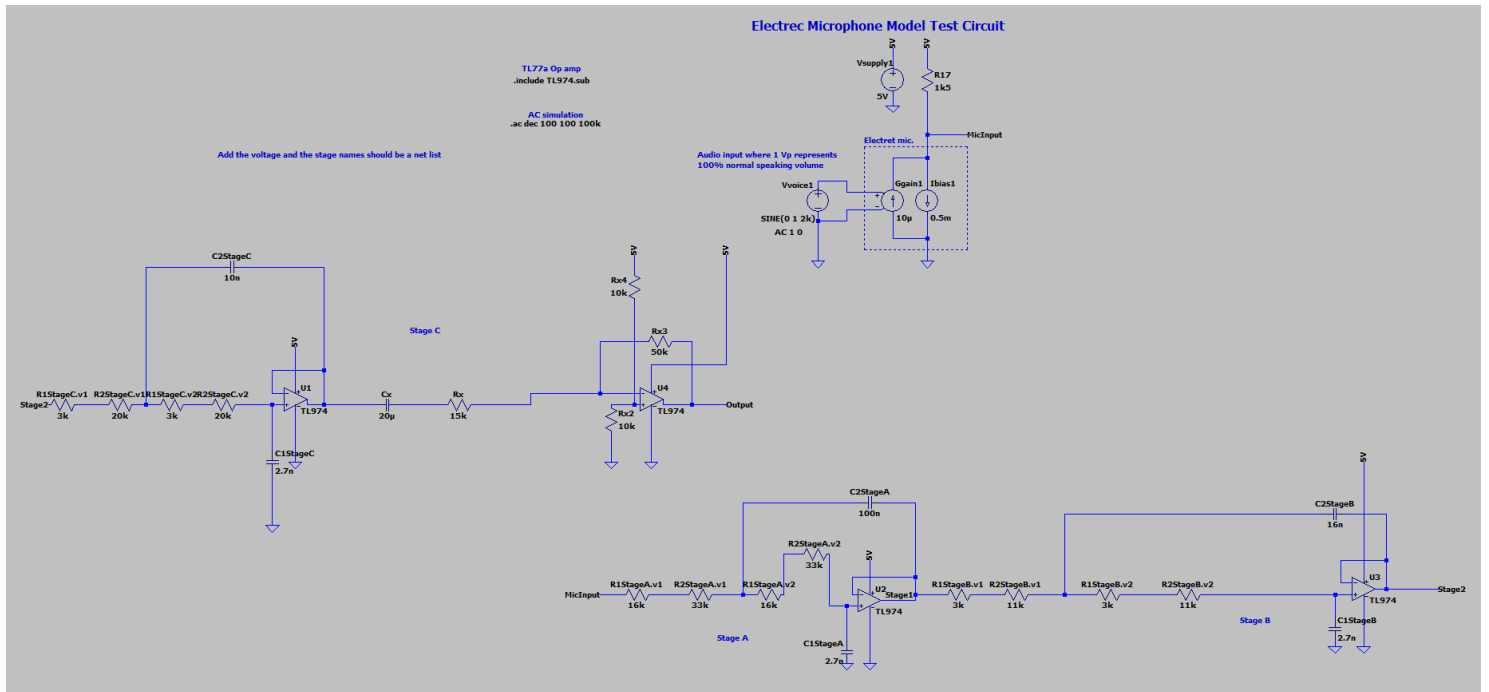


Fig. 18 LTspice simulation

Stage A and Stage B

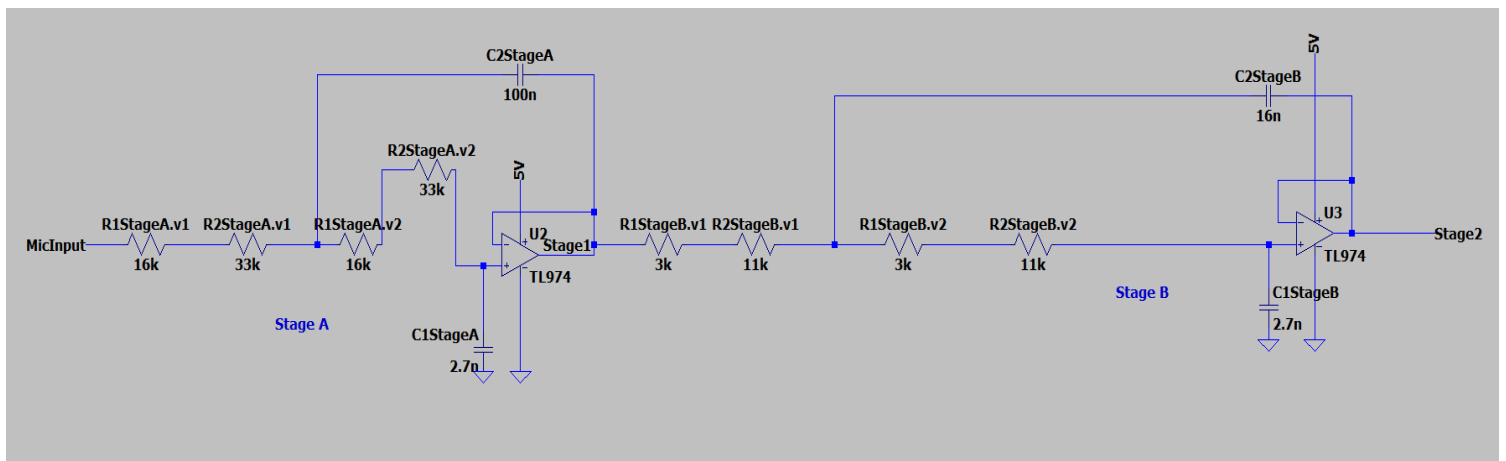


Fig. 19 LTspice simulation

Stage C

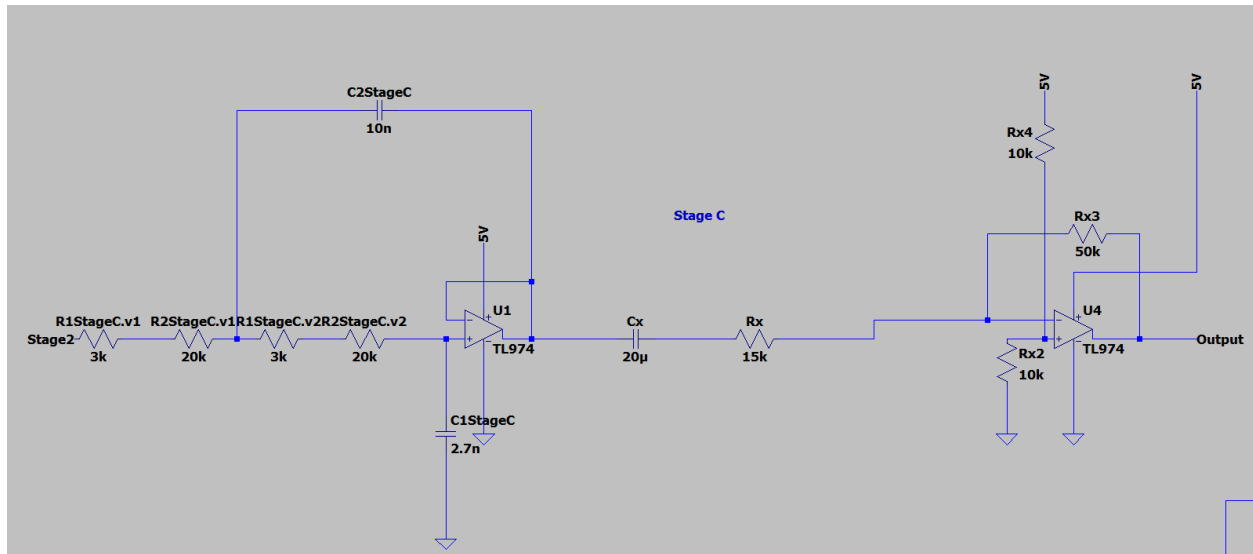


Fig. 20 LTspice simulation

Microphone

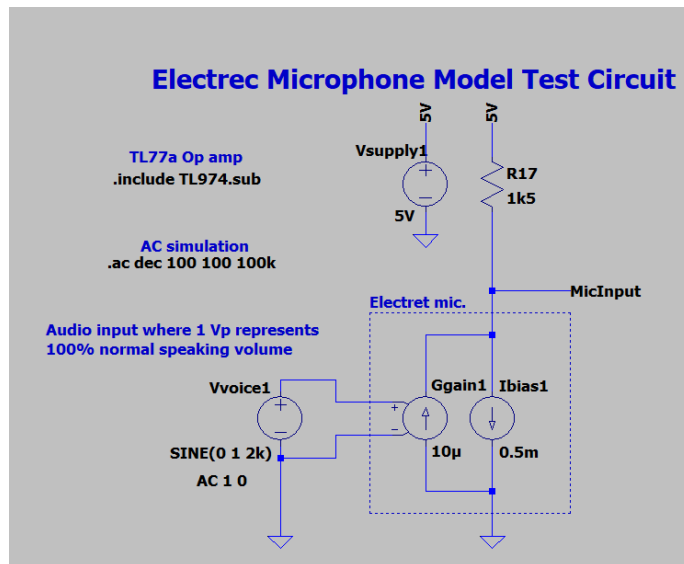


Fig. 21 LTspice simulation

Verification of simulation

The simulations performed in LTspice are shown below for the output, stage 2, the input and a final graph of all waveforms.

As seen by the results the filter works as intended.

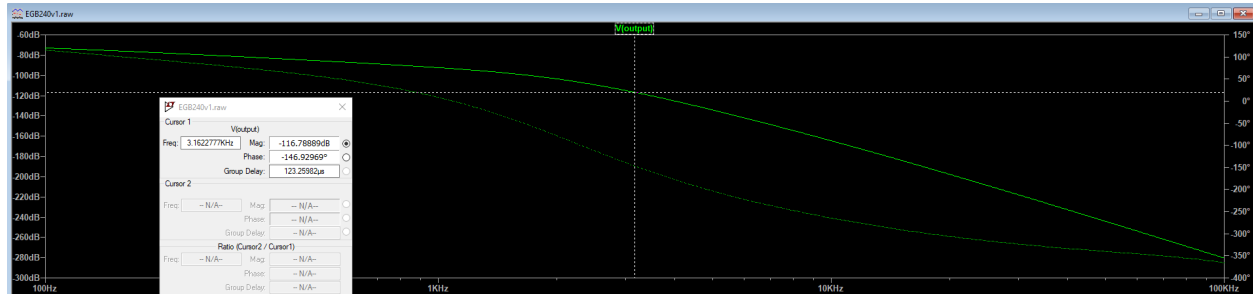


Fig. 22 LTspice simulation (output)

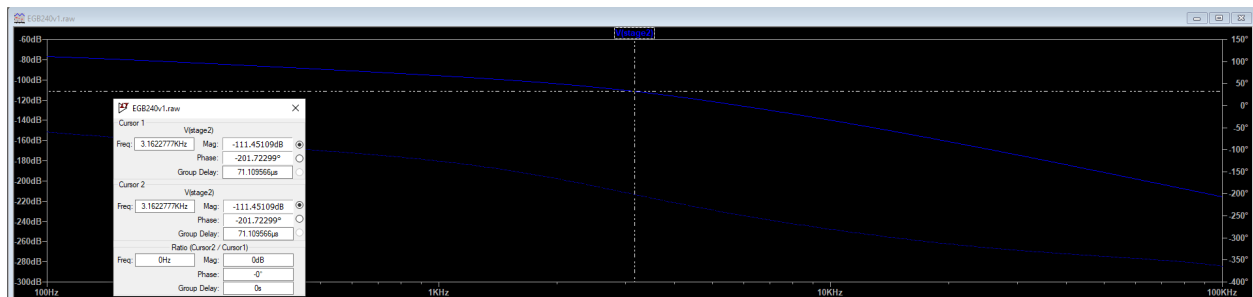


Fig. 23 LTspice simulation (stage 2)

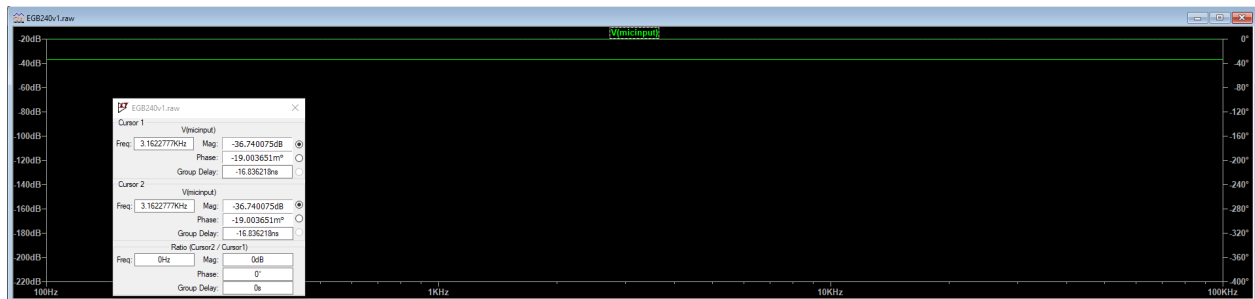


Fig. 24 LTspice simulation (input)

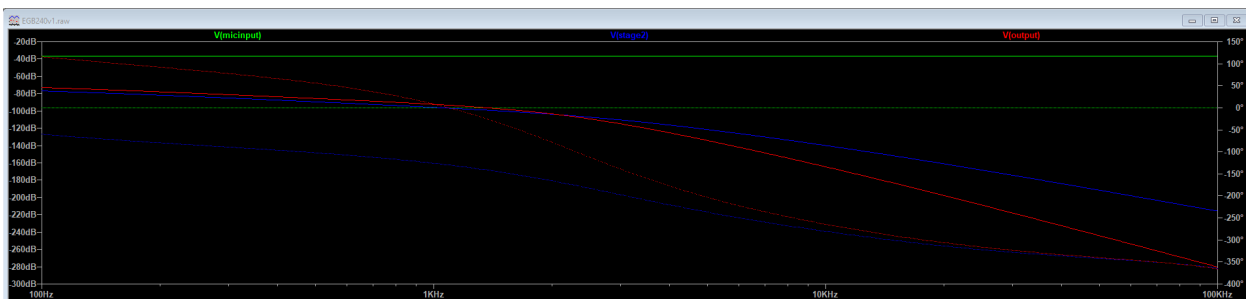


Fig. 25 LTspice simulation

Conclusion

The filter works as intended but for future projects the use of a Butterworth filter or different order of Chebyshev can be explored and different set ups so that less error occurs throughout the process.

References

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- [9] *Ele.uri.edu*, 2020. [Online]. Available: https://www.ele.uri.edu/courses/ele314/handouts/YS06_Classicfilters.pdf. [Accessed: 02- Jun- 2020].

Appendices

Appendix A – Excel sheets used

Q	1.00				
wn	26509				
C1	2.7	nF			
	m	n	C2 (nF)	R1 (kΩ)	R2 (kΩ)
0	1.00	4.00	10.80	6.99	6.99 *
1	1.10	4.01	10.82	6.65	7.32
2	1.21	4.04	10.90	6.32	7.65
3	1.33	4.08	11.03	5.99	7.98
4	1.47	4.15	11.20	5.66	8.31
5	1.62	4.23	11.43	5.34	8.63
6	1.78	4.34	11.72	5.03	8.94
7	1.96	4.47	12.06	4.72	9.25
8	2.15	4.62	12.47	4.43	9.54
9	2.37	4.79	12.94	4.14	9.83
10	2.61	4.99	13.48	3.87	10.10
11	2.87	5.22	14.10	3.61	10.36
12	3.16	5.48	14.79	3.36	10.61
13	3.48	5.77	15.57	3.12	10.85
14	3.83	6.09	16.45	2.89	11.08
15	4.22	6.45	17.43	2.68	11.29
16	4.64	6.86	18.51	2.48	11.49
17	5.11	7.30	19.72	2.29	11.68
18	5.62	7.80	21.06	2.11	11.86
19	6.19	8.35	22.55	1.94	12.03
20	6.81	8.96	24.19	1.79	12.18
21	7.50	9.63	26.01	1.64	12.33
22	8.25	10.38	28.01	1.51	12.46
23	9.09	11.20	30.23	1.39	12.59
24	10.00	12.10	32.67	1.27	12.70

Q	0.65				
wn	26509				
C1	2.7	nF			
	m	n	C2 (nF)	R1 (kΩ)	R2 (kΩ)
0	1.00	1.70	4.60	10.70	10.70 *
1	1.10	1.71	4.61	10.19	11.22
2	1.21	1.72	4.64	9.68	11.73
3	1.33	1.74	4.70	9.17	12.23
4	1.47	1.77	4.77	8.67	12.73
5	1.62	1.80	4.87	8.18	13.22
6	1.78	1.85	4.99	7.70	13.70
7	1.96	1.90	5.14	7.24	14.17
8	2.15	1.97	5.31	6.79	14.62
9	2.37	2.04	5.51	6.35	15.06
10	2.61	2.13	5.74	5.93	15.48
11	2.87	2.22	6.01	5.53	15.88
12	3.16	2.33	6.30	5.14	16.26
13	3.48	2.46	6.63	4.78	16.63
14	3.83	2.60	7.01	4.43	16.97
15	4.22	2.75	7.42	4.10	17.30
16	4.64	2.92	7.89	3.79	17.61
17	5.11	3.11	8.40	3.50	17.90
18	5.62	3.32	8.97	3.23	18.17
19	6.19	3.56	9.61	2.98	18.43
20	6.81	3.82	10.31	2.74	18.67
21	7.50	4.10	11.08	2.52	18.89
22	8.25	4.42	11.93	2.31	19.09
23	9.09	4.77	12.88	2.12	19.28
24	10.00	5.15	13.92	1.95	19.46

Q	2.88				
wn	26509				
C1	2.7	nF			
	m	n	C2 (nF)	R1 (kΩ)	R2 (kΩ)
0	1.00	33.16	89.54	2.43	2.43 *
1	1.10	33.24	89.75	2.31	2.54
2	1.21	33.47	90.37	2.19	2.66
3	1.33	33.86	91.41	2.08	2.77
4	1.47	34.40	92.88	1.97	2.89
5	1.62	35.11	94.79	1.86	3.00
6	1.78	35.99	97.17	1.75	3.11
7	1.96	37.05	100.02	1.64	3.21
8	2.15	38.29	103.39	1.54	3.31
9	2.37	39.74	107.30	1.44	3.41
10	2.61	41.40	111.78	1.34	3.51
11	2.87	43.29	116.88	1.25	3.60
12	3.16	45.42	122.64	1.17	3.69
13	3.48	47.82	129.12	1.08	3.77
14	3.83	50.51	136.38	1.00	3.85
15	4.22	53.51	144.48	0.93	3.92
16	4.64	56.85	153.50	0.86	3.99
17	5.11	60.56	163.52	0.79	4.06
18	5.62	64.68	174.64	0.73	4.12
19	6.19	69.24	186.95	0.67	4.18
20	6.81	74.28	200.57	0.62	4.23
21	7.50	79.86	215.62	0.57	4.28
22	8.25	86.02	232.25	0.52	4.33
23	9.09	92.82	250.61	0.48	4.37
24	10.00	100.32	270.87	0.44	4.41

>> Htest

Htest =

3.244e26

$s^6 + 8.029e04 s^5 + 3.961e09 s^4 + 1.243e14 s^3 + 2.768e18 s^2 + 3.889e22 s + 3.244e26$

Continuous-time transfer function.

Name	Value	Size	Class
a	[1,1.9135e...	1×6	double
a1A	3.6544e+03	1×1	double
a1B	9.5673e+03	1×1	double
a1C	4.0614e+04	1×1	double
a1D	5.1419e+0...	1×1	double (co...
a2A	5.5059e+08	1×1	double
a2B	2.3192e+08	1×1	double
a2C	7.0271e+08	1×1	double
a2D	6.6033e+0...	1×1	double (co...
Amax	1.5000	1×1	double
Amin	-48.1648	1×1	double
b	[0,0,0,0,7...	1×6	double
C1A	2.7000e-09	1×1	double
C1B	2.7000e-09	1×1	double
C1C	2.7000e-09	1×1	double
C1D	2.7000e-09	1×1	double
C2A	6.8000e-08	1×1	double
C2B	1.0000e-08	1×1	double
C2C	3.9000e-09	1×1	double
C2D	2.7000e-09	1×1	double
denA	[1,1.0893e...	1×3	double
denB	[1,2.8667e...	1×3	double
denC	[1,4.2735e...	1×3	double
denD	[1,5.6980e...	1×3	double
fp	3800	1×1	double
fs	7.8125e+03	1×1	double
fsamp	15625	1×1	double
G	1×1 tf	1×1	tf
H	1×1 tf	1×1	tf
h	1×1 bodeplot	1×1	resppack.b...

Htest	1×1 tf	1×1	tf
HtestA	1×1 tf	1×1	tf
HtestB	1×1 tf	1×1	tf
HtestC	1×1 tf	1×1	tf
HtestD	1×1 tf	1×1	tf
k	7.5503e+20	1×1	double
K1test	1	1×1	double
K2test	1	1×1	double
K3test	1	1×1	double
K4test	1	1×1	double
n	1.0311 - 0...	1×1	double (co...
n_butt	9	1×1	double
n_chsb	5	1×1	double
numA	[0,0.74714...	1×3	double
numB	[0,0.65844...	1×3	double
numC	[0,0.65949...	1×3	double
numD	[0,0.8.1168...	1×3	double
p	[-1.8272e+...	5×1	double (co...
QA	2.8794	1×1	double
QB	1.0000	1×1	double
QC	0.6527	1×1	double
QD	0.5077 - 0...	1×1	double (co...
R1A	2700	1×1	double
R1B	7500	1×1	double
R1C	12000	1×1	double
R1D	13000	1×1	double
R2A	2700	1×1	double
R2B	7500	1×1	double
R2C	12000	1×1	double
R2D	13000	1×1	double
wn_butt	2.8509e+04	1×1	double
wn_chsb	2.3876e+04	1×1	double
wnA	2.8509e+04	1×1	double
wnB	2.8509e+04	1×1	double

Appendix B – Extension



You recently submitted a request to QUT. This request is now closed. If needed, this enquiry can be reopened within the next 30 days. Thank you for your enquiry.

Subject

EN01 10496262 Kaluarachchi, Don Misura Minduwara - EGB240 Assignment Extension (EXT) [Science and Engineering Faculty]

Response By Email (Jasmine) (08/06/2020 07:35 PM)

Dear Don Misura Minduwara,

Your request for an Assignment Extension has been **approved**. The details of your extension are provided below:

Unit: EGB240 (2) - Electronic Design
Assignment Title: Assessment 2: Digital Voice Recorder
Original Submission Due Date: 05-Jun-2020
Revised Approved Submission Due Date: 08-Jun-2020

Please submit your assignment using the normal submission process as outlined in your unit's Blackboard site.

You are required to attach a copy of this email when submitting your assignment as it is confirmation of your approved extension.

If you do not submit your assignment by the extended due date your work will not be marked and you will receive a grade of 1 or 0% against the assessment item.

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